

APRIL • 1955

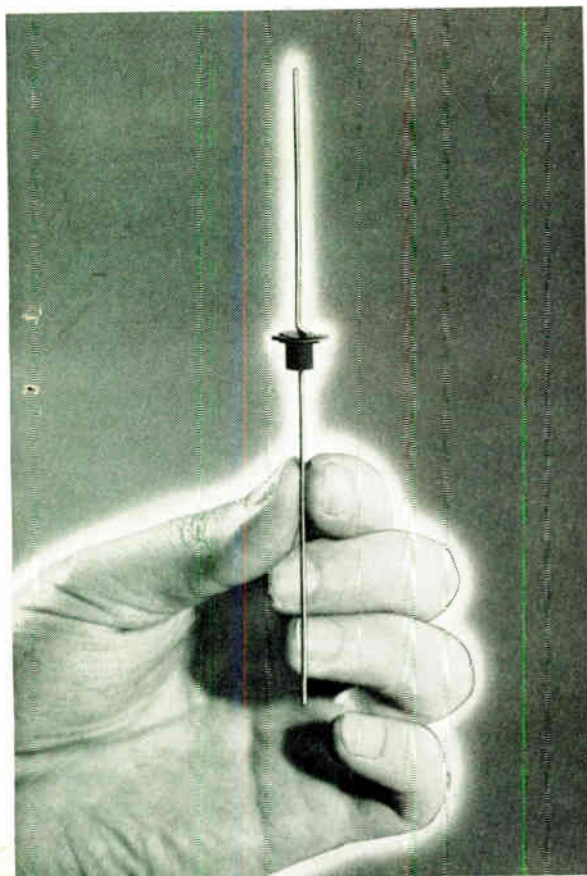
Proceedings



OF THE I R E

A Journal of Communications and Electronic Engineering

SILICON RECTIFIERS



Bogue Electric Manufacturing Co.

The rapid strides which recently have been made in the field of solid state materials and devices are typified by the development of the silicon power rectifier shown above.

Volume 43

Number 4

PART I

Survey of Magnetic Amplifiers
"M"-Type Carcinotron Tube
Power Flow in Electron Beam Devices
Germanium Surface Recombination
Graphical Filter Analysis
Noise Figure of Backward-Wave Amplifier
Induced Grid Noise
Amplification in Electron Streams
RLC Transfer Functions
Mode Control of Interdigital Magnetrans
Minority Carrier Lifetime Measurement
Transactions Abstracts
Abstracts and References
Annual Index to Transactions
Annual Index to Convention Record

TABLE OF CONTENTS INDICATED BY BLACK-AND-WHITE MARGIN, FOLLOWS PAGE 80A

The Institute of Radio Engineers

**LEADERS IN
MINIATURIZATION
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TWENTY YEARS...**



MINIATURIZED TRANSFORMER COMPONENTS

**FROM
STOCK**

Items below and 650 others in our catalog A.

HERMETIC SUB-MINIATURE AUDIO UNITS

These are the smallest hermetic audios made.

Dimensions . . . 1/2 x 11/16 x 29/32 . . . Weight .8 oz.

TYPICAL ITEMS

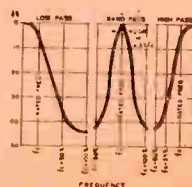
Type No.	Application	MIL Type	Pri. Imp. Ohms	Sec. Imp. Ohms	DC in Pri MA	Response ± 2 db (Cyc.)	Max. level dbm
H-30	Input to grid	TF1A10YY	50*	62,500	0	150-10,000	+13
H-31	Single plate to single grid, 3:1	TF1A15YY	10,000	90,000	0	300-10,000	+13
H-32	Single plate to line	TF1A13YY	10,000*	200	3	300-10,000	+13
H-33	Single plate to low impedance	TF1A13YY	30,000	50	1	300-10,000	+15
H-34	Single plate to low impedance	TF1A13YY	100,000	60	.5	300-10,000	+6
H-35	Reactor	TF1A20YY	100 Henries-0 DC, 50 Henries-1 Ma. DC, 4,400 ohms.				
H-36	Transistor Interstage	TF1A15YY	25,000	1,000	.5	300-10,000	+10

*Can be used with higher source impedances, with corresponding reduction in frequency range and current



COMPACT HERMETIC AUDIO FILTERS

UTC standardized filters are for low pass, high pass, and band pass application in both inter-stage and line impedance designs. Thirty four stock values, others to order. Case 1-3/16 x 1-11/16 x 1-5/8 — 2-1/2 high . . . Weight 6-9 oz.

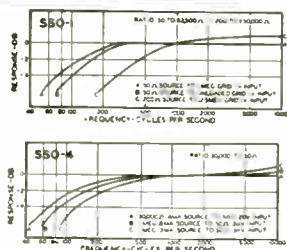
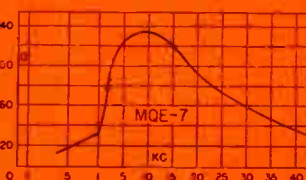


HERMETIC MINIATURE HI-Q TOROIDS

MQE units provide high Q, excellent stability and minimum hum pickup in a case only. 1/2 x 1-1/16 x 17/32 . . . weight 1.5 oz.

TYPICAL ITEMS

Type No.	Inductance	DC Max.
MQE-1	7 mhy.	135
MQE-3	20 mhy.	80
MQE-5	50 mhy.	50
MQE-7	100 mhy.	35
MQE-10	.4 hy.	17
MQE-12	.9 hy.	12
MQE-15	2.8 hy.	7.2



SUB-SUBOUNCER AUDIO UNITS

UTC Subouncer and sub-subouncer units provide exceptional efficiency and frequency range in miniature size. Constructional details assure maximum reliability. SS0 units are 7/16 x 3/4 x 43/64 . . . Weight 1/50 lb.



Type	Application	Level	Pri. Imp.	MA O.C. in Pri.	Sec. Imp.	Pri. Res.	Sec. Res.
*SS0-1	Input	+4 V.U.	200 50	0	250,000 62,500	13.5	3700
SS0-2	Interstage /3:1	+4 V.U.	10,000	0-.25	90,000	750	3250
*SS0-3	Plate to Line	+20 V.U.	10,000 25,000	3 1.5	200 500	2600	35
SS0-4	Output	+20 V.U.	30,000	1.0	50	2875	4.6
SS0-5	Reactor 50 HY at 1 mil. D.C. 4400 ohms D.C. Res.						
SS0-6	Output	+20 V.U.	100,000	.5	60	4700	3.3
*SS0-7	Transistor Interstage	+10 V.U.	20,000 30,000	.5 .5	800 1,200	850	125

* Impedance ratio is fixed, 1250:1 for SS0-1, 1:50 for SS0-3. Any impedance between the values shown may be employed.

SUBOUNCER (WIDE RANGE) AUDIO UNITS

Standard for the industry for 15 yrs., these units provide 30-20,000 cycle response in a case 7/8 dia. x 1-3/16 high. Weight 1 oz.

TYPICAL ITEMS

Type No.	Application	Pri. Imp	Sec. Imp
O-1	Mike, pickup or line to 1 grid	50, 200/250, 500/600	50,000
O-4	Single plate to 1 grid	15,000	60,000
O-7	Single plate to 2 grids, D.C. in Pri.	15,000	95,000
O-9	Single plate to line, D.C. in Pri.	15,000	50, 200/250, 500/600
O-10	Push pull plates to line	30,000 ohms plate to plate	50, 200/250, 500/600
O-12	Mixing and matching	50, 200/250	50, 200/250, 500/600
O-13	Reactor, 300 Hys.—no D.C.; 50 Hys.—3 MA. D.C., 6000 ohms		

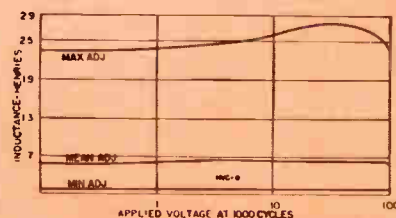
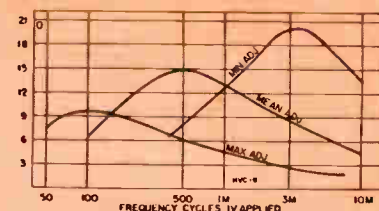


HERMETIC VARIABLE INDUCTORS

These inductors provide high Q from 50-10,000 cycles with exceptional stability. Wide inductance range (10-1) in an extremely compact case 25/32 x 1-1/8 x 1-3/16 . . . Weight 2 oz.

TYPICAL ITEMS

TYPE No.	Min. Hys.	Mean Hys.	Max. Hys.	DC Ma
HVC-1	.002	.006	.02	100
HVC-3	.011	.040	.11	40
HVC-5	.07	.25	.7	20
HVC-6	.2	.6	2	15
HVC-10	7.0	25	70	3.5
HVC-12	50	150	500	1.5



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CABLES: "ARLAB"

World Radio History

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Director, National Advisory Committee
for Aeronautics will talk about

Dr. W. A. Wildhack, Luncheon Speaker
National Bureau of Standards
will treat the subject of

Problems in Ultra High Speed Flight

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ISA

PROCEEDINGS OF THE I.R.E., April, 1955, Vol. 43, No. 4, Part 1, Published monthly by the Institute of Radio Engineers, Inc., at 1 East 79 Street, New York 21, N.Y. Price per copy: members of the Institute of Radio Engineers, \$1.00; non-members \$2.25. Yearly subscription price: to members \$9.00; to non-members in United States, Canada and U.S. Possessions \$18.00; to non-members in foreign countries \$19.00. Entered as second class matter, October 26, 1927, at the post office at Menasha, Wisconsin, under the act of March 3, 1879. Acceptance for mailing at a special rate of postage is provided for in the act of February 28, 1925, embodied in Paragraph 4, Section 412, P. L. and R., authorized October 26, 1927.

Table of Contents will be found following page 80A

World Radio History

Never before!

NOT 2: motor + gear train
BUT ONE homogeneous unit
New Power Motor-Gear-Train

1. Unique: *Not 2* separate units but a single entity. An entirely new principle—another OSTER "first."

2. More Versatile: Any output speed from 10,000 to .3 RPM.

3. Extremely High Torque Capacity: e.g., 100 #-in. at 523:1 and 1600 #-in. at 10,500:1.

4. Lighter-Smaller: e.g., 10-1/2 oz., 1-1/2" dia., 3-1/2" long at 523:1 ratio.

5. Available in 28 V and 115 V DC or in 28 V and 115 V 400 cycle AC

6. 1.50" dia. (Type 3094) illustrated. Variations include 1.25" dia. (Type 3101), 1.062"* dia. (Type 3200) and 1.75"* dia. (Type 2487).

*Available soon.

For a precision speed reducer with low backlash and low composite error at a moderate price specify an OSTER Power Motor-Gear-Train adapted to your individual application. Write for further information TODAY.



Actual Size
1.50" dia. (Type 3094)

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AVIONIC DIVISION
RACINE, WISCONSIN

Your Rotating Equipment Specialist

Other products include Actuators, AC Drive Motors, DC Motors for Special Applications, Fast Response Resolvers, Servo Torque Units, Low Inertia Servo Motors, Synchro Differentials, Two-Phase Reference Generators, Tachometer Generators and Motor Driven Blower and Fan Assemblies.

THIS IS IT!



NEW 3-WATT Blue Jacket[®] miniaturized axial-lead wire wound resistor

This power-type wire wound axial-lead Blue Jacket is hardly larger than a match head *but it performs like a giant!* It's a rugged vitreous-enamel coated job—and like the entire Blue Jacket family, it is built to withstand severest humidity performance requirements.

Blue Jackets are ideal for dip-soldered sub-assemblies . . . for point-to-point wiring . . . for terminal board mounting and processed wiring boards. They're low in

cost, eliminate extra hardware, save time and labor in mounting!

Axial-lead Blue Jackets in 3, 5 and 10 watt ratings are available without delay in any quantity you require. ★ ★ ★

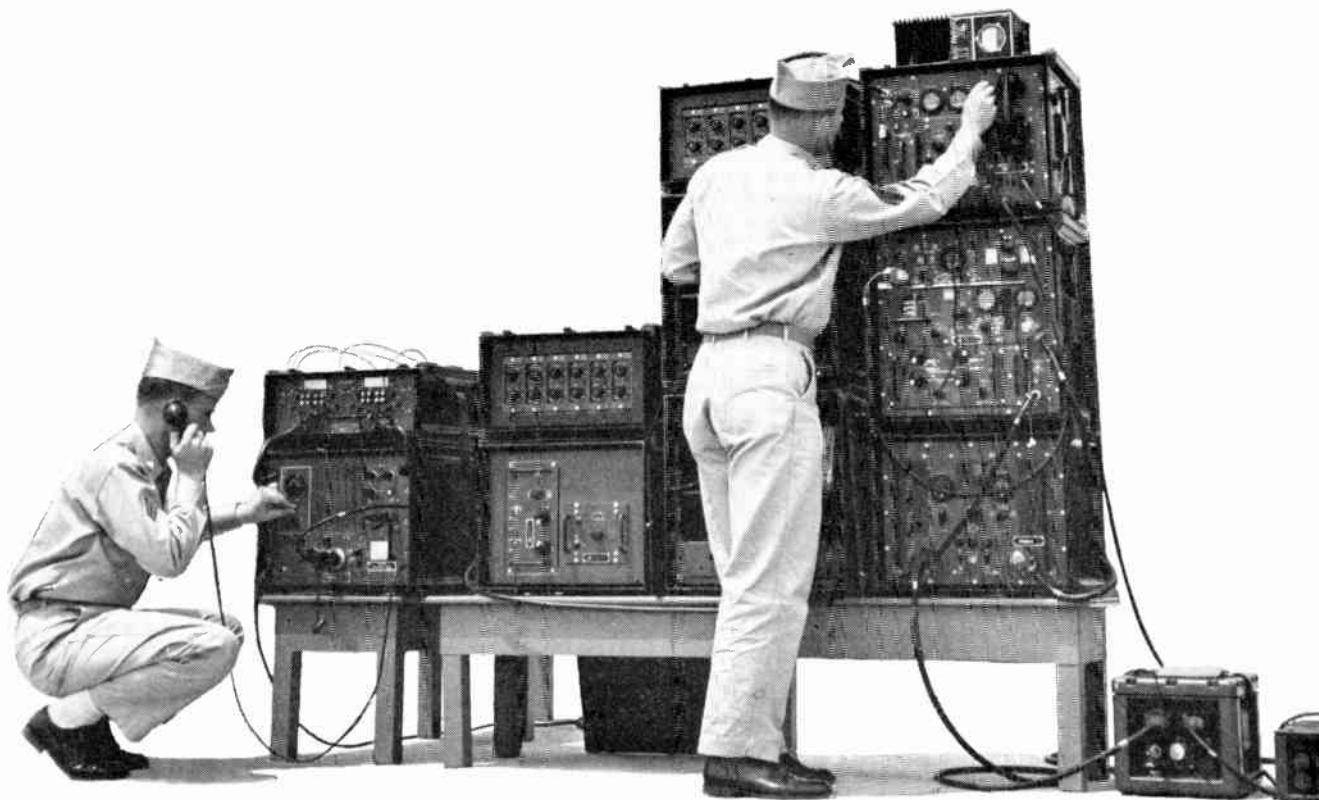
SPRAGUE TYPE NO.	WATTAGE RATING	DIMENSIONS L (inches) D		MAXIMUM RESISTANCE
151E	3	1 $\frac{1}{2}$	1 $\frac{1}{4}$	10,000 Ω
27E	5	1 $\frac{1}{8}$	$\frac{3}{16}$	30,000 Ω
28E	10	1 $\frac{1}{8}$	$\frac{3}{16}$	50,000 Ω

Standard Resistance Tolerance: $\pm 5\%$

SPRAGUE

WRITE FOR ENGINEERING BULLETIN NO. 111 B

SPRAGUE ELECTRIC COMPANY • 235 MARSHALL ST. • NORTH ADAMS, MASS.



Tuning in for radio transmission. Each item of equipment is not much bigger than a suitcase.

A leapfrog telephone system for the Armed Forces!

A new communications system, which takes to the air when water or rough terrain prevents the stringing of wires, has been developed for the U.S. Signal Corps by Bell Telephone Laboratories.

The system uses cable and radio relay interchangeably over a 1000-mile range. It is easily portable, unaffected by climate, and rugged enough for global use. Twelve voices travel at once over a pair of wires or radio waves—as clearly and naturally as over the regular telephone system.

This is the first time a completely integrated wire and radio system of this large a channel capacity has been available for tactical use by the Armed Forces. It is already in production at Western Electric, manufacturing and supply unit of the Bell System.

The new system is a joint achievement of the Signal Corps, Bell Laboratories and Western Electric . . . one of the many results of long and fruitful co-operation. It shows again how techniques which the Laboratories develop contribute to our national strength.

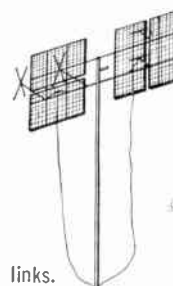


Bell Telephone Laboratories

*Improving telephone service for America provides careers
for creative men in scientific and technical fields*



Amplifiers like this are used every 5¾ miles in the cable portions of the system. They are weatherproof, can be used on a pole or the ground, and will even work under water. The system uses a spiral wound cable developed by the Signal Corps.



Easily raised antennas
send or receive for the radio links.

choose from this complete line of

MINIATURE PULSE TRANSFORMERS



Type 10Z
tubular pulse transformer



Type 15Z
miniature bathtub pulse
transformer



Type 20Z
drawn-shell bathtub pulse
transformer



Type 40Z
plug-in pulse transformer

NOW YOU CAN CHOOSE from eighteen standard pulse transformers in four major construction styles, all in quantity production at Sprague. The standard transformers covered in the table below offer a complete range of characteristics for computer circuits, blocking oscillator circuits, memory array driving circuits, etc.

These hermetically sealed units will meet such stringent military specifications as MIL-T-27, and operate at temperatures up to 85°C. Special designs are available for high acceleration and high ambient temperature operation. In addition, the electrical counterparts of each transformer can be obtained in lower cost housings designed for typical commercial environment requirements.

Complete information on this high-reliability pulse transformer line is provided in Engineering Bulletin 502A, available on letterhead request to the Technical Literature Section, Sprague Electric Company, 235 Marshall Street, North Adams, Massachusetts.

ELECTRICAL CHARACTERISTICS OF SPRAGUE PULSE TRANSFORMERS

Type No.	Turns Ratio	Pulse Width μ seconds	Rise Time μ seconds	Primary Inductance	Leakage Inductance	Repetition Rate	Load and Output	Typical Applications
10Z1	5:1	0.1	0.04	200 μ H	5 μ H	1 to 2 MC	15 volts 100 ohms	Used in digital computer circuitry for impedance matching and inter-stage coupling. Pulses are of sine wave type.
10Z2	4:1	0.07	0.03	200 μ H	20 μ H	1 to 2 MC	20 volts 100 ohms	
10Z3	1:1	0.07	0.03	125 μ H	12 μ H	1 to 2 MC	20 volts 200 ohms	
10Z4	3:1	0.07	0.03	160 μ H	15 μ H	1 to 2 MC	20 volts 100 ohms	
10Z6	4:1	0.1	0.04	200 μ H	6 μ H	1 to 2 MC	17 volts 100 ohms	
10Z12	1:1	0.25	0.02	200 μ H	2 μ H	12KC	100 volts	Blocking Oscillator
10Z13	1:1	0.33	0.07	240 μ H	2 μ H	2KC	50 volts	Blocking Oscillator
10Z14	7:1:1	0.50	0.05	1.2 mH	20 μ H	1MC	25 volts	Impedance Matching
15Z1	3:1	5.0	0.04	7.5 mH	22 μ H	10 KC	10 volts 100 ohms	Impedance Matching and Pulse Inversion
15Z2	2:1	0.5	0.07	6 mH	15 μ H		40 volts	Blocking Oscillator
15Z3	5:1	10.0	0.04	12 mH	70 μ H	10 KC	10 volts	Impedance Matching
15Z4	1:1.4	6.0	0.1	16 mH	15 μ H	0.4 KC	15 volts	Blocking Oscillator
20Z1	5:5:1 Push-Pull	1.5	0.25	4.0 mH	0.3 MH		5 volts 10 ohms	Memory Core Current Driver
20Z3	6:1	1 to 4	0.22	18 mH	0.8 MH	250 KC (max.)	21 volts 200 ohms	Current Driver
20Z4	6:1:1	1 to 7	0.25	55 mH	0.3 MH	50 KC (max.)	22 volts 400 ohms	Current Driver and Pulse Inversion
20Z5	3:3:3:1 Push-Pull	2.4	0.2	2.8 mH	0.2 MH		2.5 volts 6 ohms	Memory Core Current Driver
20Z6	11:1	6.0	0.2	90 mH	0.2 MH	50 KC (max.)	10 volts 75 ohms	Current Transformer
40Z1	7:1:1	0.50	0.05	1.2 mH	20 μ H	1 MC	25 volts	Impedance Matching

Sprague, on request, will provide you with complete application engineering service for optimum results in the use of pulse transformers.

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PERKIN...HAS A STANDARD POWER SUPPLY FOR YOUR EVERY NEED
IMMEDIATE DELIVERY!!



PERKIN

TUBELESS!!
MAGNETIC AMPLIFIER
REGULATED DC
POWER
SUPPLIES

MODEL
 MR 532-15
 5 TO 32 V.
 @ 15 AMP.
 (CONT.)



REGULATION: $\pm 1\%$ (a) from 5-32V DC (b) from 1.5 to 15 amps. (c) from 105-125V AC. (single phase, 60 cps.)
RIPPLE: 1% rms @ 32V and full load, increases to max. of 2% rms @ 5V and full load. **RESPONSE:** 0.2 sec.
METERS: 4 1/2" AM and VM, 2% accuracy.
MOUNTING: Cabinet or 19" rack panel.
FINISH: Baked Grey Wrinkle.
WEIGHT: 150 lbs.
DIMENSION: 22" x 17" x 14 1/2"

MODEL
 M60 VMC
 0 TO 32 V.
 @ 25 AMP.
 (CONT.)



REGULATION: $\pm 1\%$ (a) at 28V DC; increases to 2% max. over the range 24-32V; does not exceed 2V regulation over the range 4-24V DC (b) from 1/10 full load to full load (c) at a fixed AC Input of 115V.
RIPPLE: 1% rms @ 32V and full load; 2% rms max. @ any voltage above 4V
AC INPUT: 115V, single phase, 60 cps.
FINISH: Baked Grey Wrinkle.
WEIGHT: 130 lbs.
DIMENSIONS: 22" x 15" x 14 1/2"

MODEL
 MR 1040-30
 10 TO 40 V.
 @ 30 AMP.
 (CONT.)



REGULATION: $\pm 1\%$ (a) from 10 to 40V DC (b) from 100 to 130V AC (c) from 3 to 30 Amps DC. **RIPPLE:** 1% rms.
AC INPUT: 100-130V, 1 phase, 60 cycles.
RESPONSE: 0.2 sec. **METERS:** 4 1/2" AM and VM.
MOUNTING: Cabinet with 19" rack panel.
FINISH: Baked Grey Enamel.
WEIGHT: 200 lbs.
DIMENSIONS: 22" x 15" x 23"

MODEL
 MR2432-100X
 24 TO 32 V.
 @ 100 AMP.
 (CONT.)



REGULATION: $\pm 1/2\%$ (a) from no load to full load. (b) from 24-32V DC. (c) for 230* (or 460) V $\pm 10\%$.
DC OUTPUT: 24-32V @ 100 amps.
AC INPUT: 230 or 460V $\pm 10\%$, 3 phase, 60 cycles.
RIPPLE: 1% rms. **RESPONSE TIME:** 0.2 sec.
MOUNTING: Cabinet or 19" rack panel.
WEIGHT: 250 lbs.
DIMENSIONS: 25" x 15" x 15"
*This unit will be supplied for 230V AC Input unless 460V is specified.

ALSO AVAILABLE: Standard 6 and 115 volt models; Ground and Airborne Radar and Missile Power Supplies — Write for Perkin Bulletins.

PERKIN

ENGINEERING CORP.

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Meetings with Exhibits

● As a service both to Members and the industry, we will endeavor to record in this column each month those meetings of IRE, its sections and professional groups which include exhibits.

Δ

April 15-16, 1955

Ninth Annual Spring Technical Conference, Cincinnati Section, IRE, Engineering Society of Cincinnati Bldg., Cincinnati, Ohio

Exhibits: Mr. Clyde G. Haehnle, Crosley Broadcasting Corp., 140 West Ninth St., Cincinnati 2, Ohio

April 27-29, 1955

Seventh Regional Technical Conference & Trade Show, Hotel Westward Ho, Phoenix, Ariz.

Exhibits: Mr. George McClarathan, 509 East San Juan Cove, Phoenix, Ariz.

May 9-11, 1955

National Conference on Aeronautical Electronics, Biltmore Hotel, Dayton, Ohio.

Exhibits: Mr. William Klein, 1472 Earlham Drive, Dayton, Ohio

May 18-20, 1955

National Telemetry Conference, Morrison Hotel, Chicago, Ill.

Exhibits: Mr. Kipling Adams, General Radio Company, 920 S. Michigan Ave., Chicago, Ill.

June 2-3, 1955

I.R.E. Materials Symposium, Convention Hall, Philadelphia, Pa.

Exhibits: Mr. Merritt A. Rudner, United States Gasket Co., 611 North Tenth St., Camden 1, N.J.

Aug. 24-26, 1955

Western Electronic Show & Convention, Civic Auditorium, San Francisco, Calif.

Exhibits: Mr. Mal Mobley, 344 N. LaBrea, Los Angeles 36, Calif.

Sept. 12-16, 1955

Tenth Annual Instrument Conference & Exhibit, Shrine Exposition Hall & Auditorium, Los Angeles, Calif.

Exhibits: Mr. Fred J. Tabery, 3443 So. Hill St., Los Angeles 7, Calif.

October 3-5, 1955

National Electronics Conference, Sherman Hotel, Chicago, Ill.

Exhibits: Mr. G. J. Argall, c/o DeVry Technical Institute, 4141 Belmont Ave., Chicago 41, Ill.

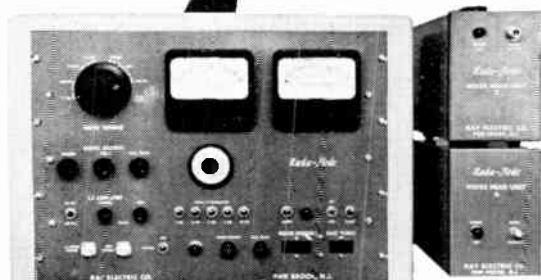
Note on Professional Group Meetings: Some of the Professional Groups conduct meetings at which there are exhibits. Working committeemen on these groups are asked to send advance data to this column for publicity information. You may address these notices to the Advertising Department, and of course listings are free to IRE Professional Groups.

KAY

PRECISION PACKAGED

NOISE MEASUREMENT

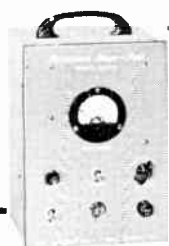
IN EVERY RANGE



There's a Kay instrument to answer most needs in noise testing—with good measure! Each *Mega-Node* type affords accurate measurement of noise figure and receiver gain in a specific frequency range, while the *Rada-Node* can be obtained with all elements required for complete noise figure measurement, 5 mc to 26,500 mc, including power supplies. Thus you may trust your noise test work to one precision line of uniform high quality.

KAY *Rada-Node* (Illustrated)

Complete radar noise figure measuring set for IF and RF, including attenuators, detector and noise sources. Provides production and lab measurement of noise figure and receiver gain. Complete with power supplies. *Freq. Range:* 5-26,500 mc. *Noise Figure:* Range, 0-21 db, accurate to ± 0.25 db. Prices on request.



KAY *Microwave Mega-Nodes*

Calibrated random noise sources in the microwave range, used to measure noise figure and receiver gain and to calibrate standard signal sources in radar and other microwave systems. Available in the following waveguide sizes to cover the range of 1200-1400 mc and 2600-40,000 mc:

RG-69/U, RG-48/U, RG-49/U, RG-50/U,
RG-51/U, RG-52/U, RG-91/U, RG-53/U

Available with fluorescent or inert gas (argon or neon) tubes. Noise output fluorescent tubes 15.8 db ± 0.25 db; argon gas tubes 15.2 db ± 1 db*; neon tubes 18.0 db $\pm .5$ db*.

*Noise output of inert gas tubes, independent of operating temperature.

RG-69/U...\$500
RG-48/U.... 295
RG-49/U.... 295
RG-50/U.... 295
RG-51/U.... 295
RG-52/U.... 295
RG-91/U.... 350
RG-53/U.... 350

ALL PRICES INCLUDE POWER SUPPLY

KAY *Mega-Node*

Calibrated random noise source reading direct in db, for measurement of noise figure, receiver gain, and for indirect calibration of standard signal sources. *Freq. Range:* 5-220 mc. *Output impedances:* Unbalanced—50, 75, 150, 300, Infinity. Balanced—100, 150, 300, 600, Infinity. *Noise Figure Range:* 0-16 db at 50 ohms; 0-23.8 db at 300 ohms. \$295.00 f.o.b. factory.

KAY *Mega-Node-Sr.*

Same uses as *Mega-Node*. *Freq. Range:* 10-3000 mc. *Output Impedance:* 50 ohms unbalanced into Type N Connector. *Noise Figure Range:* 0-20 db. \$995.00 f.o.b. factory.

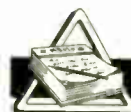
KAY

ELECTRIC COMPANY

Dept. 1-4 14 Maple Ave. Pine Brook, N. J.



Write for Technical Data Sheets and copy of Kay 1954-55 Catalog.



FCC ACTIONS

The Federal Communications Commission recently finalized, with certain changes, its proposal in Docket 11031 and amended part 18 of its Rules Governing Industrial, Scientific and Medical Services by redefining "miscellaneous equipment" to include apparatus that applies radio frequency energy to materials to produce physical, biological or chemical changes but which does not involve communications or the use of radio receiving equipment. The new rules became effective on March 1. A new subpart classifies ultrasonic equipment as a special type of miscellaneous equipment which includes any apparatus generating radio frequency energy on frequencies above 20 kc and utilizing that energy to excite or drive an electro-mechanical transducer for the production and transmission of ultrasonic energy for industrial, scientific, medical or other purposes. New technical requirements are established and a type approval and certification procedure is provided. . . . Following comments filed by RETMA with the Federal Communications Commission in Docket 9288, in which it was pointed out that many imported receivers do not comply with the proposed restrictions on spurious radiation, the FCC directed the following letter to all known importers of FM and TV receivers: "Your attention is invited to the attached Notice of Proposed Rule Making (Docket 9288) which proposes limits of oscillator radiation for FM and television broadcast receivers. The radio manufacturing industry in the United States has been cooperating with this Commission in a program of reduction of oscillator radiation. It has come to the attention of the Commission that foreign made receivers imported into this country may be in violation of standards which may be adopted by the Commission and that, in such event, steps toward the enforcement of the Commission's rules may be necessary. It is suggested that you give this problem your earnest consideration and convey this information to the manufacturers of any foreign receivers which you may be importing. The Commission will welcome any comments you may wish to submit in connection with this matter."

INDUSTRY STATISTICS

The Radio-Television-Manufacturers Association of Canada has announced the projected production of 207,256 television sets in Canada during the months of January, February and March. December production amounted to 93,928 sets and sales of 89,078 units were reported. A total of 593,856 television receivers were produced during the year 1954, with sales for the year reported to be 619,428 receivers by the Canadian RTMA. . . . The pro-

(Continued on page 14A)

* The data on which these NOTES are based were selected by permission from *Industry Reports*, issues of January 24, 31, and February 7, published by the Radio-Electronics-Television Manufacturers Association, whose helpfulness is gratefully acknowledged.

AMPHENOL

"AN" CONNECTORS for POTTING



Potting is the modern method of moisture-proofing AN connectors—and also the most effective. Briefly defined, Potting is the injection of a synthetic rubber sealant around the wired terminals on the back of a connector; the sealant is contained and shaped by a mold form which may be removed in 24 hours after the sealant has set.

What are the advantages of Potting and why does AMPHENOL present it as the most effective moisture-proofing method?

- 1 AN connector assembly terminals are completely enclosed by the sealant.
- 2 The sealant is completely resistant to moisture of any sort: water, fuel oil, salt-spray—any and all of the usual causes of AN connector failure.
- 3 Potting replaces the back-shell and cable clamp of the AN connector—reduces weight, cost and size of every assembly.
- 4 The method of Potting is easy to learn and easy to master; AMPHENOL offers full assistance.

AN connectors for Potting at your plant and complete Potted AN connector assemblies and harnesses may be ordered from AMPHENOL. Check with our nearest representative or with the home office for details.

AMPHENOL

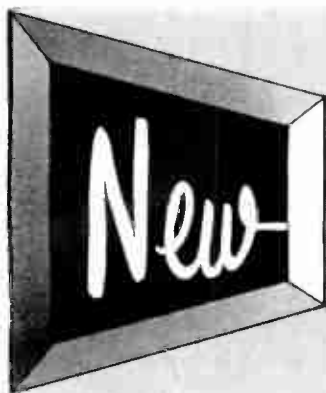
For additional information request Bulletin 2555

AMERICAN PHENOLIC CORPORATION

Chicago 50, Illinois

In Canada: AMPHENOL CANADA LIMITED, Toronto





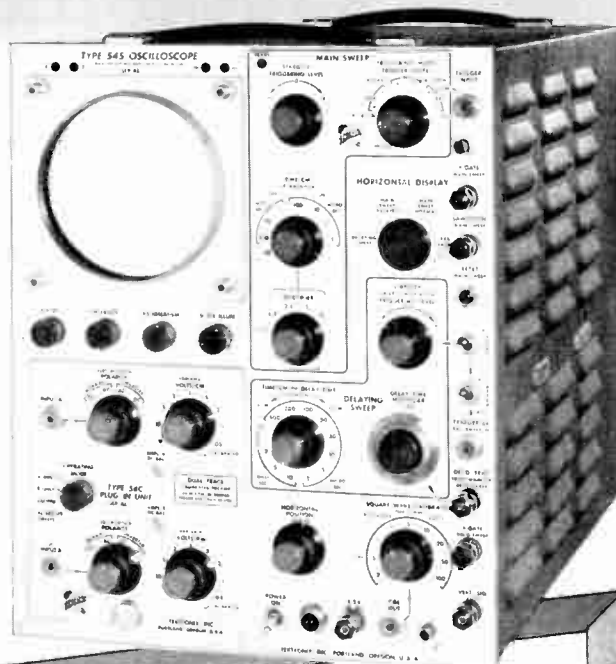
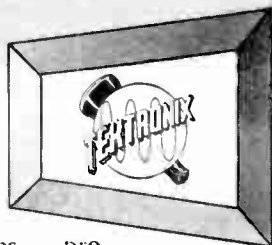
for fast-rise applications

(12 MILLIMICROSECONDS)

Tektronix Type 545 and Type 541 CATHODE-RAY OSCILLOSCOPES

TYPE 545—This new high-speed laboratory oscilloscope, in combination with the new Type 53K/54K Fast-Rise Plug-In Unit ... opens the way to quicker, easier analyses of fast-rising waveforms ... providing faithful displays and accurate measurement facilities well beyond the range of previous oscilloscopes of its size and cost. The Type 545-Type 53K/54K combination offers a vertical-amplifier passband of dc to 30 mc (12-millimicrosecond risetime) at calibrated sensitivities to 0.05 v/cm, with a full 4-cm linear vertical deflection. A wide range of calibrated sweeps, with calibrated sweep delay from 1 μ sec to 0.1 sec, and high accelerating potential, 10 kv, fully complement this greatly extended vertical-amplifier range.

The Type 545 is the most versatile oscilloscope ever made, for it can be quickly converted to many other applications. By merely plugging in the appropriate Type 53/54 Plug-In Preamplifier you are ready for wide-band, wide-band high gain, dual-trace, high-gain differential, microvolt-sensitivity, or wide-band differential applications. It's a rare oscilloscope application that isn't easily handled by this modern method.



Type 545 Oscilloscope Characteristics

Vertical-Amplifier Characteristics with Type 53K/54K Unit Plugged In

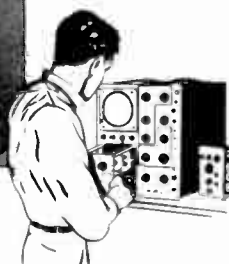
Transient Response—Risetime, 12 millimicroseconds.

Frequency Response—Passband, dc to 30 mc (down 3 db $\frac{1}{2}$ db at 30 mc, only 6 db at 45 mc).

Input impedance 20 μ f, 1 megohm.

Sensitivity—0.05 v/cm to 20 v/cm in 9 calibrated steps.

Price—\$125



LOW INPUT CAPACITANCE

With Accessory Probes for Type 53K/54K

Probe	Input Impedance	Maximum Sensitivity
P405	11.5 μ f, 5 megohms	0.25 v/cm
P410	7.5 μ f, 10 megohms	0.5 v/cm
P420	4.5 μ f, 10 megohms	1 v/cm
P450	2.5 μ f, 10 megohms	2.5 v/cm
P4100	2.5 μ f, 10 megohms	5 v/cm

Wide Sweep Range

24 Calibrated sweeps from 0.1 μ sec/cm to 5 sec/cm, accurate within 3%. Accurate 5-x magnifier extends calibrated range to 0.02 μ sec/cm. Continuously variable from 0.02 μ sec/cm to 12 sec/cm.

Wide Sweep-Delay Range

Additional delaying-sweep circuitry provides conventional, or triggered jitter-free delay, 1 μ sec to 0.1 sec in 12 calibrated ranges. Range accuracy within 2%. Incremental accuracy within 0.2% of full scale.

Versatile Triggering

Internal or external, with amplitude-level selection or AUTOMATIC TRIGGERING. High-frequency synchronization up to 30 mc.

Square-Wave Amplitude Calibrator

0.2 mv to 100 v in 18 steps, accurate within 3%.

New Cathode-Ray Tube

Tektronix T54P 5" precision metallized crt provides 4-cm vertical and 10-cm horizontal linear deflection, 10-kv regulated accelerating potential.

Balanced Delay Network

0.15 μ sec vertical signal delay.

DC-Coupled Unblanking

Uniform unblanking at all sweep speeds and repetition rates.

Electronic Voltage Regulation

All voltages affecting calibrations are fully regulated.

CRT Beam Position Indicators

Type 545—\$1450 plus price of desired plug-in units.

Type 541—Same characteristics, less delayed-sweep facility—\$1145 plus price of desired plug-in units.

Prices f.o.b. Portland (Beaverton), Oregon

Please call your Tektronix Field Engineer or Representative for complete specifications.

Tektronix, Inc.

P. O. BOX 831B • PORTLAND 7, OREGON

CYPRESS2-2611

CABLE: TEKTRONIX

GENERAL ELECTRIC ANNOUNCES

Vac-u-Sel RECTIFIERS

New Line of G-E Component Rectifiers Achieves 3 Performance Highs

- 63 VOLT PEAK INVERSE
- 130 C AMBIENT OPERATION
- 60,000 HOUR LIFE EXPECTANCY

General Electric's new line of Vac-u-Sel Component rectifiers offer greater application flexibility than any other rectifiers in history. You can now obtain a rectifier cell with a peak inverse rating of 63 volts, or a rectifier which will operate up to 130 C ambient temperature, or a rectifier which has a life expectancy of 60,000 hours.

New G-E Vac-u-Sel rectifiers now make it possible to match performance requirements for life expectancy, ambient operating temperature, and atmospheric protection, as well as electrical characteristics.

THREE NEW RECTIFIER CELLS make up the new line of Vac-u-Sel rectifiers; a 26-volt low temperature cell, a 26-volt high temperature cell, and a 45-volt high temperature cell. All three are produced by the vacuum evaporation process described at the right, but special variations in the manufacturing give them distinctly different electrical characteristics.

26-VOLT LOW TEMPERATURE CELL is the standard industrial cell, used on applications where ambient operating temperature will not exceed 55 C. Rectifiers using this cell have a life expectancy of 60,000 hours at normal current rating.

26-VOLT HIGH TEMPERATURE CELL can meet operating requirements up to 130 C at full voltage. Current need not be derated where shorter life is acceptable. Life expectancy at 130 C is 1000 hours.

45-VOLT HIGH TEMPERATURE CELL has a 63-volt peak inverse voltage. Unlike most 45-volt rectifiers, this is a true, long-life industrial cell. Frequently this rectifier may be substituted for ones employing 26-volt cells. Since fewer cells are required, savings of up to 30% in cost, and up to 35% in the size of the stacks are possible. Life expectancy of this 45-volt cell is 40,000 hours, and the cells can be used at ambient temperatures up to 110 C.

ALL VAC-U-SEL RECTIFIERS operate with exceptionally low forward voltage drop and low reverse leakage, and their margin of superiority in these characteristics increases in service. All Vac-u-Sel rectifiers undergo extensive testing and grading, and matched cells are used in assembling stacks. A variety of finishes and mounting arrangements are available to meet virtually any requirements.

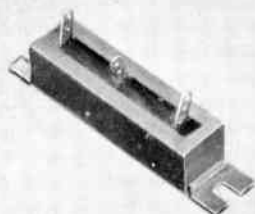
MORE INFORMATION on these new Vac-u-Sel rectifiers is available from your nearest General Electric Apparatus Sales Office, or by writing Section 461-36, General Electric Co., Schenectady 5, N. Y.

Progress Is Our Most Important Product

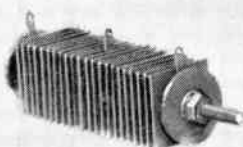
GENERAL ELECTRIC

RECTIFIER DEPARTMENT

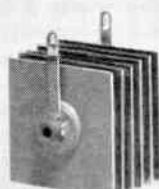
METALLIC RECTIFIER COMPONENTS FOR PRACTICALLY EVERY DC REQUIREMENT



VAC-U-SEL



SELENIUM



COPPER-OXIDE



GERMANIUM



WHAT IS VAC-U-SEL?

Vac-u-Sel is the General Electric trademark for a new line of metallic component rectifiers with exceptional electrical characteristics. The name Vac-u-Sel is used because these new rectifiers are produced by a vacuum evaporation process.

WHAT IS THE VAC-U-SEL PROCESS?

In the Vac-u-Sel process, cells are loaded on the inside of a sphere and a vacuum of over one millionth of an atmosphere is drawn on the sphere. Selenium is then evaporated inside the sphere onto the rectifier plates.

WHY USE A VACUUM?

The vacuum permits greatest control over contaminants which affect the quality of the finished rectifier. There are more than 100 variables involved in making selenium rectifiers.

WHY IS THIS PROCESS BETTER?

Vacuum evaporation provides greater uniformity between individual cells and stacks. It produces a more even deposition over the entire surface of each cell, eliminating cracks and thin spots. The electrical characteristics of a rectifier are highly dependent on the thickness of the deposit, and vacuum evaporation permits greater accuracy in controlling this thickness. In addition, it fosters better crystalline orientation, allowing natural rather than pressed formation.

ARE VAC-U-SEL RECTIFIERS BETTER?

The electrical characteristics of new General Electric Vac-u-Sel rectifiers are greatly superior to ordinary commercial selenium rectifiers. Full details of Vac-u-Sel rectifier performance are given on opposite page.

GENERAL  ELECTRIC



DATA FOR



NEW RCA TRANSISTOR RCA-2N104 (FOR LOW-POWER AF SERVICE)

Hermetically sealed type for low-power af service . . . features extreme stability and excellent uniformity of characteristics—initially and during life.

This new germanium alloy-junction transistor (p-n-p) type is intended for low-power af service. It utilizes an insulated metal envelope and a lineotetrap 3-pin base. Maximum noise factor—only 12 db. The design of the 2N104 features low base-lead resistance which minimizes ohmic losses, improves frequency response, and insures high input-circuit efficiency. In a common-emitter circuit, the 2N104 has a collection-to-base current amplification ratio of 44, a matched-impedance, low-frequency power gain of 40 db, and a collector-to-emitter alpha frequency cutoff of 700 kc.

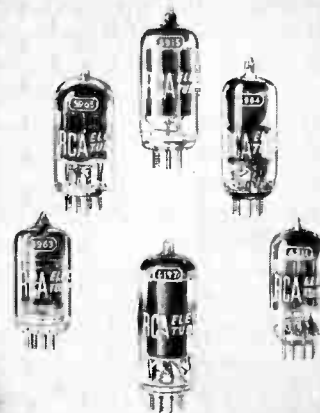


NEW RCA STORAGE TUBE (FOR COMPUTER SYSTEMS)

Designed especially for use in binary-digital computer systems, this 3-inch storage tube is of the single-beam type, has electrostatic focus and deflection, and employs "redistribution writing" and "capacitance-discharge reading". Outstanding design features of the tube include: a storage surface having relatively uniform secondary emission to prevent "bad spots" on which information can not be stored; a focused beam having an exceptionally small effective area including the fringe of low-density beam current and a well-defined boundary; and a separate external connection for the collector to permit flexibility in circuit operation.

NEW RCA MULTIPLIER PHOTOTUBES (FOR HEADLIGHT DIMMER SERVICE)

Having instantaneous response to light, RCA-6328 and 6472 are your answer for "road-proved" multiplier phototubes that meet the exacting timing requirements of headlight control. Both tubes have high luminous sensitivity—for operation with amplifiers of relatively low input impedance. Both combine stability with long life. Identical in characteristics to the 6328, RCA-6472 is built with flexible leads—for use in printed circuits.



YOUR CHOICE OF COMPUTER TUBES RCA-5915, 5963, 5964, 5965, 6197, 6211 . . . Dependable performance, a must in computer applications, is accomplished in these six RCA tubes—by using production controls correlated with typical electronic computer conditions. RCA computer tubes feature controlled cutoff for switching applications, low-grid current for applications utilizing high values of grid resistance, high zero-bias plate current, special cathode material to minimize interface, and low leakage.



RCA HIGH-VOLTAGE THYATRON (FOR DC POWER CONTROL AND LOAD-CIRCUIT PROTECTION)

Having a negative control characteristic, this high-voltage 3-electrode, mercury-vapor thyatron is primarily designed for dc power-control applications, but is also useful in load-circuit protection. For example, in power-control application, three RCA-5563-A's in a half-wave, 3-phase circuit can handle up to 45 kw—at a dc output voltage up to about 9500 volts. Six of these tubes in a series, 3-phase circuit can handle up to 143 kw at a dc output voltage up to 19,000 volts (approx.). In protection applications, the 5563-A may be operated as a grid controlled rectifier to remove the dc load voltage by blocking action of the grid, or as an electronic switch across the rectifier output for instant removal of the load voltage in case of a load fault.

**ELECTRON TUBES—SEMICONDUCTOR DEVICES—BATTERIES—
TEST EQUIPMENT—ELECTRONIC COMPONENTS**

DESIGNERS

RCA SMALL-SIZED UHF POWER TUBES

Well-suited for fixed and mobile uhf applications up to 470 mc, these unique twin beam power tubes offer designers unusual advantages—as balanced push-pull rf power amplifiers or frequency triplers. RCA-6524 delivers approx. 20 watts (ICAS) in push-pull class C amplifier service—at 462 mc! Max. plate dissipation, 25 w (ICAS). RCA-5894 delivers approx. 55 watts (CCS) at 470 mc. Max. plate dissipation: 40 watts (CCS).

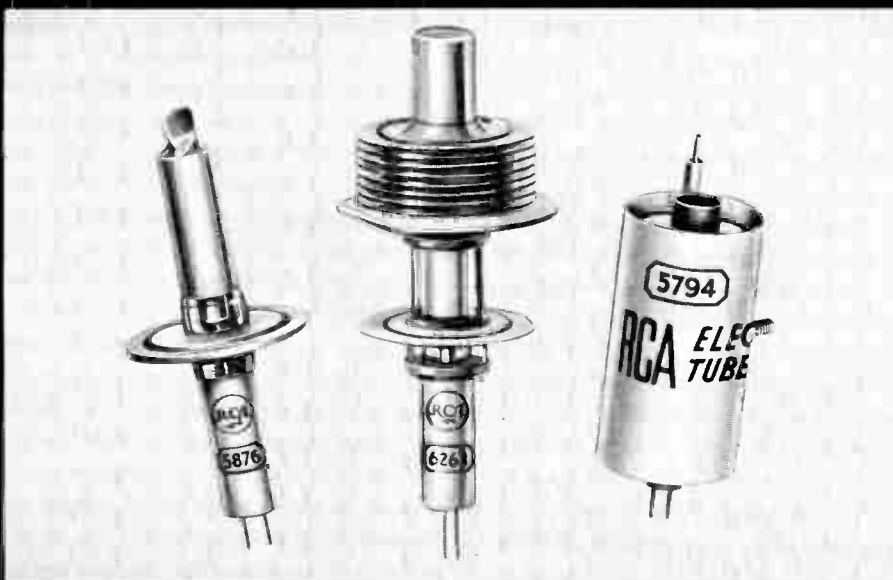


For technical information, write—specifying tube types in which you are interested—to RCA, Commercial Engineering, Section D35R, Harrison, N.J., or call your RCA Representative:

EAST _____ Humboldt 5-3900
744 Broad St.
Newark, N. J.

MIDWEST _____ Whitehall 4-2900
589 E. Illinois St.
Chicago 11, Ill.

WEST _____ Madison 9-3671
420 S. San Pedro St.
Los Angeles 13, Calif.



RCA "PENCIL" TUBES FOR UHF

Available in a choice of types for uhf applications, RCA "Pencil Tubes" are designed to have minimum transit time, low lead inductance, and low interelectrode capacitances. Features include small size, light weight, low heater wattage, and good thermal stability. RCA-6263 with external plate radiator is intended for rf power amplifier and oscillator services. 6264 is like the 6263 but is well-suited for frequency-multiplier service. Additional RCA "Pencil Tubes" include 5674, 5794, 5876, 5173.



NEW 5" PROJECTION KINESCOPE (FOR CLOSED-CIRCUIT INDUSTRIAL TV)

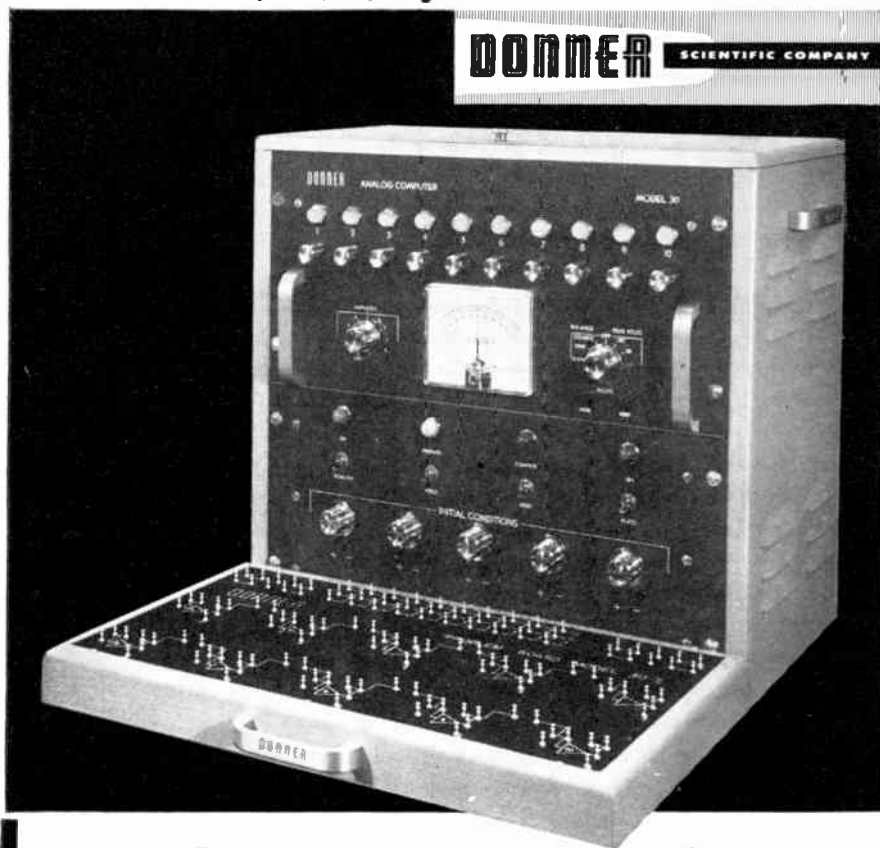
Providing a clear, bright, projected picture about eight feet by six feet when used with a suitable reflective optical system, the RCA-5AZP4 is especially useful for closed circuit industrial TV. Contributing to the brightness of the "auditorium-size" picture of high-efficiency, aluminized screen having very good color stability under varying conditions of screen current, and an unusually high operating ultor voltage (40,000 volts max.) for a tube of this type.



RADIO CORPORATION of AMERICA
TUBE DIVISION
HARRISON, N. J.

A new source of high performance instrumentation!

DONNER SCIENTIFIC COMPANY



analog computer \$995

MODEL 30

f.o.b. Berkeley, California

(with one 30-3 problem board as illustrated)

Here is a compact, economically priced analog computer designed for service as a personal tool of the engineer, mathematician, and scientist. Model 30 computers make electronic computation economically possible wherever differential equations are used. Typical applications include analysis and synthesis of physical systems and simulation of transfer characteristics. Flexibility and economy make the computer ideal for instructional use in schools and colleges and for individual use of the industrial scientist.

features

- A** "Slide Rule" versatility and simplicity—anyone who can translate physical problems into corresponding differential equations can use the Model 30 . . . even without specialized knowledge of electronics.
- B** Accuracy of solutions to better than 1% is determined by the precision of components selected
- C** Two types of inexpensive plug-in problem boards . . . Model 30-3 with solder terminals for components . . . Model 30-4 with plug-in connectors for components.

- D** Ten stable, high gain, single pentode D.C. amplifiers.
- E** Five isolated power supplies to set initial condition voltages.

PHYSICAL SPECIFICATIONS

Computer—height 19", width 21",
depth 12", weight 75 lbs.
Problem Boards—height 2", width 21",
depth 13".

Write for technical bulletin #301-A

DONNER

SCIENTIFIC COMPANY

2829 Seventh Street • Berkeley 10, California



Industrial Engineering Notes

(Continued from page 84)

duction of television receivers during 1954 was at the second highest point on record. Over 7.3 million TV sets and 10.4 million radios were produced during the year. Total television set production was reported as 7,346,715 units during the year, compared with 7,215,827 sets manufactured in 1953 and 7,463,800 TV receivers turned out during the record year, 1950. The radio production for 1954 was reported as 10,400,530 units compared with 13,368,556 receivers manufactured a year earlier.

MOBILIZATION

The first of four former Liberty ships to be converted by the Navy to ocean radar station ships will be commissioned on February 1 at the Naval Shipyard, Norfolk, Va. The ships will be employed in the continental air defense system. The conversion work primarily involved the installation of bulkheads which created spaces to accommodate communication and electronics equipment, as well as providing additional berthing and messing facilities for the crews. The electronics equipment includes air and surface search radar. Also being installed is a combat information center for evaluating radar information and controlling action of U. S. fighter aircraft against enemy targets, the Navy said.

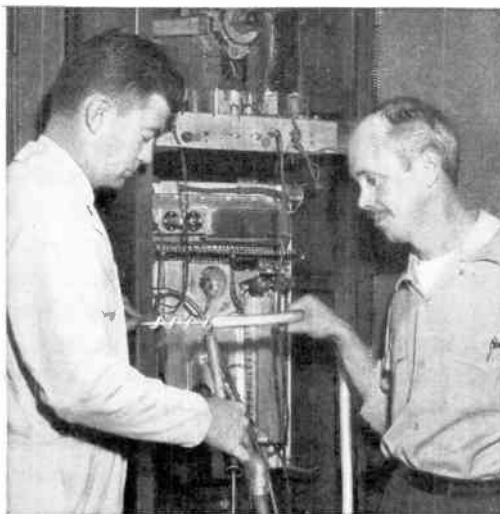
TECHNICAL

The National Bureau of Standards has announced the publication of a new circular which describes its Central Radio Propagation Laboratory's facilities atop Cheyenne Mountain, Colo., and gives sample results of the tropospheric propagation research carried out there. The circular, "Cheyenne Mountain Tropospheric Propagation Experiments," prepared by A. P. Barsis, J. W. Herbstreit and K. O. Hornberg, is available from the Government Printing Office, Washington 25, D. C., for 30 cents per copy. The announcement stated that the mountain site was established for use in studies of tropospheric radio-wave propagation in the VHF and UHF region of the radio spectrum. These facilities include high-power, continuous-wave transmitters on five frequencies, from 92 to 1046 mc. In the circular the new theory of tropospheric propagation, embodying the Booker-Gordon scattering principles as extended by Staras, is related to the measurements. . . . Men of medicine and industry met recently to discuss how the newest communications medium, color television, can best be used in medical education and also the diagnosis and treatment of disease during a symposium held Jan. 17-19 under the sponsorship of the Armed Forces Institute of Pathology. The climax of the three day meeting was the performance of an operation which brought pathologists from different cities together for consultation. Four closed circuit television presentations were received at the

(Continued on page 24A)

Styroflex Coaxial Cable

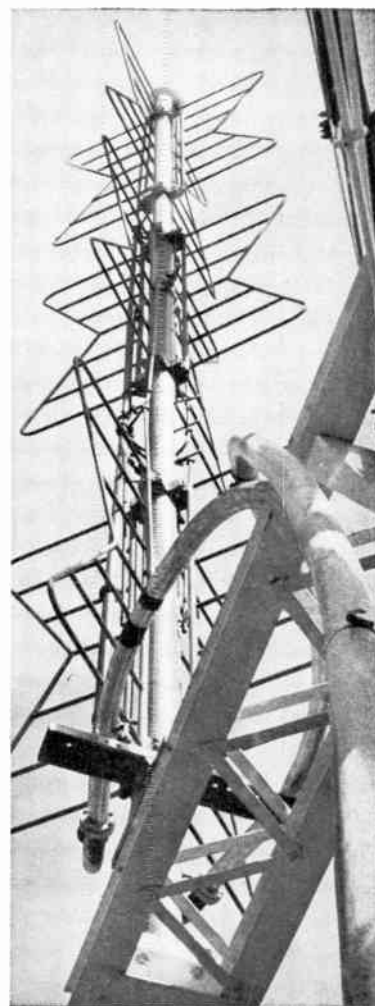
has many applications in the communication field!



Mobile Pickup



Community Antenna



Television Broadcast

Microwave

Styroflex, with its unique design, provides efficient, dependable service . . . has won the enthusiastic support of engineers as a result of successful installations throughout the communication field.

PHELPS DODGE COPPER PRODUCTS
CORPORATION

40 WALL STREET, NEW YORK 5, N. Y.



Frequency Meters



COMPLETELY
SELF-CONTAINED
FIELD TEST INSTRUMENTS

by

FREQUENCY STANDARDS

These precision-built field test instruments were designed by Frequency Standards to provide rapid and accurate means of frequency measurement in the field. Frequency is determined by means of a micrometer dial. This reading is translated to frequency by accurate individual calibration charts or curves. Transducers, fittings, and cables can be supplied to meet the requirements of customers and convenient storage space for these items is provided in the lid of the instruments.



MODEL	FREQUENCY RANGE	ACCURACY
912-4	900-1200 MC	.01%
1217-4	1200-1700 MC	.02%
1723-4	1700-2300 MC	.02%
2335-4	2300-3500 MC	.02%
3545-4	3500-4500 MC	.01%
4458-4	4400-5800 MC	.01%
5882-4	5800-8200 MC	.01%

SPECIAL CAVITY DESIGNS

Frequency Standards maintains complete facilities for the design and manufacture of Reference Cavities, Preselector Cavities, and Filters to customers' specifications or blueprints. Our facilities also permit quantity production of complex waveguide assemblies.

ILLUSTRATED
BULLETINS
SENT ON
REQUEST

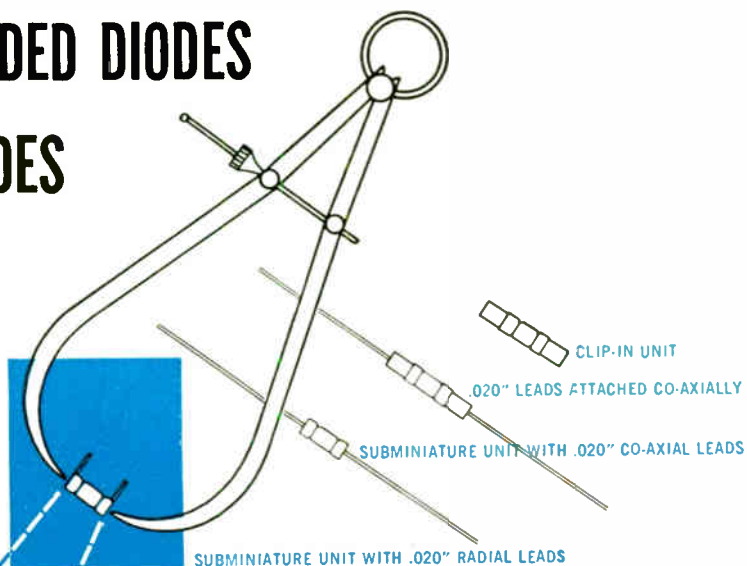
Frequency Standards
ASBURY PARK, NEW JERSEY

Address inquiries to
BOX 504

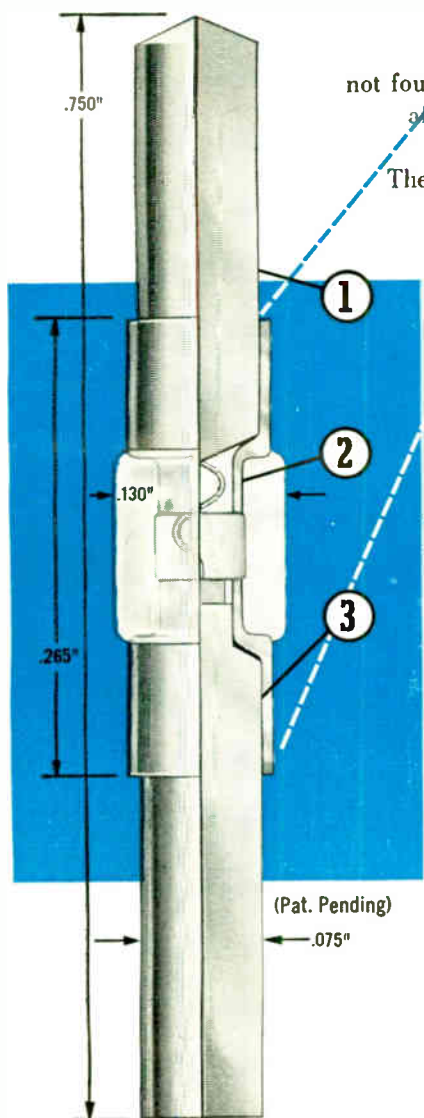


GERMANIUM GOLD BONDED DIODES SILICON JUNCTION DIODES

in the new **PSI** DIODE PACKAGE



PSI's revolutionary new package, with advantages not found in any other commercially available diodes, was designed only after an exhaustive survey of user requirements. Space limitations, environmental demands, even assembly procedures became factors in the final design. The result: diodes with demonstrably superior performance, greater versatility, top all-around utility.



CHECK THESE FEATURES...

1. VERSATILE LEAD ARRANGEMENT... for maximum adaptability, diodes may be obtained in a variety of configurations.

2. GLASS-TO-METAL SEAL... for positive moisture resistance, PSI uses a true fusion seal.

3. WELDED CONSTRUCTION... for greater strength and freedom from contamination; no low melting point solders are used.

and your net benefit from all these features...

**NEW STANDARDS OF
RELIABILITY AND STABILITY**

Typical PSI Gold Bonded Diode Characteristics @ 25°C

Forward Current @ 1v (ma)	Inverse Current (μa)	Inverse Working Voltage (volts)
100	100 (-20v)	35
35	10 (-50v)	80
15	25 (-50v) 200 (-200v)	220

Typical PSI Silicon Junction Diode Characteristics

E _s /E _t (volts)	Forward Current @ 1v (ma)	Back Current	
		at 25°C	at 150°C
30/29	80	.01μa (-15v)	5μa (-15v)
55/53	40	.01μa (-30v)	5μa (-30v)
150/145	15	.01μa (-75v)	5μa (-75v)
300/290	5	.01μa (-150v)	5μa (-150v)

a: The saturation voltage (E_s) is measured at 500μa; the transition voltage (E_t) is measured at 20μa.

b: Recovery time: after switching from 5ma forward current to ½E_s for all these types, back resistance reaches or exceeds 50K in 1μsec.

For complete product specifications, application data and quotations, address inquiries to Dept. S-12.

PACIFIC SEMICONDUCTORS, INC.

10451 WEST JEFFERSON BOULEVARD
CULVER CITY, CALIFORNIA

RAYTHEON

TRANSISTORS

MILLIONS OF THEM!!

RAYTHEON IS FIRST AND FOREMOST IN

— **mass production.** Raytheon is long past the experiment and development stage in Germanium PNP Junction Transistors — for over 2 years has had the quantity production and quality control techniques and resources

— **proved reliability in commercial application,** based on billions of hours of actual field performance and a record of success exceeding that of many reliable vacuum tubes

— **range of characteristics.** Look at the chart. You'll find one or more Raytheon Transistors that meet your specific requirements, however exacting.

LOW FREQUENCY TRANSISTORS — PLASTIC CASE

TYPE	Collector			Emitter	Base	Base Current Ampl. Factor	Max. Noise Factor db	Alpha Freq. Cutoff mc.	Max. Junction Temp. °C	Temp. Rise °C/mW
	Volts	Meg. ohms	Cutoff μ A	MA	ohms					
CK721	-6	2.0	6	-1.0	700	45	22	0.8	70	0.25
CK722	-6	2.0	6	-1.0	350	22	25	0.6	70	0.25
CK725	-6	2.0	6	-1.0	1500	90	20	1.2	70	0.25
CK727	-1.5	1.0	6	-0.5	700	45	12	0.8	70	0.25

LOW FREQUENCY TRANSISTORS — HERMETICALLY SEALED CASE

TYPE	Collector			Emitter	Base	Base Current Ampl. Factor	Max. Noise Factor db	Alpha Freq. Cutoff mc.	Max. Junction Temp. °C	Temp. Rise °C/mW
	Volts	Meg. ohms	Cutoff μ A	MA	ohms					
2N63	-6	2.0	6	-1.0	350	22	25	0.6	85	0.58
2N64	-6	2.0	6	-1.0	700	45	22	0.8	85	0.58
2N65	-6	2.0	6	-1.0	1500	90	20	1.2	85	0.58
2N106	-1.5	1.0	6	-0.5	700	45	12	0.8	85	0.58

HIGH FREQUENCY TRANSISTORS — HERMETICALLY SEALED CASE

TYPE	Collector		Emitter	Extrin. Base Resis. ohms	Base Current Ampl. Factor	Alpha Freq. Cutoff mc.	Max. Junc. Temp. °C	Temp. Rise °C/mW	Coll. Capac. μ f	Gain at 455kc db	Gain at 2 mc db	Rise time* μ secs	Decay time* μ secs
	Volts	Cutoff μ A	MA										
CK760	-6	1	-1.0	75	40	5	85	0.62	14	32	18	0.05	0.06
CK761	-6	1	-1.0	75	45	10	85	0.62	14	33	20	0.04	0.05
CK762	-6	1	-1.0	75	65	20	85	0.62	14	33	22	0.02	0.03

*measured in circuit which will be supplied on request

Note: above characteristics are average except where noted

There are more — several times more

RAYTHEON TRANSISTORS

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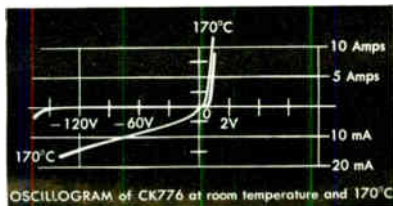
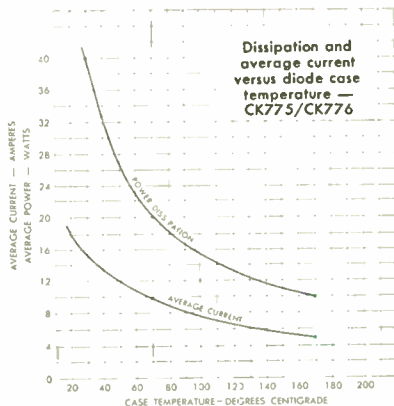
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with **95 to 99%** **EFFICIENCY**



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HIGH TEMPERATURE-175°C
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REDUCED COOLING REQUIRED
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better than 100kc

RAYTHEON SILICON POWER RECTIFIER CHARACTERISTICS

TYPE CK775	MAXIMUM VOLTAGE		MAXIMUM CURRENT		TYPICAL DISSIPATION WATTS
	RMS VOLTS	PEAK VOLTS	PEAK AMPERES	AVERAGE AMPERES	
CASE TEMP. 30°C*	40	60	50	15	40
CASE TEMP. 170°C*	40	60	15	5	10
NO HEAT RADIATOR					
AMBIENT TEMP. 25°C	40	60	6	2.0	3.0
AMBIENT TEMP. 170°C	40	60	2.0	0.5	2.0
TYPE CK776					
CASE TEMP. 30°C*	125	200	50	15	40
CASE TEMP. 170°C*	125	200	15	5	10
NO HEAT RADIATOR					
AMBIENT TEMP. 25°C	125	200	6	2.0	3.0
AMBIENT TEMP. 170°C	125	200	2.0	0.5	2.0

*maintained by external heat radiator

ADDITIONAL RATINGS (25°C)

Bath CK775 and CK776 have maximum drop at 5 amperes of 1.5 volts
 CK775 has maximum reverse current at -60 volts of 25 mA
 CK776 has maximum reverse current at -200 volts of 25 mA

RAYTHEON MANUFACTURING COMPANY

Semiconductor Division — Home Office: 55 Chapel St., Newton 58, Mass. Bigelow 4-7500

For application information write or call the Home Office or: 4935 West Fullerton Avenue, Chicago 39, Illinois, National 2-2770
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The history of radio communications has been marked by continual, relentless progress toward the upper reaches of the frequency spectrum. By the early 1940's services which utilized frequencies as high as 150 mc were coming into quite general use, and considerable experimental work was going on in the uhf and microwave portions of the spectrum.

World War II and the advent of radar gave a tremendous boost to this upward climb through the frequency domain, so that today the microwave field has blossomed to a position of prominence—almost pre-dominance—in the radio engineering art.

On March 7, 1952 this thriving field of activity received another important boost when the IRE formed a Professional Group on Microwave Theory and Techniques. For the first time, this important branch of the field was provided with its own organization for channeling and spreading vital, specialized knowledge to its own members and for stimulating a planned program of service tailored to the needs of the field. The value of the services being performed by the Group is attested to by the fact that more than 1600 engineers have joined and paid the modest \$2 assessment fee.

The most important service provided by the Group is its *TRANSACTIONS*, which is issued quarterly to all Group members. In addition, special issues of *TRANSACTIONS* are published occasionally in order to give members the complete proceedings of important conferences which are held in the microwave and related fields.

The Group has also been very active in sponsoring conferences throughout the year. In addition, local meetings are held by Chapters of the Group in Albuquerque, Baltimore, Boston, Buffalo, Chicago, Long Island, and Philadelphia.

If the activities of the PGMTT are not history-making they are assuredly history-changing, for already these activities have altered the course and speed of progress in this field.

W. R. G. Baker

Chairman, Professional Groups Committee



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-hp- 608D VHF Signal Generator

Presenting...

VHF
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-hp- 608C VHF Signal Generator

**New premium-quality
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Residual FM less than 1 kc

Drift less than 0.005%

High power output

All types of modulation

Models 608D and 608C are designed to be the best commercial instruments of their type, and to set new standards of VHF generator convenience, applicability and performance. They are the redesigned and improved successors to over 3,000 -hp- 608A/B VHF generators now in use throughout the world.

The premium quality -hp- 608D

-hp- 608D is the ultimate in VHF signal generators. It offers the highest stability attained in production equipment of its type. There is almost complete absence of incidental FM or frequency drift. There is a calibrated output from 0.1 μ v to 0.5 v throughout the frequency range, 10 to 420 mc. A built-in crystal calibrator provides a frequency check accurate within 0.01% every 5 mc throughout range.

These unique advantages are made possible in large part by new master oscillator, intermediate and output amplifier circuit design. Other features to improve stability include a regulated filament supply, a new variable condenser design and a completely new coil turret and circuit housing. The result is the most convenient, accurate and effective instrument available for testing and aligning VHF aircraft communications and other receivers having extreme selectivity.

The all-purpose -hp- 608C

The -hp- 608C is a high power, stable and accurate VHF signal generator for general laboratory and field use. Employing a master oscillator-power amplifier circuit, -hp- 608C offers 1 v maximum power and a broad frequency coverage of 10 to 480 mc. The instrument provides outstanding convenience for measuring gain, sensitivity, selectivity and image rejection of receivers, IF

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HEWLETT-PACKARD

two completely new

SIGNAL GENERATORS

amplifiers, broad band amplifiers and other VHF equipment. Its 1 v output is more than sufficient to drive bridges, slotted lines, transmission lines, antennas, filter networks and other circuits.

Outstanding features in both

Both *-hp-* 608D and 608C have broadest possible modulation capabilities. There is AM modulation to 80%, and flat response 20 cps to 1 mc which provides high quality internal and external pulse modulation. RF leakage is negligible, and sensitivity measurements to 0.1 μ v are possible. Internal impedance is 50 ohms constant, and VSWR is a maximum of 1.2.

Both instruments also feature new mechanical design and quality construction throughout. New aluminum castings and

cabinets reduce weight. Circuitry is particularly clean and accessible. Dial, condenser and turret drives are ball-bearing. Variable condensers are specially manufactured by *-hp-* and feature electrically welded Invar low temperature steel plates to minimize drift. Scaled transformers are used throughout, and construction is militarized.

Data subject to change without notice. Prices f.o.b. factory

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SPECIFICATIONS

-hp- 608D VHF Signal Generator

Frequency Range: 10 to 420 mc, 5 bands.

Calibration Accuracy: $\pm 0.5\%$ full range.

Resetability: Better than $\pm 0.1\%$ after warm-up.

Crystal Calibrator: Frequency check points every 5 mc through range. Headphone jack for audio frequency output.

Frequency Drift: Less than 0.005% over 15 minute interval after warm-up.

Output Level: 0.1 μ v to 0.5 v into 50-ohm load. Attenuator dial calibrated in v and dbm. (0 dbm equals 1 mw in 50 ohms.)

Voltage Accuracy: ± 1 db full range.

Generator Impedance: 50 ohms, maximum VSWR 1.2.

Modulation Percentage: 0 to 80% indicated by meter.

Envelope Distortion: Less than 2.5% at 30% sine wave modulation.

Internal Modulation: 400 cps $\pm 10\%$ and 1,000 cps $\pm 10\%$.

External Modulation: 0 to 80%, 20 cps to 100 kc. For RF output above 100 mc, 0 to 30% to 1 mc.

External Pulse Modulation: 10 v peak pulse required. Good pulse shape at 1 μ sec.

Residual FM: Less than 1,000 cycles at 30% AM for RF output frequencies above 100 mc. Less than 0.001% below 100 mc.

Leakage: Negligible; permits sensitivity measurements to 0.1 microvolt.

Filament Regulation: Provides highest possible oscillator and amplifier stability for line voltage change.

Power: 115/230 volts $\pm 10\%$, 50/1,000 cps. Approx. 150 watts.

Size: 13 $\frac{3}{8}$ " wide x 16" high x 20 $\frac{1}{2}$ " deep.

Weight: 70 lbs. Shipping weight, approx. 100 lbs.

Price: \$1050.00.

-hp- 608C VHF Signal Generator

Same as *-hp-* 608D, except:

Frequency Range: 10 to 480 mc, 5 bands.

Crystal Calibrator: In Model 608D only.

Frequency Drift: Less than $\pm 0.01\%$ over 10 minute interval after warm-up.

Output Level: 0.1 μ v to 1.0 v.

Residual FM: Less than 0.0025% at 30% amplitude modulation for RF output frequencies 21 to 480 mc.

Filament Regulation: In Model 608D only.

Price: \$950.00.

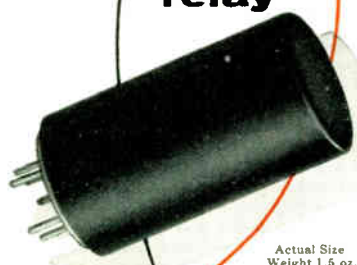


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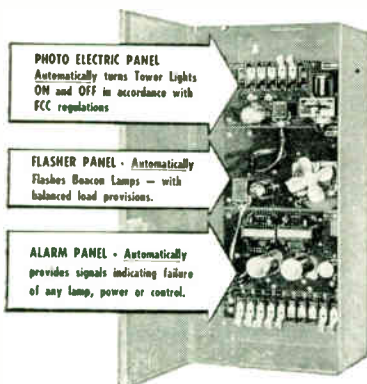
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TOWER LIGHTING CONTROL UNIT
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Models also available with separate
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Industrial Engineering Notes

(Continued from page 14A)

Institute from the hospital of the University of Pennsylvania, from studios in Baltimore, and from the National Naval Medical Center at Bethesda, Md. The Television Committee of the Armed Forces Institute of Pathology announced the names of some of the participants in the meetings. Included among the communications experts were Dr. A. N. Goldsmith, Chief Consultant for RCA; Dr. Peter C. Goldmark, Vice-President, Columbia Broadcasting System Laboratories; Edward W. Allen, Chief Engineer of the FCC; Dr. Axel G. Jensen, Director of Television Research of the Bell Telephone Laboratories, and Maj. Gen. G. I. Back, the Army's Chief Signal Officer. . . . A large reduction in engineering time and costs for selecting proper components to be used in electronic equipment is claimed for a mechanized system for storing and searching engineering data, the Office of Technical Services, Commerce Department, has announced. Details of this system, entitled the "Electronic Component Information Center (ECIC)," are contained in a research report recently made available to industry by the OTS. This report of research, by Battelle Memorial Institute under an Air Force contract, describes the elements of a machine-sorted punched-card system for recording, searching and tabulating data on an electronic component. Its importance is pointed up by the fact that proper selection of the most reliable and effective components is often the most costly and difficult step in the development of complex electronic systems. Complete details on the new system are available from the Office of Technical Services, Commerce Department, Washington 25, D. C., for \$4.25 each and should be ordered by number (PB 111548). . . . The Atomic Energy Commission . . . released 21 additional patents, including six in the electronics field. Non-exclusive, royalty-free licenses on the listed patents, as part of its program to make non-secret technological information available for use by industry, will be granted by the commission. Commission-held patents and patent applications released for licensing now total 747. Applicants for licenses should apply to the Chief, Patent Branch, Office of the General Counsel, U. S. Atomic Energy Commission, Washington 25, D. C., identifying the subject matter by patent number and title. The following six patents of interest to the electronics industry were released: High-Voltage Bushing, 2,692,297; Pulse Analyzer, 2,694,146; Electrostatic Amplifier, 2,696,530; Dual Circuit Electrical Safety Device, 2,696,539; Radio Electric Generator, 2,696,564, and Ion Source, 2,697,788. . . . The Office of Technical Services, Commerce Department, has listed studies in the field of electronics in its December-January issues of the "Bibliography of Technical Reports." The following government-sponsored research reports may be purchased from the Photoduplication Section, Library of Congress, Washington 25, D. C. "Basic Methods for the Calibration of

(Continued on page 29A)

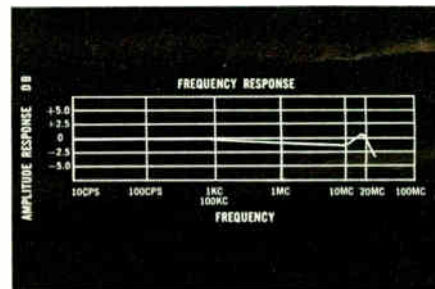
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10 cps to 20 mc

An oscilloscope deflection amplifier for measuring and analyzing pulses! Extremely wide band with extended low frequency response down to 10 cps. Will accurately analyze television signals. Excellent to increase the amplitude range of your vacuum tube voltmeters and signal generators.



MODEL VT



The Polarad Wide Band Video Amplifier offers an extremely wide band coverage: flat within $\pm 1\frac{1}{2}$ db from 10 cycles to 20 megacycles per second. It has a time delay of 0.02 microseconds and assures extreme stability because of its associated electronically regulated unit. A low capacity input probe is provided.

See other Polarad equipment advertised on pages 39A, 56A & 100A.



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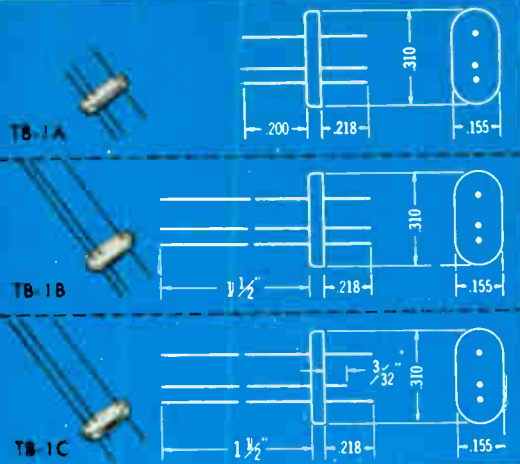
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Three electrode hermetically sealed Kovar bases supplied with closures. Lead lengths and pin layouts as illustrated. Cases are available in three types. Closures are press-fit to bases.



ALL CASES .300"
IN LENGTH



TC-1A
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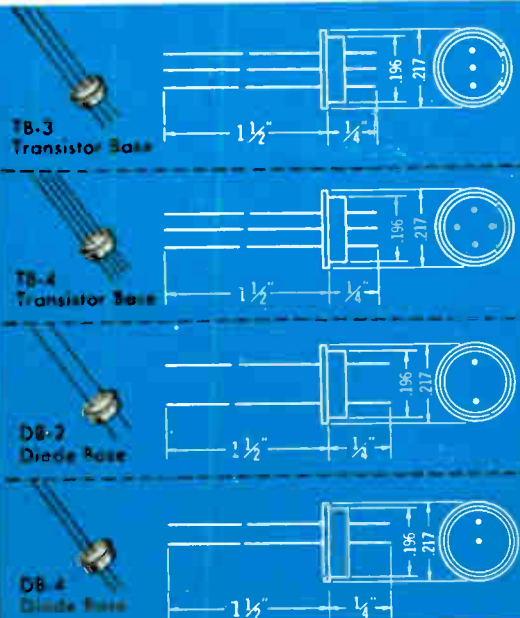


TC-1C
with hole
and dimple

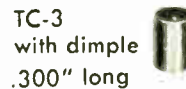
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CASES AVAIL-
ABLE WITH OR
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DIMPLE



TC-3
with dimple
.300" long



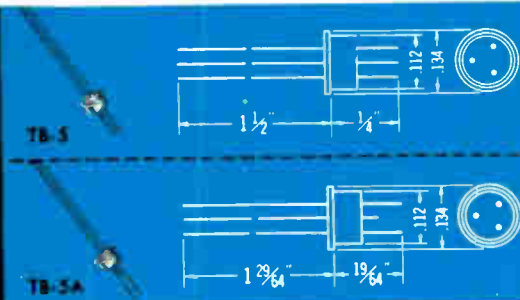
TC-3A
plain
.340" long

Where Special cases are required, EI will quote on your requirements on receipt of your drawings or specifications.

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SILVER CASES

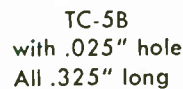
Available as illustrated. Cases are press-fit to bases.



TC-5
with dimple



TC-5A
plain



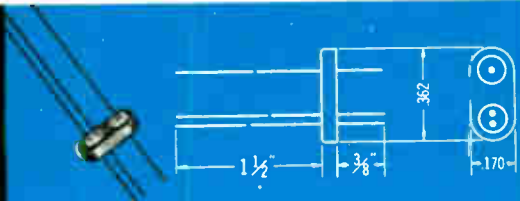
TC-5B
with .025" hole
All .325" long



TC-5C
plain .240" long

Type TB-6 TRANSISTOR BASE

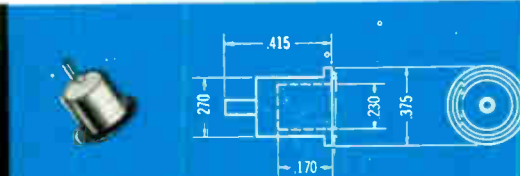
AVAILABLE WITH
TC-6 CLOSURE



TC-6 CLOSURE
Plain case .300"
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With welding projec-
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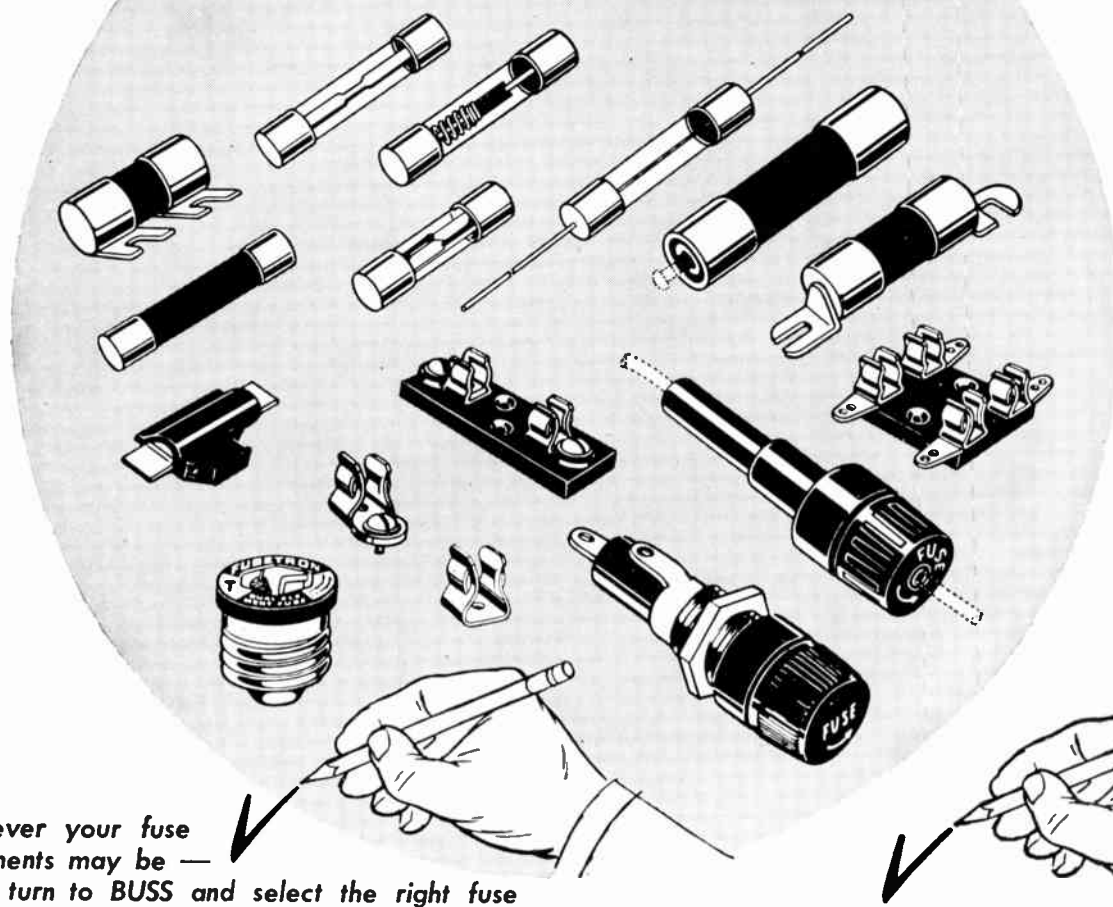
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★ These units are ideally suited for aircraft and guided missile requirements. Other typical applications that can benefit from their superior characteristics are power rectifiers in commercial equipment, magnetic amplifiers, clipping, meter protection and counter circuits. Anxiety over temperatures is completely eliminated when they are used in digital computers. Automation and control engineering suggest additional fields.

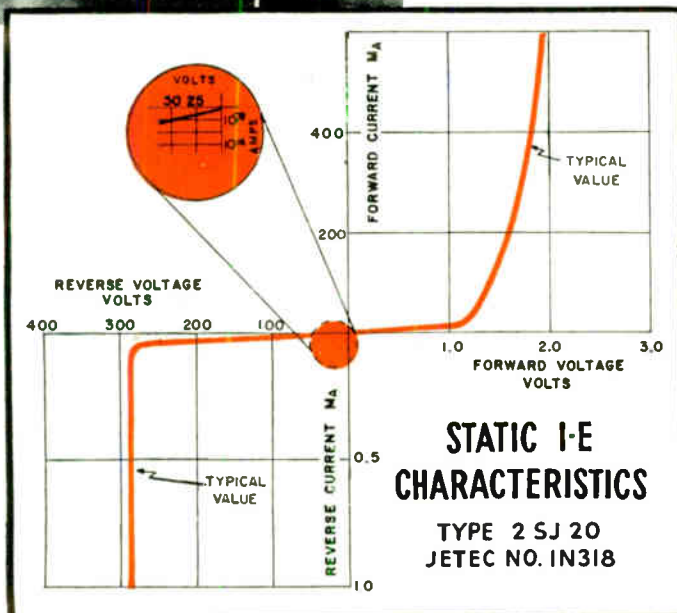
Performance:

1. Rectification Efficiency Over 99%
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7. Practically Flat Zener Characteristics

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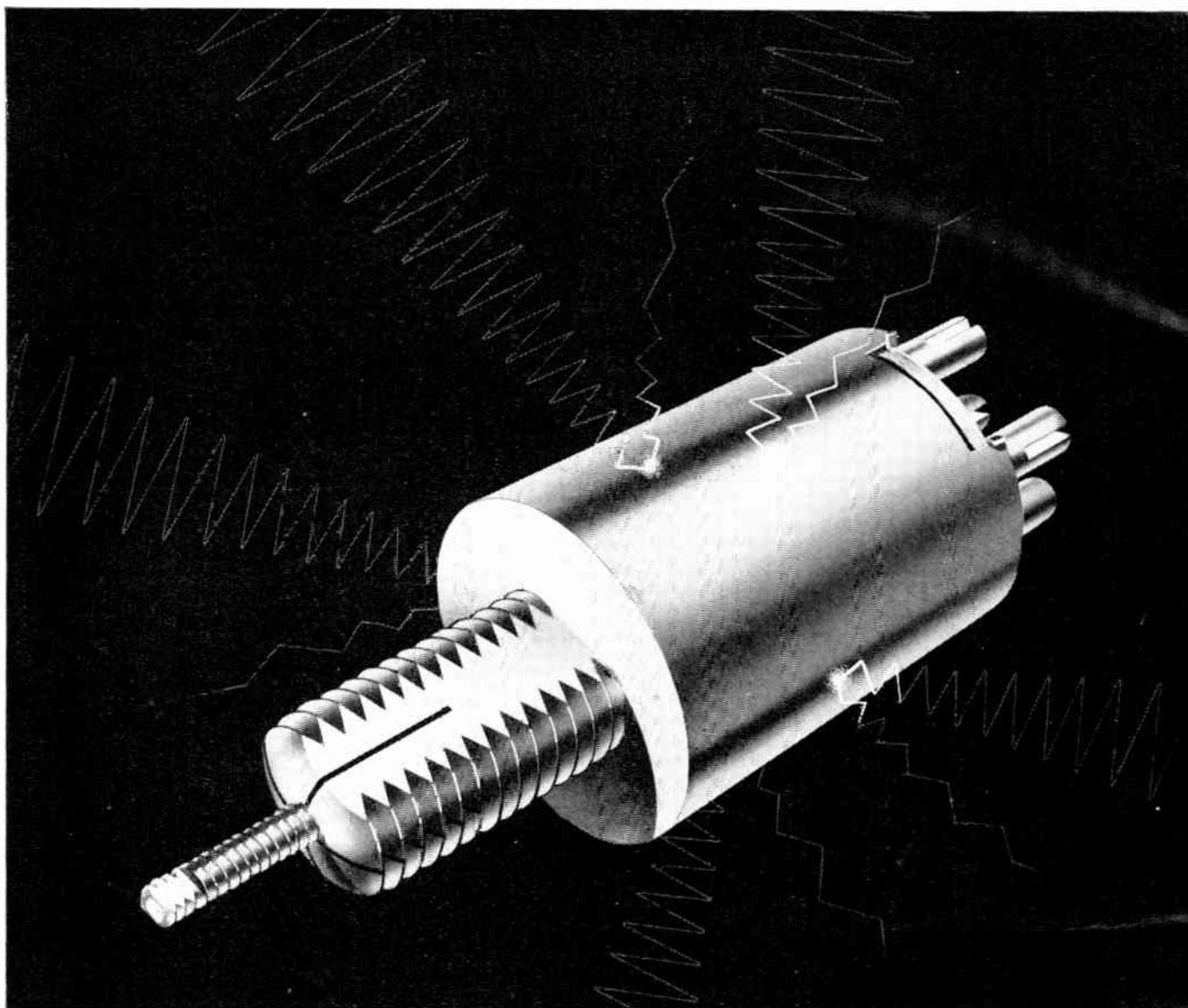
Write for fully illustrated and
informative Bulletin SR-18-3



Jetec No.	TYPE	Forward Drop @ 200 MA	Forward Current Continuous	Power Current Peak	Peak Inverse
IN 316	2SJ5	2V Max	200 MA	2A	50V
IN 317	2SJ10	2V Max	200 MA	2A	100V
IN 318	2SJ20	2V Max	200 MA	2A	200V
IN 319	2SJ30	2V Max	200 MA	2A	350V
IN 320	2SJ50	2V Max	200 MA	2A	500V

Units with peak inverse rating of 850 volts available in sample quantities.

BOGUE
BOGUE ELECTRIC
MANUFACTURING COMPANY
PATERSON 3, NEW JERSEY



Built for close "combat" in tight spots

Into the construction of this coil form goes C.T.C.'s rigid *quality control* to highest production standards.

The result is another C.T.C. *first* — a miniaturized coil form ($\frac{1}{16}$ " diameter by $\frac{1}{2}$ " high when mounted) that is shock-resistant and exceptionally rugged — shielded against radiation, electrically, and therefore ideal for "close quarter" use in I.F. strips and numerous designs where adjacent mounting is necessary.

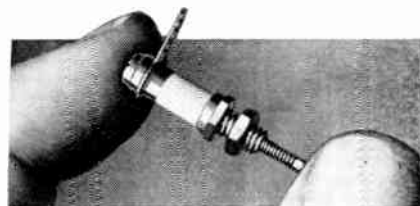
C.T.C.'s policy of continuous step-by-step quality control in the manufacture of every component means guaranteed performance. Already certified materials are doubly checked before manufacture.

Whatever your component need — let C.T.C. solve your problem — with either custom or standard designs of *quality-controlled, guaranteed* components — including insulated terminals, coil forms, coils, swagers, terminal boards, diode clips, capacitors and a wide variety of hardware items.

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C.T.C. now. For samples, specifications and prices — write today to Sales Engineering Dept., Cambridge Thermionic Corporation, 456 Concord Ave., Cambridge, Mass. On West Coast, contact E. V. Roberts, 5068 West Washington Blvd., Los Angeles 16 or 988 Market St., San Francisco, California.

Coil Form Data: C.T.C.'s IS-9 coil form has a brass shell enclosing a powdered-iron cup-core, tuning slug, phenolic coil form and silicone fibreglas terminal board. Three terminal boards are available with choice of two, three or four terminal layout. Forms, unassembled, may be had *without windings* . . . or wound and assembled to your specifications.



Capacitor: New CST-50 variable ceramic capacitor surpasses range of capacitors many times its size. Stands only $\frac{1}{2}$ " high when mounted, is less than $\frac{1}{4}$ " in diameter and has an 8-32 thread mounting stud. A tunable element of unusual design practically eliminates losses due to air dielectric giving large minimum to maximum capacity range (1.5 to 12MMFD).



CAMBRIDGE THERMIONIC CORPORATION

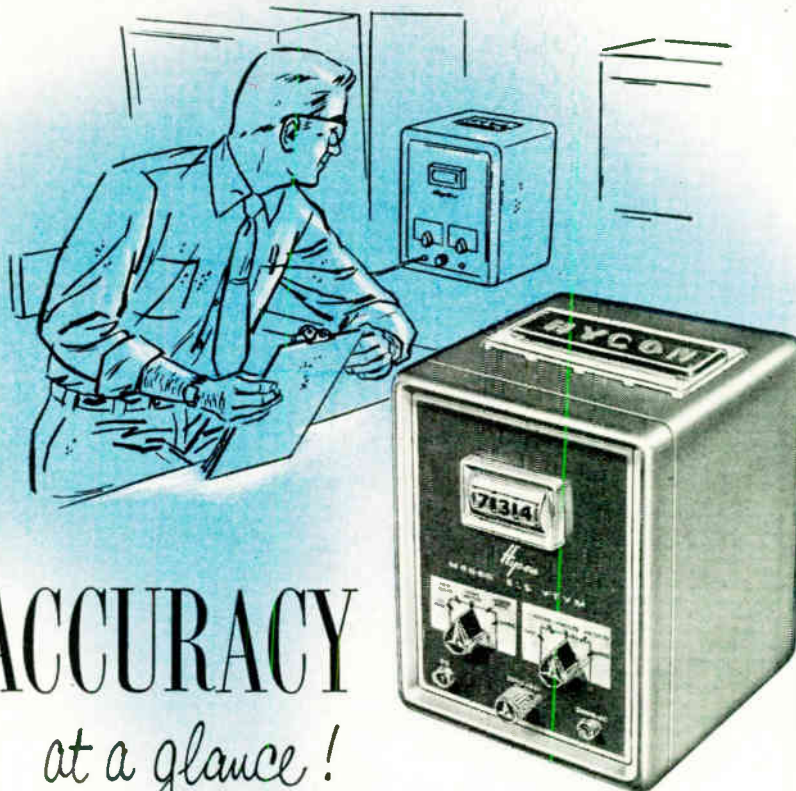
*makers of guaranteed electronic components,
custom or standard*





(Continued from page 24A)

Sonar Equipment," PB 115485, microfilm, \$4; enlargement print, \$12.75. "Development of Harmonic Mode Crystals. Final Report," PB 115059, microfilm, \$4; photocopy, \$11.50. "Diffraction of Pulses by a Circular Cylinder," PB 115215, microfilm, \$2.75; photocopy, \$6.50. "Dual-Channel Rotary Joint for 3300 mc," PB 115441, microfilm, \$1.50; photocopy, \$1.50. "Electronic Structure of Solids. II: The Perturbed Periodic Lattice," PB 115261, microfilm, \$4; photocopy, \$11.50. "IFF Antenna for Mounting on the Wing of a TBM Torpedo Bomber," PB 104816, microfilm, \$2; photocopy, \$2.75. "Interaction of Electrons and R-F Fields. Technical Report No. 1," PB 115243, microfilm, \$2.50; photocopy, \$5.25. "Recommended Designations of Radar Indicator Types," PB 110303, microfilm, \$1.50; photocopy, \$1.50. "Theory of Electromagnetic Corrections to Geometrical Optics," PB 115227, microfilm, \$1.50; photocopy, \$1.50. "Research Services and Investigations on Subminiature Multielement Diodes and Bistable Elements for Microtronic Circuit," PB 115274, microfilm, \$2.50; photocopy, \$5.25. "Services, Facilities and Materials Required for Research and Development of Accurate Fixed, Nonwire-Wound Resistors. Final Progress Report," PB 115407, microfilm, \$4.75; photocopy, \$14. "Studies and Investigations of a 100 Watt CW X-band Klystron," PB 115585, microfilm, \$3.25; photocopy, \$9. "Research on Electromagnetic Reflections from Surfaces of Complex Shape," PB 115547, microfilm, \$2; photocopy, \$2.75. "Propagation of Plane Electromagnetic Waves Past a Shoreline," PB 115630, microfilm, \$3.25; photocopy, \$9. "Reflection and Transmission of Electromagnetic Waves by a Spherical Shell," PB 115629, microfilm, \$2; photocopy, \$2.75. "Multiple Scattering of Radiation," PB 115643, microfilm, \$3.75; photocopy, \$10.25. "Mechanical Resonant Scanner," PB 115605, microfilm, \$2.25; photocopy, \$4. "Microwave Research," PB 115618, microfilm, \$2.25; photocopy, \$4. "Addition Theorems for Spherical Waves," PB 115650, microfilm, \$2.25; photocopy, \$4. "Characteristics of Ridge Waveguide. Space Charge Effects in Reflex Klystrons," PB 110058, microfilm, \$2; photocopy, \$2.75. "Diffraction of Electromagnetic Waves by a Plane Wire Grating, II," PB 115647, microfilm, \$2; photocopy, \$2.75. "Diffraction of an Arbitrary Pulse by a Wedge," PB 115649, microfilm, \$2.25; photocopy, \$4. "Linear Ordinary Differential Operators of the Second Order," PB 115648, microfilm, \$2.25; photocopy, \$4. "Maxwell's Equations in Spherically Symmetric Media," PB 115636, microfilm, \$2.25; photocopy, \$4. "On the Scattering Effect of a Rough Plane Surface," PB 115646, microfilm, \$2; photocopy, \$2.75. "Electronic Cursor for AN/AP5-15," PB 106688, microfilm, \$2; photocopy, \$2.75. "Intermodulation Distortion in Mixers," PB 115589, microfilm, \$2.50; photocopy, \$5.25. "New Ring Counter for Junction Transistors and Vacuum Tubes," PB 115588, microfilm, \$2.25; photocopy, \$4.



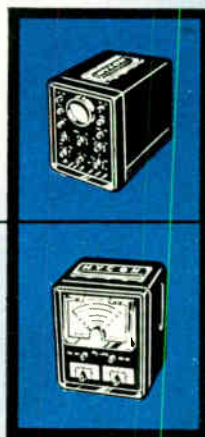
ACCURACY at a glance!

Just select the range you want... Hycon's new Model 615 Digital VTVM does the rest... gives you a *direct* reading in numerical form, complete with decimal point and polarity sign. There's no interpolation, no chance of reading the wrong scale. Even inexperienced personnel find the Model 615 easy to use... you just *can't* read it incorrectly!

Ideal for both laboratory and production-line testing, here's what the Model 615 offers...

- ... 1% accuracy on DC and ohms; 2% on AC
- ... 12 ranges... 0 to 1000 volts DC and AC; 0 to 10 megohms
- ... Illuminated 3-digit scale, with decimal point and polarity sign
- ... Response (with auxiliary probes) to 250 mc
- ... Shielded case; rugged, bench-stacking design; lightweight

Two more Hycon test instruments... designed for tomorrow's circuitry... ready for color TV...



MODEL 617 3" OSCILLOSCOPE...

Accurate enough for research, rugged enough for servicing. Features high deflection sensitivity (.01 v/in rms); 4.5 mc vertical bandpass, flat ± 1 db; internal 5% calibrating voltage. **SPECIAL FLAT 3" CRT FOR UNDISTORTED TRACE FROM EDGE TO EDGE.**

MODEL 614 VTVM...

Maximum convenience combined with unprecedented low cost. Plus features include: 21 ranges (28 with p-p scales); 6½" meter; 3% accuracy on DC and ohms, 5% on AC; response (with auxiliary probe) to 250 mc. **TEST PROBES STOW IN CASE, READY TO USE.**

See these Hycon instruments
... all in matching, bench-
stacking cases... at your
local electronic jobber.

Hycon Mfg. Company

2961 EAST COLORADO STREET
PASADENA 8, CALIFORNIA

"Where accuracy counts"

BASIC ELECTRONIC RESEARCH • ORDNANCE • AERIAL CAMERAS • ELECTRONIC SYSTEMS
ELECTRONIC TEST INSTRUMENTS • GO NO-GO MISSILE TEST SYSTEMS • AERIAL SURVEYS

STACKPOLE Fixed **RESISTORS**



... dependable, easy-to-solder molded composition types

Stackpole 1/2-, 1- and 2-watt resistors not only meet exacting performance standards, but save assembly time thanks to their highly-tinned, easily-soldered leads.

MIL-R-11A TYPES—in styles RC20, RC30, RC31, and RC42 available. Write for data on all MIL types.

STACKPOLE Variable **RESISTORS**



with versatile switching

Single, ganged and concentric shaft dual types in smallest sizes consistent with real dependability offer long, and trouble-free performance for today's requirements. Gold plated "ring spring" contactors assure low noise level. A complete array of unique midget line switches offers practically any desired switching arrangement, with types for both civilian and military use.

New!



Tab-mounting Bakelite shaft control

Just right for rear-of-chassis or concealed front panel controls in TV receivers . . . especially in high voltage circuits. Measures only 0.894" in diameter, yet handles a full .5-watt. Write for data on Stackpole Type LR-6.

AVAILABLE THROUGH PARTS DISTRIBUTORS! For name of nearest distributor stocking Stackpole resistors, switches and "EE" iron cores write: Distributors' Division, Stackpole Carbon Co., 26 Rittenhouse Place, Ardmore, Pa.

... A dependable source
of reliable components
for over 30 years

STACKPOLE Composition **CAPACITORS**

Cost-saving, low-value,
fixed types

Originated by Stackpole, these tiny units not only represent the simplest, most inexpensive capacitor design yet produced—but likewise have characteristics that make them more desirable than larger, more costly capacitors for many uses. 47 standard types, 0.1 to 10.0 mmf. Write for Stackpole GA Capacitor Bulletin.



STACKPOLE Iron **CORES**

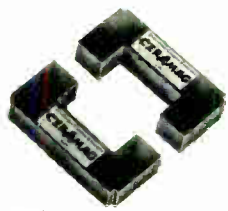


... to match any electrical or mechanical specification
Pioneers in modern iron core development, Stackpole offers practically any desired style and with assured uniformity of both electrical and mechanical characteristics.
Write for Iron Core Bulletin.

New "EE" Engineered Economy Cores

... standardized to meet 80% of all requirements at low cost.
Write for data on any type.

STACKPOLE *Ceramag*® **CORES** (Ferromagnetic)



for real uniformity! Wherever ferromagnetic cores are used, Stackpole Ceramag Cores have set the quality standards. But proved superiority in essential characteristics is only part of the story. Even more important is the fact that Stackpole Ceramag core characteristics are maintained with remarkable uniformity regardless of size, shape or production quantity. *The sample matches your specification "on the nose"—and each production unit is exactly like the sample!* Write for Ceramag Bulletin RC-9A including details on available grades and latest characteristic curves.

STACKPOLE Molded **COIL FORMS**



Cut Assembly Costs!

Reduce coil sizes and cut assembly costs with simplified point-to-point wiring and fewer soldered connections. Over 35 new types available in phenolic, iron, or phenolic with iron center sections. Axial or "hairpin" leads. Write for complete specifications on all types.

STACKPOLE Slide **SWITCHES**



... the economy switches of 1001 uses!

Over 20 types of these inexpensive little Stackpole slide switches cover just about every mechanical and electrical switching requirement for radio and television equipment, small motors, appliances, electrical toys, instruments, etc. For complete details, write for Stackpole Switch Bulletin RC-9B.

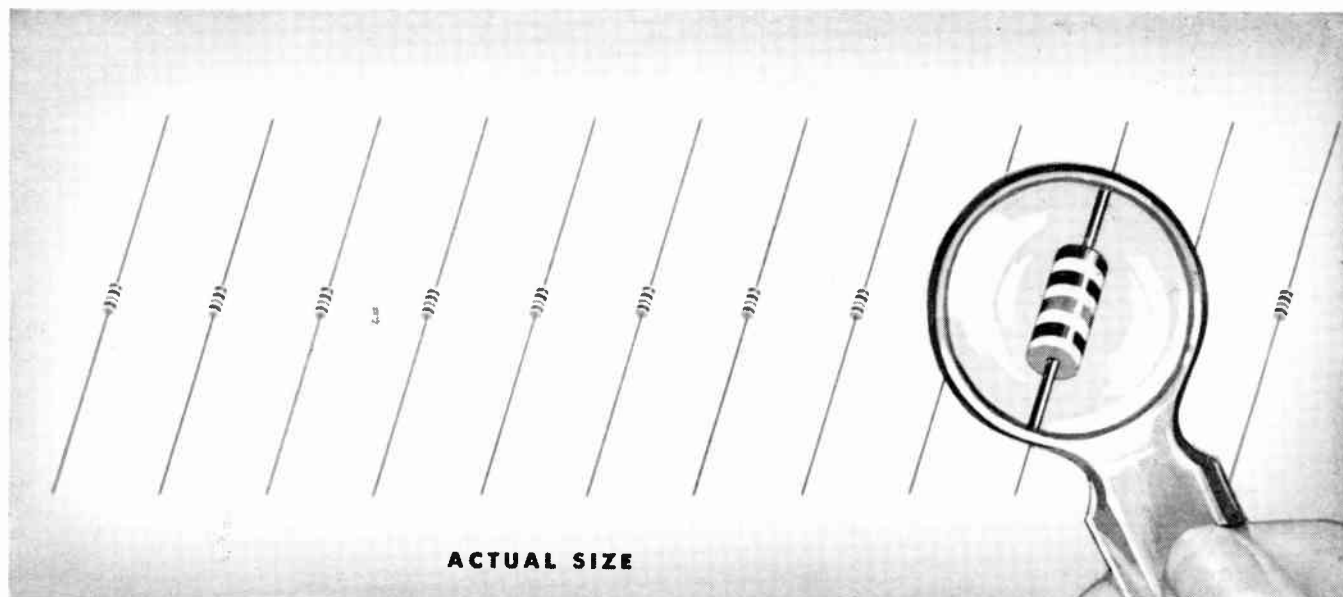


Engineering Samples are proof of the pudding!

Engineering samples of standard Stackpole components are available to quantity users. Send details of your requirement for recommendation by Stackpole engineers.

ELECTRONIC COMPONENTS DIVISION
STACKPOLE CARBON COMPANY, St. Marys, Pa.

STACKPOLE



Type TR "tiny" resistors shown in natural size—One unit is magnified to show color-code bands

NEW! TINY ALLEN-BRADLEY FIXED RESISTORS

**Type TR—Length—0.140 in. Diameter—0.067 in.
1/10th Watt—In all RETMA values and tolerances.**

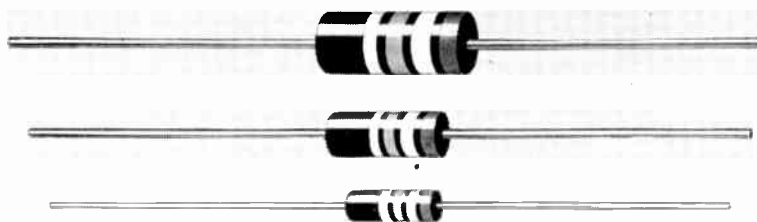
For electronic applications, where small size is a major consideration, the new Allen-Bradley Type TR "tiny" resistors are the ideal answer. While these 1/10th watt, miniaturized resistors are extremely small in size, they are a **QUALITY** product in construction and performance.

Because of their low noise level, they are especially suited for hearing aids and compact, portable receivers.

Type TR resistors have an insulating coating which affords a conservative insulation strength of 200 volts DC for continuous operation. These tiny resistors can be supplied in all standard RETMA, JAN-R-11A, and MIL-R-11A resistance values from 10.0 ohms to 22.0 megohms, inclusive in 5%, 10%, and 20% tolerances. If you build miniaturized electronic equipment, take advantage of Allen-Bradley Type TR **QUALITY** resistors.



**The Sign of
QUALITY**

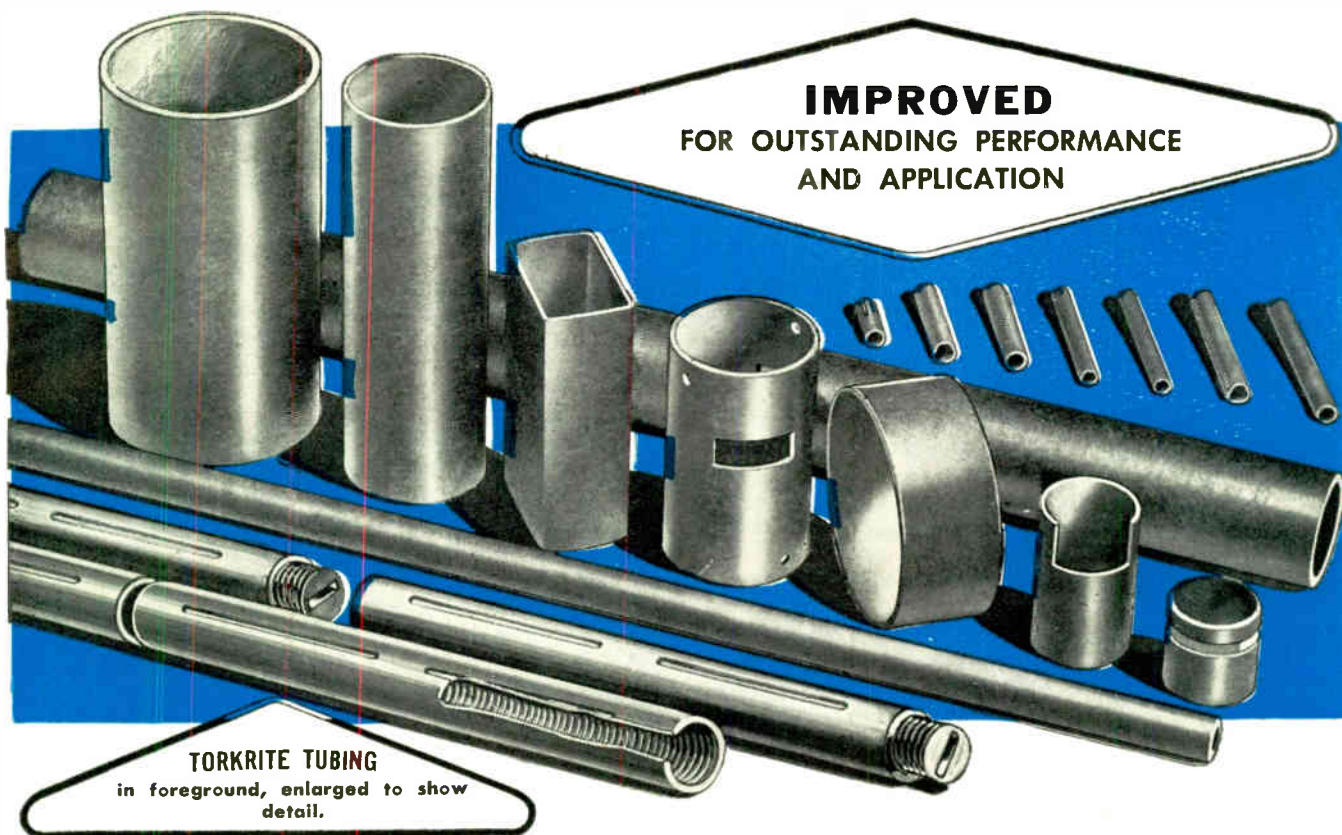


Allen-Bradley Types HB (2-watt), GB (1-watt), and EB (1/2-watt) solid molded fixed resistors are shown actual size in the above illustration.

Allen-Bradley Co., 114 W. Greenfield Ave., Milwaukee 4, Wis. • In Canada—Allen-Bradley Canada Limited, Galt, Ont.

ALLEN-BRADLEY

RADIO, ELECTRONIC AND TELEVISION COMPONENTS



IMPROVED
FOR OUTSTANDING PERFORMANCE
AND APPLICATION

TORKRITE TUBING
in foreground, enlarged to show
detail.



**TORKRITE
POSSESSES MANY
ADVANTAGES**

Torkrite affords unmatched recycling ability. After a maximum diameter core has been recycled in a given form a reasonable number of times, a minimum diameter core can be inserted and measured at 1" oz. approximately.

Torkrite has no hole or perforation through the tube wall. This eliminates the possibility of cement leakage locking the core or cores.

Torkrite permits use of lower torque as it is completely free of stripping pressure.

With Torkrite, torque does not increase after winding, as the heavier wall acts to prevent collapse and core bind.

Improved new Torkrite is now available in various diameter tubes. Lengths from 3/4" to 3 1/8" are made to fit 8-32, 10-32, 1/4-28 and 5/16-24 cores.



**WRITE for your copy
of our new
CLEVELITE folder**

CLEVELITE^{*}

LAMINATED PAPER BASE PHENOLIC TUBING

In seven specific grades, Clevelite is one of the finest and most complete lines of tubing available to the electronic and electrical industries.

Grade	Application
Grade E	Improved post-cure fabrication and stapling
Grade EX	Special grade for TV yoke sleeves
Grade EE	Improved general purpose
Grade EEX	Superior electrical and moisture absorption properties
Grade EEE	Critical electrical and high voltage application
Grade XAX	Special grade for government phenolic specifications
Grade SLF	Special for very thin wall tubing having less than .010 wall

High performance factors, uniformity and inherent ability to hold to close tolerances, make Clevelite outstanding for Coil Forms, Collars, Bushings, Spacers and Cores. Competent Research and Engineering facilities are always available to aid in solving those tough and stubborn design and fabrication problems. May we help you?

Fast, Dependable Delivery at all times.

WHY PAY MORE? For Good Quality . . . call CLEVELAND!

^{*}Reg. U. S. Pat. Off.

The **CLEVELAND CONTAINER Co.**
6201 BARBERTON AVE. CLEVELAND 2, OHIO

PLANTS AND SALES OFFICES at Chicago, Detroit, Memphis, Plymouth, Wisc., Ogdensburg, N. Y., Jamesburg, N. J.
ABRASIVE DIVISION at Cleveland, Ohio
CANADIAN PLANT: The Cleveland Container, Canada, Ltd., Prescott, Ontario



REPRESENTATIVES

NEW YORK AREA R. T. MURRAY, 604 CENTRAL AVE., EAST ORANGE, N. J.
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CHICAGO AREA PLASTIC TUBING SALES, 5215 N. RAVENSWOOD AVE., CHICAGO
WEST COAST IRV. M. COCHRANE CO., 408 S. ALVARADO ST., LOS ANGELES



April 1955

Servo Systems Brochure

Feedback Controls, Inc., 1332 N. Henry St., Alexandria, Va., has available a detailed and illustrated brochure describing their complete line of servos and associated equipment. It is a standardized system of units, adaptable to many problems.

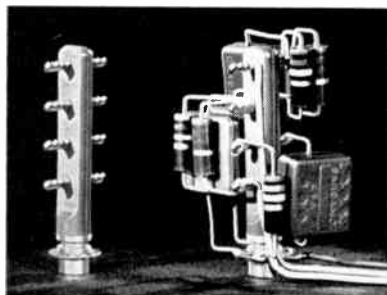
Casting Resin

A two-part casting resin which is easy to use and requires no additional catalyst is announced by **Emerson & Cuming, Inc.**, 869 Washington St., Canton, Mass.

Stycast 2340M has excellent electrical and mechanical properties. Warming the two components to about 125°F facilitates mixing and pouring the material which curves to a tack-free, brown, opaque, resin which is tough, and flexible. Its adhesion to metals, plastics, glass, etc. is also good. Stycast 2340M may be machined and is usable over a temperature range of -100°F to $\pm 400^\circ\text{F}$ without loss of physical or electrical properties.

Component Mount

A new component mounting post called the Tote-m-pole has been developed by **Sangamo Electric Co.**, Springfield, Ill., to improve the "bug resistance" of model and production wiring in government and industrial gear. It provides ideal mounting support for small components such as resistors, capacitors, diodes and transistors at their operating point. Critical leads to grid suppressor resistors, for example, can be reduced to pigtailed.



The device assists the engineer to get near optimum component density and point-to-point wiring. Fewer leads, cables and soldered joints are necessary. Users report as much as 5 feet of wire saved by each Tote-m-pole. Ventilation of parts mounted with this wiring aid is excellent. A melamine pole gives it its low tracking, heat-resistant properties. Post illustrated has 5 resistors and 4 capacitors attached.

The Tote-m-pole mounts with a single chassis drill hole. It can be reused many times for model mock-up or component replacement. It is adapted to jig wiring practices whether the jig is of cardboard for design study or a production type.

These manufacturers have invited PROCEEDINGS readers to write for literature and further technical information. Please mention your I.R.E. affiliation.

Interference-Field Intensity Meter

The NM-30A radio interference-field intensity meter developed by **Stoddart Aircraft Radio Co., Inc.**, 6644 Santa Monica Blvd., Hollywood 38, Calif., is a precision made equipment for the accurate measurement of field intensities of signals and rf disturbances within the frequency range of 20 to 400 mc.



Radio signals or interference, either radiated or conducted, may be measured through the use of accessories which are available for the equipment. Sine wave, pulsed rf, impulsive and random noise may be readily measured. Average, quasi-peak or peak values of complex waveforms can be selected. The NM-30A may also be used as a two-terminal frequency selective voltmeter.

Field intensity surveys, antenna radiation pattern studies and interference location and measurement are but a few of the many uses of the versatile NM-30A.

The NM-30A operates from either 105 to 125 volts or 210 to 250 volts ac, single phase, at any frequency between 50 and 1,000 cps.

Miniature Actuator Motor

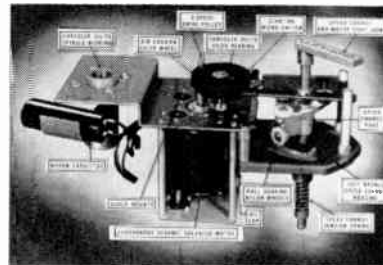
American Electronic Mfg., Inc., Instrument Div., American Electronics, Inc., 9503 W. Jefferson Blvd., Culver City, Calif., is now in production on a small actuator motor measuring 1.705 inches od $\times 2\frac{9}{16}$ inches long.



The motor operates on 400 cps. It is excited with 115 v on the fixed phase and 24 v on the control phase. Torque at stall is 2.9 inch/ounces with a power factor at stall torque of 50 per cent. No load speed is 5,100 rpm. Temperature range: -55°C to 90°C. Weight is 13.7 ounces. Additional data is available from the manufacturer.

Three-Speed Turntable

Gates Radio Co., Quincy, Ill., has a new three-speed turntable for continuous broadcast duty.



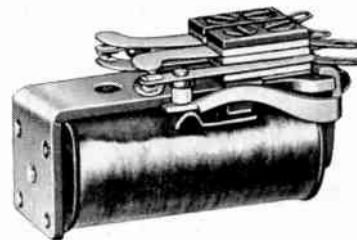
All three speeds, plus motor starting, are controlled by one-flip-type lever.

Speed change mechanism by the use of ball-bearing nylon wheels on a diagonal aluminum casting, running against a cast bronze cam, provides smooth, quiet operation for all three speeds. Increased torque is developed through a synchronous motor that operates the drive arrangement. It is claimed that the timing is exceedingly accurate and slippage practically eliminated. Three diameters on the motor shaft engage with the neoprene idler wheel which in turn drives inside platter rim.

Recession in center of platter with large spindle accommodates 45 R.P.M. records and eliminates the necessity for spindle change for 33 $\frac{1}{3}$ and 78 recordings. Size of cabinet is 29 $\frac{1}{2}$ \times 21 $\frac{1}{2}$ \times 21 $\frac{1}{2}$ inches.

Relay

Available in 1 to 5 amperes contact ratings, and in contact combinations from SPST to 6 PDT, the new relay, announced by the **Advance Electric & Relay Co.**, 2435 N. Naomi St., Burbank, Calif., offers a maximum sensitivity of 15 milliwatts per pole in the dpdt combination. This is an optimum combination, withstanding 10 G's vibration from 10 to 500 cps. When power is increased to 40 milliwatts per pole, vibration resistance rises to 30 G's. Sensitivity and vibration resistance decrease as additional contact combinations are added with the same amount of power.



Friction-creating hinge pins are eliminated, and the use of a beryllium copper armature retaining spring insures positive contact between armature and pivot point at all times.

Cross-bar palladium-type contacts are
(Continued on page 36A)

For maximum resistance in minimum space!

NEW Lollypop Precision Resistor Davohm Type 1273

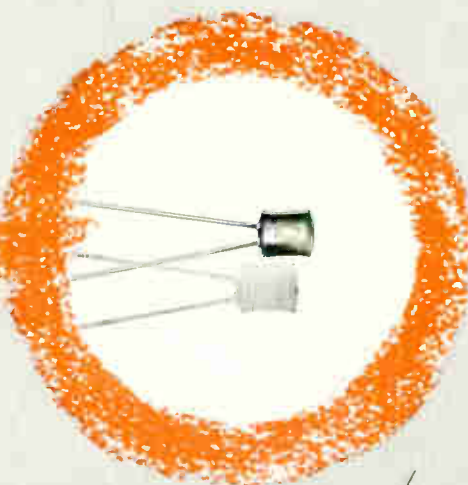
It's no trick at all with Daven's unique and extremely small size resistor to achieve ease of mounting in new printed circuit and transistor assemblies. The trick is **inside** this tiny unit . . . it's a completely new specialized winding technique developed by Daven, which enables them to use extremely fine sizes of resistance wire to obtain two or three times the resistance value that was previously supplied on a bobbin of this size.



You can't lick Daven's new wire-wound Lollypop Resistor

Only 1/4" in diameter by 5/16" long, yet is available in values as high as 400,000 ohms:

- * Fully encapsulated
- * Exceeds all humidity, salt water immersion and cycling tests as specified in MIL-R-93A, Amendment 2
- * Operates at 125°C continuous power without de-rating
- * Can be obtained in tolerances as close as $\pm .02\%$
- * Standard temperature coefficient is ± 20 PPM/ $^{\circ}\text{C}$. Special coefficients can be supplied on request



Below are other miniature encapsulated Daven resistors, part of the largest selection of precision wire-wound resistors available:

	Type 1250	Type 1170	Type 1195
Max. Ohms	450K	2 Meg.	760K
Dia.	1/4	1/2	1/4
Length	1/2	1/2	3/4
Max. Watts	1.8	1.3	1.4

All Daven resistors can be operated at 125°C continuous power without de-rating.



Write for complete resistor catalog.



THE **DAVEN** CO.

191 Central Avenue, Newark 4, New Jersey

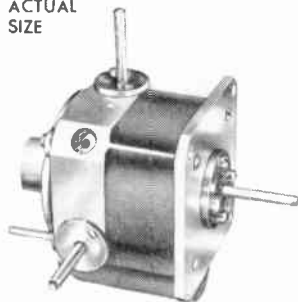
WORLD'S LARGEST MANUFACTURER OF ATTENUATORS

World Radio History

TWO NEW KEARFOTT COMPUTER COMPONENTS

MINIATURE MECHANICAL RESOLVER

1/2 ACTUAL
SIZE



An extremely compact unit measuring only 1 15/16" high, 1 3/4" wide and 2 1/8" long. It combines the functions of a ball and disc integrator and a spherical resolver. Will integrate the sine and cosine functions of an angle or resolve a vector displacement into its horizontal and vertical components.

INTEGRATING FILTER

Used to integrate a voltage signal from a specified minimum integration period to one approaching an infinite period of time. Available for DC to AC or AC to AC applications. These units eliminate harmonic and quadrature voltages to the servo motor driving a tachometer generator. Permits the use of a low gain, non-critical amplifier by effectively providing infinite gain.

DIMENSIONS:

AC-AC Filter 1.437" diam. x 2.484" long.
DC-AC Filter 1.969" diam. x 2.938" long.



1/2 ACTUAL SIZE

The close attention to details that has made Kearfott one of the leading producers of servo system components goes into the design and production of these devices. Detailed descriptions sent on request.

KEARFOTT COMPONENTS

INCLUDE:

Gyros, Servo Motors, Synchros, Servo and Magnetic Amplifiers, Tachometer Generators, Hermetic Rotary Seals, Aircraft Navigational Systems, and other high accuracy mechanical, electrical and electronic components.

ENGINEERS:

Many opportunities in the above fields are open—please write for details today.



A SUBSIDIARY OF GENERAL PRECISION EQUIPMENT CORPORATION

KEARFOTT COMPANY, INC., LITTLE FALLS, N. J.

Sales and Engineering Offices: 1378 Main Avenue, Clifton, N. J.

Midwest Office: 188 W. Randolph Street, Chicago, Ill. South Central Office: 6115 Denton Drive, Dallas, Texas

West Coast Office: 253 N. Vineland Avenue, Pasadena, Calif.



News-New Products

These manufacturers have invited PROCEEDINGS readers to write for literature and further technical information. Please mention your I.R.E. affiliation.

(Continued from page 31:1)

always properly aligned. A long nylon bobbin, fastened to the frame and core assembly by means of a snap-ring, permits the selection of many types of wire insulation, and allows, in addition, the winding of any type of coil including multiple and matched windings. The unit is insulated with Silicon glass, Kel-F, or Teflon tubing. Movable blades are actuated by ceramic bumpers, with nylon or linen-base bakelite optional.

The SQ withstands the Signal Corps tumbling tests and shock for mechanical damage in excess of 200 G's with operating characteristics unimpaired.

Arbor Listing

Precision Paper Tube Corp., 2035 W. Charleston St., Chicago 47, Ill., has published a new arbor list. It contains specifications on over 2,000 coil forms in all shapes, sizes, id's and od's.

Miniature Relays



Pacific Relays, Inc., 6819 Melrose Ave., Los Angeles 38, Calif., announces the new subminiature "CPL" series of miniature relays. It is designed for application where size, sensitivity and low and high temperature are a major factor. These units are hermetically sealed and are 3/4 inch x 1 1/8 inch x 1 3/8 inches and weigh 1 ounce. It is available to spdt (CPL-1) and in dpdt (CPL-2), contacting ratings to 5 amperes resistive at 28 vdc 115 vac or 3 amperes inductive. Temperature range is 55° to +125°, Vibration—15 G's through 500 cps, Shock—50 G's. Operational life is in excess of 1 million cycles under 1 ampere resistive load. For further information, write to the manufacturer.

(Continued on page 121:A)

Raytheon — World's Largest Manufacturer of Magnetrons and Klystrons



2J32 MAGNETRON



2K28 KLYSTRON



HELPING ESTABLISH RELIABILITY RECORDS

Raytheon Magnetrons and Klystrons in proved Gilfillan ASR-1 Radar

Civil Aeronautics Administration reports record-breaking reliability of Gilfillan airport surveillance radar. Boston International Airport had 8,760 hours continuous performance with only 7½ hours involuntary outage—less than 1/10 of 1%—from their Gilfillan installation.

Check these performance records of Raytheon tubes in the Gilfillan ASR-1. Average life, 2J32 Magnetron: 4,000 hours. Average life, 2K28 Klystron: 2,500 hours.

Your microwave and radar equipment offers extra reliability when you specify Raytheon Magnetrons and Klystrons. Use these rugged, reliable tubes in your present and proposed systems. Contact Power Tube Sales to take advantage of Raytheon's Application Engineer Service, without obligation. Write for free Tube Data Booklets.

Condensed Typical Operating Data							
	Power Output	Frequency Range, mc	Reflector Voltage	Resonator Voltage	Maximum Temp. Coef.	Tuning	Cavity
2K28	140 mw	1200-3750	-140 v. to -300 v.	300 v.	±.15	Mech. Inductive	Ext.
	Power Output	Frequency Range, mc	Anode kv	Anode Amps.	Pulse Width	P. R. R.	
2J32	285 kw min.	2780-2820 Fixed freq.	20	30	1 μsec	1,000	



*Excellence
in Electronics*

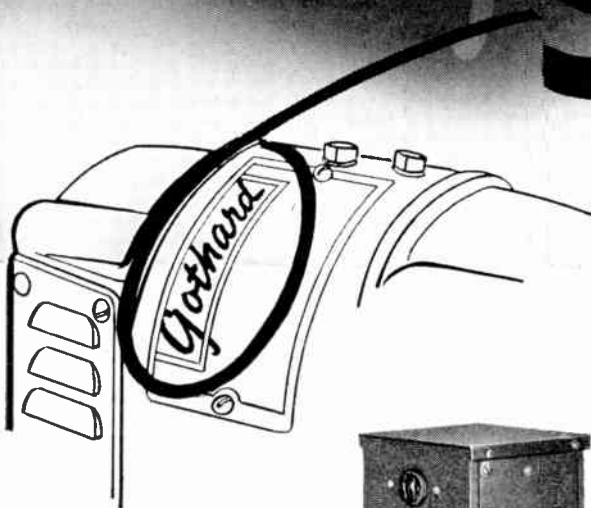
RAYTHEON MANUFACTURING COMPANY, Microwave and Power Tube Operations, Section PL-19, Waltham 54, Massachusetts

World Radio History

WE ARE CONVINCED —
that Gothard Converters, Dyna-
motors, Motor-Generators, Gen-
erators and special DC Motors
have been engineered to the
finest standards of any on the
market.

NOW **SANGAMO** GENERATORS, INC.

takes over!



**SANGAMO DC to AC
ROTARY CONVERTERS**
50 or 60 cycles, 115 or
230 volts AC from
available DC supply of
6, 12, 24, 28, 32, 48,
64, 115 or 230 volts.



**SANGAMO DC to DC
DYNAMOTORS**

Series "S" — Military
Series "G" — Commer-
cial and Mobile
Input voltages
6 to 115 volts
Output voltages
to 750 volts



**Special
Motor-Generators,
Generators,
Etc.**

Yes, after testing and retesting by
our own engineers — careful analysis
of Gothard engineering reports — and
most important, investigating the de-
gree of satisfaction that these products
are capable of providing to you, the
user — we decided this line is truly
worthy of the name Sangamo. We will
continue this line in larger quantities
and greater varieties under the new cor-
porate name of Sangamo Generators,
Inc. Now you will receive the added
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Electric Company.

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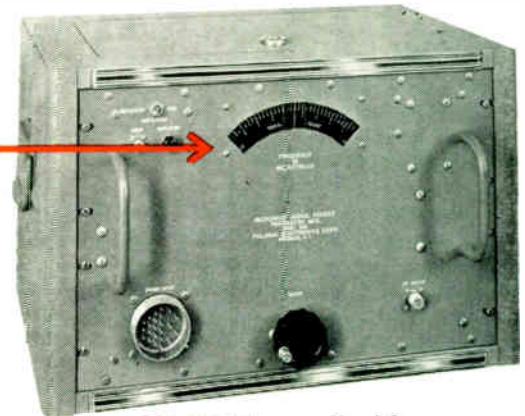
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*as fast
as you
can turn
the dial*



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Model KX, Klystron Power Supply, especially designed for Polarad Signal Sources. Works with all 5 models. Has special 1000 cps square wave output for modulating purposes.

Make microwave measurements rapidly.

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There are 5 models available, covering the range — 650 to 10,750 mc . . . each has approximately a 2:1 frequency range with continuous tuning . . . power output: 10 to 100 mw . . . external modulation: square wave or fm . . . temperature compensated klystron tube.

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MINIMUM POWER AVAILABLE FROM POLARAD SIGNAL SOURCES IN THE RANGE OF 650 TO 10,750 MC

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	650-1300MC	1050-2350MC	2200-4550MC	4350-8250MC	8000-10,750MC
MINIMUM POWER AVAILABLE (mw)	150	80	15	10	13
	400	150	60	70	30
	100	100	40	15	10

Signal Sources in the range 10,750 to 50,000 mc available on special order.



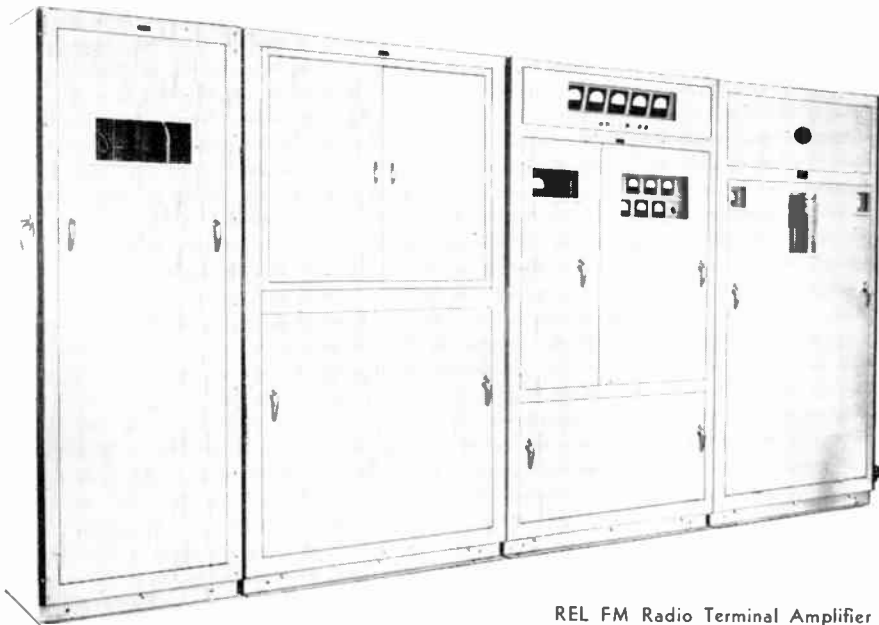
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Radio Engineering Laboratories uses Eimac klystrons in high power, beyond-the-horizon communication equipment



Eimac 3K50.000L klystron in klystron section of REL 10kw power amplifier.



REL FM Radio Terminal Amplifier employing Eimac klystron has frequency range of 400-1050mc.

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For further information on Eimac high power amplifier klystrons, contact our Technical Services Department.



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S A N B R U N O • C A L I F O R N I A



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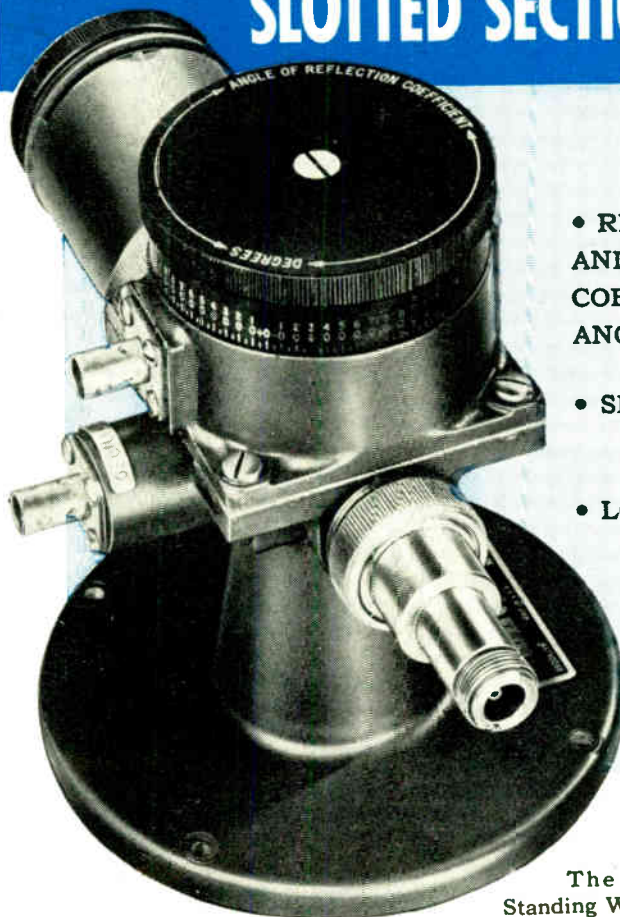
Abraham, W. G., Varian Associates, 611 Hansen Way, Palo Alto, Calif.
 Arams, E. R., RCA, Bldg. 55-1, Harrison, N. J.
 Bell, J. F., 618 Meadow Dr., Glenview, Ill.
 Bihernan, L. M., 703A, Lexington, China Lake, Calif.
 Borders, C. R., 2929 Broadway, New York, N. Y.
 Bristol, T. R., RD 1, Ballston Lake, N. Y.
 Buggy, R. V., 5934 N. Seventh St., Philadelphia, Pa.
 Chelgren, A. E., 576 Fairview Ave., Elmhurst, Ill.
 Crothers, H. H., c/o Electrical Engineering Dept., University of Illinois, Urbana, Ill.
 Eannarino, J. M., 610 Highland Ave., Rome, N. Y.
 Galagan, S., 136 Fessenden St., Newtonville, Mass.
 Goldstone, L. O., 226-17 Manor Rd., Queens Village, N. Y.
 Hawkins, W. G., 4506 Atwood, Fort Wayne, Ind.
 Hogan, D. L., 12512 Epping Ct., Silver Spring, Md.
 Hoglund, R. H., 1825 E. Lynn St., Seattle, Wash.
 Hutton, W. I., 35 Gilmore Blvd. North, Wappingers Falls, N. Y.
 Kahrilas, P. J., 9345 Loyola Blvd., Los Angeles, Calif.
 Kirby, T. H., Cow Hill Rd., Mystic, Conn.
 Klawnsnik, F., 142 Norwood Ave., Brooklyn, N. Y.
 Lapin, S. P., 1214 W. Jarvis Ave., Chicago, Ill.
 Manning, L. A., 649 Alvarado Row, Stanford, Calif.
 Marion, T. M., 1699 Carling Ave., Ottawa, Ont., Canada
 Mattingly, R. L., Bell Telephone Labs., Whippany, N. J.
 Meek, T. J., Jr., 1001 McLeod Bldg., Edmonton, Alta., Canada
 Meyer, A., 4280 Orchard La., Cincinnati, Ohio
 Mooney, V. J., 104 Carnation Ave., Florak Pk., L. I., N. Y.
 Pankove, J. I., RCA Labs., Princeton, N. J.
 Pelc, T., 2775 Delta Ave., Long Beach, Calif.
 Pihl, G. E., 46 Elm, Abington, Mass.
 Powers, A. B., Box 2117, Riverside, Calif.
 St. John, E. E., 4931 W. 122 St., Hawthorne, Calif.
 Serota, R. M., 1861 Burnette Ave., E. Cleveland, Ohio
 Shankweiler, R. G., Plainsboro Rd., Cranbury, N. J.
 Smith, H. M., Electron Tube Lab., Hughes Aircraft Co., Culver City, Calif.
 Smith, M. C., 312 Inverness Dr., Pasadena, Calif.
 Ulmer, H. W., 302 N. Clementine St., Oceanside, Calif.
 Whitcraft, W. A., Jr., 60 Division St., Malden, Mass.
 Woodrow, G. V., Jr., 1530 Providence Rd., RD 6, Towson, Md.
 Wroblewski, T., 7 Belgian Rd., Danvers, Mass.

Admission to Senior Member

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 Alma, G., c/o Bibliotheekcentrale, N. V. Philips' Gloeilampenfabrieken Eindhoven, Holland
 Arlowe, H. H., 1707 N. 50 St., Seattle, Wash.
 Azgapetian, V., 40 Shelter La., Roslyn Hgts., L. I., N. Y.
 Bennett, B. J., Stanford Research Inst., Stanford, Calif.
 Doehler, O., 21 Bld. de L'Ermitage, Montmorency S.e.O, France
 Edgerton, H. E., 205 School St., Belmont, Mass.
 England, W. B., 46 Lynwood Dr., Rochester, N. Y.
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(Continued on page 45A)

SUPERSEDES 100-1000 MC SLOTTED SECTIONS!



• READS VSWR
AND REFLECTION
COEFFICIENT
ANGLE DIRECTLY

• SMALL AND
COMPACT

• LOW IN COST

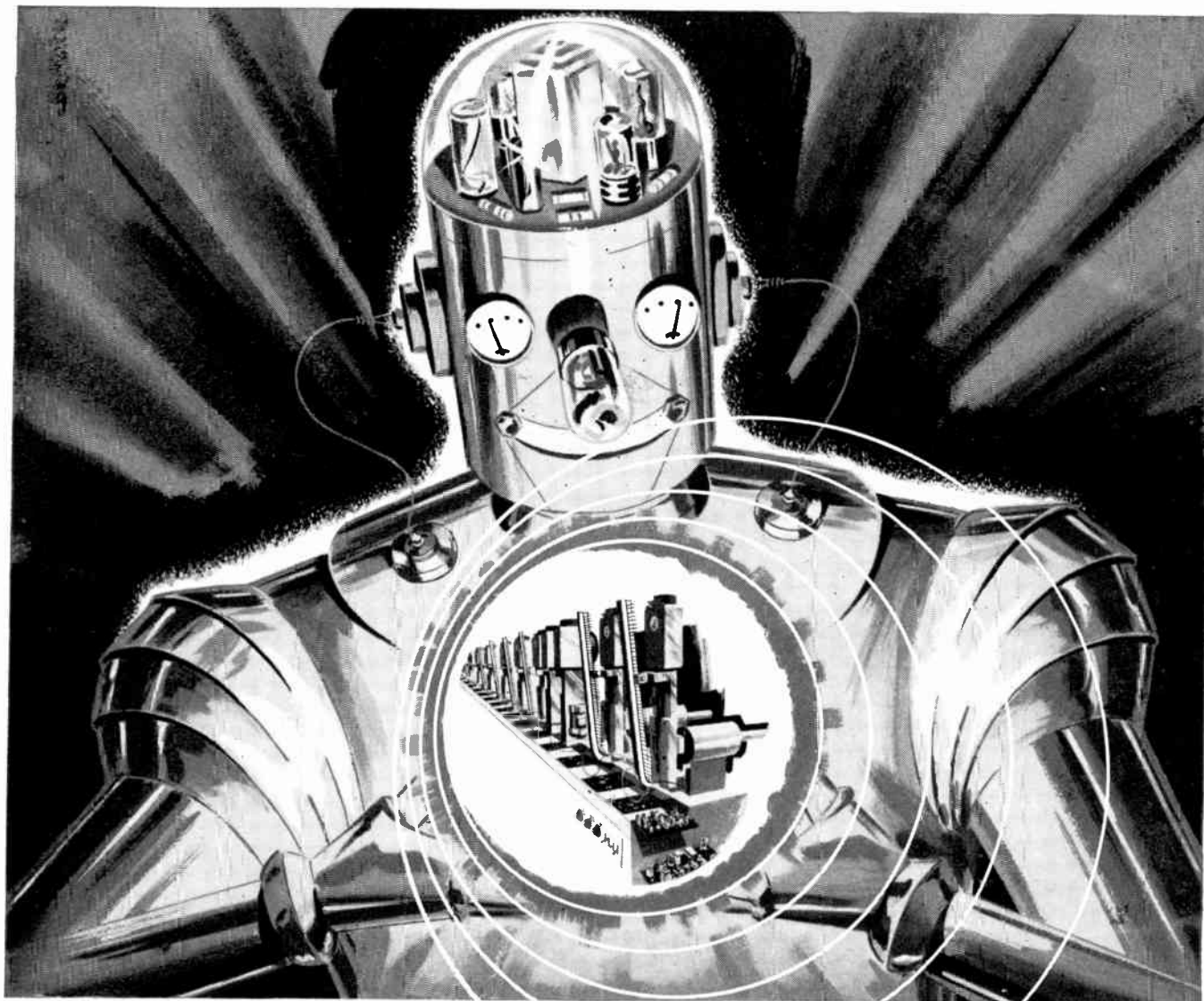
SPECIFICATIONS

Frequency Range:
100 to 1000 mc/s
 Residual VSWR:
Less than 1.05
 Accuracy of Reflection
Coefficient Angle:
Better than $\pm 5^\circ$
 Characteristic Impedance:
50 ohms
 Output Terminals:
Type N jack.
Other interchangeable
connectors
 Min. Input Signal:
Approx. 1 volt
at 100 mc/s,
0.1 volt at 1000 mc/s
 Dimensions:
8" l. x 5" w. x 5 3/4" h.
 Weight:
4 1/2 lbs.

The PRD Type 219 Standing Wave Detector is the small package, low cost solution for making measurements easily and accurately in the 100 to 1000 mc/s region. By connecting the output to a VSWR indicator, such as the PRD Type 277, VSWR may be read directly on the indicator meter. No special detection equipment is required. The reflection coefficient angle is easily determined merely by rotating the top drum dial to a minimum indication on the meter and reading the angle on the dial directly in electrical degrees. No calculations are required. The probe and crystal detector are self-contained.

Usually it is more convenient to work with VSWR and reflection coefficient angle directly instead of with other components of the measured impedance. When other quantities are also of interest, they can easily be read from a conventional impedance chart. Only \$475 f.o.b. N.Y. Write for PRD Reports, Vol. 3, No. 2, and for 1955 catalog.

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NOTE: COLOR SOUND FILM on Automation available for showing to technical or business groups. Film runs 9 minutes. Address requests to Public Relations Director, Admiral Corporation, Chicago 47, Ill.

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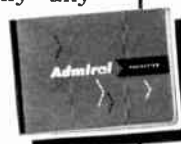
CODERS and DECODERS

DISTANCE MEASURING

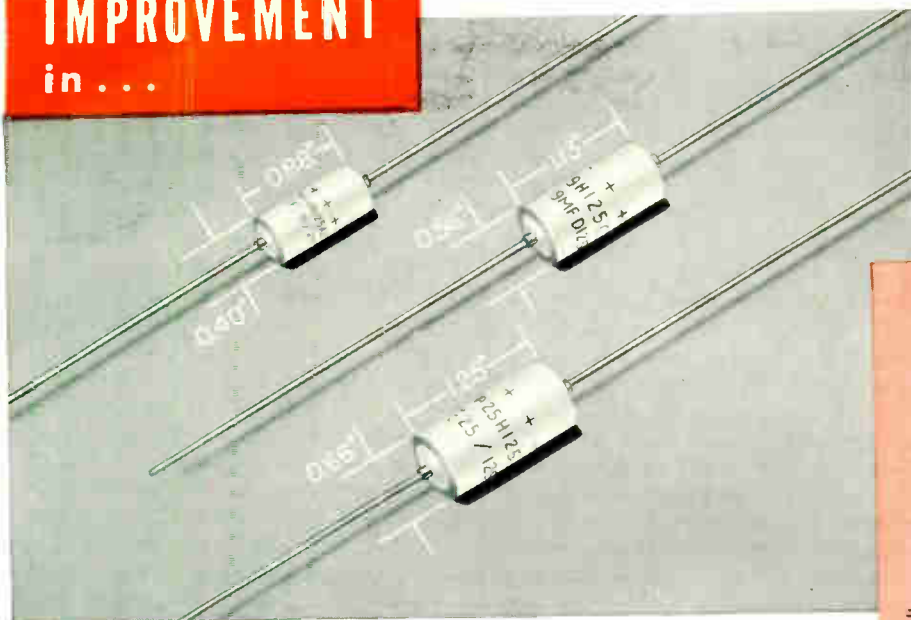
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in ...**



Hermetically Sealed
Fansteel Tantalum
Capacitors are made
in 3 sizes,
29 ratings.

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Now ...

Hermetically Sealed for High Temperature Operation

Wider Temperature Range: Continuous operation in ambient temperatures up to 125°C, with working voltage derated to 85% of nominal. Low temperature limit, -55°C.

Vastly Improved Leakage Characteristics: Precision construction results in lowest d-c leakage of all tantalum capacitors. Maximum leakage ranges from 1 to 8 microamperes as shown in table.

Closer Capacity Tolerances: All Fansteel Grade 1 Hermetically Sealed Tantalum Capacitors are manufactured to capacity tolerances of -15%, +20%. Grade 2 capacitors, also available, are -15%, +50%.

Rugged Construction: These capacitors have an actual metal to glass hermetic seal. The sturdy, plated steel case is insulated from the capacitor. They have passed rigorous tests for vibration, impact, humidity, reduced barometric pressure and thermal shock.

If your product requires capacitors of long life, small space and exceptionally stable characteristics over a wide temperature range, Fansteel Tantalum Capacitors may be the answer. Engineering samples may be ordered from the list at right.

CONDENSED LIST OF AVAILABLE CAPACITORS

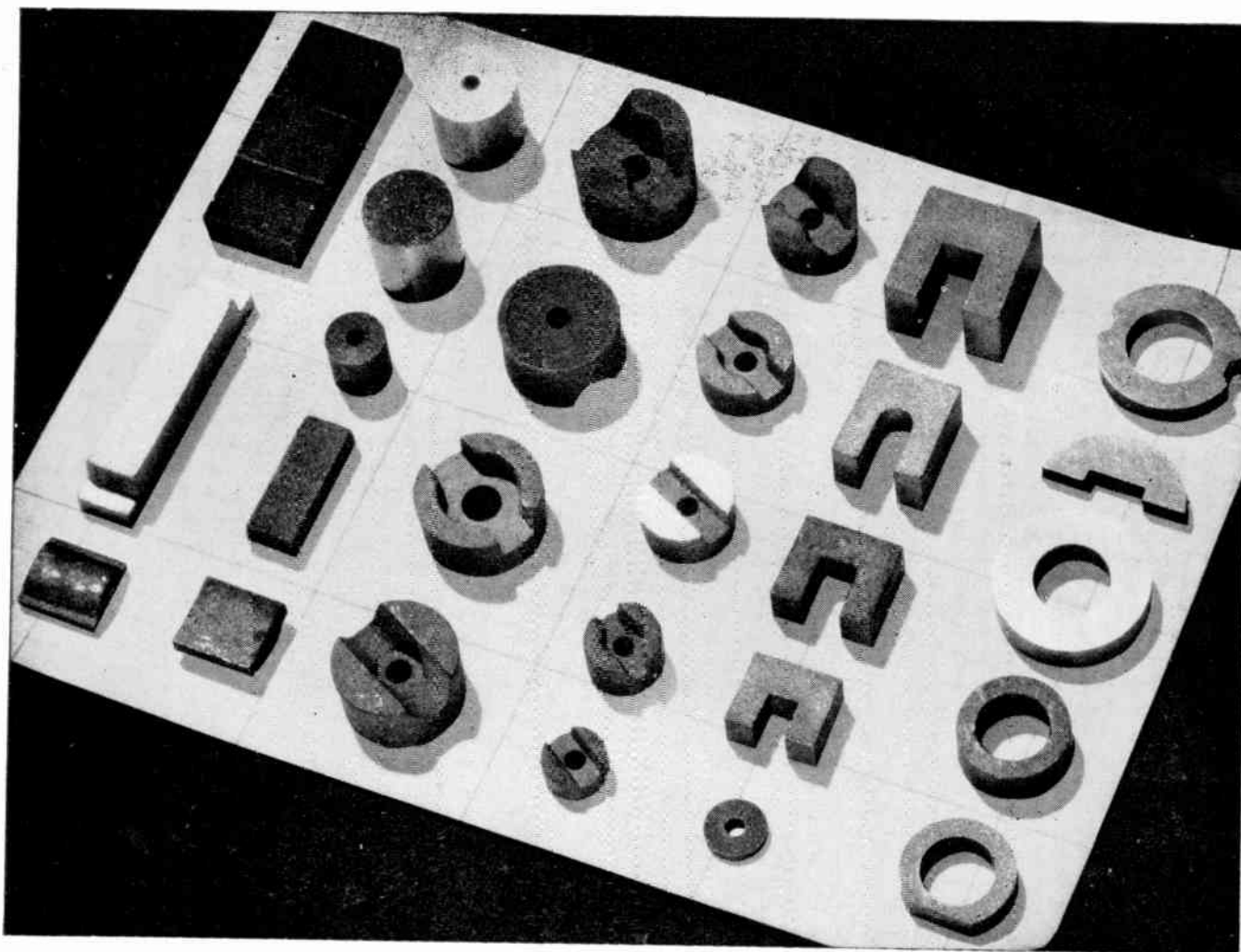


CATALOG NUMBER	CAPACITY MFD.	WORKING VOLTAGE, D-C	MAXIMUM D-C LEAKAGE ²
PP30H6A1	30	6	1.0
PP25HBA1	25	8	1.0
PP20H10A1	20	10	1.0
PP15H15A1	15	15	1.5
PP10H25A1	10	25	2.0
PP8H30A1	8	30	2.0
PP5H50A1	5	50	3.0
PP4H60A1	4	60	3.0
PP3.5H75A1	3.5	75	3.0
PP2H100C1	2	100	3.0
PP1.75H125C1	1.75	125	3.0
PP140H6A1	140	6	2.0
PP100H10A1	100	10	2.0
PP70H15A1	70	15	3.0
PP40H30A1	40	30	4.0
PP25H50A1	25	50	5.0
PP20H60A1	20	60	5.0
PP15H75A1	15	75	6.0
PP11H100C1	11	100	7.0
PP9H125C1	9	125	7.0
PP325H6A1	325	6	3.0
PP250H10A1	250	10	3.0
PP175H15A1	175	15	4.0
PP100H30A1	100	30	5.0
PP60H50A1	60	50	6.0
PP50H60A1	50	60	6.0
PP40H75A1	40	75	7.0
PP30H100C1	30	100	8.0
PP25H125C1	25	125	8.0

¹ -15%, +20% at 120cps, 25°C

² Microamperes, at 25°C

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Page-full of ideas for you

on *Sintered Magnets*



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Contains handy data on various types of Alnico Magnets, partial lists of stock items, and information on other permanent magnet materials. Also includes valuable technical data on Arnold tape-wound cores, powder cores, and types "C" and "E" split cores in various tape gauges and core sizes.

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(Continued from page 41A)

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 Haberstroh, A., 9 Black Horse La., Cohasset, Mass.
 Hillier, J., RCA Labs., Princeton, N. J.
 Kelly, J. H., 161 W. 16 St., New York, N. Y.
 Keyser, J. H., Jr., 4725 Hampton Rd., La Canada, Calif.
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 Souder, C. W., 2 Sycamore Ave., Glen Cover, L. I., N. Y.
 Welsh, J. P., 544 Lisbon Ave., Buffalo, N. Y.
 Zinn, W. H., Box 299, Lemont, Ill.

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 Benner, B., 930 Huron River Dr., Belleville, Mich.
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 Britton, C. C., Dept. Electrical Engineering, Colorado A & M College, Fort Collins, Col.
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 Campbell, R., 6352 —49 St., San Diego, Calif.
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(Continued on page 48A)



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Miniature HIGH Voltage - LOW Current POWER SUPPLIES

Developed by ATR
in cooperation with
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The ATR HIGH Voltage - LOW Current Power Supplies utilize ATR miniature vibrators and are ideally suited for flash-light cell operation in conjunction with:

- RADIATION MEASURING DEVICES.
- PHOTO-MULTIPLIER CELLS.
- INFRA-RED DETECTION EQUIPMENT.

SPECIFICATIONS

Five (5) basic ATR HIGH Voltage - LOW Current flash-light cell operated Power Supplies are available as follows:

ATR TYPE NO.	DC INPUT VOLTAGE	DC OUTPUT VOLTAGE	DC OUTPUT CURRENT
ZOD-451	1.5 VDC	800 VDC	50 μ a.
ZOD-455	1.5 VDC	900 VDC	100 μ a.
ZOD-471*	1.5 VDC	10,000 VDC	1 μ a.
ZOD-463	6 VDC	1,000 VDC	3 ma.
ZOD-443	6 VDC	16,000 VDC	1 μ a.

*AS FEATURED ABOVE

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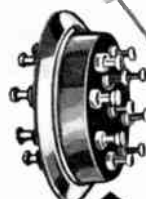
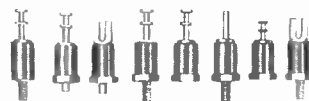
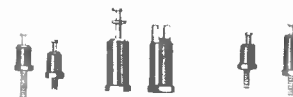
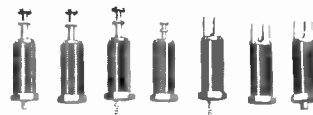
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FOR HIGH FREQUENCY
HIGH VOLTAGE

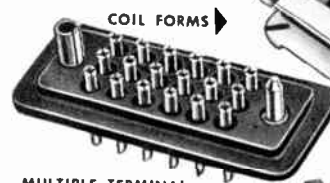
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BROAD-BAND MIXER CRYSTAL

TYPE IN286 covering the frequencies from
10,000 to 22,000 mc

Its broad-band characteristics make the new Sylvania Type IN286 especially useful in tunable radar systems and counter-measure devices. The IN286 is a coaxial, point-contact silicon crystal diode designed for use as a mixer in the frequency range from 10,000 to 22,000 mc.

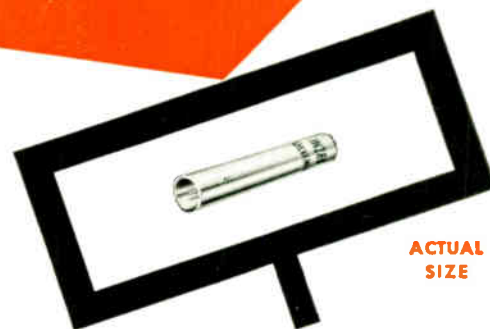
RF IMPEDANCE

The RF impedance of the IN286 is designed to match a 65-ohm load over its entire frequency range.

CRYSTAL HOLDERS

A variety of crystal holders may be used with the IN286

- standard X, Ku, K-band waveguide holders to cover appropriate segments of the band.
- WR-51 waveguide holder to cover the range from 15,000 to 22,000 mc.
- WR-75 waveguide holder to cover the frequency range from 10,000 to 15,000 mc.



ACTUAL
SIZE

SPECIFICATIONS

Conversion Loss8.5 db max.
Output Noise2.5 times max.
IF Impedance250—450 ohms
RF Impedance3.0 VSWR max.
Burnouteach crystal subjected to 20 mw (cw) at 10,000 mc.

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RELIABLE ELECTRON TUBES



With electronic controls taking over more and more operational functions in military and industrial applications, it is becoming increasingly important that the electron tubes used be dependable under extremely severe conditions. This applies particularly to installations in aircraft where tubes must operate reliably at high altitudes, while subjected to continuous vibration, varying voltages and frequent shock. Because of their advanced design and construction . . . born of never-ceasing research and special production skills . . . Bendix Red Bank Reliable Electron Tubes have the dependability necessary to meet these severe operating conditions. You can depend on our long, specialized experience to give you the right answer . . . for all types of regular as well as special-purpose tube applications. Tubes can be supplied to both commercial and military specifications. Call on us for full details.

Manufacturers of Special-Purpose Electron Tubes, Inverters, Dynamotors, Voltage Regulators and Fractional D. C. Motors

DESIGNATION AND TYPE					TYPICAL OPERATING CONDITIONS		
Type	Proto-type	Bendix No.	Description	Base And Bulb	Heater Voltage	Plate Voltage Per Plate	M.A. Load
5838	6X5	TE-3	Full Wave Rectifier	Octal T-9	12.6	350.	70.
5839	6X5	TE-2	Full Wave Rectifier	Octal T-9	26.5	350.	70.
5852	6X5	TE-5	Full Wave Rectifier	Octal T-9	6.3	350.	70.
5993	6X4	TE-10	Full Wave Rectifier	9-Pin Miniature	6.3	350.	70.
6106	5Y3	TE-22	Full Wave Rectifier	Octal T-9	5.0	350.	100.

Type	Proto-type	Bendix No.	Description	Base And Bulb	Heater Voltage	Plate Voltage	Screen Voltage	Grid Voltage	Gm	Plate Current	Power Output
5992	6V6	TE-8	Beam Power Amplifier	Octal T-9	6.3	250.	250.	12.5	4000	45. MA	3.5 W
*6094	6AQ5 6005	TE-18	Beam Power Amplifier	9-Pin Miniature	6.3	250.	250.	12.5	4500	45. MA	3.5 W
6385	2C51 5670	TE-21	Double Triode	9-Pin Miniature	6.3	150.	—	-2.0	5000	8. MA	—

*Tube Manufactured with Hard (Nonex) Glass for High Temperature Operation (Max. Bulb Temp. 300°C.)



DIVISION OF



EATONTOWN, N. J.

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Canadian Distributor: Aviation Electric Ltd., P.O. Box 6102, Montreal, P. Q.



(Continued from page 45A)

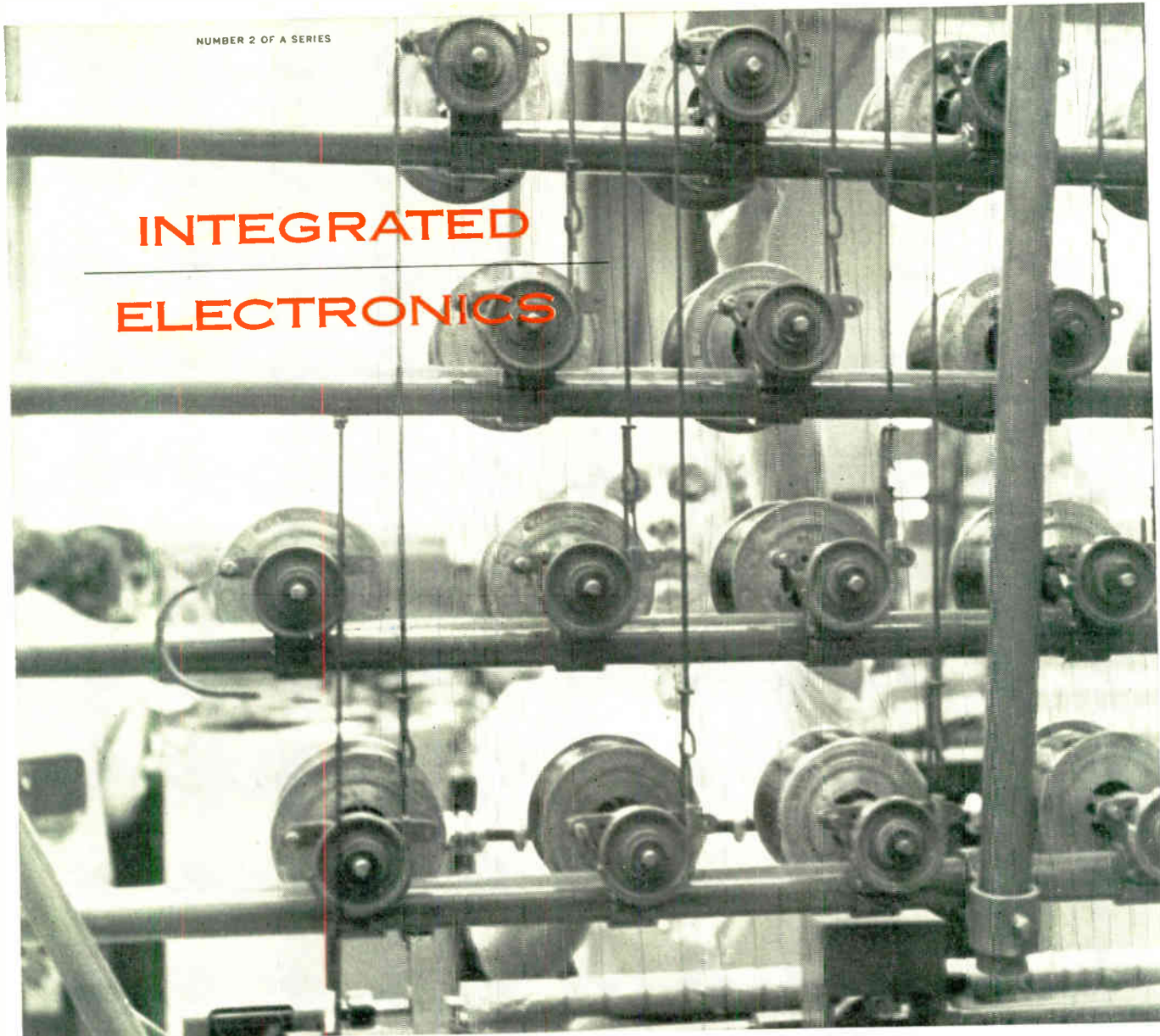
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Sleman, G. R., Nova Scotia Tech. College, Halifax, Nova Scotia, Canada
Spear, W. G., 1519 McPherson, Richland, Wash.
Starzer, F. J., 6837 W. Imlay St., Chicago, Ill.
Thompson, D. G., 4620 Knox Rd., Apt. 4, College Pk., Md.
Tuma, W. S., 4407 Grantwood Dr., Parma, Ohio
Turnage, H. C., 1107 Country Club Rd., Warwick, Va.
Weber, S. E., 205 W. Second St., Arcanum, Ohio
Weisman, I., 2811 Exterior St., Bronx, N. Y.
Wilkinson, J. H., 1438 Elkgrove Circle, Venice, Calif.
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Wood, C. E., 6650 S. Cicero Ave., Chicago, Ill.
Yeh, L. P., Elec. Dept., Colorado A & M College, Ft. Collins, Col.

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(Continued on page 52A)

INTEGRATED ELECTRONICS



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THE IMAGINATION FOR RESEARCH + THE SKILL FOR PRODUCTION

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 Missile Guidance &
 Control Systems
 Radar
 Noise Reduction
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 Communications
 Terminal Equipment
 Transistor Application



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SINCE 1929

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BOROHM® Deposited Boro-Carbon Resistors



Bulletin L-33

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CASTOHM® Ceramic Power Resistors



Bulletin L-29

Unusually light-weight wirewound power resistors with a unique integral core and coating having exceptional resistance to thermal shock and excellent heat conductivity. Ten humidity-resistant, tab-terminal styles available with ratings from 8 to 225 watts at 350°C. hot-spot. Meet MIL-R-10566, Amendment 1.

CMP and MP Miniature Power Wirewounds



Bulletin L-36

Lead-mounting, miniature power wirewounds for crowded chassis or printed circuits. MP types enclosed in a Fiberglass sleeve and coated with silicone-impregnated ceramic. CMP types encased in ceramic tube with ends hermetically sealed with silicone cement. Designed to MIL-R-26B. 3 to 10 watt sizes available.

SPECIALS



Bulletin L-37

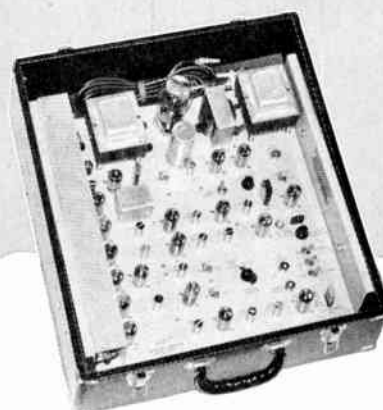
Hermetically-sealed Steatite resistors, Ayrton-Perry resistors, high-voltage surge resistors, card-type resistors, multi-section bobbin resistors, and many other special types are regularly produced to individual specifications.

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655XC

HICKOK

TELEVISION color bar generator



MODEL 655XC

This generator produces a standard 100% fully saturated NTSC color bar pattern on color TV sets. Regardless of future color television receiver design this color bar generator will be compatible.

Produces same type of signal that is transmitted over the air. All literature and alignment data is published around this standard NTSC signal. The Model 655XC provides signal for complete color alignment. When alignment is made with this type of signal, operator is sure that proper program colors will be displayed on the TV receiver.

The color bars appear on the TV screen in the following order from left to right: green, yellow, red, magenta, white, cyan, blue and black.

OUTPUT. . . EITHER R F. or VIDEO

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LOOK TO

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designed for specific applications

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Rated for 125°C operation, Transitron's silicon rectifiers provide high power handling ability and reliability at high temperature. They are specifically designed for magnetic amplifier and power supply applications. Send for Bulletin TE-1321.

Specifications and Ratings at 125°C

POWER SUPPLY TYPES			MAGNETIC AMPLIFIER TYPES		
TYPE	P.I.V.* (volts)	I _{dc} ** (ma)	TYPE	P.I.V.* (volts)	I _{dc} ** (ma)
1N341	400	400	1N332	400	400
1N343	300	400	1N334	300	400
1N345	200	400	1N336	200	400
1N347	100	1000	1N338	100	1000

* Peak Recurrent Inverse Voltage at full load
** Maximum Average Forward Current at full load



ACTUAL SIZE

SILICON JUNCTION DIODES

Transitron's silicon junction diodes are characterized by superior forward conductance and reliable operation up to 150°C. They are specifically designed for applications requiring extremely high inverse resistance at high temperatures. Send for Bulletin TE-1322.

TYPE	Forward Current at +1 V (ma)	Inverse Current at Specified Voltage (ua)		Maximum Working Voltage (volts)
		at 25°C	at 125°C	
1N137A	3	.03 at 20V	—	36
1N138A	5	.01 at 10V	—	18
1N137B	20	.03 at 20V	5 at 20V	36
1N138B	40	.01 at 10V	2 at 10V	18
1N350	20	.03 at 60V	5 at 60V	70
1N351	8	.03 at 100V	5 at 100V	120
1N352	5	.05 at 150V	10 at 150V	170
1N353	3	.10 at 200V	20 at 200V	225
1N354	1	.10 at 300V	20 at 300V	325



ACTUAL SIZE

SILICON BONDED DIODES

Transitron's silicon bonded diodes are specifically designed for high frequency and very fast switching applications at high temperatures. They are particularly useful in detector, discriminator and pulse circuitry. Send for Bulletin TE-1308.

TYPE	Forward Current at +1 V (ma)	Inverse Current at Specified Voltage (ua)	Inverse Breakdown Voltage
S4	1	1 at 10V	15
S5	1	.1 at 10V	20
S6	4	.5 at 5V	10
S7	2	1 at 10V	20
S8	1	1 at 10V	10

Operating frequency range 0-500 mc. Average Shunt Capacitance 0.8 uufd



ACTUAL SIZE

Transitron's special engineering group is available to assist you with specific applications. Inquiries concerning your particular design problems are invited.

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Glass Diodes



Silicon Diodes



Germanium Diodes



Transistors



Silicon Rectifiers

VHF

... Very High Frequencies



**RADIO INTERFERENCE
and FIELD INTENSITY *
measuring equipment**

Stoddart NM-30A • 20mc to 400mc

Commercial Equivalent of AN/URM-47

PRINTED CIRCUITRY... Modern printed circuits offer many advantages over conventional wiring, lighter weight, more compact units and freedom from many of the troubles normally encountered in conventionally-wired electronic equipment. Vibration becomes even less of a problem with printed circuits, adding to the many portable features already available with Stoddart equipment.

ADVANCED DESIGN... Specialized engineering and modern production techniques have produced one of the most advanced instruments for the accurate measurement, analysis and interpretation of radiated and conducted radio-frequency signals and interference ever manufactured. Designed to laboratory standards, rugged, and with matchless performance, the versatile NM-30A is an outstanding example of modern instrumentation. Its frequency range includes FM and TV bands.

SMALLER SIZE... A wider frequency range and higher standard of performance is incorporated into an equipment whose size is one-third that of any similar equipment ever manufactured.

SENSITIVITY... Sensitivity ranges from one to ten microvolts-per-meter, depending upon frequency and antenna in use.

APPLICATIONS... Field intensity surveys, antenna radiation pattern studies, interference location and measurement for checking radiation from virtually any mechanical or electrical device capable of generating or radiating radio-frequency signals or interference.

Stoddart RI-FI* Meters cover the frequency range 14kc to 1000mc

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cies.

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NM-20B, 150kc to 25mc
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AN/PRM-1A. Self-contained
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Includes standard broadcast
band, radio range, WWV, and
communications frequencies.
Has BFO.

UHF

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includes Citizens band and
UHF color TV-band.

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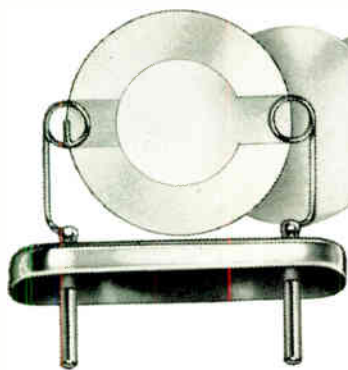
(Continued from page 48A)

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head, Mass.
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grade were approved to be effective as of
February 1, 1955:**

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Ault, J. C., 5 Marblehead Dr., Brentwood 17, Mo.
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Barsamian, A. S., 16579 Lilac Ave., Detroit 21,
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Bergen, S., Commercial Radio Equipment Co., 1319
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Bichara, M. R. E., Transvaal Chamber of Mines,
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(Continued on page 58A)



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That one quality is simply the highest that modern methods and machines can produce. It's assured by Midland's system of critical quality control—exact inspection and test procedures through every step of processing.

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*Whatever your Crystal need, conventional or highly
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WORLD'S LARGEST PRODUCER OF QUARTZ CRYSTALS

G.E. MECHANIZED PRODUCTION AT LOWER COST...ASSURES

Both types offer high reliability at temperatures

Take a close look at the transistor values G.E. now offers. Because production lines are now mechanized, these transistors are made in *less time* at *reduced cost*. Machine methods today assure strictest adherence to the top quality standards demanded of all General Electric Germanium Products.

Mechanization results in CONTROLLED CHARACTERISTICS, removing any inaccuracy on the part of the operator. Narrow limits are built into production transistors giving



TYPE 2N43A

a more uniform product.

In military and commercial applications these G-E transistors offer precision quality, topmost reliability at mass-volume prices!

General Electric's P-N-P junction transistor, 2N43A, is the first to be written into Air Force specifications! MIL-T-25096 (USAF) was actually written around this G-E product which was developed for the military. Now it serves an ever-increasing number of commercial as well as military applications.

APPLICATIONS AND SPECIFICATIONS

TYPICAL USES: Audio and Intercom Amplifiers, Servo Amplifiers, Carrier Current Amplifiers, Test Equipment, Fuel Gauges.

SPECIFICATIONS OF THE 2N43A and USAF 2N43A

Absolute Maximum Ratings:

Collector Voltage (Referred to base)	-45 volts
Collector Current	-50 ma
Collector Dissipation	150 mw
Storage Temperature	100° C
Collector Cutoff Current (-45 volts)	-10 microamps

DESIGN FEATURES:

STURDY CONSTRUCTION... meets critical military tests for shock, vibration, humidity, life.

SEALED JUNCTION... contamination gases permanently eliminated!

HIGH POWER OUTPUT... case design makes possible a collector dissipation of 150 mw.

HERMETIC SEAL... unaffected by moisture.

LONG LIFE... no change in characteristics during life of equipment.

MAKES TRANSISTORS AVAILABLE CONTROLLED CHARACTERISTICS

up to 100°C...are now available in production lots!

HIGH FREQUENCY TRANSISTOR

A new, revolutionary manufacturing technique, the exclusive G-E rate-growing process, coupled with the all-welded hermetic seal, now makes possible extra long life, and noticeably-reduced manufacturing costs by —

- Making 2000 or more transistors from one rate-grown crystal.
- Achieving uniform characteristics in all 2000 transistors—*eliminating wasteful rejects.*



TYPE 2N78

APPLICATIONS

For pulse and switching circuits, RF and IF amplifiers; high-frequency test equipment; telephone repeaters.

SPECIFICATIONS

Collector Voltage (Referred to Base)	15 V
Collector Current	20 ma
Emitter Current	— 20 ma
Storage Temperature	100° C.
High Frequency Gain at 2 mc	13 db

- For further details on specifications and prices, write *General Electric Co., Section X5245, Germanium Products, Electronics Park, Syracuse, N. Y.*



Billet of germanium is removed from furnace, prior to cutting into enough tiny pellets for 2000 transistors.

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VERTICAL MOUNT REGULATED POWER UNITS

MODEL PT-112

- Precise electronic voltage regulation
- Low ripple content
- Does not utilize electrolytic condensers
- Sturdy construction
- Provisions for rack mounting.



World Radio History

OUT OF THE LAB...

INTO THE LIGHT

Another
Hughes semiconductor
development,
available now
—the new,
subminiature
photocell,
Type
HD 2501.

SUBMINIATURE—smallest over-all volume of any photoelectric detector (approx. 1/1000 cu. in.).

FUSION-SEALED—only subminiature photocell with true glass-to-metal seal.

FAST—response at 20 kc down less than 5 per cent.

VERSATILE—non-directional sensitivity (360°) and photovoltaic properties lend unusual flexibility in equipment design.

RUGGED—welded whisker construction withstands severe shock, vibration, and acceleration.

RELIABLE—packaged in the famous Hughes one-piece glass envelope, impervious to moisture and external con-

tamination. A 100% testing ensures uniformity of characteristics.

Hughes Type HD 2501 germanium point-contact photocell can be used as a light detector in card readers, binary encoding and decoding wheels, motion picture sound—and for near infrared applications. Because of this infrared response, tungsten light sources can be

operated at voltages below normal and their effective life increased accordingly.

For other diode applications in high and low temperature ranges, be sure to check the growing family of Hughes semiconductors. Scores of types of germanium point-contact and silicon junction diodes are available in RETMA, JAN, and Special listings.

HUGHES

Aircraft Company, Culver City, Calif.

SEMICONDUCTOR SALES DEPARTMENT



New York Syracuse
Philadelphia Chicago

Photocell dimensions, glass envelope
Length: c.263-inch, maximum
Diameter: c.086-inch, maximum

TYPE HD 2501 PHOTOCCELL—SOME CHARACTERISTICS AT 25° C.

Dynamic Breakdown Voltage: 175 Vo is. minimum. Minimum Sensitivity: 1mA/L at 50 Volts and 25 ML.
Maximum Dark Current: 20 μ A at 50 Volts. Dynamic Resistance: 1 megohm at 50 Volts and 25 ML.



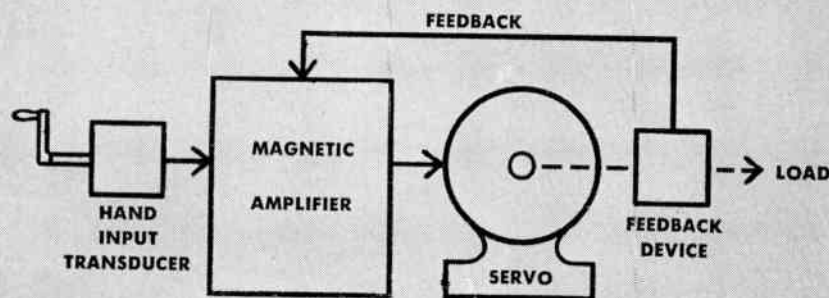
SINCE 1915 LEADERS IN AUTOMATIC CONTROL



Membership

(Continued from page 52A)

Magnetic Amplifier SERVO SYSTEMS In the Horsepower Range



The task called for a rugged, reliable drive of a motor which would deliver up to four horsepower on acceleration, and at least 1½ horsepower continuously. Maintenance requirements to be at a minimum. The drive must be able to stand high shock and operate under several G's. It must operate in temperatures from -65° to 165°F .

Ford engineers developed such a drive in a magnetic amplifier servo system. It could be made for position control or rate control, and it operated smoothly and accurately under an unbalanced load condition. The gain or current-output/current-input (with motor stalled) = 60,000; with a maximum output of over 90 amps.

This is typical of the solution of engineering problems in the field of servomechanisms by the Ford Instrument Company. Should you have a problem such a solution may answer for you, write and indicate your needs. Ford Instrument Company's forty years of experience in developing, designing and manufacturing special devices in the field of automatic control will help you find the answer.



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Busch, K. J., Bell Telephone Laboratories, Inc., Murray Hill, N. J.
Caldwell, A., 11024—100 Ave., Edmonton, Alta., Canada
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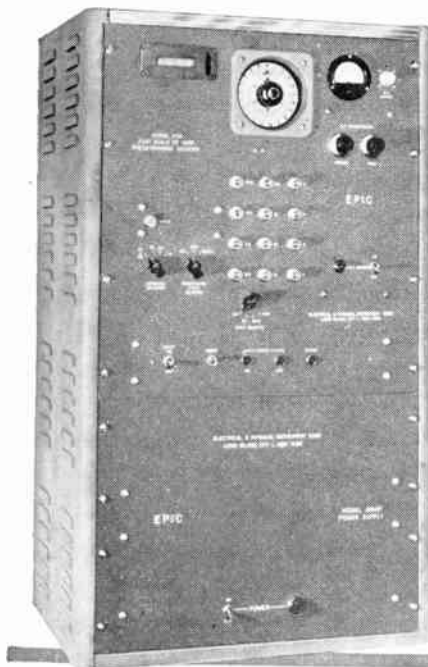
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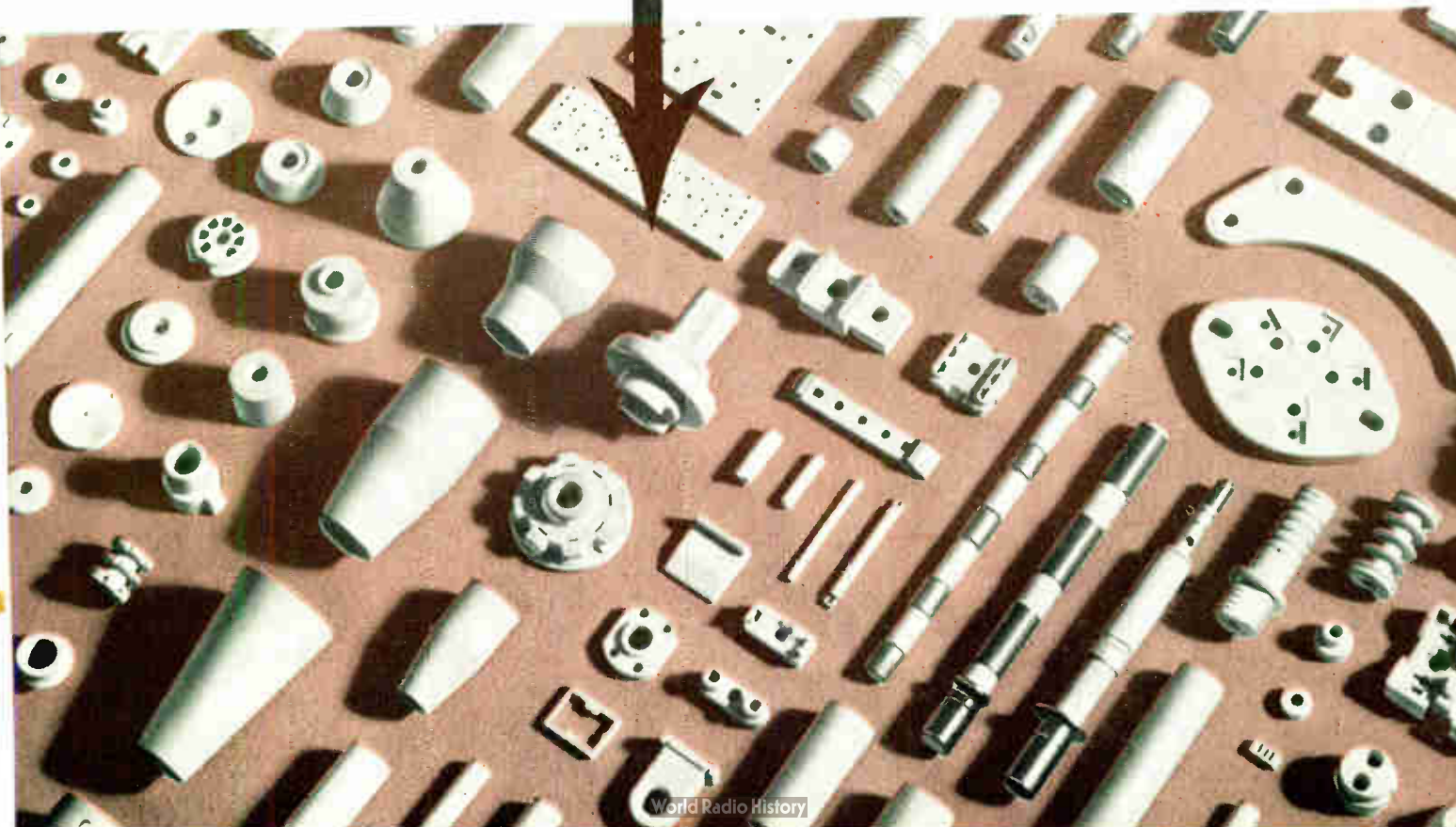
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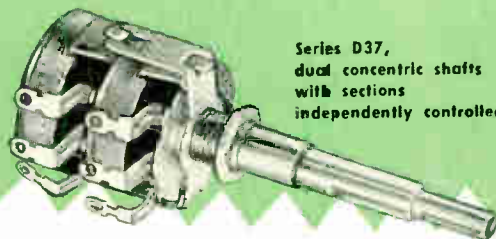
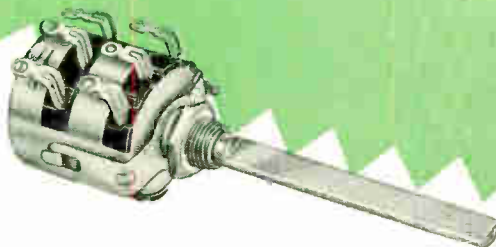
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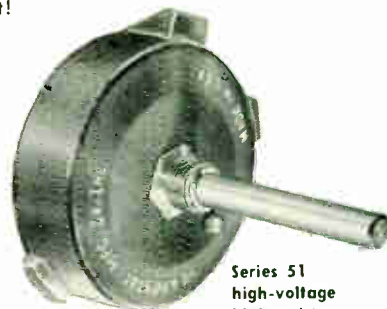
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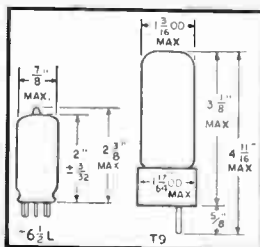
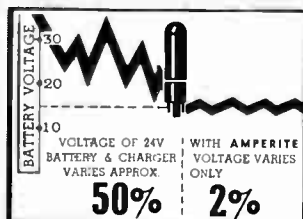


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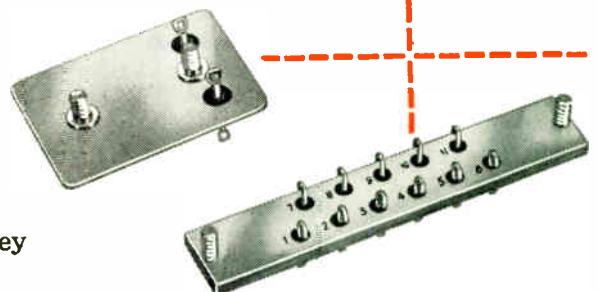
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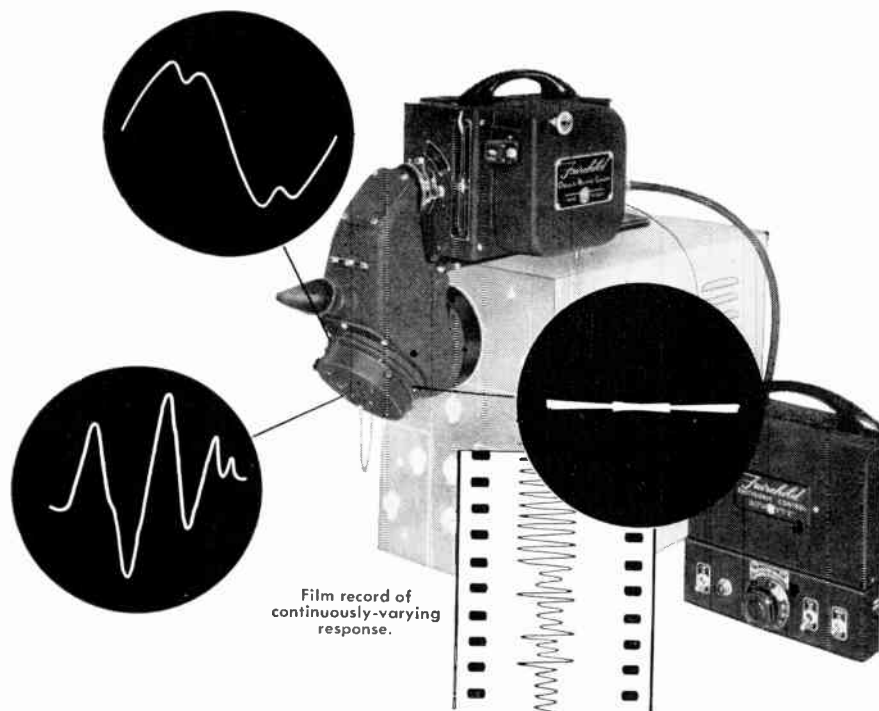




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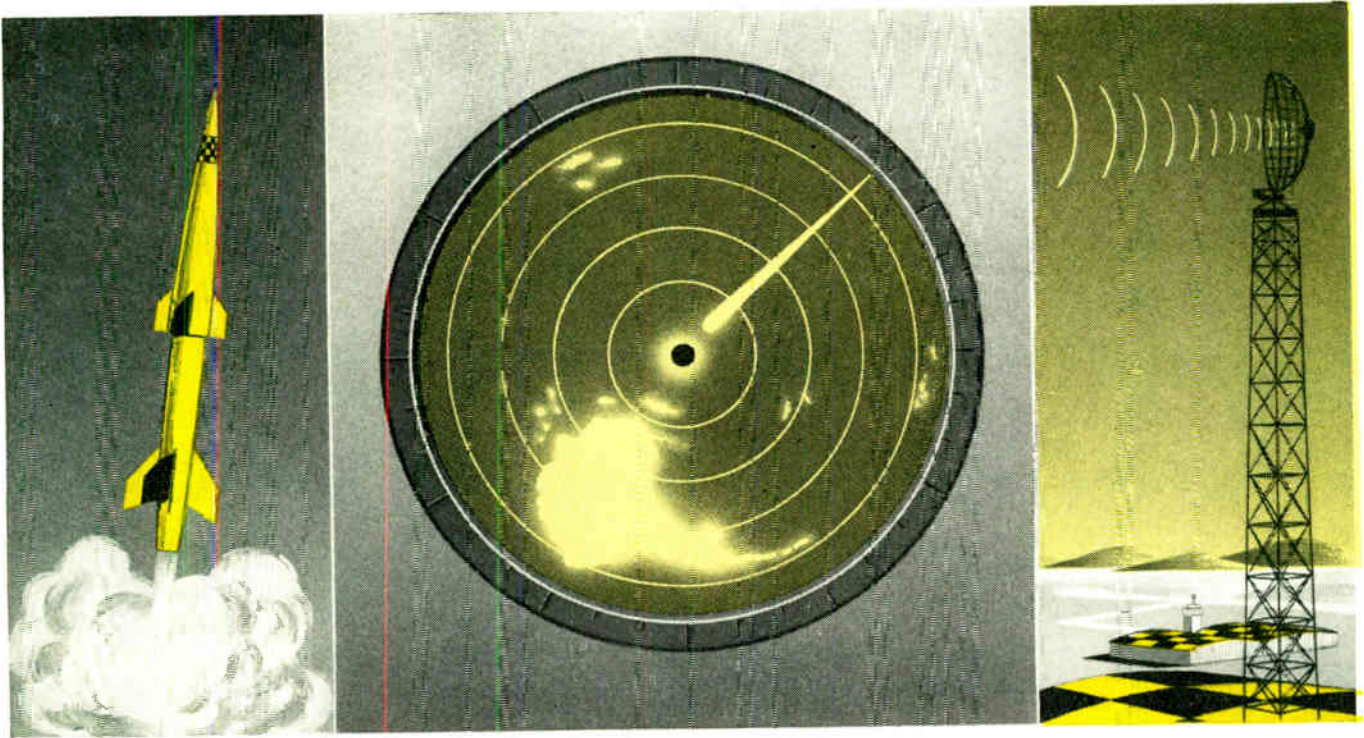
Fairchild-Polaroid® Oscilloscope Camera

You can produce a print of any stationary or single-transient pattern in one minute with this Fairchild camera. The trace reads from left to right and is reduced to exactly one-half life size for easy measurement. Two images may be exposed on each $3\frac{1}{4} \times 4\frac{1}{4}$ print.

For more information on Fairchild oscilloscope cameras and how they can assist you in engineering and research analysis, write Fairchild Camera and Instrument Corporation, 88-06 Van Wyck Expressway, Jamaica, N. Y., Department 120-2211.

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VARIAN'S NEW VA-94



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Frequency	16.5 kmc
Resonator Voltage	300 v
Resonator Current	38 ma
Reflector Voltage	-150 v
Power Output (VSWR < 1.1)	40 mw
Electronic Tuning	65 mc

Varian now offers the most advanced reflex klystron ever developed for airborne radar local oscillator and beacon service. The VA-94 provides a minimum power output of 20 mw throughout its range of 16 to 17 kmc ... to give you absolutely reliable operation at any altitude without pressurization.

Exclusive Varian features include a unique brazed-on external tuning cavity ... to assure you of excellent frequency stability, extremely low microphonics, slow tuning rate and long tuning life. Its single screw tuner adapts easily to motor tuning. The VA-94 weighs only four ounces and mates directly with standard waveguide flanges.

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with the NEW

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0.05cps to 30KC
down to 0.01cps with corrections

VOLTAGE RANGE

0.02 to 200V peak to peak
lowest reading corresponds to
7.07mv rms of a sine wave

ACCURACY

3% throughout ranges
and for any point on meter

IMPEDANCE

10 megohm by an average
capacitance of 30 μ f

OPERATION

Unaffected by line variation
100 to 130V, 60 cycle, 45 watt

APPLICATIONS

The Ballantine Infrasonic Voltmeter Model 316 has been introduced to satisfy a growing need for an instrument to facilitate the measurement of ultra-low frequency potentials as are encountered in low frequency servomechanisms, geophysics, biological research, and in loop analysis of negative feedback amplifiers. Among many other uses, it will serve as a very satisfactory monitor for the output of commercially available ULF signal generators most of which are not fitted with an output indicator.



MODEL
316

PRICE: \$290

FEATURES

- Pointer "flutter" is almost unnoticeable down to 0.05cps, while at 0.01cps the variation will be small compared to the sweep observed when employing the tedious technique of measuring infrasonic waves with a dc voltmeter.
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- The reading stabilizes in little more than 1 period of the wave.
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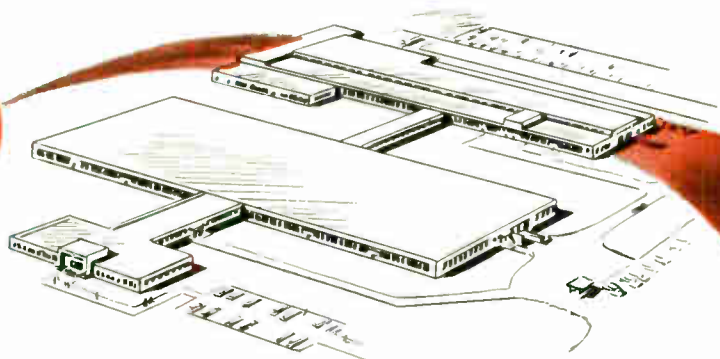
(Continued from page 66A)

The following transfers and admissions were approved to be effective as of March 1, 1955:

Transfer to Senior Member

Aldrich, D. F., 1030 Hoffman St., Elmira, N. Y.
Bawer, L. I., 16-27 Eberlin Dr., Fair Lawn, N. J.
Beier, M. G., 440 Fairfield Ave., Elmhurst, Ill.
Benson, M. C., Box 1316, Shreveport, La.
Bidlack, C. S., 14 Gregory Hall, Urbana, Ill.
Boltz, H. A., 18069 Outer Dr., Dearborn, Mich.
Brachman, M. K., 2237 Republic National Bank Bldg., Dallas 1, Tex.
Brent, L. L., 1330 N. Newstead Ave., St. Louis 13, Mo.
Brooks, J. F., Box 1042, U. S. Naval Station, Key West, Fla.
Carman, W. H., 2701 Parsifal, N.E., Albuquerque, N. Mex.
Chapin, E. W., 6 Fairfield Dr., Catonsville 28, Md.
Cleveland, W. C., Jr., 4644 Dana Dr., La Mesa, Calif.
Conrad, E., 9027 Molly Woods Ave., La Mesa, Calif.
Copeland, W. H., 11188 Ophir Dr., Los Angeles 24, Calif.
Cotsworth, A., III, 809 Linden Ave., Oak Park, Ill.
Criss, G. R., Box 208, R.F.D. 2, Lansdale, Pa.
Deerhake, W. J., Watson Laboratory, 612 W. 115 St., New York 27, N. Y.
Githens, T. A., Zenith Radio Corp., 6001 W. Dickens Ave., Chicago 39, Ill.
Golden, N. J., 161 Topsfield Rd., Ipswich, Mass.
Gorbunoff, A., 219 N. Vale Ave., Villa Park, Ill.
Graber, R. E., Box 783, San Fernando, Calif.
Graham, R. W., Graham Electronic Products Co., 6502 E. Cooper St., Tucson, Ariz.
Gray, R. O., 2305 S. 16 Ave., Broadview, Ill.
Green, J. S., 6422 Wynkoop St., Los Angeles 45, Calif.
Guerrero, E. S., 115 Witmer St., Los Angeles 26, Calif.
Hayt, W. H., Jr., Electrical Engineering School, Purdue University, W. Lafayette, Ind.
Kups, E. F., 552 S. 19 St., Newark 3, N. J.
Langford, R. C., 920 Adams Ave., Elizabeth, N. J.
Larkin, K. T., Box 359, Wayland, Mass.
Larsen, F. J., Honeywell Research Center, 500 Washington Ave., S., Hopkins, Minn.
Lindeman, B., 419 Randolph St., Huntsville, Ala.
Loth, P. A., 11 Wishing La., Hicksville, L. I., N. Y.
Loughlin, R. G., 31 Laurel Dr., New Hyde Park, L. I., N. Y.
Moynahan, G. F., Jr., Box 582, Fort Huachuca, Ariz.
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O'Donnell, R. J., 412 Westgate Rd., Baltimore 29, Md.
Peterson, A. W., 150 Elmwood Ct., Emporium, Pa.
Raybin, M. W., 1373 E. Kingsley Ave., Pomona, Calif.
Reedy, P. H., 1742 Grevalia St., Apt. E, S. Pasadena, Calif.
Reiche, H., 235 Cooper St., Ottawa, Ont., Canada
Riepka, H. C., Kaiserstr. 23, Porz/Rhein (22c), Germany
Robinson, L. P., 1438 Loma Vista, Pasadena, Calif.
Rohde, L., 7 Tassiloplatz, Muenchen 9, Germany
Ruze, J., 231 Beacon St., Boston, Mass.
Salz, N. P., Sylvania Electric Products, Inc., Bay-side, L. I., N. Y.
Scheneman, E. E., 412 Oak Ct., Baltimore 28, Md.
Scheraga, M. G., 29 Westview Rd., Verona, N. J.
Snelling, E. A., 5930 S. Kingsington Ave., La-Grange, Ill.
Steen, W. J., 5442 N. Lamont Ave., Chicago 30, Ill.
Stewart, C., 2711 Wilson Ave., S.W., Cedar Rapids, Iowa

(Continued on page 71A)



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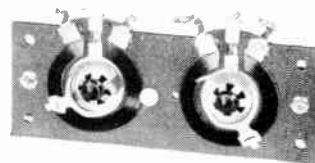
In addition, because of the radically simplified design, Mallory is able to offer multiple units at materially lower cost than that of corresponding numbers of conventional single controls.

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Rear view shows simple, rugged design, with resistance wafer attached directly to phenolic panel.

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(Continued from page 68A)

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 Vincent, H. F., 305 Edgewood Dr., Huntsville, Ala.
 Wade, E., 5337 E. Falls View Dr., San Diego 15, Calif.
 Wilkinson, W. C., RCA Laboratories, Princeton, N. J.
 Winkler, E. H., 1004 Elm St., Ronie, N. Y.

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 Brough, J. R., Borton Landing Rd., Moorestown, N. J.
 Clementson, G. C., 316 Dawnview Ave., Dayton 3, Ohio
 Eaves, H. H., 1575 E. Valley Rd., Santa Barbara, Calif.
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 Feldmann, F., 17 Hillside Ave., Roslyn Heights, L. I., N. Y.
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 Hammack, C. M., 1651 Tulane Dr., Mountain View, Calif.
 Jenssen, M., Norges tekniske hogskole, Trondheim, Norway
 Johnson, J. K., 1968 Park Pl., Lancaster, Pa.
 Jones, H. J., 6000 Lemmon Ave., Dallas 9, Tex.
 Kott, W. O., 235 E. Bruce Ave., Dayton 5, Ohio
 Lee, Y. W., Electrical Engineering Department, Massachusetts Institute of Technology, Cambridge, Mass.
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 McDavid, J. A., Headquarters, Northeast Air Command, (Dir./Comm), APO 862, New York, N. Y.
 Morgan, S. P., Bell Telephone Laboratories, Murray Hill, N. J.
 Pelham, K. F., 34 Bobrich Dr., Rochester 10, N. Y.
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 Wood, M. R., 108 E. 38 St., New York 16, N. Y.
 Young, C. M., General Electric Co., 1 River Rd., Bldg. 269, Schenectady 5, N. Y.

Transfer to Member

Allison, D. B., 727 E. Third St., Hastings, Nebr.
 Barringer, E. E., 4612 E. Eastland, Tucson, Ariz.
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(Continued on page 72A)

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A versatile, wide-range instrument for measuring resistance, capacitance, inductance, dissipation factor and Q. Full-scale ranges (in decade steps) are 1 ohm to 10 meg for resistance, 100 mmf to 100 mf for capacitance, 1 mh to 100 h for inductance. Accuracy is ± 2% up to extreme high or low ranges. Interpolation by two large concentric dials gives readings as low as 1/1000 full scale.

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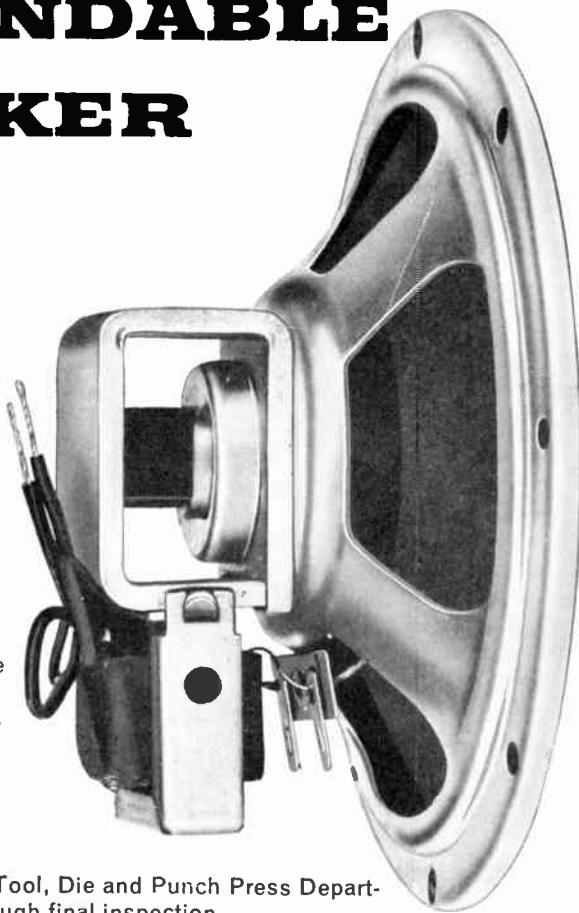
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(Continued from page 71A)

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Bussgang, J. J., 185 Hancock St., Cambridge 39, Mass.
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Christiansen, D., 12 Hay St., Newbury, Mass.
Coble, R. B., 645-34 St., S.E., Cedar Rapids, Iowa
Colletti, N., 783 S. Hawkins Ave., Akron 20, Ohio
Connally, R. E., 515 Cottonwood Dr., Richland, Wash.
Cottle, D. W., Doyle Rd., R.F.D. 3, Baldwinsville, N. Y.
Craine, W. P., 3758 N. Pacific Ave., Chicago 34, Ill.
Crane, R. L., 528 Ann La., Levittown, L. I., N. Y.
Dale, P. R., RCA Service Co., Government Service Division, Gloucester, N. J.
De Janovich, C. R., 5034 Farwell Ave., Skokie, Ill.
Dickinson, I. E., McClatchy Broadcasting Co., Seventh and Eye Sts., Sacramento, Calif.
Dresser, S. R., Jr., 104 Elmwood Ct., Emporium, Pa.
Ellis, P. V., 6334 N. Missouri Ave., Portland 11, Ore.
Ertman, R. J., 127 Galveston St., S.W., Washington 24, D. C.
Espanlaub, W. C., 56 Hillside La., E., Syosset, L. I., N. Y.
Falkenbach, G. J., Lee Laboratories, Inc., Genesee, Pa.
Flashner, G., 24 East St., Beverly, Mass.
Friedman, D., "Caprice" Long Hill Rd., Oakland, N. J.
Glomb, W. L., 39 Surrey La., Clifton, N. J.
Hamann, K. R., Cleveland Recording Co., 1515 Euclid Ave., Cleveland 15, Ohio
Hevesh, A. H., 79-50 Langdale St., New Hyde Park, L. I., N. Y.
Hiebert, R. D., 1792-37 St., Los Alamos, N. Mex.
Horowitz, J., 2682 W. Second St., Brooklyn 23, N. Y.
Howard, D. D., 4230 Oak La., S.E., Washington 22, D. C.
Johnson, L. B., Range Instrumentation Dept., NAMTC, Point Mugu, Calif.
Johnson, W. F., 1062 Portola Dr., Monterey, Calif.
Larsen, R. L., 5943 N. Warnock St., Philadelphia 41, Pa.
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Mitoma, E. Y., 6658-13 St., N.W., Washington 12, D. C.
Montgomery, D. N., 4219 Lynd Ave., Arcadia, Calif.
Moyer, J. N., 3235 Norwalk Dr., Dallas 20, Tex.
Norwood, C. A., 830 Gales Ave., Winston Salem, N. C.
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Ossman, E. A., 3 Juniper St., Rochester 10, N. Y.
Piskor, J., 7000 Capitol Ave., Hartford 6, Conn.
Puckett, T. H., 410 N.E. 14 St., Oklahoma City, Okla.
Quist, W. G., 103 Williston, Wheaton, Ill.
Rabin, R., 105-34-65 Ave., Forest Hills, L. I., N. Y.
Reiling, G. E., 2226 Winston Ave., Louisville 5, Ky.

(Continued on page 75A)

THE PILOT BAILED OUT... BUT *Bendix-Pacific* TELEMETERING "STAYED WITH THE SHIP"

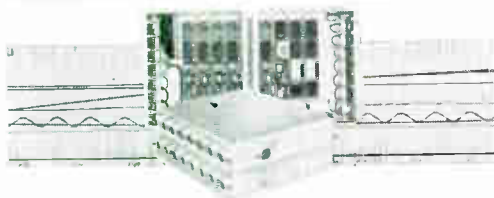


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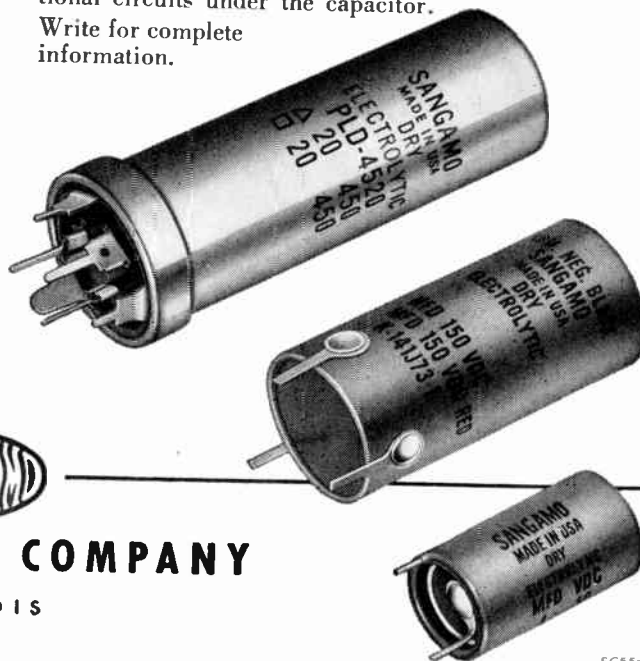
PLUG-IN TUBULAR PAPER CAPACITORS

These plug-in paper tubulars incorporate all the internal design features of the famous Sangamo Telechief. They come in a molded bakelite case with a moisture resistant end fill, and leads cut and properly spaced to fit. They are available in a range of popular sizes for almost any application.

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Write for complete information.



SANGAMO ELECTRIC COMPANY

MARION, ILLINOIS



(Continued from page 72A)

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 Simmons, D. J., 4305 Fairfax, Fort Worth, Tex.
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 Stone, R. M., 1816 Puritan Dr., Irving, Tex.
 Sutton, A. T., 3913 Cross Creek Rd., Nashville 12, Tenn.
 Tannenbaum, D. A., 1025 Collings Ave., W. Collingswood, N. J.
 Thayer, J. W., 4553 W. 150 St., Cleveland 11, Ohio
 Vehslage, E. P., 14 Stoner Ave., Apt. 33K, Great Neck, L. I., N. Y.
 Vick Roy, J. W., 223 Pembroke Ave., Morristown, N. J.
 Walance, C. G., 200 Via Colorin, Palos Verdes Estates, Calif.
 Ward, D. L., 1914 Beach St., Winston-Salem, N. C.
 Williams, R. L., 2705 University Dr., Durham, N. C.
 Willsey, R. H., 1877 Chaucer Dr., Cincinnati 37, Ohio
 Yarosh, N. P., Goodyear Aircraft Co., Akron, Ohio

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 Brewster, P. J., 92 E. Logan St., Philadelphia 44, Pa.
 Brown, D. J., Hq. NEAC D Comm Opns, APO 862, c/o Postmaster, New York, N. Y.
 Brown, E. R., Jr., Philco Corp., Tioga and "C" Sts., Philadelphia, Pa.
 Buckner, G. O., Jr., 1628 Old Spanish Trail, Houston 25, Tex.
 Bugeja, A. A., 6, Leighton Pl., London N.W. 5, England
 Burgess, E. G., Jr., 83 47 110 St., Kew Gardens, L. I., N. Y.
 Burns, W. E., Rombout Ridge, Poughkeepsie, N. Y.
 Busch, C. W., 19 Holbeinstrasse, Bremen, American Zone, Germany
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(Continued on page 76A)

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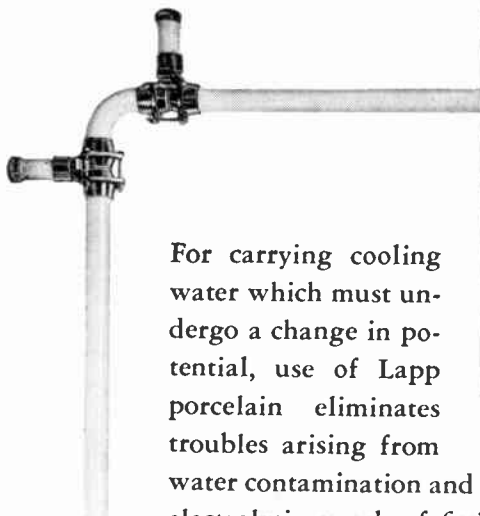
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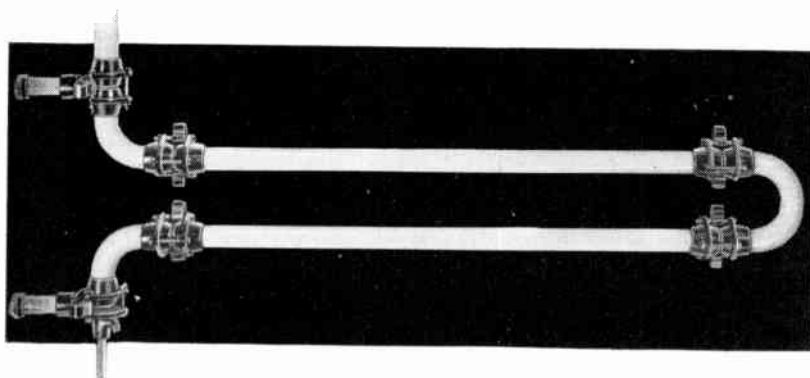
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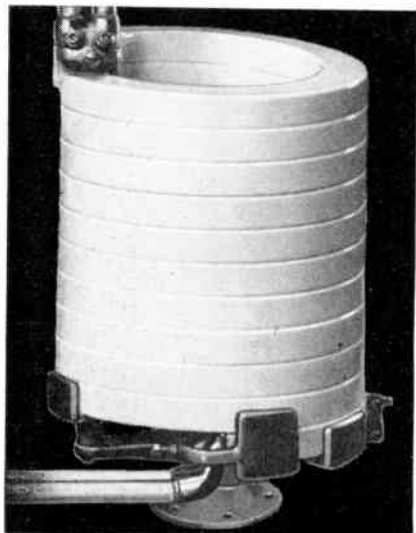


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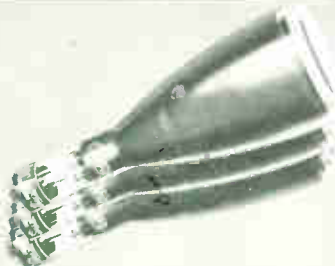
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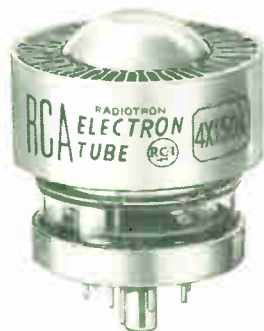
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CONTENTS

John F. Byrne, Director, 1955.....	402
Index to Abstracts and References.....	<i>The Managing Editor</i> 403
5204. A Survey of Magnetic Amplifiers.....	<i>Carroll W. Lucy</i> 404
5205. The "M"-Type Carcinotron Tube.....	<i>R. R. Warnecke, P. Guénard, O. Dochler, and B. Epsstein</i> 413
5206. Power Flow in Electron Beam Devices.....	<i>W. H. Louisell and J. R. Pierce</i> 425
5207. The Effective Surface Recombination of a Germanium Surface with a Floating Barrier.....	<i>A. R. Moore and W. M. Webster</i> 427
5208. A Chart for Analyzing Transmission-Line Filters from Input Im- pedance Characteristics.....	<i>Harvel N. Dawirs</i> 436
5209. Concerning the Noise Figure of a Backward-Wave Amplifier.....	<i>T. E. Everhart</i> 444
5210. The Nature of the Uncorrelated Component of Induced Grid Noise.....	<i>T. E. Talpey and A. B. Macnee</i> 449
5211. On the Possibility of Amplification in Space-Charge-Potential- Depressed Electron Streams.....	<i>Walter R. Beam</i> 454
5212. PLC Lattice Transfer Functions.....	<i>A. D. Fialkow and Irving Gerst</i> 462
5213. Modes and Operating Voltages of Interdigital Magnetrons.....	<i>Amarjit Singh</i> 470
5214. Measurement of Minority Carrier Lifetime and Surface Effects in Junction Devices.....	<i>S. R. Lederhandler and L. J. Giacoletto</i> 477
Correspondence:	
5215. Understanding the Gyrator.....	<i>L. M. Vallese</i> 483
5216. Effect of Heisenberg's Principle on Channel Capacity.....	<i>R. J. Solomonoff</i> 484
5217. On Entropy Equivalence in the Time- and Frequency-Domains.....	<i>Robert Price</i> 484
5218. Beam-Hugging Plates for Unlimited Cathode Ray Deflection.....	<i>H. E. Kallmann</i> 485
5219. Single-Sideband Transmission without Transient Distortion.....	<i>H. E. Kallmann</i> 485
5220. Quasi-Fraunhofer Gain of Parabolic Antennas.....	<i>R. F. H. Yang</i> 486
5221. On Fourier Transforms in the Theory of Cathode-Ray Tubes.....	<i>E. Folke Bolinder</i> 487
5222. Intrinsic Barrier Transistor.....	<i>W. C. Hittinger, J. W. Peterson, and D. E. Thomas</i> 487
5223. Checking Codes for Digital Computers.....	<i>Joseph M. Diamond</i> 487
5224. Valve Noise Produced by Electrode Movement.....	<i>John J. Glauber</i> 488
5225. Rebuttal.....	<i>P. A. Handley and P. Welch</i> 488
5226. Russian Vacuum-Tube Terminology.....	<i>I. G. Maloff</i> 488
5227. Continuous Radar Echoes from Meteor Ionization Trails.....	<i>V. R. Eshleman, P. B. Gallagher and A. M. Peterson</i> 489
5228. A Mathematical Technique for the Analysis of Linear Systems.....	<i>Rubin Boxer</i> 489
5229. High-Voltage Silicon Diodes.....	<i>L. G. Rubin and W. D. Straub</i> 490
Contributors.....	491

IRE News and Radio Notes:

IRE Awards, 1955.....	493
New Fellows.....	494
IRE Southwestern Conference.....	504
Professional Group News and Technical Committee Notes.....	506
5230-5335. Books.....	507
National Conference on Aeronautical Electronics.....	509
1955 IRE Convention Record.....	511
National Telemetering Conference.....	512
5336. Abstracts of Transactions.....	513
5337. Abstracts and References.....	516
Annual Index to 1954 Convention Record.....	Follows Page 530
Annual Index to 1954 IRE Transactions.....	Follows Convention Record Index
Meetings with Exhibits.....	6A
Industrial Engineering Notes.....	8A
News and New Products.....	34A
Section Meetings.....	92A
Professional Group Meetings.....	101A
IRE People.....	108A
Positions Wanted.....	122A
Positions Open.....	137A

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John F. Byrne

DIRECTOR, 1955

John F. Byrne was born on October 26, 1905, in Cincinnati, Ohio. He attended Ohio State University, receiving the B.S. degree in Engineering Physics in 1927 and the M.S. degree in Electrical Engineering in 1928.

After a year with the Bell Telephone Laboratories, Mr. Byrne returned to Ohio State as a faculty member; he was Assistant Professor of Electrical Engineering when he left the university in 1937 to join the Collins Radio Company. In 1942, he became associated with the newly formed Radio Research Laboratory at Harvard University, the first laboratory organization to devote its time exclusively to the development of electronic countermeasures equipment and techniques. He was appointed Associate Director of the laboratory in January, 1945. From 1946 to 1950 Mr. Byrne was Vice-President in charge of research and engineering at the Airborne Instruments Laboratory in Mineola, New York. With Motorola since 1950, he was first Director of Engineering for the

Communications Division, and is now General Manager of the Riverside Research Laboratory at Riverside, California.

Mr. Byrne has served on several government committees; he was Chairman of the Electronic Countermeasures Panel of the Research and Development Board, 1949-1951, and is currently a member of the Advisory Council for the Army Electronic Proving Ground at Fort Huachuca.

For his work during World War II, Mr. Byrne received the U. S. Navy Certificate of Commendation and the Presidential Certificate of Merit. He is a member of Tau Beta Pi and Eta Kappa Nu.

Mr. Byrne became a Senior Member of the IRE in 1945, and received the Fellow Award in 1950, "for his development of a system of polyphase broadcasting and for effective engineering administration in connection with countermeasures during the war." He has served on various IRE committees including Tellers, 1949, and Awards, 1953-1955.



Index to Abstracts and References



To keep himself reasonably well informed today, the radio engineer must overcome difficulties which could be characterized both as gastronomical and astronomical. He must continually digest large quantities of information about myriad technical developments in a vast and complex field if he is to sustain his professional health and not wither on the vine.

His chief source of nourishment is the technical literature. But here his troubles multiply. There are in existence today at least 1,000 publications in which technical papers related to radio engineering might appear. Hence, he has even greater difficulty in finding the particular nourishment his diet requires than he has in assimilating it.

One of the few outstanding and comprehensive guides to the technical literature is Abstracts and References, which has been reprinted monthly in the PROCEEDINGS since June, 1946 from *Wireless Engineer* in England. This material is compiled from over 200 leading journals by the Department of Scientific and Industrial Research in London for *Wireless Engineer*. Its appearance in PROCEEDINGS has provided readers with an extremely valuable digest of a major portion of the significant contributions to the technical literature.

As valuable as this service has been, its usefulness has been only transitory. The abstracts can be read to great advantage as each issue appears, but once read, their usefulness ceases. There is no ready way of referring back to them at a later date and finding specific information. Thus, a glittering treasure of 30,000 abstracts now lies buried on the bookshelf, beyond the reach of the average reader.

In order that the 3,700 abstracts published last year may be of permanent reference value in the future, an annual index has been reprinted from the March, 1955 issue of *Wireless Engineer* and is published as Part II of this issue of PROCEEDINGS. The index, which is separately bound, has been mailed together with the regular issue (Part I) to all IRE members and subscribers. Since the abstracts are reprinted in PROCEEDINGS one month after they appear in *Wireless Engineer*, the index covers those abstracts which were published in the February, 1954 through January, 1955 issues of PROCEEDINGS.

We are grateful to W. T. Cocking, Editor of *Wireless Engineer*, for his co-operation in providing the material for the index. It is felt that the wide distribution thus afforded the index will add very substantially to the value of an already outstanding service.

—The Managing Editor

A Survey of Magnetic Amplifiers*

CARROLL W. LUFKY†

The following paper is one of a planned series of invited papers, in which men of recognized standing will review recent developments in, and the present status of, various fields in which noteworthy progress has been made.

—The Editor

Summary—This paper was written to present the subject of magnetic amplifiers to those scientists and engineers who have not had an opportunity to observe the progress which has taken place in this field. No detailed technical discussions have been attempted and many aspects of magnetic amplifier operation and applications are only briefly mentioned.

The basic operation, along with certain fundamental circuits which represent present and potential applications, are discussed with the view in mind of indicating to the reader the range and usefulness of magnetic amplifiers.

HISTORICAL DEVELOPMENT

THE FIRST practical application of a magnetic amplifier in which actual power amplification was achieved was reported in a paper presented before the Institute of Radio Engineers in 1916.¹ This paper, by Dr. Alexanderson, described the use of such a device to amplify the current from a carbon microphone to control the output of a high frequency alternator for radio telephone transmission. As a result of Alexanderson's developments magnetic amplifier controlled alternators were incorporated in many low frequency transmitting stations constructed during World War I. Many of these installations are in operation in various parts of the world today and are still a major factor in present long-range radio communication.

By the close of World War I the vacuum tube amplifier had established itself as a powerful tool and the magnetic amplifier was pushed into the background. For many years the vacuum tube reigned supreme. Between World War I and the close of World War II, despite a few publications on magnetic amplifiers and issuance of several patents on magnetic amplifier circuitry, very little was done in this country by way of its commercial utilization. Developments in this field were carried forward elsewhere, however, most notably in Germany, with the result that by the end of World War II magnetic amplifiers of good quality were being used extensively in their military equipment. The appearance of such units as servo controllers in Luftwaffe planes, voltage regulators in the V-1 "buzz bombs," and

in the stabilization equipment of German naval fire control systems, spurred further development in this field in the post-World War II years. The result is that today the magnetic amplifier has emerged as a device of considerable importance in both military and industrial control systems, and shows promise of taking an ever increasing position of importance in the developments and designs of the future.

The small interest in the application of magnetic amplifiers in this country was not due to a lack of suitable circuitry for, indeed, the patent literature contains a wealth of information on circuits and applications thereof, which dates back to the early 1920's. Rather, the almost complete absence of suitable core materials in commercial quantities, plus the lack of a suitable dry-disc-type rectifier, made the performance of magnetic amplifiers constructed from the available components inadequate for most purposes. The use of superior core materials and the development of the selenium dry rectifier largely account for the present successful utilization of principles and circuits which have been known for years.

BASIC PRINCIPLES OF OPERATION

Fundamentally the magnetic amplifier is a device which utilizes the change in inductive impedance of a winding placed upon a magnetic core when the magnetic core becomes saturated. By using a core material having a highly rectangular B-H loop characteristic, this change in inductive impedance can be made to be quite large and very abrupt. In this manner it is possible, through proper procedures, to make such a reactor—when placed in series with an ac power source and a load—act as a switch between the two. The result is a controller which releases power to the load in a manner analogous to the well-known thyatron-type controller.

The exact method of effecting the control of the reactor flux level, which in turn will determine the time at which saturation occurs, can be quite varied. Also the exact manner of inter-connections between the saturable reactor, main ac supply, and load can assume many configurations. The basic principles of operation, however, remain unchanged.

* Original manuscript received by the IRE, January 7, 1955.

† U. S. Naval Ordnance Laboratory, White Oak, Md.

¹ E. F. W. Alexanderson, "A magnetic amplifier for radio telephony," *Proc. I.R.E.*, vol. 4, pp. 101-120; April, 1916.

A simple magnetic amplifier circuit, the principles of operation of which are easily followed, is shown in Fig. 1. From inspection of this figure it is seen that N_p is the winding on the reactor which controls the flow of power from the main ac power source E_s to the load R_L . This control is effected by a second winding on the reactor N_c which is connected to a control source E_c . It is immediately seen that N_c and N_p are closely coupled by the reactor magnetic circuit; therefore any control signal on N_c must operate against the reflected impedance

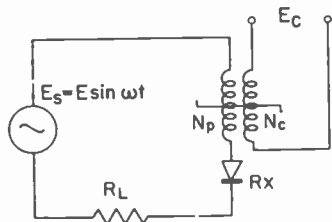


Fig. 1—A simple half-wave magnetic amplifier circuit.

from the winding N_p in effecting a desired flux change in the reactor core. This reflected impedance would normally be very low, thus requiring considerable power from E_c , if it were not for the inclusion of the rectifier element R_X in series with N_p . With R_X in the circuit there will be one half-cycle of the main ac power source during which the circuit containing N_p is open. During this half-cycle, the impedance reflected from N_p to N_c is very high, and it is possible for a signal from E_c to effect readily a change in flux in the reactor with a small expenditure of power. During the next half-cycle conduction through rectifier R_X is permitted. If the internal impedance of E_s and R_L are low, the amount of conduction permitted during this half-cycle will be determined primarily by the inductive impedance of the winding N_p . If the reactor core has a rectangular B - H loop characteristic as shown in Fig. 2, this inductive impedance will be extremely high as long as the reactor core is unsaturated, and very low when saturated. It is immediately apparent that power flowing from E_s through R_L can be controlled by fixing the time during this half-cycle when saturation of the core occurs. This may be accomplished by the control circuit, through N_c , during the preceding half-cycle or "control" period when conduction through R_X is prohibited.

If output is obtained during the conducting or "operating" half-cycle, the reactor must be driven into saturation. Then at the beginning of the next half-cycle (or following control period) the reactor flux will return to its remanence position, which is very near saturation for rectangular loop core materials. This is shown as point A in Fig. 2(a). At this time the main power circuit is again opened by rectifier R_X (Fig. 1), and power from the signal source E_c may be made to force the flux of the reactor down the loop from the remanence point. The amount of signal can be adjusted to position or "reset" the reactor flux by an amount $\Delta\phi$ to any given

point on the loop in accordance with Faraday's Law:

$$\Delta\phi = \frac{1}{N_c} \int E_c dt.$$

Thus, at the beginning of the following half-cycle (or next output half-cycle) the voltage from the supply source which will appear across N_p will cause the reactor to proceed again toward saturation from the level established or reset by the control action. If very little flux reset was accomplished during the control period [as to point B in Fig. 2(a)] the reactor very quickly saturates and most of the supply voltage appears across the load. If a large reset action has taken place [Fig. 2(b)] saturation will occur only during the latter portion of the half-cycle and very little supply voltage will appear across the load. Indeed, if sufficient reset action has occurred, the entire supply source volt-time integral may be absorbed by the reactor and no voltage will appear across the load [Fig. 2(c)]. This represents the cut-off condition and is not exactly zero because a small magnetizing current will always flow through the load in a circuit such as is being discussed.

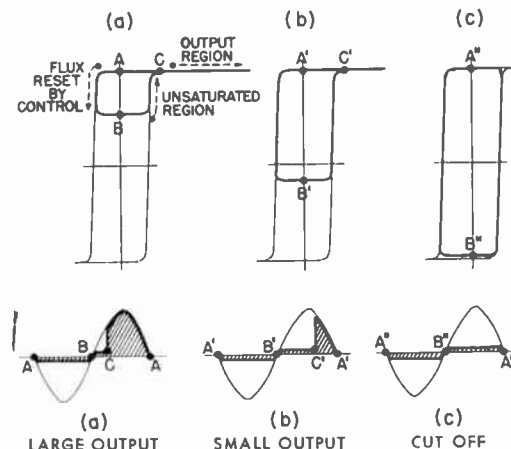


Fig. 2—Flux control and output waveform characteristics of a simple half-wave magnetic amplifier.

MAGNETIC AMPLIFIER CIRCUITS

The simple circuit of Fig. 1 is rarely used as shown. The more commonly used, practical circuits may, however, be easily built from it. The changes made are usually to circumvent certain of its inherent difficulties or shortcomings. For example, two reactors are generally used in which the N_c windings on each are in series opposition. This cancels fundamental supply frequency voltage which is induced into the control winding by transformer action. It is also evident that only half-wave output is obtained from the circuit of Fig. 1. Full-wave output may be obtained by placing two such circuits back to back. A circuit in which these two changes have been incorporated is shown in Fig. 3 on the following page. This circuit, called the full-wave "doubler" circuit,² is one of the most commonly used building

² F. G. Logan, "Electric Controlling Apparatus," U. S. Patent 2,126,790, issued August 16, 1938 (application filed June 23, 1936).

blocks in the magnetic amplifier field today.

If a phase reversing output is desired, as for example in a servo controller, two reactors in a bridge arrangement³ may be used for half-wave output as shown in Fig. 4, or four reactors⁴ if full-wave output is desired, as shown in Fig. 5. It is readily seen that the circuit of Fig. 4 is basically two circuits such as in Fig. 1 in a bridge arrangement, while Fig. 5 is two circuits such as in Fig. 3 in a bridge arrangement.

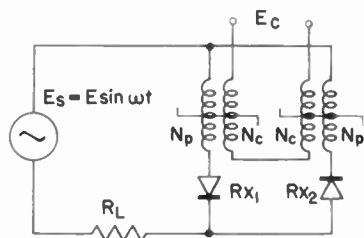


Fig. 3—Full-wave doubler magnetic amplifier circuit.

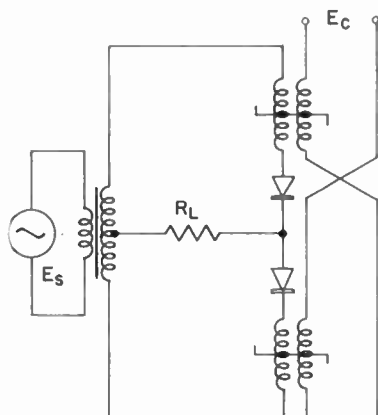


Fig. 4—Half-wave magnetic amplifier circuit with phase reversing output.

Where increased gain or power handling capacity is required these circuits may be cascaded. In a "multi-stage" amplifier the load of the first stage becomes the control circuit of the second, etc. While larger reactors and rectifiers will generally be used in each succeeding stage, the operation remains basically the same. A typical magnetic amplifier will contain two or three cascaded stages. The exact connections of both the control and power windings depend upon whether an ac or dc signal source is used and upon whether an ac or dc output is desired.

Some requirements may be met by circuits in which the rectifier element (RX in Fig. 1) is absent. In this case control is more difficult because it must be effected in the face of a much lower reflected impedance from

the N_p windings, as well as induced voltages from the main power source. Such a circuit is shown in Fig. 6. Magnetic amplifiers of this type have low gain but do have an extremely linear transfer characteristic.⁵

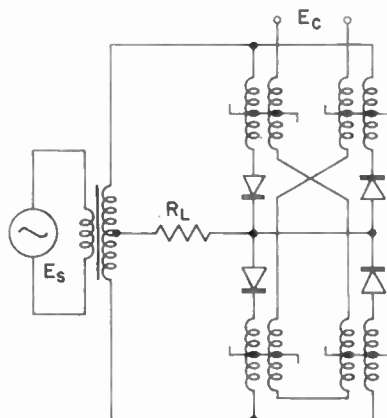


Fig. 5—Full-wave magnetic amplifier circuit with phase reversing output.

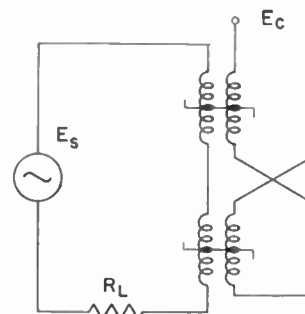


Fig. 6—Simple saturable reactor magnetic amplifier circuit.

Most magnetic amplifiers with a rectifier element in the power winding circuit will, in the absence of any control or bias, immediately proceed into complete saturation and full output. They are therefore called self-saturating amplifiers. In amplifiers without the rectifier element, since both half-cycles of the ac power source voltage appear across the reactor, saturation is brought about only as a result of control action. These are referred to as saturable-reactor amplifiers or "transducers."^{6,7} Self-saturating amplifiers are characterized by high gain, a somewhat nonlinear control characteristic, and are sensitive to control voltage polarity (or phase). Saturable-reactor amplifiers have low gain, are very linear, and are insensitive to polarity of control voltage.

It should not be assumed that the inclusion of a rectifier element in the power winding of a reactor is necessary to the basic operation or control of a high-

³ C. S. Hudson, "Improvements in or Relating to Magnetic Amplifiers," British patent 598,285, issued February 13, 1948 (application filed October 26, 1945).

⁴ W. A. Geyger, "Grundlagen der magnetischen Verstärker für die Mess- und Regeltechnik" (Fundamentals of magnetic amplifiers for measurement and control purposes), *Wissenschaftliche Veröffentlichungen aus den Siemens-Werken*, vol. 19, p. 233.

⁵ W. J. Dornhoefer and V. H. Krummenacher, "Applying magnetic amplifiers," *Elec. Mfg.*, vol. 45, p. 94; March, 1951; p. 112, April, 1951 is an example.

⁶ A. U. Lamm, "Some fundamentals of a theory of the transducer or magnetic amplifier," *Trans. AIEE*, vol. 66, pp. 1078-1085; 1947.

⁷ U. H. Krabbe, "The Transducer Amplifier," Lindhska Boktryckeriet, Örebro, Sweden; 1947.

gain magnetic amplifier. Any means whereby the reflected impedances into the control windings may be increased will result in an increase in gain characteristics. Recent circuitry advances have been made in which a combination of a pulse ac source with selective filters replacing the rectifiers has given excellent results.⁸

SPEED OF RESPONSE

From the operation of the circuit in Fig. 1 it is evident that magnetic amplifiers of this type have a limitation on their speed of response which is fundamental. The device controls power, operating on an ac or pulse-type source in such a way that each period of output must be preceded by a period which establishes, through a flux-setting control action, the amount of output to be delivered. These control and output periods are usually one half-cycle in duration, but may both be within the same half-cycle.⁹ If they are one half-cycle in duration there will be a minimum delay of one half-cycle of the source frequency per stage of amplification.^{10,11} Other factors may contribute to extend this delay over a considerably longer period but in no event can a certain inherent "dead time" be avoided. If faster response is required it is obtained usually by increasing the power source frequency. This may be done by static frequency multiplication¹² of a basic line frequency, by high frequency converters, pulse generating circuits, etc. However, the delay will be decreased only in the same ratio as the amplifier source frequency is increased.

Generally, half-wave circuits will exhibit the minimum dead time or delay of one half-cycle. Full-wave circuits (two half-wave circuits back to back), however, exhibit this minimum delay only under certain conditions. The principal difficulties arise from the fact that it becomes necessary to couple at least two reactors together, either by the control windings, bias windings, or output windings, or combinations thereof, which are in different half-cycles of their basic operation. For example, when one reactor is being controlled or reset the other is being driven into saturation or is delivering output. Voltages induced into windings on the reactor that is in its operating half-cycle will be coupled into the other reactor that is in its control half-cycle. This "feedback" from one reactor to the other is usually in such a direction as to act as an additional aiding control, and hence is positive feedback. This positive feedback action has the advantage of increasing the gain of the circuit

but, since it may require several cycles of operation to stabilize, will add to its over-all response time. The full-wave connections also offer many possibilities for circulating currents to flow. These currents will have definite control actions and may have long L/R time constants associated with them. The circuit parameters, core material and rectifier quality determine to a considerable extent the gain versus time delay characteristics of such circuits. It is therefore convenient to express an amplifier's quality in terms of the ratio of its gain to its speed of response. This ratio is called the "figure of merit" of the amplifier,¹³ and with present-day core materials and rectifiers can be as great as 5,000 or more.

For half-wave circuits the term "figure of merit" has a limited meaning since the gain and the inherent fixed delay discussed above are not related. The maximum gain of a half-wave circuit, in view of the absence of any positive feedback effects, will be much lower than the gains achievable with a full-wave circuit using identical components. By cancelling the inherent positive feedback effects of the normal full-wave circuit,¹⁴ its response time may be dropped to the same minimum value, but its gain will generally also be dropped to the same order of magnitude as that of the half-wave circuit. The decision as to whether a half-wave or full-wave circuit is desirable will depend largely on the specific application and should be determined only after careful consideration by qualified engineers.

COMBINATION CIRCUITS

Combinations of vacuum tubes or transistors and magnetic amplifiers are frequently used to accomplish results which would not be possible with magnetic amplifiers alone. Almost without exception these combination circuits use vacuum tube or transistor input stages driving magnetic amplifier output stages. In this way gain or high input impedance requirements may be easily met with a suitable input stage while power output requirements are met with a magnetic amplifier stage. Since most of the vacuum tube failures encountered in practice are in the power handling or output stages, a marked increase in over-all reliability may be achieved. Furthermore, the requirements for a large B+ or plate supply source is eliminated by using magnetic amplifier power stages which operate directly from the main ac power source. In many instances this will result in over-all decrease in amplifier size and weight. Fig. 7 (next page) shows typical combination circuit where use of push-pull vacuum tube driver stage controls saturable reactor-type magnetic amplifier output stage. Transistor input stages may be used to advantage in place of vacuum tube stages. A typical transistor-magnetic amplifier

⁸ R. E. Morgan and J. B. McFerran, "Pulse Relaxation Amplifier—A Low Level D-C Magnetic Amplifier," AIEE Technical Paper 54-198 presented at the AIEE Northeastern District Meeting, Schenectady, N.Y., May 5-7, 1954.

⁹ F. Hill and J. A. Fingerett, "Fast-response magnetic servo amplifier," *Electronics*, vol. 27, pp. 170-173; October, 1954.

¹⁰ R. A. Ramey, "On the mechanics of magnetic amplifier operation," *Trans. AIEE*, vol. 70, part II, pp. 1214-1223; 1951.

¹¹ C. W. Lufcy, A. E. Schmid, and P. W. Barnhart, "An improved magnetic servo amplifier," *Trans. AIEE*, vol. 71, part I, pp. 281-289; 1952.

¹² J. J. Suozzi and E. T. Hooper, "An all magnetic audio amplifier system," *Trans. AIEE*, Paper 55-70, for presentation at the Winter General Meeting, New York City, January 31-February 5, 1955 is an example.

¹³ J. T. Carleton and W. F. Horton, "The figure of merit of magnetic amplifiers," *Trans. AIEE*, vol. 71, part I, pp. 239-245; 1952.

¹⁴ W. A. Geyger, "Magnetic amplifiers of the self-balancing potentiometer type," *Trans. AIEE*, vol. 71, part I, pp. 383-395; 1952. also D. G. Scorgie, "Fast response with magnetic amplifiers," *Trans. AIEE*, vol. 72, part I, pp. 741-749; 1953.

combination appears in Fig. 8 below, where transistor input drives a half-wave bridge phase-reversing, self-saturating magnetic amplifier stage.¹⁵

Both the vacuum tube and transistor stages require a source of dc power to obtain best performance. They are usually less efficient over-all and at present are less reliable than a magnetic amplifier. These factors, nevertheless, are not serious for low-level stages. Such combinations therefore represent an ever increasing field of development and application.

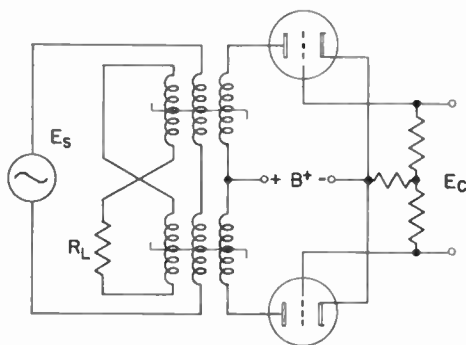


Fig. 7—A combination vacuum tube-saturable reactor magnetic amplifier circuit with the output isolated from the main power source.

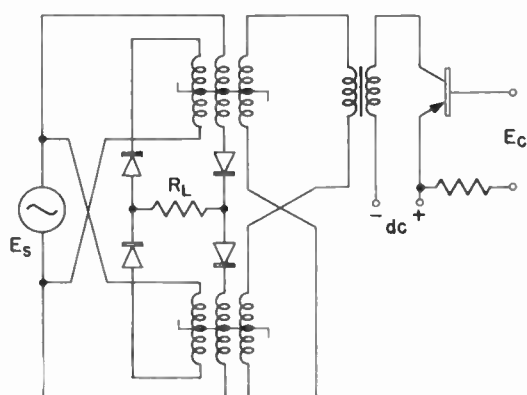


Fig. 8—A combination transistor-magnetic amplifier circuit.

The recent improvement in transistor quality and power-handling capacity have pointed the way to improved designs which in many instances appear to be better than either an all magnetic amplifier or combination transistor-magnetic amplifier system. In many of the lower power applications this may prove to be the case. For control of larger power, however, the magnetic amplifier is unmatched. Where efficiencies are obtained of the order of 50 per cent to 60 per cent with vacuum tube stages, and 60 per cent to 70 per cent with transistor stages, the magnetic amplifier will give 80 per cent to 95 per cent. Neither the transistor nor the vacuum tube at present enjoy the reputation for ruggedness, long life, and low maintenance that the magnetic amplifier has.

THEORY

Circuit-wise the magnetic amplifier is a relatively simple device. This simplicity is very misleading, however, when a formal mathematical analysis of its operation is attempted. Because of the extremely nonlinear characteristics of the core materials and dry rectifiers, linear circuit theory may be applied only to carefully selected periods of its operation. For example, the control period will require one set of assumed conditions, the saturating or output period another. Effects of interwinding and rectifier capacitance, rectifier forward and reverse impedance, induced voltage and current transients, their time constants, etc. need to be carefully considered. If this is rigorously done, a complexity of terms and equations results that is extremely difficult to handle. To simplify the analysis it is common practice to make the assumptions of perfect core material and ideal rectifiers. These assumptions unfortunately do not always give resultant mathematical expressions of sufficient accuracy to predict the performance of actual practical circuits. No satisfactory general analysis of magnetic amplifier operation has yet been done although many of an approximate and specialized nature have been published.¹⁶

It is as yet difficult to design magnetic amplifiers on paper as one might design a vacuum tube amplifier. Thus most magnetic amplifier design today requires considerable past experience and engineering skill. It is reasonable to expect this situation to improve rapidly as more and more effort is put into this field. A survey of published papers on magnetic amplifiers indicates a very healthy increase each year.¹⁷ In the colleges and universities little attention has been given the magnetic amplifier, per se, until very recently. At the present time a few institutions have courses dealing specifically with such circuitry.

COMPONENTS

Core Materials

In order to obtain well-defined control characteristics from a magnetic amplifier it is necessary to change the saturable reactor's inductive impedance abruptly from an extremely high value to a very low value. In this way a true switching action may be closely approached. This is achieved only by careful selection and application of suitable core materials. The more rectangular the core B - H loop the better will be the switching operation obtained. It is also desirable that the core material have as low a coercive force as possible since the control action must overcome at least the coercive force (H_c) of the core before any control of flux level can be exercised. Thus the lower the H_c the less control power required. Taken together these two requirements dictate a high B/H ratio or high permeability core. Fig. 9 shows the

¹⁵ J. J. Suozzi, "A Half-Wave Transistor Magnetic Amplifier," Master's thesis submitted to the School of Engineering and Architecture of the Catholic University of America, Washington, D. C., February 28, 1954.

¹⁶ J. G. Miles, "Bibliography of magnetic amplifier devices and the saturable reactor art," *Trans. AIEE*, vol. 70, Part II, pp. 2104-2123; 1951.

¹⁷ W. A. Geyger, "Magnetic Amplifier Circuits," McGraw-Hill Book Co., Inc., New York, N.Y.; pp. 6-18, 1954.

B-H loops of several core materials which, due to their rectangularity, are suitable for magnetic amplifier service. From inspection of Fig. 9, it would appear that 4-79 molypermalloy, which has very high permeability, would be the best. From the standpoint of power output, however, a high saturation flux density is also desired because for a given core volume the total volt-time integral which the core can absorb or given power source voltage it can hold back is directly proportional to the saturation flux density. Thus it is seen that the grain-oriented 50 per cent Ni-50 per cent Fe Orthonol is an excellent material having all the desirable features. For this reason a major percentage of all high-performance magnetic amplifiers built today use a core material of this type. Where very high gain is necessary a permalloy core would, of course, be used. If gain is secondary to the power to be controlled, a 4750 alloy or grain-orientated silicon iron core could be used.

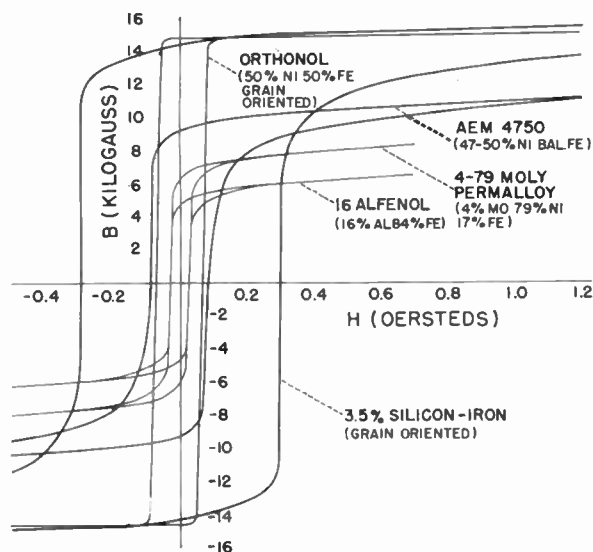


Fig. 9—*B-H* loops of typical core materials for magnetic amplifier service.

All high permeability magnetic materials are extremely strain-sensitive; therefore cores for magnetic amplifier use are usually encased in a protective box that isolates the core from any strains which might be introduced in winding and handling.

Core Types

Because the toroidal configuration has advantages over the stacked *E-I* or double-backed *U*-type core from a magnetic circuit standpoint, the toroid is to be preferred where high performance is paramount. Not only does the tape-wound toroidal core take advantage of the grain orientation of materials, such as the orientated Ni-Fe and Si-Fe, but it can be easily fabricated in very thin gauges. On the other hand a stacked core must be made from laminations carefully punched with the proper orientation if the grain-orientated materials are used. Also it cannot be assembled after anneal, if the laminations are below about 0.007 inch in thickness,

without extreme precautions to avoid strain of the material. It is possible, of course, to build excellent magnetic amplifiers using a stacked laminated core, but general commercial availability of a large variety of high-grade toroidal cores (Fig. 10 below) coupled with improved toroidal winding equipment and facilities brought about a definite trend to the toroid reactor, especially in the lower power magnetic amplifier field.

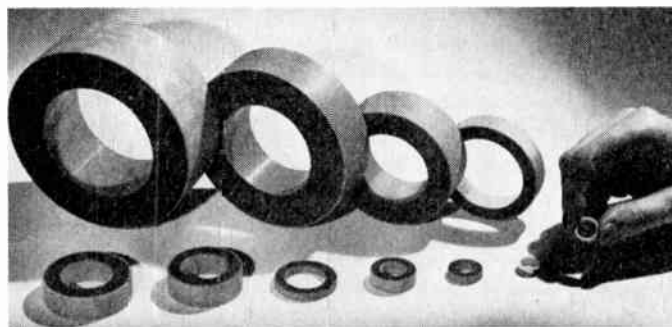


Fig. 10—High permeability toroidal cores for magnetic amplifier use. The cores are encased in protective covers to avoid loss of permeability due to strain which might occur during winding and handling. (Courtesy Magnetics, Inc.)

For very large reactors, where large amounts of power are to be handled, stacked laminations of the poorer grades of core materials such as Si-Fe are frequently used. First the cost of a large core of high quality material becomes excessive and secondly the poorer gain and response characteristics which results are usually secondary to power handling capacity in such applications.

Rectifiers

Dry-disc-type rectifiers are conventionally used in magnetic amplifiers for decoupling the power windings from the control circuits, for producing dc bias, for conversion of ac to dc in the output, etc. Their use is dictated by their long life and ruggedness, their ability to handle large currents, and their relatively high efficiency. For general application it can safely be stated that the best rectifier for magnetic amplifier usage would be one having infinite back impedance and zero forward impedance. If this statement should at first sound facetious, it is pointed out that some circuits are designed to take advantage of commercial rectifier deficiencies and direct substitution of an ideal rectifier in such a circuit would result in a definite loss of operating characteristics. However, as a result of dry rectifier back leakage or forward impedance or both, it is generally necessary to reach a compromise circuit design. The decision as to whether back leakage will be sacrificed for lower forward resistance or vice versa depends upon the circuit's intended application. It is evident, referring again to the simple circuit of Fig. 1, that back leakage in the power circuit rectifiers will appear not only as a decrease in reflected impedance to the control circuit but also as an additional control action on the reactor. Hence it is most important that this leakage either be carefully controlled or kept to an absolute minimum.

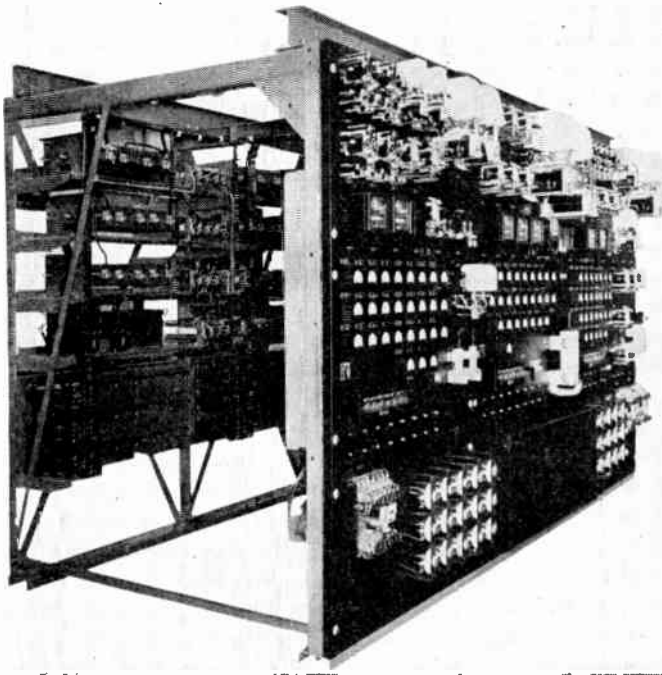


Fig. 11—Magnetic amplifier equipped control panel for a 48-inch four-high metal foil mill. A magnetic amplifier in combination with an amplidyne generator excites and controls the 800 kw main mill generator. A 4.7 KVA magnetic amplifier is used as a regulating exciter for the 1,000 hp mill motor field. The wind-up reel motors—100 hp unwind and 200 hp wind-up—employ 4.7 KVA magnetic amplifiers as field regulators. Magnetic preamplifiers regulate each of the reel motor amplidyne generators. The power section of one 4.7 KVA magnetic amplifier is shown on the lower rear rack. (Courtesy Gen. Elec. Co.)

Unfortunately dry rectifiers tend to change their characteristics with temperature and age. Thus many of the limitations of stability of the magnetic amplifier are centered in the rectifier elements. This is particularly true with respect to temperature drift. Nevertheless, by careful matching and aging of commercial quality rectifiers, their effect on the over-all amplifier can be controlled to the point where very low drift and stable systems are practical. Recent advances in the development of semi-conductor-type rectifiers offer promise of considerable improvement in this respect.

APPLICATIONS

When one considers that a few years ago the magnetic amplifier was practically unknown, its widespread usage today can truly be described as phenomenal. Few people realize to what extent this device has become a part of the control, regulation and instrumentation fields. Despite the fact that it is fundamentally a device to control the impedance in an ac circuit and is primarily limited to this function alone, its potential applications have just begun to be realized. The examples given here represent but a selected few of the uses to which magnetic amplifiers have been placed. They will, however, indicate the flexibility and range of services in which it may be used.

The magnetic amplifier is one of the most efficient and reliable methods of controlling large amounts of power which we have today. In effect a properly de-

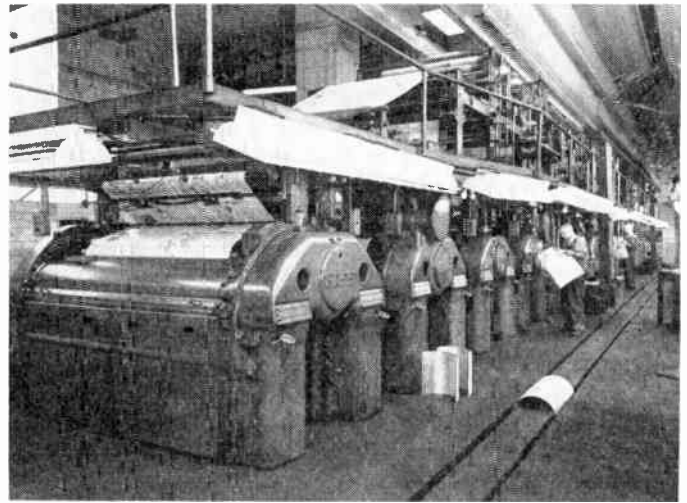


Fig. 12—A newspaper press driven by several dc motors powered from a magnetic amplifier-controlled 300 kw power rectifier. Magnetic amplifiers are also used to supply and regulate the press motor fields. (Courtesy Gen. Elec. Co.)

signed unit will behave very much like a controlled switch having no moving parts and no contacts, yet requiring relatively small signals to actuate. As a result of its proven reliability and efficiency it has become one of the standard controllers for large power installations. For example, in the electric utilities industry the requirements for continuity of service are so severe that no compromise can be made on the reliability of the main generating equipment. For this reason the regulation of these systems was one of the first big industrial applications of the magnetic amplifier. The acceptance of the magnetic amplifier was accomplished with a very short "proving in" period, and today practically all generating equipment built is equipped with magnetic amplifier voltage and frequency regulators.^{18,19} Indeed many systems use magnetic amplifiers all the way from the low-level voltage and frequency sensing stages through to the main generator field exciters, having replaced the conventionally used rotary amplifiers and dc exciter generators normally used.

The control of speed and tension in steel rolling mills, the rolling and unrolling rates, and tension of steel sheet in cleaning and pickling lines; as paper and textile mill speed regulators; as voltage and current regulators in dc supplies for arc welding; as power controllers for draw bridges, ship steering, gun turret drives, etc., are examples of but a few of the in-service applications of magnetic amplifiers in the heavy-duty control field today (see Figs. 11, 12, 13). Most of the power controlled in such heavy-duty applications is dc; hence self-saturating full-wave circuitry is commonly applied. In this way ac is converted to dc power in the same rectifiers used with the reactors to obtain high gain in the amplifier.

¹⁸ H. F. Storm, "Voltage regulator with magnetic amplifiers for large alternators," *Proc. NEC*, vol. 7, pp. 247-253; October, 1951.

¹⁹ E. L. Harder, "Power control with magnetic amplifiers," *Electronics*, vol. 25, pp. 115-117; October, 1952.

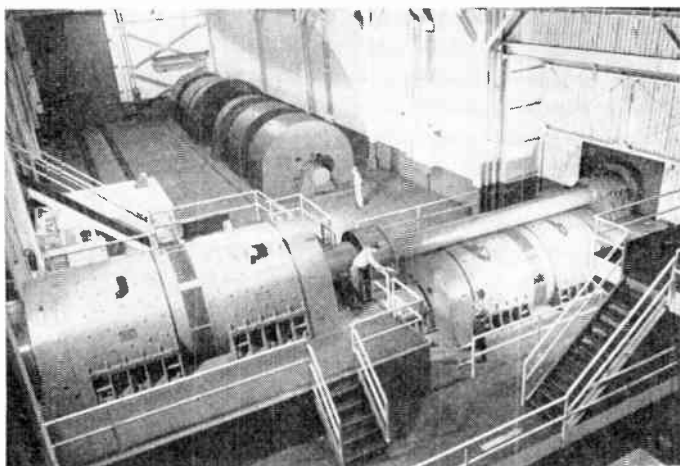


Fig. 13—A 12,000 hp blooming mill drive (with *M/G* set in background) which uses a magnetic amplifier to amplify current limit signals. The magnetic amplifier, in combination with amplidyne regulators on the motor and generator fields, permits rapid reversals of one second at base motor speed and $2\frac{1}{2}$ seconds at top motor speed. (Courtesy Gen. Elec. Co.)

The magnetic amplifier is by no means limited to the field of heavy equipment. In the regulation of low power generators excellent performance has been obtained (see Fig. 14). In the small instrument-type servo field, great strides have been made until today, within its limitations of input impedance and bandwidths (due to the inherent time delays previously discussed), dynamic performance equal to that produced with vacuum tube amplifiers is not uncommon.²⁰ For example, with a 60-cycle ac supply source bandwidths of from 6 to 10 cycles may be achieved, while with a 400-cycle supply 15- to 20-cycle bandwidths are possible.

Recent advances in applying conventional servo compensation techniques to magnetic amplifiers make possible design of systems incorporating lead, lag, lead-lag, integral and lead-integral compensation within the amplifier itself.²¹ Using these techniques in servo controllers it is now possible to replace high-performance vacuum tube controllers in many precision servos that but a few years ago were considered completely out of the range of the magnetic amplifier. The potentials of its applications in this field are indeed far-reaching. The complexity of many of our present-day control systems has become so great and continuity of service so important that failure in any component can scarcely be tolerated. Magnetic servo controllers give the reliability required. They are thus becoming standard servo controller equipment in the guided missile field, in aircraft, submarines, and on shipboard (see Figs. 15, 16, 17, 18 on the following page).

Because of certain of its inherent properties the magnetic amplifier is ideally suited for many instrumentation applications. For example, the saturable-reactor-type circuit possesses current transformer character-

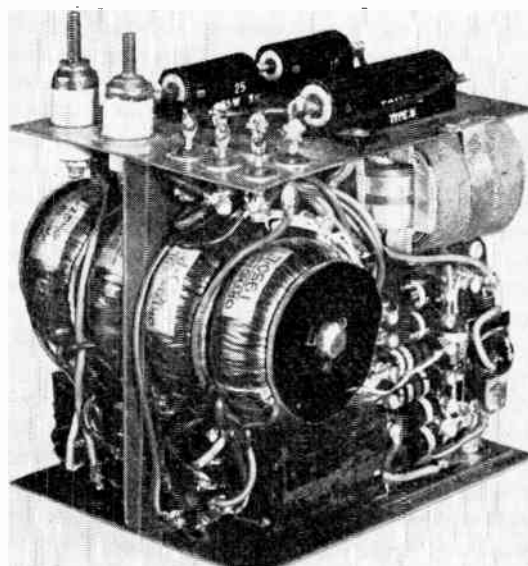


Fig. 14—A magnetic amplifier voltage and frequency regulator for a 750 volt-ampere inverter for use in a guided missile. Regulation of 2 per cent on voltage and 1 per cent on frequency is obtained over extreme conditions of temperature, input voltage and load. (Official U. S. Navy Photo.)

istics in that control dc ampere turns are reproduced as average output ampere turns. Thus by using a single-turn control winding carrying large dc currents with a multi-turn low-level ac power winding it is possible to match huge control current ampere turns with small power circuit current ampere turns in a completely isolated circuit.²² Their use, therefore, in metering large dc currents—as, for example, in big electroplating installations, electro-processing of aluminum, etc.—is evident. Input levels as low as 10^{-12} watts are capable of controlling magnetic amplifiers. Consequently for thermocouple inputs and the like they are well-suited as metering amplifiers. With compound feedback circuitry it is possible to build magnetic amplifiers which behave as voltmeters having extremely high impedance or ammeters having extremely low impedance.¹⁴ Use of magnetic amplifiers as tubeless audio amplifiers has been shown to be practical²³ although the lack of suitable high frequency power sources to obtain the necessary bandwidth is a definite drawback to their use in such service.

In many special applications magnetic amplifier circuitry is being applied to perform operations which are not primarily of amplification. Perhaps the most outstanding example is the bi-stable magnetic decision element²⁴ which has become a powerful tool in the design and construction of digital computer systems. The ease with which multiple inputs may be mixed in a single reactor makes their use in analog computers ideal.²⁵

²² W. F. Horton, "Isolation metering of d-c bus currents," *Proc. NEC*, vol. 7, pp. 260-262; October, 1951.

²³ J. J. Suozzi and E. T. Hooper, *op. cit.*

²⁴ A. Wang, "Magnetic delay-line storage," *Proc. I.R.E.*, vol. 39, pp. 401-407; April, 1951; also R. A. Ramey, "The single core magnetic amplifier as a computer element," *Trans. AIEE*, vol. 71, part I, pp. 442-446; 1952.

²⁵ B. E. Davis and I. H. Swift, "An analog computer technique using magnetic amplifiers," *Trans. AIEE*, Paper 54-389, presented at Fall General Meeting, Chicago, Ill., October 11-15, 1954.

²⁰ C. W. Lufcy, A. E. Schmid, and P. W. Barnhart, *op. cit.*, p. 110.
²¹ H. H. Woodson, C. V. Thrower, and A. E. Schmid, "Compensation of a magnetic amplifier servo system," *Proc. NEC*, vol. 8, pp. 158-165; 1952.



Fig. 15—A packaged high performance magnetic amplifier servo controller for airborne applications. This amplifier, which will control a 5-watt ac motor, is completely self-contained. (Courtesy Specialties, Inc.)

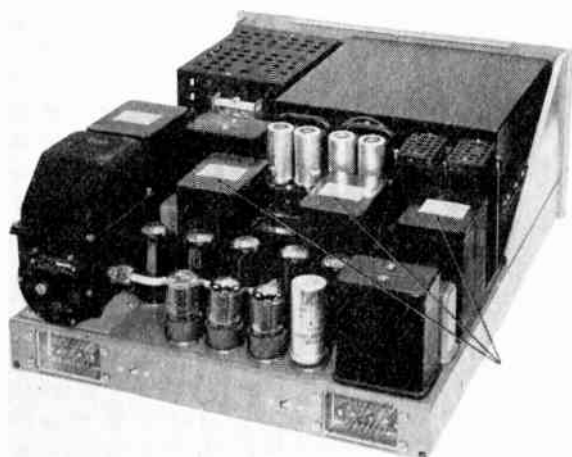


Fig. 16—A commercial auto-pilot three-channel servo controller with combination vacuum tube input-magnetic amplifier output. Arrows point out the three magnetic amplifier output stages. (Courtesy Eclipse-Pioneer Div., Bendix Aviation Corp.)

With sufficient positive feedback an ordinary magnetic amplifier can be made unstable to the point where a small additional control current will result in its going from zero output to full output or vice versa, thus performing the operations of a relay without moving parts or contacts.²⁶

CURRENT STATUS

Rapid as the growth of the magnetic amplifier field has been, it is still retarded by a lack of engineers trained in the design and utilization of such circuitry. A major portion of the potential applications is in the field of servomechanisms, which is in itself a highly specialized field. Generally the servo engineer has a background in electronics and electron tube design but little or no experience with magnetic circuitry. As a re-

²⁶ A. U. Lamm, "The Transducer, D-C Pre-Saturated Reactor, with Special Reference to Transducer Control of Rectifiers," ("Transducer Locking Relay") 2nd. ed., pp. 19-20, Esselte Aktiebolag, Stockholm, Sweden; 1948.

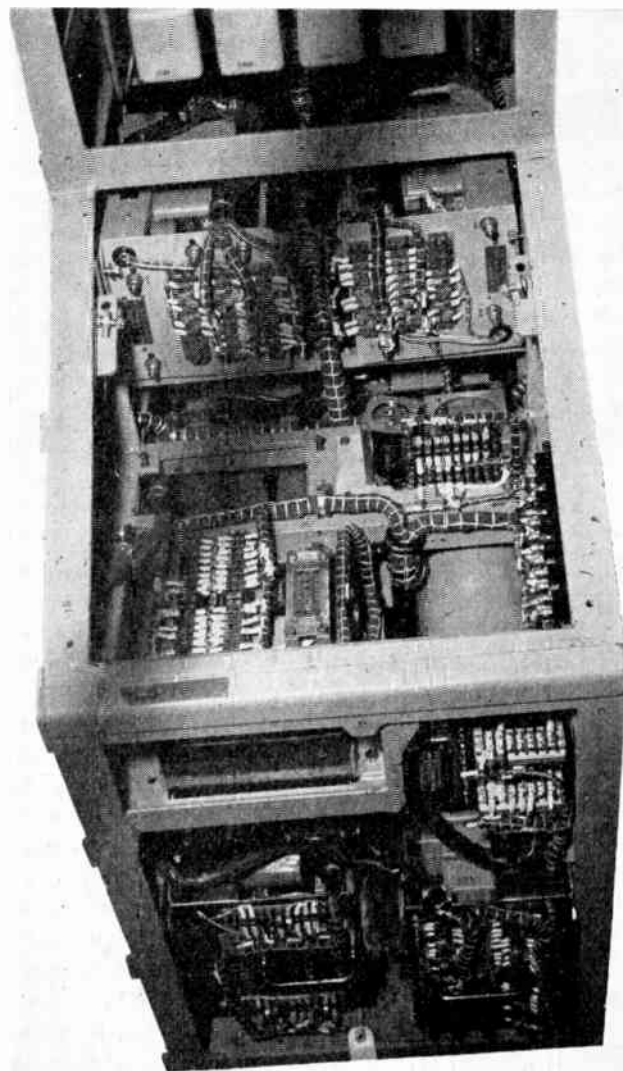


Fig. 17—A cubicle containing several magnetic servo amplifiers used in control of a submarine atomic power plant. These units were designed for long life, very high degree of reliability, and ability to withstand high shock and vibration. (Courtesy Gen. Elec. Co.)

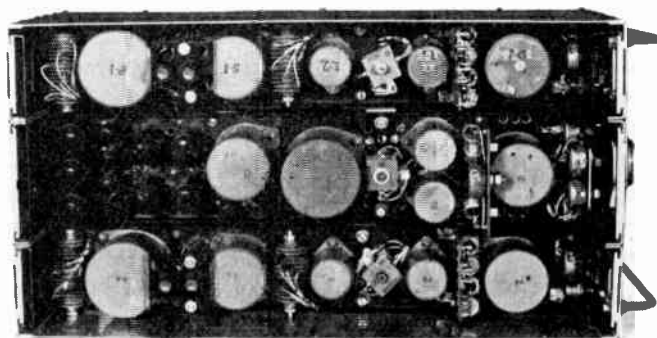


Fig. 18—A three-channel magnetic amplifier servo controller for a small search-track antenna system. One channel controls azimuth, the second controls elevation, and the third controls switching to govern the search pattern. (Courtesy Gen. Elec. Co.)

sult many applications which could be handled with magnetic amplifiers are instead solved by other methods, for, without some knowledge of the magnetic amplifier's characteristics, it is difficult to specify or balance their capabilities against known requirements. Other

practical considerations, such as available production facilities and engineering costs, also frequently exclude their use. These factors, however, will become less important as more engineering know-how and production experience are gained. The relative simplicity of the magnetic amplifier makes for easier, cheaper and faster assembly once a system has been properly designed and engineered.

The magnetic amplifier is no longer a laboratory device but has been tested and proven in actual military and industrial service until today it is a major factor in the design and development of an ever-increasing number of new and improved systems.

Reliability is by no means the only virtue of the magnetic amplifier. In many applications it has proven to be superior in performance, size and cost to vacuum tube

amplifiers, rotary amplifiers and even mechanical devices. No other amplifier offers such a combination of long life; low maintenance; ruggedness; resistance to extreme conditions of vibration, shock and temperature; no warm-up time; and high efficiency, in a completely static device which may be permanently sealed or potted. The day of the magnetic amplifier has arrived.

ACKNOWLEDGMENT

The author gratefully acknowledges the assistance of the many people in the magnetic amplifier field who so freely gave their time and advice in the initial preparation of this paper. Special thanks are given to E. T. Hooper and W. A. Geyger of the Naval Ordnance Laboratory for assistance in the final preparation of the manuscript.

The "M"-Type Carcinotron Tube*

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Summary—This paper presents theoretical and experimental results concerning the "M Carcinotron." In particular, the influence of space charge has been considered, thereby permitting an explanation for the measured values of starting current, the influence of the coupling impedance on efficiency and the existence of parasitic oscillations. The "rising sun effect" which should be present in these tubes, as it is in the magnetron, has been investigated theoretically and experimentally. The experimental results exhibit a decrease of efficiency in the predicted range of operation.

INTRODUCTION

THE Carcinotron¹ tubes are backward-wave oscillators. Their structure is characterized by the following features:

1. An electron beam is in interaction with a backward space harmonic of a delay line.
2. The power output is located at the gun end of the interaction space.
3. Means for absorbing rf energy reflected by possible output mismatch are introduced in the rf field of the delay line near the collector end generally inside the tube.

This structure gives a very wide electronic tuning range, and frequency insensitivity to load impedance.

Two types of Carcinotron tubes have been investigated: the "O" type, in which the beam travels in an interaction space at constant dc potential, as in the classical traveling-wave tube, and the "M" type, where the beam travels perpendicularly to crossed electric and magnetic fields, as in the magnetron amplifier.

This paper deals with the "M" Carcinotron. The structure of this tube and the results of a simplified small signal theory were given in a short note published in 1952.² The methods used to obtain these results were more fully described later.^{3,4}

The aim of this paper is to describe with more detail the properties of the "M" type Carcinotron, gathering and completing the information given previously.⁵⁻⁷

A small signal theory taking into account space-charge effects is established, allowing expressions for the starting current and frequency, build-up time and frequency pulling to be derived. The results of this theory are checked against experimental data. Some effects typical of the "M" Carcinotron (rising sun effect, parasitic oscillations) are explained. The practical interest of the "M" Carcinotron is best shown by the per-

* P. Guénard, O. Doehler, B. Epszstein, and R. Warnecke, "Nouveaux tubes oscillateurs à large bande d'accord électronique pour hyperfréquences," *C. R. Acad. Sci. (Paris)*, vol. 235, pp. 235-236; July, 1952. They were given previously, together with experimental results, by Epszstein at the 10th Conference on Electron Tube Research, Ottawa, Can., June 1952, in a discussion on R. Kompfner's paper "Backward waves," presented at this conference.

† R. Warnecke and P. Guénard, "Some recent work in France on new types of valves for highest radio-frequencies," *Proc. IEE*, vol. 100, part III, p. 351; November, 1953.

‡ R. Warnecke, P. Guénard, and O. Doehler, "Phénomènes fondamentaux dans les tubes à onde progressive," *L'Onde Electrique*, no. 325; April, 1954.

§ P. Guénard, R. Warnecke, O. Doehler, and B. Epszstein, "A new wide electronic tuning high efficiency microwave oscillator, the M Carcinotron," paper presented at the 11th Conference on Electron Tube Research, Stanford University, Stanford, Calif., June 1953.

¶ P. Guénard, "On some results obtained with O and M type Carcinotron," paper presented at the 12th Conference on Electron Tube Research, University of Maine, Orono, Me., June, 1954.

‡ O. Doehler, "Space charge effects in traveling wave tubes using crossed E-H fields," paper presented at the Symposium on Modern Advances in Microwave Techniques, New York, N. Y., October, 1954.

* Original manuscript received by the IRE, November 9, 1954; revised manuscript received December 16, 1954.

† Electronics Dept., Center of Tech. Res., Compagnie Générale de Télégraphie sans fil, Paris, France.

‡ Registered trade-mark of the Compagnie Générale de T.S.F.

performances obtained on this type of tube, some of which are given at the end of the paper. These performances, together with those previously published,^{3,4,8} suggest that the "M" Carcinotron should be best suitable when high power high efficiency operation is required.

DESCRIPTION OF THE "M" CARCINOTRON

Fig. 1 shows a linear version of the "M" Carcinotron. A delay line L is perfectly matched at the end M near the collector K by means of an attenuating material to avoid reflections. N is the output. An electron beam F , produced by the gun, travels parallel to the delay line L , and the sole S in the x -direction under the influence of a constant and uniform magnetic field B in the z -direction and an electric field E_0 in the y -direction due to the voltage V_0 applied between L and S .

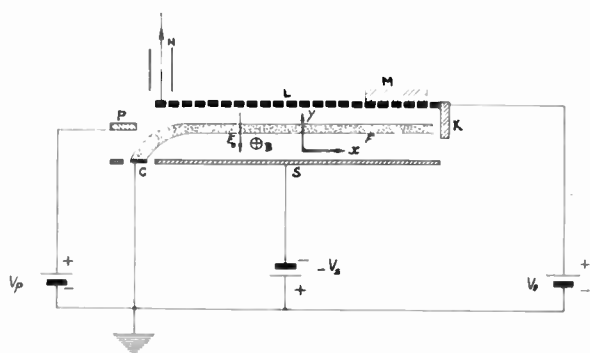


Fig. 1—Schematic structure of a linear "M" Carcinotron.

Interaction between the beam and the backward space harmonic will occur, if the phase velocity of the space harmonic is equal or nearly equal to the mean electron velocity $v_e = E_0/B$, i.e.:

$$v \cong v_e = E_0/B. \quad (1)$$

The positive feedback introduced by the beam leads to an oscillation if the current is high enough. This oscillation occurs with perfect matching at both ends of the line, the necessary feedback being furnished by the beam only. The oscillation frequency will be that for which (1) is fulfilled.

The beam focusing and the transfer of energy from the beam to the rf field occur in the same manner as in the magnetron amplifier:⁹ the transverse rf field produces a beam bunching in the favorable phase of the longitudinal rf electric field.

The energy transfer from the beam to the electromagnetic field of the line is achieved through the

longitudinal electric field. The electrons which transfer energy to the rf field, approach the anode and are brought to a higher dc potential, their velocity remaining approximately constant and equal to E_0/B .

As compared to the mechanism of operation of the "O" Carcinotron, it can be said that in the traveling-wave tubes with crossed electric and magnetic fields the potential energy of the electrons is transformed into rf energy; while in the "O" type traveling-wave tubes the kinetic energy of the electrons is transformed into rf energy.

SMALL SIGNAL THEORY

Small Signal Theory Neglecting Space Charge

In this theory, a rectilinear beam is considered. In the equations of the static trajectories

$$x = v_e \tau + a \cos(\omega_r \tau + \phi)$$

$$y = y_0 + a \sin(\omega_r \tau + \phi),$$

where τ is the transit time and $\omega_r = eB/m$, the cyclotron angular frequency, the amplitude a of the rolling circle is assumed zero, a condition which can be obtained with a proper set of initial conditions. Assuming that all rf quantities vary as $e^{j(\omega t - \Gamma x)}$, there exist, in the absence of coupling between the beam and the line, six waves

Beam	$\left\{ \begin{array}{l} \Gamma_1 = \Gamma_5 = \Gamma_e = \frac{\omega}{v_e} \\ \Gamma_{3,4} = \frac{\omega \pm \omega_r}{v_e} \end{array} \right.$	δx and δy arbitrary
		$\delta \dot{x} = 0, \delta \dot{y} = 0$
		$\delta y = \mp j \delta x$
Line	$\left\{ \begin{array}{l} \Gamma_2 = \Gamma_0 \\ \Gamma_6 = -\Gamma_0 \end{array} \right.$	$\delta \dot{x} = \mp j \omega_r \delta x, \delta \dot{y} = \mp j \omega_r \delta y$
		$E_y = j K E_x$
		$E_y = -j K E_x$

where $\delta z, \delta y, \delta \dot{x}, \delta \dot{y}$ are the rf components of electron motion, E_x and E_y the field components in the beam, and $K = \coth \Gamma y_0$.

The coupling between the beam and the line can modify significantly only those waves for which the propagation factors are near one another, i.e. the two beam waves Γ_1, Γ_5 and the line wave Γ_2 .

This assumes that one space harmonic only is considered. There is a possibility of simultaneous coupling of beam waves with different space harmonics, a question which will be discussed later (rising sun effect).

The theory (Appendix A) shows that one of the beam waves, say Γ_5 , is not coupled to the line. For this wave:

$$\Gamma_5 = \Gamma_e \quad \delta x = j K \delta y \quad \delta \dot{x} = \delta \dot{y} = 0 \quad E_x = E_y = 0.$$

The modified values of Γ_1 and Γ_2 are solutions of the equation:

$$(\Gamma - \Gamma_e)(\Gamma - \Gamma_0) = \frac{\Gamma_0^2 \Gamma_e R_c I_0 K}{E_0} = \gamma M^2, \quad (2)$$

where R_c is the coupling impedance, I_0 and E_0 the dc beam current and electric field. The two solutions can

⁸ R. Warnecke, "Sur quelques résultats récemment obtenus dans le domaine des tubes pour hyperfréquences," *Ann. Radioélect.*, vol. ix, pp. 107-135; April, 1954.

⁹ R. Warnecke, W. Kleen, A. Lerbs, O. Doehler, and H. Huber, "The magnetron type traveling-wave amplifier tube," *Proc. I.R.E.*, vol. 38, pp. 486-495; May, 1950. See also J. R. Pierce, "Traveling-wave Tubes," D. Van Nostrand Co., Inc., New York, N. Y., chap. XV; 1950.

be written:

$$\Gamma_{1,2} = \Gamma_m \pm \sqrt{\Gamma_d^2 + \gamma_M^2},$$

where

$$\Gamma_m = \frac{\Gamma_0 + \Gamma_e}{2}$$

$$\Gamma_d = \frac{\Gamma_e - \Gamma_0}{2}.$$

For these waves:

$$\delta x = \frac{K}{\Gamma - \Gamma_e} \frac{E_x}{E_0}, \quad \delta y = j \frac{\delta x}{K}, \quad \delta \dot{x} = -j(\Gamma - \Gamma_e) v_e \delta x, \quad \delta \dot{y} = j \frac{\delta \dot{x}}{K}.$$

The three other waves are only slightly modified by the coupling.

This theory can be extended to a beam of finite thickness if the trajectories are linear, which means that a nonequipotential cathode is used. It is then found that the rf charge density inside the beam is zero (see Appendix A).

In addition to assumptions already mentioned, it has been supposed that the periodicity of the line structure has no influence on the static trajectories, i.e. that the distance between the line and the sole is large as compared to the pitch of the line. A two-dimensional problem has been treated, which supposes the structure infinite in the z direction.

These six waves make it possible to satisfy the boundary conditions, i.e. the values of δx , δy , $\delta \dot{x}$, $\delta \dot{y}$ at $x=0$ (gun end of the line) and the existence of reflection factors a_0 and a_l at the ends of the line. If it is supposed that $\delta x = \delta y = \delta \dot{x} = \delta \dot{y} = 0$ for $x=0$, the amplitudes of the waves Γ_3 , Γ_4 , Γ_5 are found equal to zero and there remain only the two principal waves Γ_1 , Γ_2 and the reflected wave Γ_6 .

Small Signal Theory Taking Into Account Space Charge

It is no more possible in this case to consider an infinitely thin beam. As shown by Brillouin,¹⁰ linear trajectories are possible, assuming an equipotential cathode, if the plasma frequency Ω_0 equals the cyclotron frequency ω_r :

$$\Omega_0 = \sqrt{\frac{e}{m} \frac{\rho_0}{\epsilon_0}} = \frac{eB}{m} = \omega_r,$$

ρ_0 being the charge density in the electron beam. The electron velocity varies inside the beam as $\omega_r y$.

If a nonequipotential cathode is used, it is possible to avoid this condition; with a constant charge density in the beam, the electron velocity varies as $y\Omega_0^2/\omega$.

It has been supposed, with no other basis than the agreement between theory and experimental data, that it is possible to apply the results thus obtained to a beam

with complicated trajectories, if the charge density in the linear beam is taken equal to the average charge density in the actual beam.

The general space-charge theory leads to a transcendental equation which is difficult to solve. Therefore, the influence of space charge has been introduced as a perturbation in the theory without space charge in the following manner:¹¹ The beam travels in an rf field which is the sum of the field Φ_L guided by the line and the field Φ_e created by the space charge. The trajectories of the electrons are determined by $\Phi_L + \Phi_e$, while the transfer of energy from the beam to the line is determined by Φ_L only. The approximation consists in calculating Φ_e from trajectories determined without space charge. This leads (see Appendix A) to the following values of the propagation constants for the two principal waves:

$$T_{1,2} = \Gamma_m \pm \sqrt{\Gamma_d^2 + \gamma_M^2 \left(1 - \frac{\alpha}{\Gamma_d \mp \sqrt{\Gamma_d^2 + \gamma_M^2}} \right)}, \quad (3)$$

where

$$\alpha = \frac{\Omega_0^2 \Gamma_d \Delta}{\omega v_e}.$$

2Δ is the width of the beam.

Starting Conditions

The boundary conditions give the starting conditions for oscillations⁴

$$(\Gamma_2 - \Gamma_e)(a_0 a_l e^{i\Gamma_0 l} - e^{-i\Gamma_2 l}) = (\Gamma_1 - \Gamma_e)(a_0 a_l e^{i\Gamma_0 l} - e^{-i\Gamma_1 l}). \quad (4)$$

If the line is matched at least at one end ($a_0 a_l = 0$) and if it has no attenuation, the starting conditions are:

$$v_0 = v_e \quad (5)$$

$$\gamma_M l = \frac{\pi}{2} + 2\pi n \quad (n = 0, 1, 2, \dots) \quad (6)$$

Eq. (5) says that for every possible oscillation ($n=0, 1$, etc.) the phase velocity of the line at the oscillating frequency is equal to the electron velocity. Eq. (6) is the condition for the starting current.²

$$I_s = \left(\frac{\pi}{2} + 2\pi n \right)^2 \frac{E_0}{\Gamma_0^2 \Gamma_e l^2 R_e} \tanh \Gamma_e y_0, \quad (7)$$

l being the length of the line. The influence of an attenuated line and of reflections already have been discussed.⁴ If space charge effects are taken into account, (3) must be used, and the starting conditions are:

$$\Gamma_d = \frac{\alpha}{2} \quad (8)$$

¹⁰ L. Brillouin, "Trajectories in a single anode magnetron," *Elec. Commun.*, p. 460; 1946.

¹¹ R. Warnecke, O. Doehler and O. Bobot, "Les effets de la charge d'espace dans les tubes à propagation d'onde à champ magnétique," *Ann. Radioélect.*, vol. V, p. 279; October, 1950.

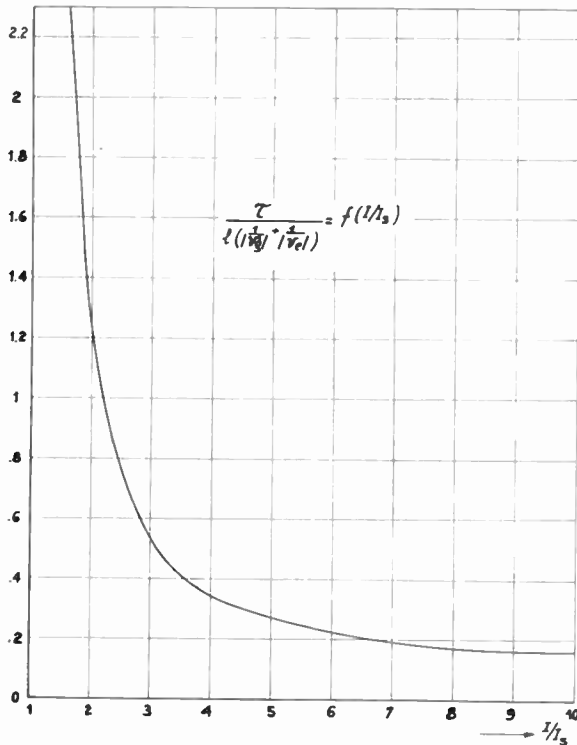


Fig. 2—Build-up time as a function of beam current.

$$\gamma_M l \left(1 + \frac{3\alpha^2}{8\gamma_M^2} \right) = \frac{\pi}{2} + 2\pi n. \quad (9)$$

Eq. (8) shows that, in the presence of space charge, the phase velocity of the line at the oscillating frequency is not the same as the electron velocity and assumes different values for the various values of n .

Eq. (9) shows that the starting current is decreased by the effect of space charge.

Build-Up Time

If the beam current is higher than the starting current, the tube oscillates and the rf amplitude starts increasing with t , the time dependent factor being:

$$e^{j\bar{\omega}t} = \alpha e^{j\omega t + t/\tau}.$$

The time τ characterizes the build-up time for small values of t and permits the determination of the approximate order of magnitude of the build-up-time.

The calculation of τ is analogous to the calculation of the starting current. But the balance of power must be modified to take into account the increase of the amplitude.

The power transferred by the beam to the rf field in a section dx of the line, and during the time dt , is the sum of:

1. The energy absorbed by the line in the length dx and during the time dt .
2. The increase of power, propagated along the line in the section dx and during the time dt .

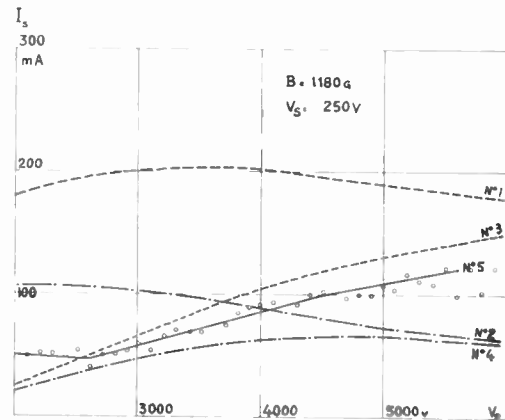


Fig. 3—Starting current as function of voltage. (1) Without space charge—linear trajectories. (2) Without space charge—cycloidal trajectories. (3) With space charge—linear trajectories. (4) With space charge—cycloidal trajectories. (5) Experimental curve.

3. The increase of electromagnetic energy dW stored in the section dx during the time dt .

$$dW = j(\bar{\omega} - \omega_0)W dx dt$$

W = stored energy per unit length,

ω_0 = angular frequency of the free wave.

If space charge is neglected, the equation which determines the propagation constant then has the form:

$$\Gamma - \Gamma_e - \frac{\bar{\omega} - \omega_0}{v_g} = \frac{\gamma_M^2}{\Gamma - \frac{\bar{\omega}}{v_e}}. \quad (10)$$

v_g is the group velocity.

The boundary conditions of rf current for $x=0$ and of electric field for $x=l$ permit determination of $\bar{\omega}$.

If there is a perfect match at least at one end of the tube and if the line has no attenuation, ω and τ are given by:

$$\omega = \omega_0 \quad (11)$$

$$\cos \left[2\gamma_M l \sqrt{1 - \left(\frac{\theta}{2\gamma_M} \right)^2} \right] = - \left[1 - 2 \left(\frac{\theta}{2\gamma_M} \right)^2 \right] \quad (12)$$

with:

$$\theta = \frac{1}{\tau} \left[\frac{1}{|v_g|} + \frac{1}{|v_e|} \right]. \quad (13)$$

Eq. (12) is a transcendental equation for θ , if γ_M , i.e. the current, is given. In Fig. 2,

$$\frac{\tau}{l \left(\frac{1}{|v_g|} + \frac{1}{|v_e|} \right)}$$

has been plotted as a function of I/I_s (I_s = starting cur-

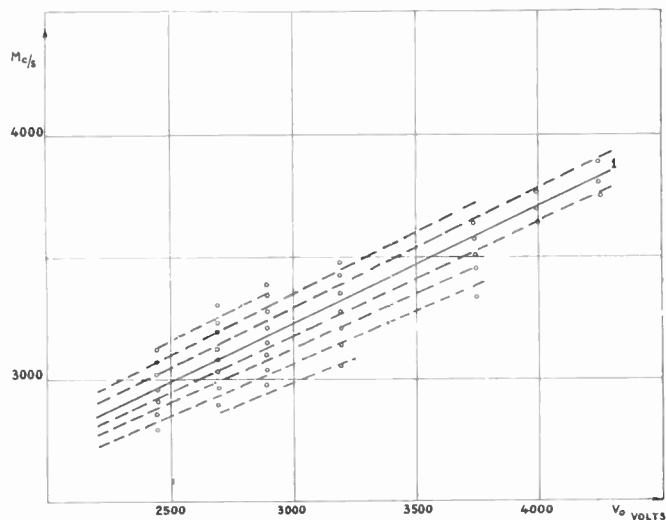


Fig. 4—Frequency of parasitic oscillations as a function of voltage (solid curve-normal oscillation).

rent). If I/I_s is large, τ is given by the approximate expression:

$$\tau = \frac{1}{\pi} \frac{1}{\sqrt{I/I_s}} \cdot l \left(\frac{1}{|v_g|} + \frac{1}{|v_e|} \right).$$

In practice, τ is of the order of 10^{-9} seconds in the "S" band tubes. The total build-up time is then of the order of 10^{-7} to 10^{-8} seconds.

In the presence of space charge, τ decreases, so that (12) and (13) give the upper limit for τ .

EXPERIMENTAL RESULTS

Starting Current

The theory for the starting current has been checked on different types of experimental tubes with a linear structure. Fig. 3 shows the results obtained with a linear tube. The starting current is a function of the shape of the trajectories, and it has been calculated for two different electron guns. For the "ideal" gun trajectories are straight lines; for the "magnetron gun" trajectories are cycloids as in the plane magnetron without space charge. In practice, the trajectories are between these two limits. Curves 1 and 2 have been calculated from (6) neglecting space charge. Curves 3 and 4 have been calculated from (9). Curve 5 has been measured. The theory neglecting space charge gives starting currents much too high, especially at low voltages; the space charge has an important influence on the starting conditions and the theory gives the correct order of magnitude.

Parasitic Oscillations

For high currents and low voltages, parasitic oscillations can be observed. Sidebands with an amplitude from 1/100 to $\frac{1}{2}$ that of the principal frequency occur.

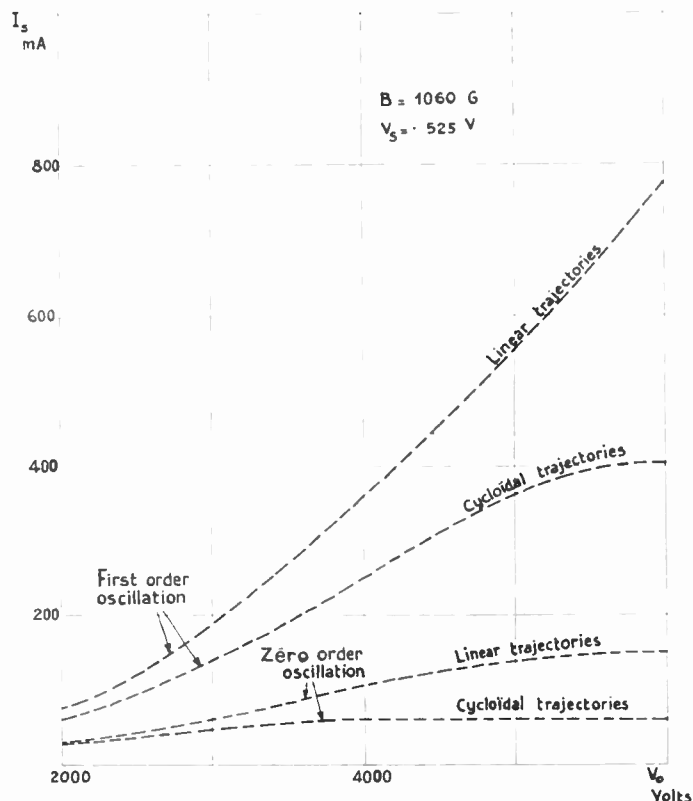


Fig. 5—Theoretical starting current for normal oscillation and first parasitic oscillation.

In Fig. 4 the main frequency and the measured sidebands as a function of voltages are shown. The difference of frequency between successive bands is approximately 60 mc. The amplitude is smaller for sidebands of higher order.

These oscillations can be attributed to the excitation of the higher orders as given by (9). In Fig. 5 the theoretical starting current of the zero order ($n=0$) and of the first order ($n=1$) have been plotted. It follows that the starting current of the first order for low voltages is only three times higher than that of the zero order. The frequency difference between these two oscillations is almost independent of voltage and, for the case of Fig. 4, has from (8) a theoretical value of 45 to 70 mc for different trajectories of the electrons.

The high frequency sidebands can be explained by intermodulation between these two oscillations, and an asymmetry must appear, as shown in Fig. 4.

Pushing Figure

If small signal theory with space charge were used to calculate the frequency as a function of current, large variations (a few per cent) of frequency should occur. But measurements show that the pushing figure is relatively low. Measurements on a tube in the "S" band have shown that variation of frequency is 2 to 3 mc for high voltages (5,000 v) and 5 mc for low voltages (2,500 v) when the power output varies from one to ten.

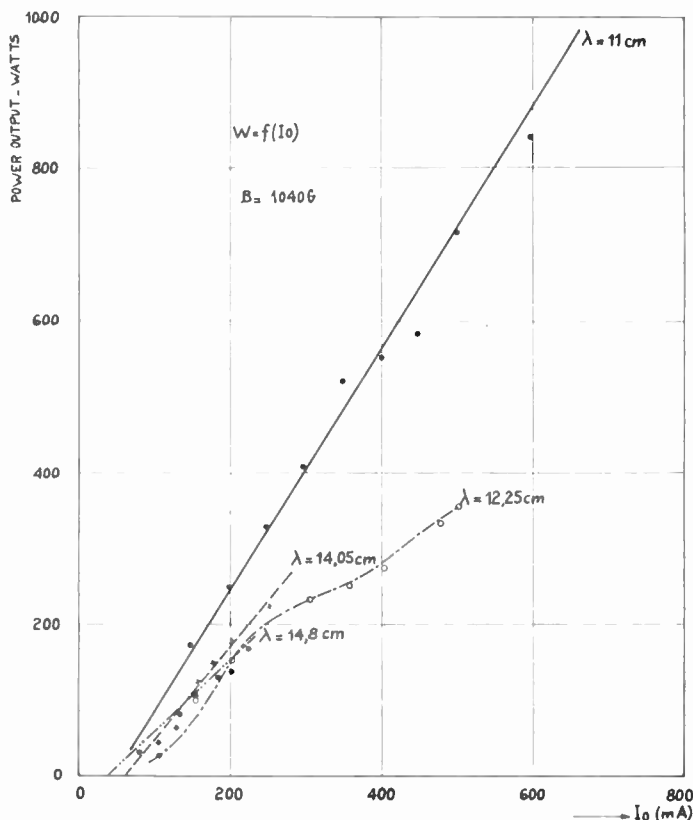


Fig. 6—Power output as function of beam current for various frequencies.

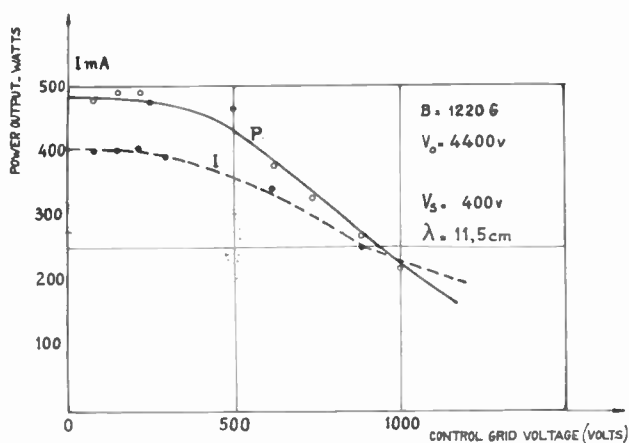


Fig. 7—Power output and beam current as functions of control grid voltage.

Amplitude Modulation

In Fig. 6, power has been plotted as a function of beam current for different frequencies. For these experiments a tungsten filament was used and the current controlled by the temperature of the filament.

Usually, these curves are straight lines and power output is of the form:

$$P = A(I - I_s),$$

I_s being the starting current and A a constant.

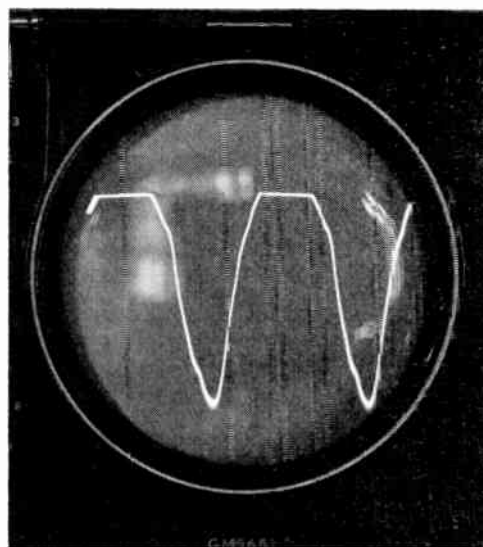


Fig. 8—Oscillogram of power output against control electrode voltage.

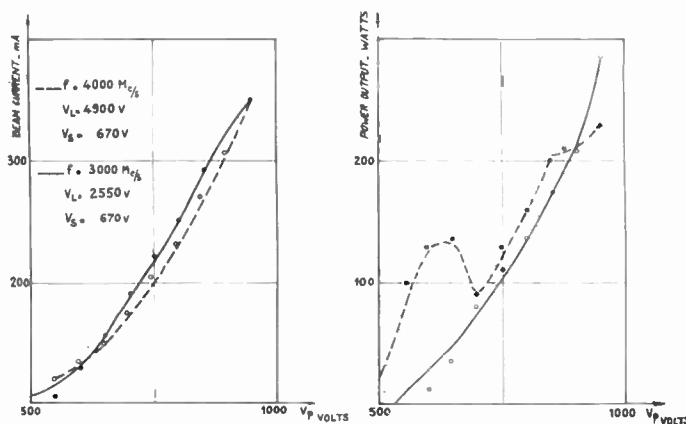


Fig. 9—Power output and beam current as a function of plate voltage.

For some frequencies, especially at low voltages, $P = f(I)$ has a more complicated form. Up to now, this has not been explained.

Fig. 7 shows rf power and beam current as a function of control grid voltage for two frequencies.

Fig. 8 shows the power output as a function of grid voltage for a 50 cps modulation.

This characteristic remains unchanged for modulating frequencies up to at least 5 mc.

The control grid is an electrode surrounding the filament. The "cutoff" is relatively high. It is also possible to control the current with the plate of the optical system. In this case the modulating voltage is lower. But as shown in Fig. 9 the power vs plate voltage curve is sometimes more complicated because the shape of the trajectories is influenced by the plate voltage.

Influence of Load

If attenuation at the end of the line near collector gives a perfect match and there is no reflection along circuit from machining irregularities, frequency will be in-

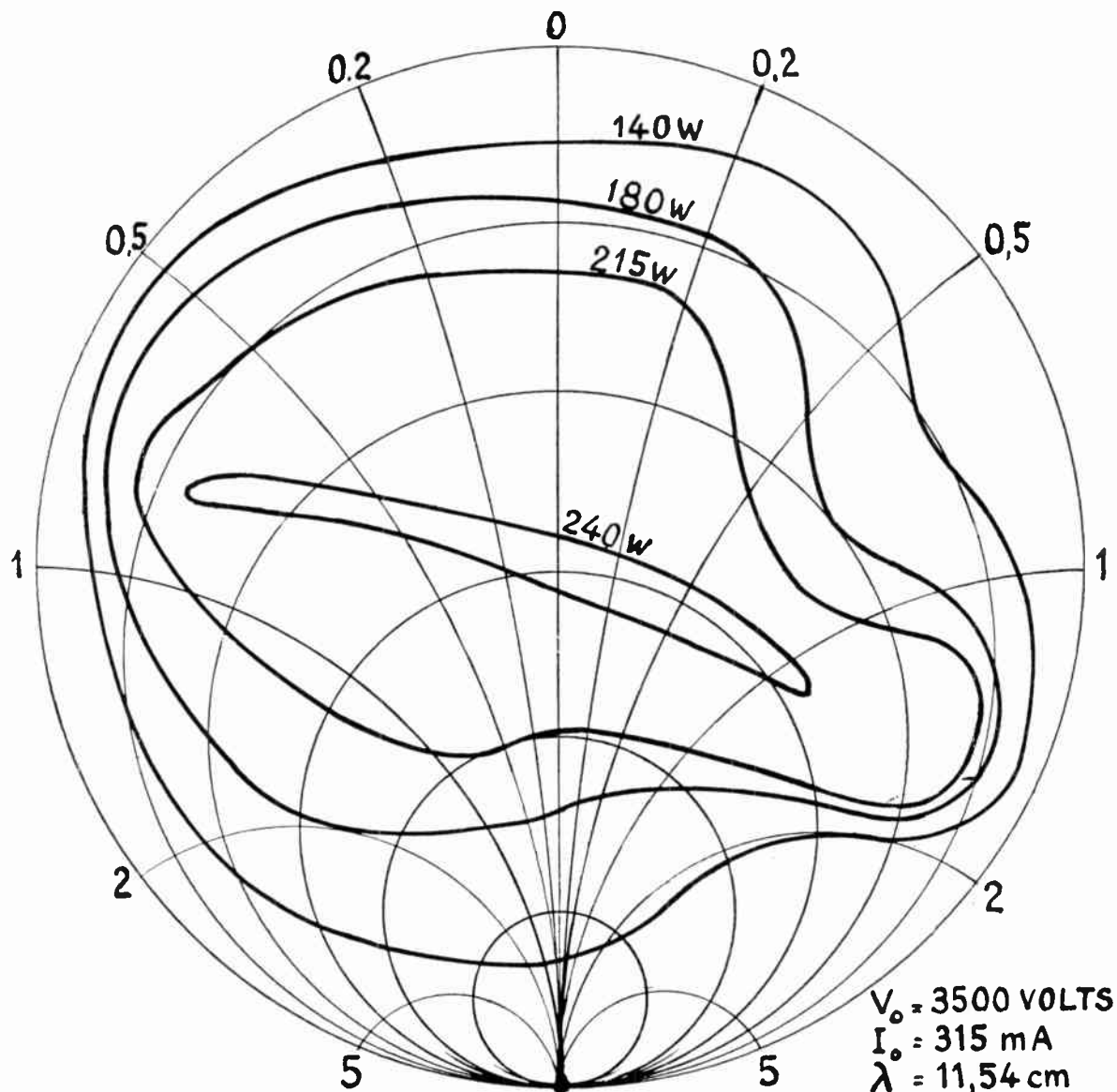


Fig. 10—Lines of constant power on a Smith chart.

dependent of load and curves of constant power must be circles about the center of the Smith chart.

Fig. 10 is the Rieke diagram for constant power. Variation of frequency was below measurement accuracy (1 mc).

Fig. 11 (next page) shows frequency vs sole-voltage for an imperfect attenuation at end of line near the collector. Parameter is vswr introduced in the output circuit. The form of the curves agrees with theory.⁵

Signal to Noise Ratio

The measurement of the signal-to-noise ratio is possible, if the "M" carcinotron itself is used as local oscillator. In this case, the variations of frequency due to the ripples in the power supplies cancel out, but it is not possible to separate the two noise sidebands. The noise of the Carcinotron was compared with a mercury-argon

noise source amplified with a low-noise traveling-wave tube.

Fig. 12 (next page) shows noise per cps to signal ratio as a function of frequency distance from oscillating frequency. Noise per cps to signal ratio is of the order of -140 db.

Fig. 13 (page 421) shows the noise in relative units as a function of current.

Efficiency

Electronic Efficiency: If space charge is neglected the electronic efficiency η_e can be calculated with the same method as in the magnetron amplifier. If all electrons are absorbed by the line, η_e is given by:

$$\eta_e = 1 - m \frac{V_f}{V_0}$$

V_f is the voltage corresponding to the drift velocity of

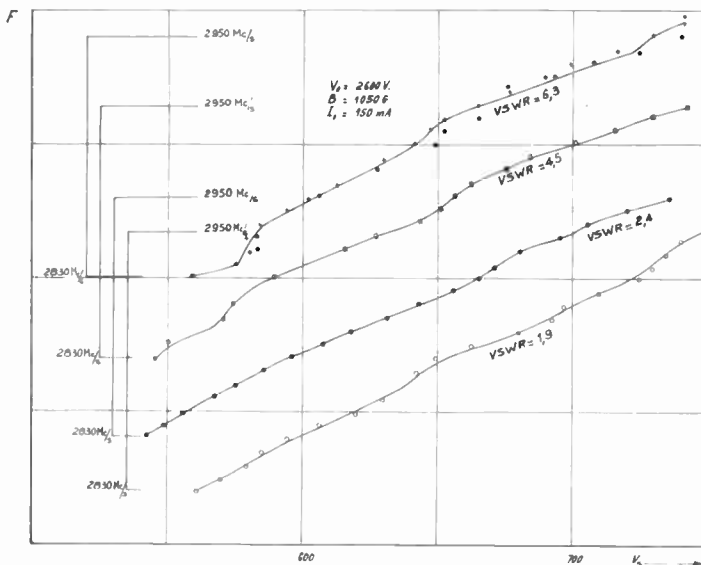


Fig. 11—Frequency vs sole voltage for various values of load vswr (ordinate curves have been translated for the various curves).

the electrons given by (1). m is a constant depending on the trajectories. If the trajectories are linear, K is unity; if they are cycloids corresponding to a plane magnetron without space charge, we have $m = 4$.

Actually, not all electrons are absorbed by the line. Measurements on distribution of collector and line current indicate that more than 80 per cent of electrons arrive on the line, and electronic efficiency should be:

$$\eta_e \geq 0.8 \left(1 - m \frac{V_f}{V_0} \right)$$

which can be transformed to:

$$\eta_e \geq 0.8 \left(1 - \frac{m}{4} \left(\frac{B_{cr}}{B} \right)^2 \right),$$

B_{cr} being the cut-off magnetic field of a plane magnetron under the same conditions.

Circuit Efficiency: The circuit efficiency is given by the Q_v of the circuit and the external Q_{ext} :

$$\eta_c = \frac{Q_v}{Q_{ext} + Q_v}$$

Q_v is defined by:

$$Q_v = \frac{\omega \int_0^L W(y) dy}{P_L}$$

and Q_{ext} by:

$$Q_{ext} = \frac{\omega \int_0^L W(y) dy}{P_0}$$

W is the stored energy per unit length, P_L the power loss in the line, and P_0 the output power. If γ_a is the

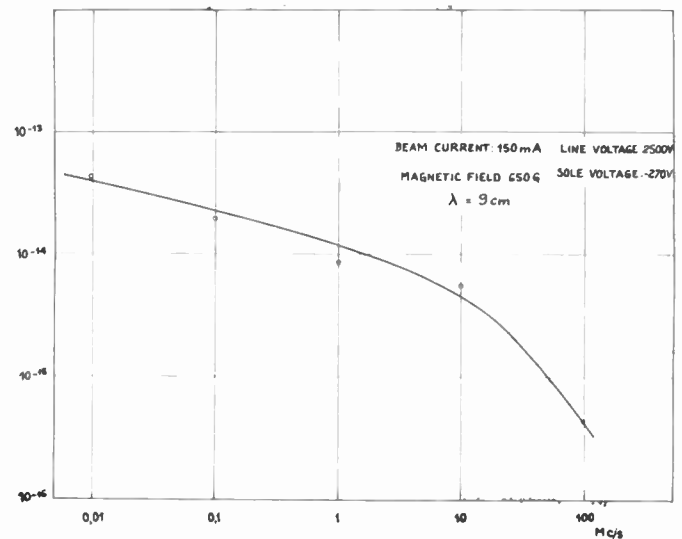


Fig. 12—Noise-to-signal ratio for a 1 cps bandwidth. The abscissa is the difference between the oscillating frequency and the frequency at which noise is measured.

attenuation of the line in Neper/cm, we have:

$$P_L = 2\gamma_a \int_0^L P'(y) dy,$$

where $P'(y)$, power propagating along the line, is related to $W(y)$:

$$P'(y) = W(y)v_g$$

and therefore:

$$Q_v = \frac{k_g}{2\gamma_a}, \quad k_g = \frac{\omega}{v_g}.$$

Q_{ext} is given by:

$$Q_{ext} = \frac{k_g \int_0^L W(y) dy}{W(0)}.$$

For small signal theory $W(y)$ can be obtained by the superposition of the waves with the propagation constants given by (2), and we obtain:

$$Q_{ext} = \frac{k_g}{2} \frac{\epsilon^{\gamma_a l} - 1}{\gamma_a} = Q_v (\epsilon^{\gamma_a l} - 1) \quad (14)$$

and:

$$\eta_c = \epsilon^{-\gamma_a l}. \quad (15)$$

If the attenuation is small (14) gives:

$$Q_{ext} = \frac{k_g}{2} l, \quad (16)$$

and found directly, if losses in the line are neglected.

Influence of the Coupling Impedance on the Efficiency: According to the theory neglecting space charge, efficiency should not depend on coupling impedance for

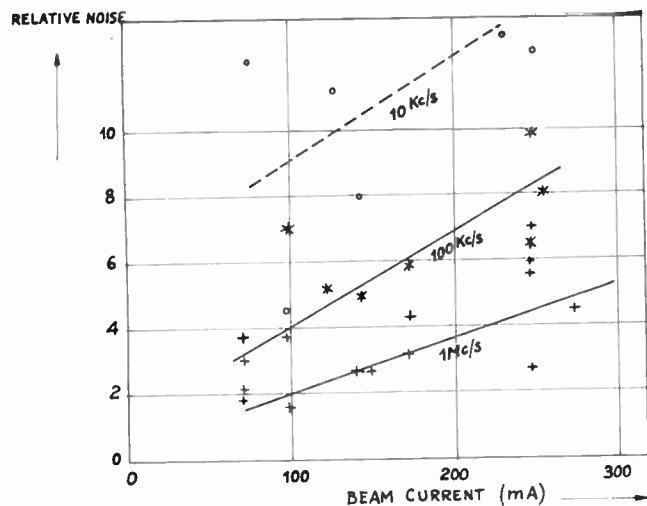


Fig. 13—Relative noise as a function of beam current.

high beam currents and should be a function only of the shape of trajectories, phase velocity of the circuit, and acceleration voltage. However, the first tubes tested have given a relatively low efficiency, much lower than predicted by theory.

It has been assumed space-charge phenomena have an important influence on efficiency (diocotron effect)¹² and if coupling impedance is high, space-charge effects should be negligible and efficiency should increase.

It can be seen that, contrary to the space-charge neglecting theory, efficiency increases rapidly with coupling impedance. The same occurs with the tuning range. As far as power is concerned, there is optimum value for the coupling impedance: an increase of coupling impedance decreases the thermal dissipation of the line and consequently the dc power which can be applied.

Rising-Sun Effect¹³

In the small signal theory, it is generally assumed that interaction takes place with one space harmonic. However there is the possibility that one space harmonic Γ_0 being coupled to the beam wave Γ_e , another space harmonic Γ_0' will be coupled to the beam wave $\Gamma_e - \omega_r/v_e$. This will occur if:

$$\Gamma_0 - \Gamma_0' = \frac{\omega_r}{v_e} \quad (17)$$

The differences $\Gamma_0 - \Gamma_0'$ between two successive space harmonics being $2\pi/p$, where p is the pitch of the line, (17) implies that:

$$\frac{\omega_r}{v_e} = \frac{2\pi}{p} \quad (18)$$

¹² P. Guénard and H. Huber, "Etude expérimentale de l'interaction par ondes de charge d'espace au sein d'un faisceau électronique se déplaçant dans des champs électrique et magnétique croisés," *Ann. Radioélect.*, vol. VII, p. 252; October, 1952.

¹³ W. E. Willshaw, G. Mourier, and G. Guilbaud, "Effet de résonance électronique dans les tubes à champs électrique et magnétique croisés," *Compt. Rend. Acad. Sci. (Paris)*, (in press).

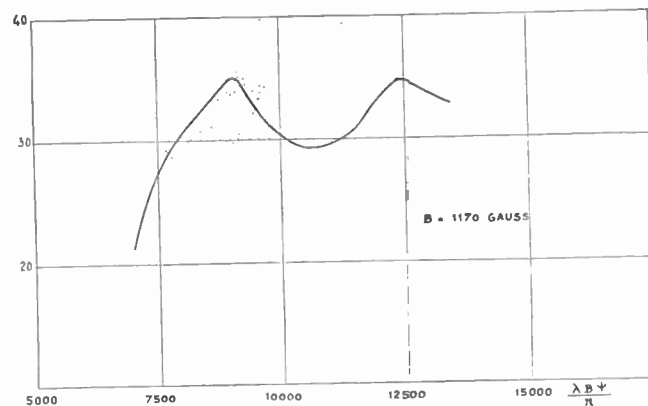


Fig. 14—Efficiency as function of voltage exhibiting rising-sun effect (average over 16 tubes).

If (18) is fulfilled, the space harmonic $\Gamma_0'' = \Gamma_0 + 2\pi/p$ will be coupled to the beam wave $\Gamma_e + \omega_r/v_e$. However, this effect can be neglected, because of the low coupling impedance of the wave Γ_0'' .

This simultaneous coupling of two space harmonics can occur in certain lines, e.g., interdigital lines, where asymmetrical and symmetrical space harmonics exist.^{14,15} The first symmetrical space harmonic being normally used, interaction can occur with the first asymmetrical space harmonic which is in fact the fundamental space harmonic. For the symmetric space harmonics, the apparent pitch of the line is half the real pitch p . ψ being the phase angle along this apparent pitch for the first symmetric space harmonic, (18) can be expressed numerically in the following form:

$$\lambda B = \frac{10700\pi}{\psi} \quad (19)$$

At the "cutoff" ($\psi = \pi$), this is the well-known condition for the "rising-sun effect" in the magnetron¹⁶ characterized by a pronounced minimum in efficiency. As shown by Fig. 14 above, same phenomenon appears in the "M" Carcinotron for values of parameters satisfying (19). It is shown in Appendix B that simultaneously the starting current must exhibit a maximum. This can be seen on Fig. 15 (next page).

CHARACTERISTICS OF AN "M" CARCINOTRON

Fig. 16 (page 422) shows an experimental setup for studying properties of "M" Carcinotron. Fig. 17 (page 422) shows same type of tube in an industrial form. To reduce the bulk of the tube, the line has been curved into a circular form. The electromagnet has been replaced by a permanent magnet. Fig. 18 (page 423) shows the performance charts measured on this tube, and Fig. 19 (page 423)

¹⁴ R. C. Fletcher, "A broad-band interdigital circuit for use in traveling-wave-type amplifiers," *Proc. I.R.E.*, vol. 40, pp. 951-958; August, 1952.

¹⁵ A. Leblond and G. Mourier, "Etude des lignes à barreaux a structure périodique," *Ann. Radioélect.*, vol. ix, p. 184; April, 1954.

¹⁶ G. B. Collins, "Microwave Magnetrons," Radiation Lab. Series, McGraw-Hill Book Co., Inc., New York, N. Y., sec. 3.3; 1948.

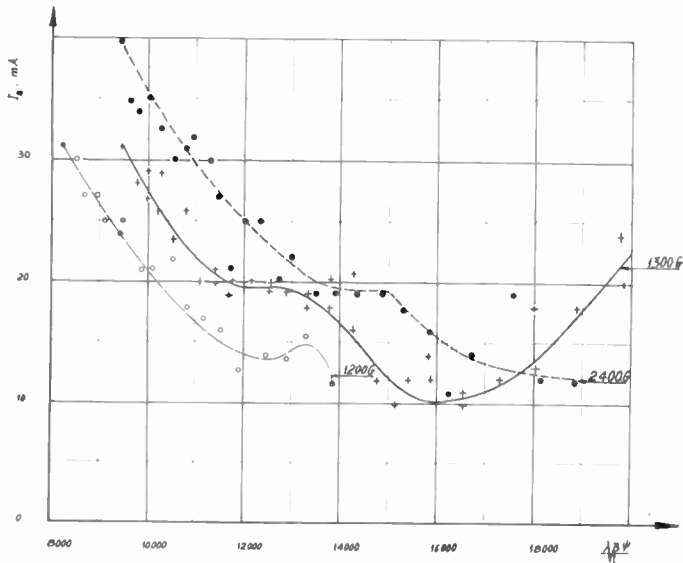


Fig. 15—Starting current curves exhibiting rising-sun effect.

the variation of power, efficiency and frequency with line voltage. These figures show possibility of obtaining, under practical conditions and for frequencies around 3,000 mc, power outputs of several hundred watts in a frequency range larger than half an octave and with efficiencies of the order of 40 per cent. In some cases, efficiencies in excess of 60 per cent have been measured.⁸

CONCLUSION

This paper shows that if the theory does not cover entirely the behavior of the "M" Carcinotron, in particular the space-charge effects, it nevertheless predicts the main features of this type of tube.

The main features of the "M" Carcinotron, as compared to other types of electronically tunable tubes, are: high efficiency, a fairly linear frequency-voltage characteristic, and a low pushing and pulling figure, together with a wide electronic range.

These features make the "M" Carcinotron particularly advantageous as a high-power electronically-tunable tube.

APPENDIX A

Without space charge, the motion of the electrons is determined by the field of the line, the components of which can be written:

$$E_x e^{j(\omega t - \Gamma x)} = -\frac{\partial \Phi_L}{\partial x}, \quad E_y e^{j(\omega t - \Gamma y)} = -\frac{\partial \Phi_L}{\partial y}$$

with:

$$\Phi_L = \phi_L \frac{\sinh \Gamma y}{\sinh \Gamma y_0} e^{j(\omega t - \Gamma x)}.$$

The equations giving the rf components of the motion are, for a linear trajectory $x = v_e \tau$, $y = y_0$:

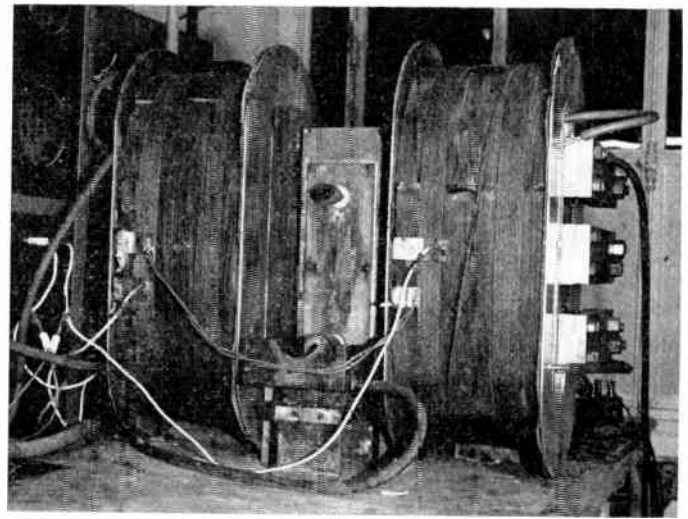


Fig. 16—Experimental setup for studying a linear "M" Carcinotron.

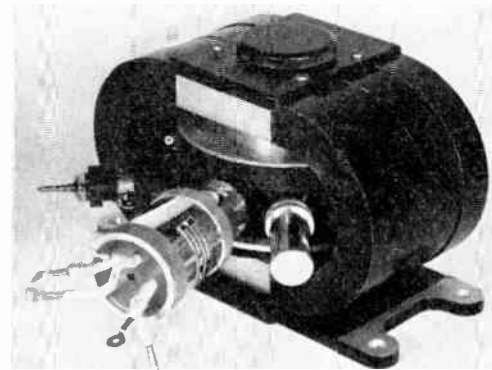


Fig. 17—Industrial model of an "M" Carcinotron.

$$-\xi^2 \delta x - j\xi \omega_r \delta y = -j\eta \Gamma \phi_L \quad (20)$$

$$j\xi \omega_r \delta x - \xi^2 \delta y = \eta \Gamma \phi_L K \quad (21)$$

$$(\xi = \omega - \Gamma v_e; K = \coth \Gamma y_0).$$

The coupling between the beam and the line is given by:

$$\Gamma - \Gamma_0 = j \frac{\vec{I} \cdot \vec{E}^*}{4P} = j \frac{\vec{I} \vec{E}^* R_c \Gamma_0^2}{2\Gamma \Gamma^* \phi_L \phi_L^*},$$

where R_c is the coupling impedance

$$R_c = \frac{E_x E_x^*}{2\Gamma_0^2 P}.$$

An infinitely thin sheet of the beam carrying the dc current dI_0 (which is taken positive) contributes to $\vec{I} \cdot \vec{E}^*$ through the following term:

$$-j\Gamma_e dI_0 [\delta x E_x^* + \delta y E_y^*] = -\Gamma_e \Gamma^* \phi_L^* dI_0 [\delta x - jK\delta y],$$

which gives, I_0 being the current carried by the beam:

$$j(\Gamma - \Gamma_0) \Gamma \phi_L = \frac{\gamma_M^2}{2} \left[\frac{\delta x}{K} - j\delta y \right] \quad (22)$$

where

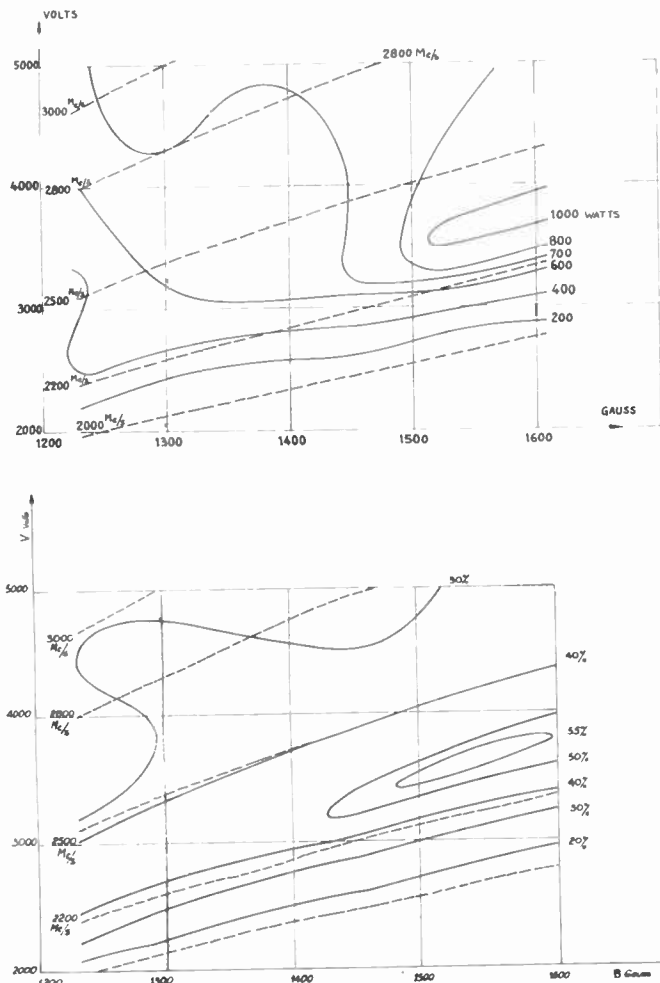


Fig. 18—Performance charts of an "M" Carcinotron: (a) Frequency and power output vs magnetic field and voltage. (b) Efficiency vs magnetic field and voltage.

$$\gamma_M^2 = \frac{I_0 \Gamma_e \Gamma_0^2 R_c K}{E_0}$$

Eqs. (20), (21) and (22) determine five waves, one of which is:

$$\Gamma_s = \Gamma_e \quad \phi_L = 0 \quad \delta x = jK\delta y \quad \delta \dot{x} = \delta \dot{y} = 0.$$

Assuming

$$|\Gamma_e - \Gamma_0| \ll \frac{\omega_r}{v_e} \quad \text{and} \quad \gamma_M^2 \ll \left(\frac{\omega_r}{v_e}\right)^2,$$

$\Gamma_{1,2}$ are the two principal waves for which $|\xi| \ll \omega_r$. The propagation constants of these two waves are given by

$$(\Gamma - \Gamma_e)(\Gamma - \Gamma_0) = \gamma_M^2 \quad (23)$$

i.e.

$$\Gamma_{1,2} = \Gamma_m \pm \sqrt{\Gamma_d^2 + \gamma_M^2} \quad (24)$$

with

$$\Gamma_m = \frac{\Gamma_e + \Gamma_0}{2}, \quad \Gamma_d = \frac{\Gamma_e - \Gamma_0}{2}.$$

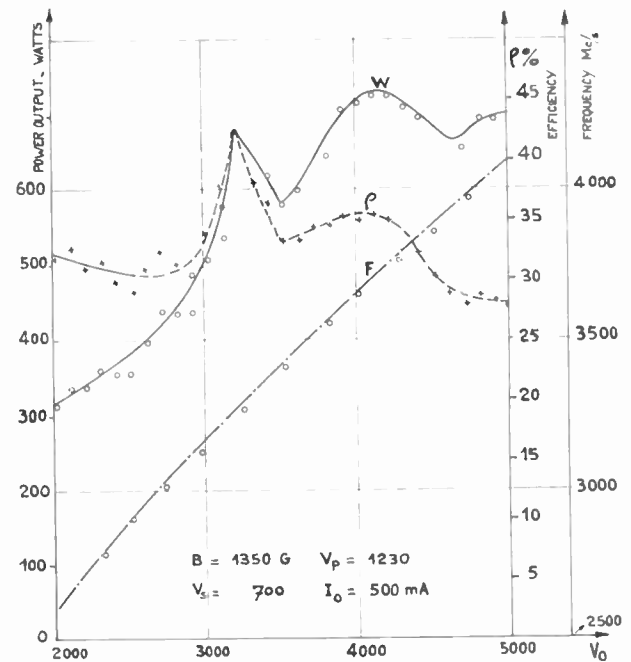


Fig. 19—Frequency, power output and efficiency as functions of line voltage.

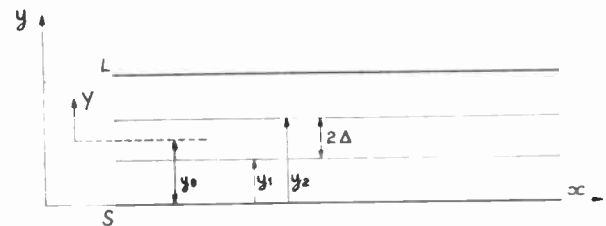


Fig. 20

The two other waves correspond to $\xi = \pm \omega_r$. As compared to the waves $\Gamma_{1,2}$, they are weakly coupled to the field of the line, as seen from (20) and (21), which relate amplitudes of the field and of the electron motion.

The potential $\phi e^{j(\omega t - \Gamma x)}$, from which derives the space-charge field of such a beam, can be calculated from the rf space-charge ρ and the surface density $-\rho_0 \delta y$.

The space-charge density:

$$\rho = +\rho_0 \left(\frac{\partial}{\partial x} \delta x + \frac{\partial}{\partial y} \delta y \right)$$

is equal to zero. The potentials ϕ_c inside, and ϕ_1 , ϕ_2 outside the beam (Fig. 20 above) satisfy Laplace equations and are thus a linear combination of $\epsilon \Gamma_v$ and $\epsilon - \Gamma_v$.

If the sole and the line are remote enough from the beam ($\Gamma y_1 \gg 1$, $\Gamma(d - y_2) \gg 1$), the potentials are:

$$\begin{aligned} \phi_1 &= A \epsilon \Gamma_v \\ \phi_2 &= B \epsilon - \Gamma_v \\ \phi_c &= C \epsilon \Gamma_v + D \epsilon - \Gamma_v. \end{aligned}$$

The boundary conditions at the two edges of the beam give

$$\phi_c = \frac{-\rho_0 \delta y_2}{2\Gamma \epsilon_0} \epsilon^{\Gamma(\nu-\nu_2)} + \frac{\rho_0 \delta y_1}{2\Gamma \epsilon_0} \epsilon^{\Gamma(\nu_1-\nu)}.$$

Assuming $\Gamma\Delta \ll 1$, ϕ_c takes the simplified form:

$$\phi_c = -\psi L \frac{\Omega_0^2 \Gamma}{\omega_r \xi} (K\Delta + Y)(1 - \Gamma\Delta). \quad (25)$$

With space charge, the equations of motion are:

$$\begin{aligned} \delta x'' - \omega_r \delta \dot{y} &= -j\eta \Gamma \left(\phi_L \frac{\sinh \Gamma y}{\sinh \Gamma y_0} + \phi_c \right) \epsilon^{j\xi' \tau} \\ \delta \ddot{y} + \omega_r \delta \dot{x} &= \Omega_0^2 \delta y + \eta \frac{\partial}{\partial y} \left[\phi_L \frac{\sinh \Gamma y}{\sinh \Omega y_0} + \phi_c \right] \epsilon^{j\xi' \tau} \end{aligned}$$

with

$$\begin{aligned} \xi' &= \omega - \Gamma v_e' - \Gamma \frac{\Omega_0^2}{\omega_r} Y \\ v_e' &= v_e - \frac{\Omega_0^2}{\omega_r} \Delta \left(1 - \frac{2y_0}{d} \right). \end{aligned}$$

Introducing in these equations the value of ϕ_c given by (25), and assuming:

$$\Gamma \frac{\Omega_0^2 \Delta}{\omega_r v_e} = \alpha \ll \gamma_M,$$

these equations give

$$(\Gamma - \Gamma_0)(\Gamma - \Gamma_e) = \gamma_M^2 \left(1 - \frac{\Omega_0^2 \Gamma \Delta}{\omega_r \xi} \right), \quad (26)$$

ξ being the value of $\omega - \Gamma v_e$, given by the theory which neglects space charge. The propagation constants of the two principal waves are thus given by:

$$\Gamma = \Gamma_m \pm \sqrt{\Gamma_d^2 + \gamma_M^2 \left(1 - \frac{\alpha}{\Gamma_d \mp \sqrt{\Gamma_d^2 + \gamma_M^2}} \right)}. \quad (27)$$

APPENDIX B

In Appendix A, it has been supposed that one space harmonic was coupled to the beam. Let us suppose now that the space harmonics are coupled to the beam corresponding to propagation constants Γ_0 and $\Gamma_0' = \Gamma_0 - 2\pi/p$, p being the pitch of the line. Both space harmonics will be strongly coupled to the beam if:

$$\Gamma_e \sim \Gamma_0, \quad \Gamma_e - \frac{\omega_r}{v_e} \sim \Gamma_0 - \frac{2\pi}{p}$$

that is if:

$$\frac{\omega_r}{v_e} = \frac{2\pi}{p}.$$

The amplitudes of these two space harmonics are related by the properties of the line, and for the second wave, the propagation constant will be $\Gamma' = \Gamma - 2\pi/p$ and the amplitude $k\phi_L$. The equations of motion (20) and (21) hold for the first space harmonic. For the second space harmonic, the equations of motion are:

$$-\xi'^2 \delta x' - j\xi' \omega_r \delta y' = -j\eta \Gamma' k \phi_L \quad (28)$$

$$j\xi' \omega_r \delta x' - \xi'^2 \delta y' = \eta \Gamma' k \phi_L K', \quad (29)$$

with:

$$\xi' = \xi + \omega_r.$$

The equation of coupling between the line and the beam must take into account both space harmonics, that is:

$$\begin{aligned} \vec{I} \cdot \vec{E}^* &= -jI_0 \Gamma_e [\delta x E_x^* + \delta y E_y^* + \delta x' E_x'^* + \delta y' E_y'^*] \\ &= -\Gamma_e \Gamma^* \phi_L^* I_0 \left[(\delta x - jK\delta y) \right. \\ &\quad \left. + \frac{\Gamma'^*}{\Gamma^*} k(\delta x' - jK'\delta y') \right]. \quad (30) \end{aligned}$$

Eq. (23) becomes

$$(\Gamma - \Gamma_e)(\Gamma - \Gamma_0) = \gamma_M^2(1 - S),$$

where

$$S = \frac{1 + K'^2}{4K} |k|^2 \left| \frac{\Gamma'}{\Gamma} \right|^2,$$

showing that the starting current is increased by the factor $1/\sqrt{1-S}$.

There are now two waves uncoupled to the beam ($\phi_L = 0$) corresponding to $\xi = 0$, $\xi' = \omega_r$, the four factors δx , δy , $\delta x'$, $\delta y'$ being related by two equations:

$$\delta x - jK\delta y + \frac{\Gamma'}{\Gamma} k(\delta x' - jK'\delta y') = 0$$

$$\delta x' = -j\delta y'.$$



Power Flow in Electron Beam Devices*

W. H. LOUISELL† AND J. R. PIERCE†, FELLOW, IRE

Summary—This paper discusses power flow in devices in which electrons are constrained to move in the z direction only. Besides the electromagnetic power flow given by Poynting's vector, there is a kinetic power flow per unit area in the z direction. In a linear system equivalent to the electron beam at low levels of operation this power flow is

$$P_R = -\frac{1}{2} \frac{m}{e} (-J_0 + J)(u_0^2 + 2u_0v).$$

Here $-J_0$ and J are the dc and instantaneous ac convection current densities and u_0 and v are the dc and instantaneous ac velocities.

The electromagnetic power must be calculated including all fields due to the presence of the beam. In the case of space-charge waves, the electromagnetic power flow adds to or subtracts from the kinetic power flow. If the electric field is purely longitudinal, H is zero, and the electromagnetic power flow is zero.

IN ELECTRON beam devices such as traveling-wave tubes and klystrons there is not only electromagnetic power flow, but also power flow associated with the kinetic energy of the electrons; we may call this latter kinetic power flow. This note discusses power flow in beam devices, and particularly power flow at low signal levels for which linearized equations can be used to describe the operation of the devices.

Maxwell's equations yield a relation which may be written

$$\nabla \cdot P_e + \frac{\partial}{\partial t} W_e + E \cdot J = 0, \quad (1)$$

where

$$P_e = E \times H \quad (2)$$

$$W_e = \frac{1}{2} \mu H \cdot H + \frac{1}{2} \epsilon E \cdot E. \quad (3)$$

Here E , H and J are vectors; they are the electric field, the magnetic field, and the convection current density. P_e is the Poynting vector, which may be interpreted as electromagnetic power flow per unit area and W_e , a scalar, is the electromagnetic stored energy per unit volume associated with the total electric and magnetic fields, in the presence of the electrons.

Consider a case in which the only convection current (other than convection current at the surface of perfect conductors, for which $E \cdot J = 0$) is due to a cloud of electrons of charge density ρ , the electrons being free to move in the z direction only and having at any point a common velocity v . The nonrelativistic equation of motion is

$$\frac{\partial v}{\partial t} + v \frac{\partial v}{\partial z} = -\frac{e}{m} E_z$$

$$E_z = -\frac{m}{e} \left[\frac{\partial v}{\partial t} + v \frac{\partial v}{\partial z} \right]. \quad (4)$$

We also have relations

$$J = J_z = \rho v \quad (5)$$

$$\nabla \cdot J + \frac{\partial \rho}{\partial t} = \frac{\partial J_z}{\partial z} + \frac{\partial \rho}{\partial t} = 0. \quad (6)$$

Consider the term $E \cdot J$ in (1); using (4) we can rewrite it

$$E \cdot J = -\frac{m}{e} \left[\frac{\partial v}{\partial t} + v \frac{\partial v}{\partial z} \right] J. \quad (7)$$

A little differentiation, together with (5) and (6), shows that

$$E \cdot J = \frac{\partial W_k}{\partial t} + \frac{\partial P_k}{\partial z}, \quad (8)$$

where

$$W_k = -\frac{1}{2} \frac{m}{e} \rho v^2 \quad (9)$$

$$P_k = (P_k)_z = -\frac{1}{2} \frac{m}{e} J v^2. \quad (10)$$

Thus, we can rewrite (1)

$$\nabla \cdot (P_e + P_k) + \frac{\partial}{\partial t} (W_e + W_k) = 0. \quad (11)$$

Here of course $\nabla \cdot P_k = \partial(P_k)_z / \partial z$. Eq. (11) represents the conservation of energy. As P_e and W_e are electromagnetic power flow and electromagnetic energy, so P_k and W_k are kinetic power flow and kinetic energy.

In dealing with the low-level operation of electron beam devices we do not use (4) and (5) but, rather, we use linearized equations and use

$$E_z = -\frac{m}{e} \left[\frac{\partial v}{\partial t} + u_0 \frac{\partial v}{\partial z} \right] \quad (4a)$$

$$J = J_z = -\rho_0 v + u_0 \rho. \quad (5a)$$

Here u_0 is the average electron velocity and v is a small ac velocity component; $-\rho_0$ is the average charge density and ρ is a small ac component. We may regard (4a) and (5a) and (6) as the equations of an "equivalent" linear system whose behavior is the same as the behavior of the actual nonlinear system at low levels of operation. We may ask, what is the correct expression for power flow and stored energy in the equivalent linear system.

In the linear system

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$$E \cdot J = -\frac{m}{e} \left[\frac{\partial v}{\partial t} + u_0 \frac{\partial v}{\partial z} \right] J. \quad (12)$$

$$E_z = -\frac{\partial V}{\partial z} = j\beta V. \quad (16)$$

We find that $(E \cdot J)$ can be expressed in the form (8) if we define

$$P_k = -\frac{1}{2} \frac{m}{e} (-J_0 + J)(u_0^2 + 2u_0v) \quad (13)$$

and

$$W_k = -\frac{1}{2} \frac{m}{e} [-\rho_0(u_0^2 + 2u_0v + v^2) + \rho(u_0^2 + 2u_0v)]. \quad (14)$$

These expressions are not the result of a linearized expansion; (13) and (14) were chosen so that (11) holds for the linear system with P_k and W_k so defined. We will run into no contradictions if we interpret P_k and W_k as kinetic power flow of the beam per unit area and kinetic energy in the beam per unit volume respectively.

In dealing with amplifiers we deal with signals which vary sinusoidally with time. In the linear system we can consider each frequency component of the ac quantities separately. Let us then think of the ac quantities as complex quantities containing a factor $e^{j\omega t}$. Then the average power flow in the z direction per unit area will be

$$P_z = \frac{1}{4} \left[(E \times H^* + E^* \times H)_z - \frac{m}{e} u_0(vJ^* + v^*J) \right]. \quad (15)$$

In this expression E , H , v and J are peak values. We can get the total power flow in the z direction by integrating the z component of P as given by (15) with respect to x and y .

Let us now consider the power flow in space-charge waves. In the case of true plasma waves, in which all the displacement current is in the direction of electron motion, the displacement current is equal and opposite to the convection current. There is no net current in the z direction and no electromagnetic power flow, since II is zero. In this case the total power flow of the wave is the kinetic power flow, the second term of (15).¹

At the other extreme, we can consider space-charge waves in an electron stream in a tube narrow compared with the space-charge wave length. In this case the displacement current in the z direction is negligible, and there is an electromagnetic power flow.

In the case of a space-charge wave, the variation of the ac quantities with respect to z can be expressed by a factor $e^{-i\beta z}$. Let us consider slow space-charge waves, for which the electric field can be expressed with adequate accuracy in terms of the gradient of a potential V , so that

Let us consider the electromagnetic power flow in the z direction per unit area, P_{e1} associated with the convection current J .²

$$P_{e1} = \frac{1}{4}(VJ^* + V^*J).$$

Using (16) and (12) to express V in terms of v , we find this to be

$$P_{e1} = -\frac{1}{4} \frac{m}{e} u_0 \left[\frac{\omega - \beta u_0}{\beta u_0} \right] [vJ^* + v^*J]. \quad (17)$$

We see from (15) that the ratio of P_{e1} to the kinetic power density P_k is

$$\frac{P_{e1}}{P_k} = \frac{\omega - \beta u_0}{\beta u_0}; \quad (18)$$

for small space charges, very nearly

$$\omega - \beta u_0 = \pm \omega_q \quad (19)$$

$$\beta u_0 = \omega. \quad (20)$$

Here ω_q is called the *effective plasma frequency*. Thus, for small space charge, approximately

$$\frac{P_{e1}}{P_k} = \pm \frac{\omega_q}{\omega}. \quad (21)$$

In terms of the traveling-wave tube parameters Q and C

$$\frac{P_{e1}}{P_k} = \pm (2\sqrt{QC})C. \quad (22)$$

For realistic space charge waves, the electromagnetic power flow will lie between the extremes of 0 (electric field purely longitudinal) and P_{e1} (longitudinal electric field negligible compared with transverse electric field). Thus, we see that when space charge is small ($\omega_q/\omega \ll 1$, $C \ll 1$) the electromagnetic power of the space-charge wave is negligible compared with the kinetic power. However, for large space charge the electromagnetic power may be important except in cases of negligible transverse electric field.

When the space charge is large, (19) does not hold unless the transverse electric field is negligible, and (20) does not hold at all. Thus, to calculate the power accurately one must integrate (15) across the system.

One can also use (15) to calculate the small-signal power flow in traveling wave tubes. Here one must include the total electromagnetic power flow associated with both beam and circuit currents as well as the kinetic power flow.

¹ The kinetic power and the fact that it is negative for the slow space-charge wave, were treated in a paper by L. J. Chu, presented at the IRE Electron Devices Conference, University of New Hampshire, June, 1951.

² S. A. Schelkunoff, "Electromagnetic Waves," D. Van Nostrand Co., New York, p. 79; 1943; S. A. Schelkunoff and H. T. Friis, "Antennas, Theory and Practice," John Wiley and Sons, New York, pp. 76-78; 1952.

There is quite a different approach toward evaluating the power at low levels, in which one deals not with linear equations describing an equivalent linear system, but with the actual nonlinear equations at low levels.³ In this case one must carry the solution beyond the first order. The final results, however, agree with those given here.

³ L. R. Walker, "Power Flow in Electron Beams," to be published in the *Jour Appl. Phys.*

R. W. Gould of the California Institute of Technology⁴ has evaluated the power flow and stored energy in the case of electrons free to move in any direction and in which $II=0$ (plasma oscillations).

The authors are much indebted to L. R. Walker for contributions at all stages in the evolution of this paper.

⁴ R. W. Gould, California Institute of Technology, Electron Tube and Microwave Laboratory Quarterly Status Report No. 5. p. 15; April 1, 1954 to June 30, 1954.

The Effective Surface Recombination of a Germanium Surface with a Floating Barrier*

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Summary—The effect of heavily doped (alloyed) *p*-type and *n*-type surface layers on *n*-type base, and of metallic plating on *n*-type base, on the surface recombination velocity *s* has been computed on the basis of one-dimensional junction theory. The results indicate that *s* should be of the order 1 cm/sec for the heavily doped surfaces, and several thousand cm/sec for the electroplated surface. The low *s* comes about for the same reason that the injection efficiency of alloy junctions is high; the alloy junction is a very efficient emitter of minority carriers into the base and a poor acceptor of majority carriers from the base because of the high doping level in the alloyed region. Since recombination in the surface layer of minority carriers from the base requires both majority and minority carriers, the restriction of the flow of either reduces the surface recombination.

However, measurements of *s* by diffusion and pulse methods on alloy junction surfaces indicate that their apparent recombination is almost the same as adjacent untreated surface, e.g., 300–500 cm/sec. It is shown that lateral current flow, due to minority carrier gradients parallel to the junction interface, and neglected in one-dimensional theory, gives rise to circulating currents which translate the minority carriers to the nearest high recombination surface. This hole translation property of the floating *p*-layer is used to explain the erroneously high lifetimes often observed by diffusion measurements on silicon and *p*-type germanium, and certain discrepancies in effective life measurement on completed transistors.

GENERAL DISCUSSION

THE RATE OF recombination of minority carriers at free surfaces often plays a dominant role in determining the characteristics of semiconductor devices. For example, the current amplification-factor of a transistor is usually determined by surface recombination more than by any other quantity.^{1,2}

* Original manuscript received by the IRE, December 3, 1954; revised manuscript received, January 11, 1955.

† RCA Labs, Princeton, N.J.

¹ A. R. Moore and J. I. Pankove, "Effect of junction shape and surface recombination on transistor current gain," *PROC. I.R.E.*, vol. 42, pp. 907–913; June, 1954.

² W. M. Webster, "On the variation of junction transistor current amplification factor with emitter current," *PROC. I.R.E.*, vol. 42, pp. 914–920; June, 1954.

Thermal generation of minority carriers, which is related to both surface and volume recombination, results in saturation current in rectifiers and transistors. In many devices, the free surfaces contribute most of this generally undesirable saturation current. In some photoconductor devices, surface recombination limits sensitivity.

Surface recombination can be expressed quantitatively through the surface recombination velocity *s*.³ The rate at which minority carriers recombine is proportional to the product of their concentration and *s*. In theory *s* is a characteristic of the surface and may have any value between zero and thermal velocity (about 10⁷ cm/sec). Experiments on germanium surfaces show that *s* depends on the surface treatment and that values ranging from about 50 cm/sec to several thousand cm/sec may be obtained by different chemical treatments.^{1,4} To reduce surface recombination, we are interested in treatments which result in very low values of *s*.

A variety of models of the surface which might give low surface recombination velocity can be imagined. Three, experimentally attainable with reasonable certainty, are illustrated in Fig. 1 (next page). They all imply the addition or production of a film on the surface which has different electrical characteristics from the bulk. These three possibilities are: (1) a metallic film, (2) a layer of opposite conductivity type, and (3) a layer of the same conductivity type but of higher conductivity. Plating techniques permit a metallic layer to be formed and the change of conductivity may be accom-

³ W. Shockley, "Holes and Electrons in Semiconductors," D. Van Nostrand Co., New York, p. 321; 1950.

⁴ E. M. Conwell, "Properties of silicon and germanium," *Proc. I.R.E.*, vol. 40, pp. 1327–1337; November, 1952.

plished by alloying⁵ or by diffusing impurities into the surface.⁶ While many other surface models are possible, these three are easily analyzed and should permit comparison of theory with experiment. The present work evaluates the possibilities for reducing s by these means in terms of a simple one-dimensional analysis and discusses some preliminary experimental results.

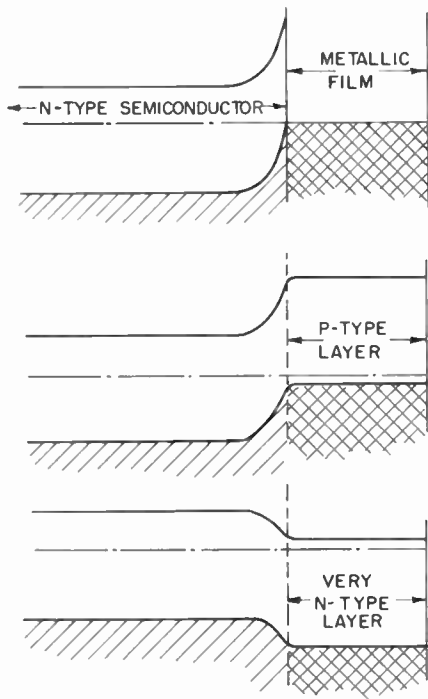


Fig. 1—Energy level diagrams for three possible types of surface barrier.

THEORY

In this section, equations are given which may be used to predict the rate of surface recombination for metallic, n , and p -type films on n -type material. The same reasoning can of course be applied to films on p -type material. The treatment is one-dimensional for simplicity and is therefore subject to the assumption that no parameters vary appreciably along the surface. The discussion of experimental results which follows shows that this condition can be troublesome in actual practice.

One can see intuitively why surface layers such as these are hopeful. In all cases, a barrier exists near the surface to one type of carrier or the other. Since both holes and electrons must be present for recombination, restraining the flow of either should reduce surface recombination.

The approach of this section is as follows. First, an equivalent surface recombination velocity s is defined

⁵ R. R. Law, C. W. Mueller, J. I. Pankove, and L. D. Armstrong, "A developmental germanium p - n - p junction transistor," *PROC. I.R.E.*, vol. 40, pp. 1352-1357; November, 1952.

⁶ R. N. Hall and W. C. Dunlap, " P - N junctions prepared by impurity diffusion," *Phys. Rev.*, vol. 80, pp. 467-468; November, 1950.

for p - and n -type surface layers and evaluated in terms of recombination in the film. Following this an equation for s for a metallic surface is given and an approximate value computed. Finally, a relation between s and γ ("emitter efficiency" of the layer if it were used as the emitter of a transistor) is demonstrated.

n - and p -Type Layers

Definition of Equivalent s : The equivalent surface recombination velocity for a surface with a semiconducting layer on it will be called s and will be defined as follows:

Consider the situation of an n -type semiconductor with a surface layer d cm thick and composed of the same material but of different conductivity. The energy band configuration for both cases of interest, and the pertinent parameters, are labeled in Fig. 2. Holes and electrons recombine in the surface layer by both surface and volume recombination. To maintain nonequilibrium steady-state densities of minority and majority carriers in the surface layer, holes and electrons must flow in equal numbers from the bulk into the surface layer.

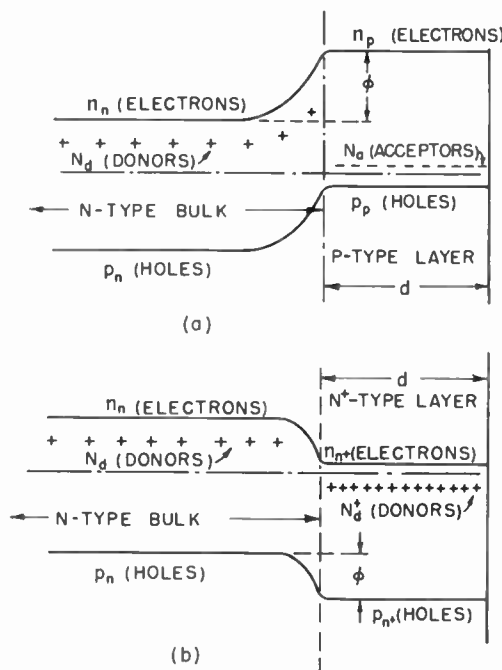


Fig. 2—(a) Detailed energy level diagram for (a) p -type layer on n -type base and (b) strongly n -type layer on an n -type base.

Further, this flow must equal the net recombination rate in the surface layer. This results in two equal current densities J_e and J_p . The effective surface recombination velocity will be defined as

$$s = \frac{J}{q(p_n - p_0)}; \quad (1)$$

where $J = J_e = J_p$, p_0 is the equilibrium value of p_n , and

q is the electronic charge ($p_n - p_0$ is the excess hole density in the bulk). This is the value of surface recombination velocity which one would measure by any of our present techniques.⁷⁻⁹

So that the general considerations may apply to either p - or n -type layers we will identify the minority carrier density in the surface layer with the letter m , and its equilibrium value with m_0 . J is determined by the minority carrier lifetime in the surface layer τ , the surface recombination velocity for minority carriers s_m at the actual outer surface [as distinguished from the effective s of (1)]; the thickness of the surface layer d , and the excess minority carrier density ($m - m_0$) in the surface layer. The minority carriers determine the recombination rate in the surface layer regardless of whether they are electrons or holes.

Two approximate expressions for J are

$$J = q(m - m_0) \left(\frac{d}{\tau} + s_m \right), \quad (2a)$$

when $d < L_m$ (i.e., for thin layers),

$$J = q(m - m_0) \left(\frac{L_m}{\tau} \right) = q(m - m_0) \frac{D_m}{L_m}, \quad (2b)$$

when $d > L_m$. D_m and L_m are the minority carrier diffusion coefficient and diffusion length, respectively, in the surface layer.

Combining (2a) and (2b) with (1):

$$s = \left(\frac{d}{\tau} + s_m \right) \frac{(m - m_0)}{(p_n - p_0)} \quad (3a)$$

when $d < L_m$, and

$$s = \frac{D_m}{L_m} \frac{(m - m_0)}{(p_n - p_0)} \quad (3b)$$

when $d > L_m$.

The ratio of the steady-state excess minority carrier densities, $(m - m_0)/(p_n - p_0)$, are now calculated for n - and p -type layers.

p -type Surface Layer: The ratio $(m - m_0)/(p_n - p_0)$ depends on the conductivities of the bulk and surface layer with acceptor density N_a as shown in Fig. 2(a). The electron density in the p -type material n_p is the minority carrier density m in (3a) and (3b). Thus, what is desired is $n_p - n_0$ to replace $m - m_0$ in the general expressions for s .

Four basic equations link the densities of holes, electrons, donors, and acceptors with the barrier height, ϕ :

$$n_n = N_d + p_n, \quad (4a)$$

$$n_p + N_a = p_p, \quad (4b)$$

$$n_n e^{-q\phi/kT} = n_p, \quad (4c)$$

⁷ L. B. Valdes, "Measurement of minority carrier lifetime in germanium," *Proc. I.R.E.*, vol. 40, pp. 1420-1423; November, 1952.

⁸ S. Lederhandler and L. Giacoletto, "Measurement of minority carrier lifetime and surface effects in junction devices," to be published.

⁹ D. T. Stevenson, and R. J. Keyes, *Bull. Am. Phys. Soc.*, vol. 29, p. 18; March, 1954.

and

$$p_n = p_p e^{-q\phi/kT}. \quad (4d)$$

The first two equations indicate charge neutrality in the surface layer and bulk, and the second pair relate the hole and electron concentrations on either side of the boundary. The only assumptions involved in applying these equations are (1) net electrical neutrality except in the depletion layer at the boundary, and (2) s reasonably small (compared to thermal velocity). Both are sufficiently satisfied. Eq. (4) may be combined to yield an expression linking n_p and p_n :

$$p_n^2 + p_n N_d + n_p^2 N_a, \quad (5a)$$

and the same form applies to the equilibrium densities:

$$p_0^2 + p_0 N_d = n_0^2 + n_0 N_a. \quad (5b)$$

Combining these to yield the form needed to calculate s is difficult. However, we may make some simplifying assumptions. In the event that N_a is very large (a very p -type surface), n_p^2 is negligible compared to $n_p N_a$. If, in addition, $N_d \gg p_n$, we can write very simply

$$\frac{n_p - n_0}{p_n - p_0} \approx \frac{N_d}{N_a}. \quad (6a)$$

N_d/N_a may now be substituted into (3a) and (3b) in place of $(m - m_0)/(p_n - p_0)$ to calculate s . Under conditions of high injected hole density in the bulk, p_n may not be negligible compared to N_d . However, it will then be great compared to p_0 and we may write

$$\frac{m - m_0}{p_n - p_0} \approx \frac{n_p - n_0}{p_n} \approx \frac{p_n + N_d}{N_a}. \quad (6b)$$

Thus, at high levels of p_n , s will increase linearly with hole density in the n -type material.

Substitution of (6a) into (3a) and (3b) yields:

$$s = \left(\frac{d}{\tau} + s_m \right) \frac{N_d}{N_a} \text{ for thin layers,} \quad (7a)$$

and

$$s = \frac{D_m}{L_m} \frac{N_d}{N_a} \text{ for thick layers (compared to } L_m). \quad (7b)$$

L_m and D_m apply to electrons in the surface layer. These equations apply for low injection levels ($p_n \ll N_d$).

n^+ Surface Layer: The form of the equivalent expressions for s for the case of a very n -type layer is similar to the foregoing and may be derived in much the same way. Here, holes are the minority carriers in both the surface layer and the bulk. The symbol $+$ refers to characteristics in surface layer as indicated in Fig. 2b.

We can write four equations similar to those used for the case of the p -type surface:

$$n_n = p_n + N_d \quad (8a)$$

$$n_{n+} = p_{n+} + N_d \quad (8b)$$

$$n_n = n_{n+} e^{-q\phi/kT} \quad (8c)$$

$$p_{n+} = p_n e^{-q\phi/kT}. \quad (8d)$$

These equations are of the same form as (4). The difference is that n_p , n_0 , and n_a of (5a) and (5b) are replaced by p_{n+} , p_{0+} , and N_d^+ , respectively. Subject to similar assumptions, the solutions will be of the same form as (7). Thus we have:

$$s = \left(\frac{d}{\tau} + s_m \right) \frac{N_d}{N_d^+}, \quad \text{when } d < L_m, \quad \text{and} \quad (9a)$$

$$s = \frac{D_m}{L_m} \cdot \frac{N_d}{N_d^+}, \quad \text{when } d > L_m, \quad (9b)$$

where D_m and L_m apply to holes in the surface layer.

General Expressions for s : In the case of a surface layer which is thick compared to a minority carrier, diffusion length within it, we have (7b) and (9b). We may now use the Einstein relationships ($D_p = kT\mu_p/q$ and $D_n = kT\mu_n/q$) and introduce $\sigma_b \approx q\mu_n N_d$ (the conductivity of the bulk material) and σ_s (the conductivity of the surface layer). For the n^+ type layer, $\sigma_s = q\mu_n N_d^+$, while for the p -type layer, $\sigma_s = q\mu_p N_a$. By manipulation, both (7b) and (9b) become

$$s = \frac{D_p}{L_m} \cdot \frac{\sigma_b}{\sigma_s}; \quad (d > L_m). \quad (10)$$

It should be emphasized that D_p is the hole diffusion coefficient in the surface layer. Because of impurity scattering this may be lower than D_p in the bulk.

Similarly, a generalized expression for (7a) and (9a) may be written which applies to thin layers:

$$s = \left(\frac{d}{\tau} + s_m \right) \frac{\sigma_b}{\sigma_s} \cdot \frac{\mu_s}{\mu_b} \quad \text{when } d < L_s. \quad (11)$$

Here, τ is the minority lifetime in the surface layer, s_m is the surface recombination velocity at the actual outer surface of the layer, μ_s and μ_b are majority carrier mobilities in the surface layer and bulk, respectively. The value μ_s will be less than the value measured in relatively pure material because of impurity scattering, and possibly scattering at the surface.

Evaluation of p and n Layers: It is difficult to evaluate (11) since τ and s_m for highly doped semiconductors have not been measured. If we assume, however, that d is sufficiently small that d/τ is dominated by s_m , and assume further that s_m is of the same order of magnitude as surface recombination values already obtained (say 1,000 cm/sec, to be conservative), then s may be of the order of 1 cm/sec when $\sigma_b/\sigma_s = 10^{-3}$. Eq. 10 is easier to consider because the term $\sigma_s L_m$ is the familiar one which enters into the expression for the efficiency of an emitter. Previous work suggests that $\sigma_s L_m$ has the value of about 1.6 mhos in alloy junctions in germanium.² Now, if σ_b is assumed to be 0.2 mhos/cm and $D_p \approx 6$ cm²/sec (consistent with heavily doped germanium), then $s = 0.75$ cm/sec. This is a very hopeful result. There is no reason

to prefer a p -type layer to an n -type layer of equal conductivity as far as values of s are concerned. However, the latter would be preferable in many cases for other reasons (e.g., there would be no tendency to produce surface short-circuit paths from emitter to collector of a transistor).

Metal Films

The above analysis only applies when the surface layer is also a semiconductor and has the same energy gap as the bulk. To add to the picture, surface layers with different energy gaps should be discussed. In particular, the case of a metal film deserves attention.

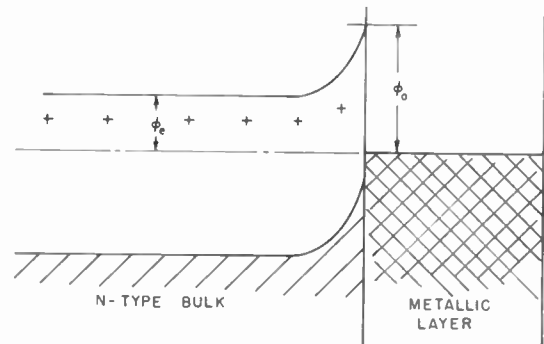


Fig. 3—Detailed energy level diagram for a metallic layer on an n -type base.

An expression for s can be computed for the case of a metal-semiconductor contact such as illustrated in Fig. 3. The derivation is straightforward and so will not be given; the result is:

$$s \approx \frac{\sigma_b^2}{\sigma_i^2} \frac{(1+b)^2}{b^2} \frac{\bar{c}}{4} e^{-q(\phi_0 - \phi_s)/kT}. \quad (12)$$

Here, σ_i is the conductivity of the intrinsic semiconductor, \bar{c} is mean thermal velocity ($\approx 10^7$ cm/sec at 300 degrees K). ϕ_0 and ϕ_s are labeled in the figure, and $b = \mu_n/\mu_p$. This equation is derived assuming diffusion flow for holes and "diode" flow for electrons crossing the barrier.^{10,11}

Evaluation of (12) requires a knowledge of $(\phi_0 - \phi_s)$. [Schwartz and Walsh have estimated this quantity to be 0.3 electron volt for 5 ohm-cm germanium in connection with the surface-barrier transistor.¹¹] If N_d for 5 ohm-cm n -type germanium and 0.3 electron volt for $(\phi_0 - \phi_s)$ are substituted into (12), a value for s of 3,500 may be computed. This calculated value is of the same order as the surface recombination velocity measured for a copper-plated surface (7,400 cm/sec.¹). Even if (12) is not

¹⁰ W. E. Bradley, "Principles of the surface-barrier transistor," *Proc. I.R.E.*, vol. 41, pp. 1702-1706; December, 1953.

¹¹ R. F. Schwarz and J. F. Walsh, "Properties of metal to semiconductor contacts," *Proc. I.R.E.*, vol. 41, pp. 1715-1720; December, 1953.

strictly applicable, due to inversion layer effects, at present the evidence suggests that metal films will give values of surface recombination velocity which are much larger than those that are easily attained with chemical treatments.

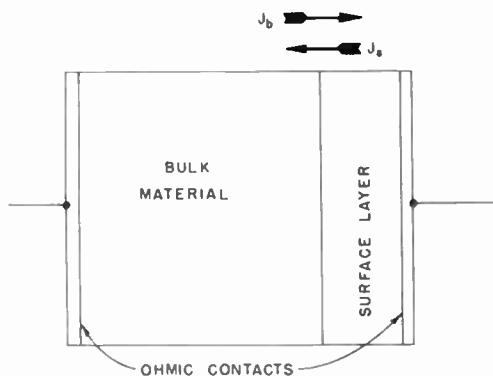


Fig. 4—Hypothetical diode formed by connection to the bulk material and to the surface layer. The arrows show the direction of carrier flow when the diode is biased in the forward direction.

Relation to Emitter Efficiency

Consider Fig. 4 which shows a diode made by connection to the semiconductor bulk and to the surface layer. The currents J_b and J_s labeled according to the direction of carrier flow when the diode is biased in the forward direction. Their ratio defines the emitter efficiency γ of a transistor which might be made using the surface layer as an emitter and the bulk material as the base region. The approximate relation is

$$\frac{1 - \gamma}{\gamma} = \frac{J_b}{J_s} \cdot \frac{W_b}{L_p}, \quad (13)$$

where W_b is the emitter-collector spacing of the transistor.¹² For a $p-n-p$ transistor,

$$\frac{1 - \gamma}{\gamma} = \frac{\sigma_b W_b}{\sigma_s L_p},$$

where L_p is the hole diffusion length in the emitter. This is also the result of multiplying (10) by W_b/D_p , since L_p and L_m are identical.

$$\frac{1 - \gamma}{\gamma} = s \frac{W_b}{D_p}. \quad (14)$$

This relation is actually quite general. It applies to the metallic film of the surface-barrier transistor and even for an n^+ surface layer and what might be called n^+-n-n^+ transistor. In fact, (14) may be derived directly from the diode equations in such a way that the relation between s and γ emerges as fundamental. While such a procedure suggests a more direct way of deriving s for different surface layers, it is less easy to consider in

physical terms than the preceding development. The connection between s and γ suggests that one may be evaluated from a measurement of the other.

It is worth pointing out that s is dependent on injection level in the same way as γ . Thus, the rate of surface recombination would be expected to go as the square of p_n when p_n becomes large compared to N_d . This is not usually observed in chemically treated samples which probably indicates that none of the simple models considered here may be applied to such surfaces.

EXPERIMENTAL OBSERVATIONS

The rather small values of s predicted by theory for a floating surface layer of opposite conductivity type should be directly measurable. A series of experiments which will be described in detail have been made to test this theory for the case of a very p -type layer on an n -type sample. The expected reduction in s is not obtained. It is by examination of these negative results that an opposing mechanism is revealed which contributes to our understanding of the floating barrier.

Lifetime Measurements

Diffusion Method: Surface recombination velocity is usually obtained by measuring the actual (often called "effective") lifetime of minority carriers in a sample of known volume lifetime and dimensions. Then s is obtained by computation for the specific geometry. For the case of a rectangular bar of cross-sectional dimensions B and C , Shockley gives the first order formula:³

$$1/\tau_m = 1/\tau_v + \nu_s, \quad (15)$$

where

$$\nu_s = 2s \left(\frac{1}{B} + \frac{1}{C} \right)$$

holds for small s .

Lifetime can be measured by the diffusion method in which the minority carrier density is measured as a function of distance from a line of light. If a thin bar (thickness \ll width) is used, the measured lifetime will be largely surface controlled, thus providing a simple measure of s on the large surfaces. If the thickness of the bar is W , then $s = W/2 \tau_m$ when s is the same on both surfaces.

A sample (see Fig. 5) was used. One side of the bar carried a large indium alloy junction. This was the surface whose effective s was to be determined. The point contact and line of light were on the opposite side of the bar. Measurements could also be made on a region of the bar which did not have the floating junction layer. One might expect that the region including the junction, which, theory predicts, will have one side where $s \approx 0$, would show a longer lifetime by a factor of about 2, compared to the rest of the sample. This is because one-half of the surface sink is effectively removed. The situation is the same as if the bar had been in-

¹² W. Shockley, M. Sparks, and G. K. Teal, "P-N junction transistors," *Phys. Rev.*, vol. 83, pp. 151-162; July, 1951.

creased to twice its thickness, thereby decreasing γ , by a factor two. Since the bar was known to have a volume lifetime in excess of several hundred microseconds, this would result in a doubled τ_m .

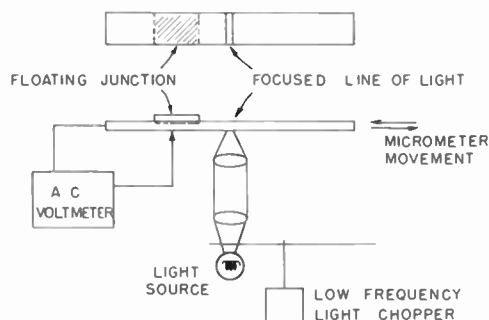


Fig. 5—Apparatus for measurement of lifetime by the diffusion method on a germanium bar carrying a floating alloy junction.

The results of such a measurement are shown in Fig. 6. The log of the open circuit probe voltage is plotted against the distance between the probe and the light line. When this spacing is less than the thickness of the bar, the slope of the curve yields a lifetime of 90 μsec . This is about the same as the value measured at probe positions far removed from the

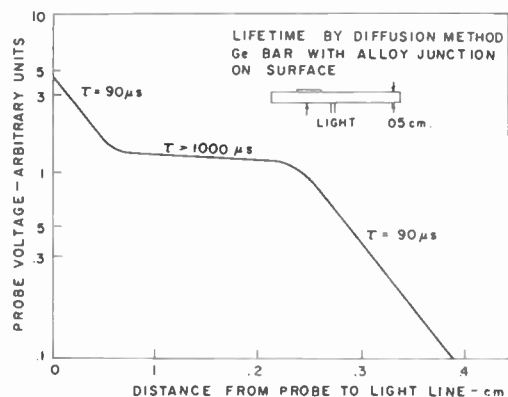


Fig. 6—Curve obtained by plotting probe voltage vs distance from probe to light line, using apparatus of Fig. 5.

floating alloy junction. As the light line is moved further than the bar thickness from the probe, the slope of the curve becomes practically zero, indicating an apparent lifetime of many thousands of microseconds. This is far in excess of the expected factor of 2. When the light line passes the edge of the floating p -layer, the curve resumes its original slope, which is characteristic, as before, of the lifetime far removed from the junction.

The apparent increase in lifetime cannot be explained on the basis of a reduced s on one surface. Additional

experiments to be described were performed on the same bar in an effort to find the reason for the discrepancy.

Pulse Photo Conductivity: Another way to determine lifetime in the bar is to measure the time decay of excess conductivity after a pulse excitation of hole-electron pairs by a short flash of light.⁹ In this measurement the conductivity variation is obtained from the voltage across a resistor in series with the sample and a bias battery. If the resistor is matched to the dark resistance of the sample, the light pulse is of low intensity so that $\Delta\sigma \ll \sigma$, and the field across the sample is sufficiently small such that the transit time of carriers due to the electric field is long compared to the decay of recombination, then

$$\frac{\Delta\sigma}{\sigma} = \text{const.} \cdot e^{-t/\tau}. \quad (16)$$

This method measures lifetime directly, instead of obtaining it through the diffusion relation $L = \sqrt{D\tau}$, as in the previous experiment.

The experimental arrangement is shown in Fig. 7. The light source is a spark discharge in air operated as a relaxation oscillator from a 5,000-volt power supply. When focused on the sample, the spark produces a line of light which was arranged perpendicular to the long axis of the sample. A micrometer screw enables the sample to be moved parallel to this axis so that the lifetime can be measured by the change in the voltage drop across a resistor connected in series with the sample and a bias battery. The measurement is made with an oscilloscope with a built-in delay line so that the entire trace can be studied. If the photoconductivity decays exponentially [as it should from (16)] the lifetime can be read directly on the oscilloscope face as the $1/e$ point.

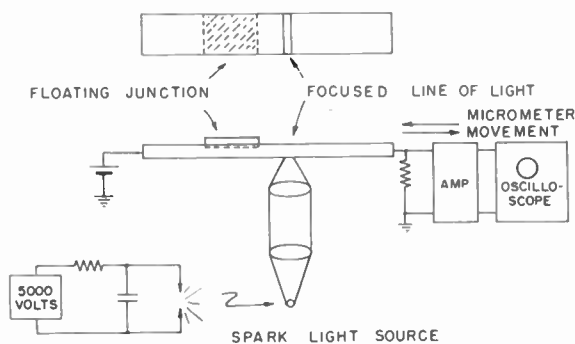


Fig. 7—Apparatus for measurement of lifetime and drift mobility by the drift method on a germanium bar carrying a floating junction.

At positions far removed from the floating junction, the decay time corresponded to 90 μsec , in agreement with the diffusion length measurement. In the floating junction region the decay departed from an exponential and could not possibly be construed as a doubling of the

lifetime. The detailed nature of this curve will be discussed later in connection with the drift-time measurement. For the present the conclusion is that the expected doubling of the decay time was not observed.

Diode Measurements: A measurement of the effective minority carrier lifetime in the base region of an alloy junction diode can be obtained by injecting minority carriers into the region with a pulse of forward current. The decay is then followed by observing the open circuit emitter voltage as a function of time.⁸ The same measurement can be made in a transistor in which the collector is allowed to float electrically. If the base wafer is thin, the effective lifetime is a measure of s . Furthermore, most of the surface recombination will take place on the surface opposite the alloy junction, provided s is the same value there as at other surfaces. Now, if we substitute a large floating collector for the surface opposite the emitter, as in an alloy transistor, and if this surface actually has a very low value of s as predicted by the junction theory already given, one would expect a marked increase in the effective lifetime. Such measurements were made on a series of transistors and diodes of essentially identical emitter-base region geometry. Both diodes and transistors gave the same value of effective lifetime within experimental error. Thus, again, the expected reduction in s was not observed.

The Feed-In Feed-Out Effect

The negative results in the above experiments require examination and modification of the theory of surface recombination velocity at a floating junction. The most logical explanation appears to be connected with the fact that the p -type germanium on the surface is a good conductor (relatively) and hence, is an equipotential. Under conditions of an applied field within the n -type bar, or when a gradient of minority carrier density exists along the interface, the floating junction assumes the dual role of emitter and collector. While the net current across the junction is zero, this need not be true at every point. In fact, the effective s may be the same as for adjacent germanium surfaces. The p -region then acts as a translator of holes rather than as a low s interface. Some additional experiments which support this view will now be described.

Floating Potential Measurements: Indirect measurements indicate that the p -layer in an alloy junction has a resistivity of the order 0.001 ohm-cm. If an electric field is maintained in a bar which carries a floating junction, and if the bar has a resistivity of the order 1 ohm-cm, the p -layer can be considered as an equipotential surface. The situation is illustrated in Fig. 8. The layer will float at some potential intermediate between the potential V_b and V_a . Part of the junction is biased in the reverse direction, collecting thermally generated holes from the n -type bar, while the rest is biased in the forward direction, injecting holes back into the bar closer to the negative electrode. The net current across the junction is zero, since the p -layer is floating. No

holes are lost in the process. Because the hole concentration in the p -type layer is much higher than the electron concentration in the n -type bar, most of the current crossing the junction in either direction will consist of holes. The hole currents will furthermore be small compared to the main electron current in the n -type bar. Hence it can be assumed that the electric field in the bar is not seriously disturbed by presence of the junction.

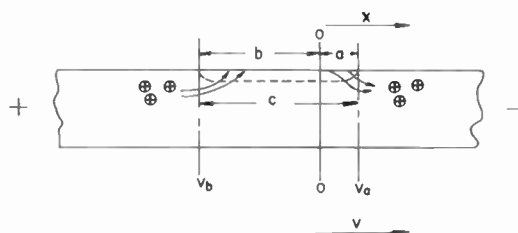


Fig. 8—Diagram for the calculation of the floating potential of the alloy junction.

The floating potential of the junction may be calculated by equating the integrated current for the forward biased region to that for the reverse biased region. One may take the position at which the potential in the germanium is equal to the p -layer potential as the zero of x and v . If the total potential drop $V_a - V_b$ is called V and the total length of the p region $a - b$ is called c , then $v = (x/c)V$. Then

$$-\int_{-b}^0 (e^{qVx/kTc} - 1)dx = \int_0^a (e^{qVx/kTc} - 1)dx.$$

One can then solve for the floating potential V_f with respect to V_a ($V_f = (a/c)V$):

$$V_f = \frac{kT'}{q} \ln \left[\frac{qV/kT'}{1 - e^{-qV/kT'}} \right]. \quad (17)$$

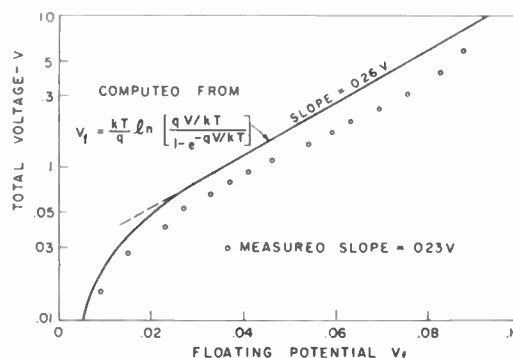


Fig. 9—Results of measurement of the floating potential comparison with (17). Measurement made at 300 degrees K.

Fig. 9 shows the result of a test of this equation. Both the floating potential and the total potential drop were measured by means of probes and a high-impedance millivoltmeter. Inasmuch as there are no adjustable constants in this equation, the agreement is considered

as satisfactory evidence of the feed-in feed-out phenomenon.

Mobility or Drift Time Measurements: The previous experiment suggests that holes can be translated towards a negative electrode through a p -type surface layer. It is presumed that this process occurs almost instantaneously, i.e., if a hole is fed into the reverse biased junction region another hole is immediately emitted at the forward biased region. This process occurs much faster ($\sim 10^{-11}$ sec) than ordinary minority carrier drift times ($\sim 10^{-5}$ sec). Thus, if the hole drift time in an electric field is measured on a bar carrying a floating junction, an artificially short drift time should be observed due to the bypassing action of the floating junction. The same pulse lifetime equipment described in connection with lifetime measurements was used to test this conjecture except that the field in the bar was increased to the point at which the drift time was shorter than the lifetime. The light pulse which acts as the hole source was focused at the center of the bar and the drift time measured for holes drifting first toward the end without the floating junction, and then, by reversal of the polarity of the electric field, for holes drifting through the part of the bar carrying the junction. Fig. 10 shows the two cases. From the dimensions of the bar and the length of the junction region, the drift time should have been halved for the second case. The measured ratio was 0.47.

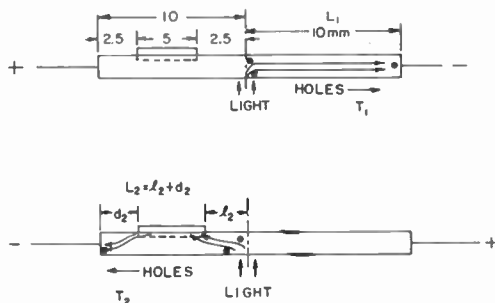


Fig. 10—Measurement of drift time in the germanium bar when (a) holes are confined to the bar volume and (b) when holes bypass the bar through the floating junction region. $T_2/T_1 = 0.47$; $L_2/L_1 = 0.50$.

Additional drift experiments using a light line movable along the bar have all confirmed the picture of instantaneous hole translation through the p -region. One point which was of special interest was the explanation of an unusual spike on the decay curve during pulse lifetime measurements in the vicinity of the floating junction. This is illustrated in Fig. 11. Initially (Region I), a sharp spike occurs, beginning immediately after the light injection pulse. Then the carriers decay exponentially and simultaneously drift in the electric field, as in Region II. Finally, carriers reach the end of the sample and the conductivity falls, as in Region III.

The pulse of light generates hole-electron pairs. In spite of the movement of the holes in an electric field, the hole charge remains neutralized by electrons in order to insure space charge neutrality. Since the translating action of the junction operates on holes only, the holes are re-emitted into the germanium without their accompanying neutralizing electrons. The spike is a consequence of redistribution of electrons which takes place in order to re-establish space charge neutrality. Part of this electron current comes up from ground through the load resistance, generating the spike voltage. The size of the spike depends on the resistance of the n -type germanium between the point of absorption and emission of the holes and the load resistance. The width depends on the capacitance in the external circuit.

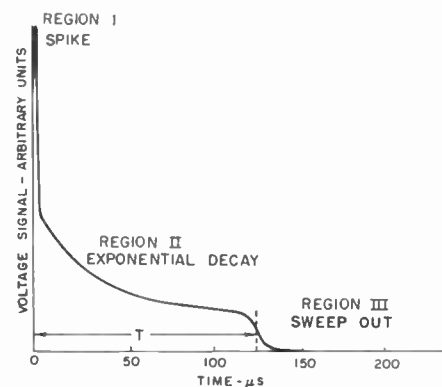


Fig. 11—Typical oscilloscope pattern during drift time measurement.

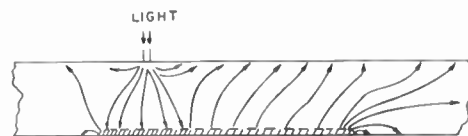


Fig. 12—Sketch of probable hole flow pattern during measurement of lifetime by diffusion method of Fig. 5.

Interpretation of Diffusion Measurements

The results of the diffusion measurement of surface recombination (See Diffusion Method, above) can be understood on the basis of hole translation. The hole flow pattern is of the type sketched in Fig. 12. Holes feed out from the source at the focused line of light. Some recombine at the surface adjacent to the source, but a larger number feed into the floating junction a distance W away. These are re-emitted about uniformly over the rest of the junction, providing a region along the upper surface in which the hole density is constant with distance along the bar. Thus the measured probe voltage becomes independent of distance after an initial dis-

tance W , as in Fig. 6, until the end of the floating junction is reached. Then there is no feed-in feed-out effect and the hole density falls off with distance in just the way to be expected for a surface controlled filament lifetime, i.e.,

$$\nu_s = \frac{2s}{W} \quad \text{or} \quad \tau_m = \frac{W}{2s}$$

Interpretation of Diode Measurements

The same hole translation effect can take place in the measurement of effective lifetime in alloy transistor structures. After the emitter injection pulse is over, holes feed into the floating collector opposite the emitter, where the hole density is high. They immediately feed out again near the edge of the junction and so are lost to the adjacent surface. Hence this surface becomes the controlling sink, just as in the case of a diode structure without any collector. The fact that experimental agreement between diode and transistor effective lifetimes are about equal may be thus explained.

CONCLUSIONS

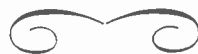
The fundamental reason for the failure of the junction analysis to predict the observed result is the assumption of a one-dimensional model. In this model the steady state hole and electron currents balance at every point along the boundary. In practice, that is for a three-dimensional case, the total currents balance, as required for steady state conditions, but they do not necessarily balance on a per unit area basis. If lateral gradients are not the same in the surface layer and in the bulk, circulating currents can exist, which destroy the effectiveness of the layer in reducing the surface recombination. Unfortunately, the requirement for low effective surface recombination is just that $\sigma_s \gg \sigma_b$, which implies that lateral conduction in the surface layer is large. This type of surface junction is therefore not suitable for reducing surface recombination in practical cases.

If it is not recognized, this effect can cause difficulty in some measurements. The measurement of lifetime in p -type germanium and in silicon by diffusion methods often yields results which are clearly too high. It is

thought that this is due to the presence of layers of opposite conductivity type (inversion layer) on the surface, resulting from certain etches. While careful etching apparently removes this layer from n -type germanium, we have found no certain method of removing it from p -type germanium¹³ or from n - and p -type silicon. The surface layer probably acts in much the same way as the floating junction in the above experiments; minority carriers from the base semiconductor bias the measuring probe through the inversion layer, thus making the probe voltage less dependent on the distance between probe and the source of hole-electron pairs. The fact that the inversion layer is on the same surface as the probe and source, rather than on the opposite side of a thin bar as in Figs. 5 and 6 does not materially effect the argument. The pulse method of Fig. 7 (See Pulse Photo Conductivity) is dependent only on the number of minority carriers actually within the bar and is independent of their special distribution, provided that the drift field is small enough to prevent sweep-out. Hence lifetime measurements on these materials are best made by the pulse drift method.

Under certain conditions, the interpretation of effective lifetime in completed transistors⁸ can be affected by hole translation. If the collector does not penetrate deeply into the base wafer, the effective lifetime is simply related to the surface recombination velocity by $\tau_s = W/2s$, where W is the wafer thickness. If the collector does penetrate far into the wafer, the feed-out of holes from the edge of the collector to the adjacent free germanium surface effectively translates the recombining surface on the collector side closer to the recombining surface on the emitter side. Thus the effective thickness of the wafer is reduced. The measured effective lifetime is less in this case. When the transistor is in use as a device, however, the hole translation effect is not operative since the collector is biased in the reverse direction rather than floating as in the test measurement. The proper value of W can be determined empirically.

¹³ Adsorption of H_2O vapor plays a part in creating the inversion layer on p -type germanium. See H. Christiansen, "Surface conduction channel phenomena in germanium," *Proc. I.R.E.*, vol. 42, pp. 1371-1376; September, 1954; also A. L. McWhorter and R. H. Kingston, *ibid.*, pp. 1376-1380.



A Chart for Analyzing Transmission-Line Filters from Input Impedance Characteristics*

HARVEL N. DAWIRS†, ASSOCIATE, IRE

Summary—Filter calculations become difficult when network elements consist of transmission-line sections, since transcendental equations are involved. It is the purpose of this paper to describe the application of familiar impedance methods and Smith chart^{1,2} techniques which simplify many of these calculations. A chart is developed by means of which the most important characteristics of a filter may be read directly from a conventional input impedance curve plotted on a Smith chart. The principles involved in these methods are equally valid for all lossless filters consisting of identical and symmetrical sections, but are particularly well-suited for use with transmission-line circuits.

INTRODUCTION

IT IS OFTEN necessary to construct filters in the frequency range where transmission-line circuits are used. In this range the elements of the filters must be sections of transmission line. The design of such filters is difficult because the usual design equations become transcendental and are difficult to solve. It is common practice, however, to manipulate such expressions arising in connection with transmission-line circuits by means of the Smith impedance chart.^{1,2} It is the purpose of this paper to show how these Smith chart techniques may be used to facilitate many important calculations encountered in the design of lossless transmission-line filters which consist of identical and symmetrical sections.

Since only lossless filters consisting of identical and symmetrical sections are considered, the input impedance of the filter is given by the expression³

$$z_i = \frac{z_l \cosh \gamma_n + z_c \sinh \gamma_n}{z_l \sinh \gamma_n + z_c \cosh \gamma_n}, \quad (1)$$

where

- z_c is the characteristic impedance of the filter,
- z_l is the impedance of the terminating load,
- $\gamma_n = \alpha_n + j\beta_n$ is the propagation constant of the filter,

and

n is the number of sections in the filter.

This expression can be put into the form

$$z_n = \frac{z_0 + \tanh \gamma_n}{1 + z_0 \tanh \gamma_n}, \quad (2)$$

where

$z_n = r_n + jx_n$ is the input impedance of the filter normalized to z_c ,

and

$$z_0 = r_0 + jx_0 = z_l/z_c. \quad (3)$$

In all further discussion let z_l be an arbitrary (but fixed) real impedance and define:

$$Z_n = R_n + jX_n = z_n/z_0, \quad (4)$$

and

$$Z_c = z_c/z_l. \quad (5)$$

Note that Z_n and Z_c are, respectively, the input impedance and the characteristic impedance of the filter, both normalized to z_l and that $Z_c = 1/z_0$. With these definitions the expression

$$Z_c = \sqrt{\frac{R_n^2 + X_n^2 - R_n}{R_n - 1}} \quad (6)$$

follows from (2) for both the pass and rejection bands.

Since, for transmission-line circuits Z_n is easily obtained by well-known procedures, (6) provides a convenient means of determining the characteristic impedance of a transmission line filter.

SIGNIFICANT IMPEDANCE RELATIONS ON THE SMITH CHART

Now consider the significance of (6) in relation to the Smith chart and to the input impedances of a filter. (It is assumed throughout the remainder of this paper that all filter input impedances are obtained with the filter terminated in z_l and that these are normalized to z_l and plotted as points or a curve on the Smith chart.)

Eq. (6) indicates that cutoff occurs when $R_n = 1$ ($Z_c = \infty$), or when $R_n = R_n^2 + X_n^2$ ($Z_c = 0$). These relations are equations of two easily identified circles on the Smith chart which will be called "CUT-OFF CIRCLES." (See Fig. 1, facing page.) At a cut-off frequency the input impedance of a filter will fall on one of these circles, and conversely, the cut-off circles will intersect the input impedance curve of a filter at any cut-off frequency.

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† Antenna Lab., Dept. Elect. Engrg., Ohio State Univ., Columbus, Ohio.

¹ P. H. Smith, "A transmission line calculator," *Electronics*, vol. 12, pp. 29-31; January, 1939.

² P. H. Smith, "An improved transmission line calculator," *Electronics*, vol. 17, pp. 130-133, 318-325; January, 1944.

³ J. J. Karakash, "Transmission Lines and Filter Networks," The Macmillan Company, New York, N. Y., p. 169; 1950.

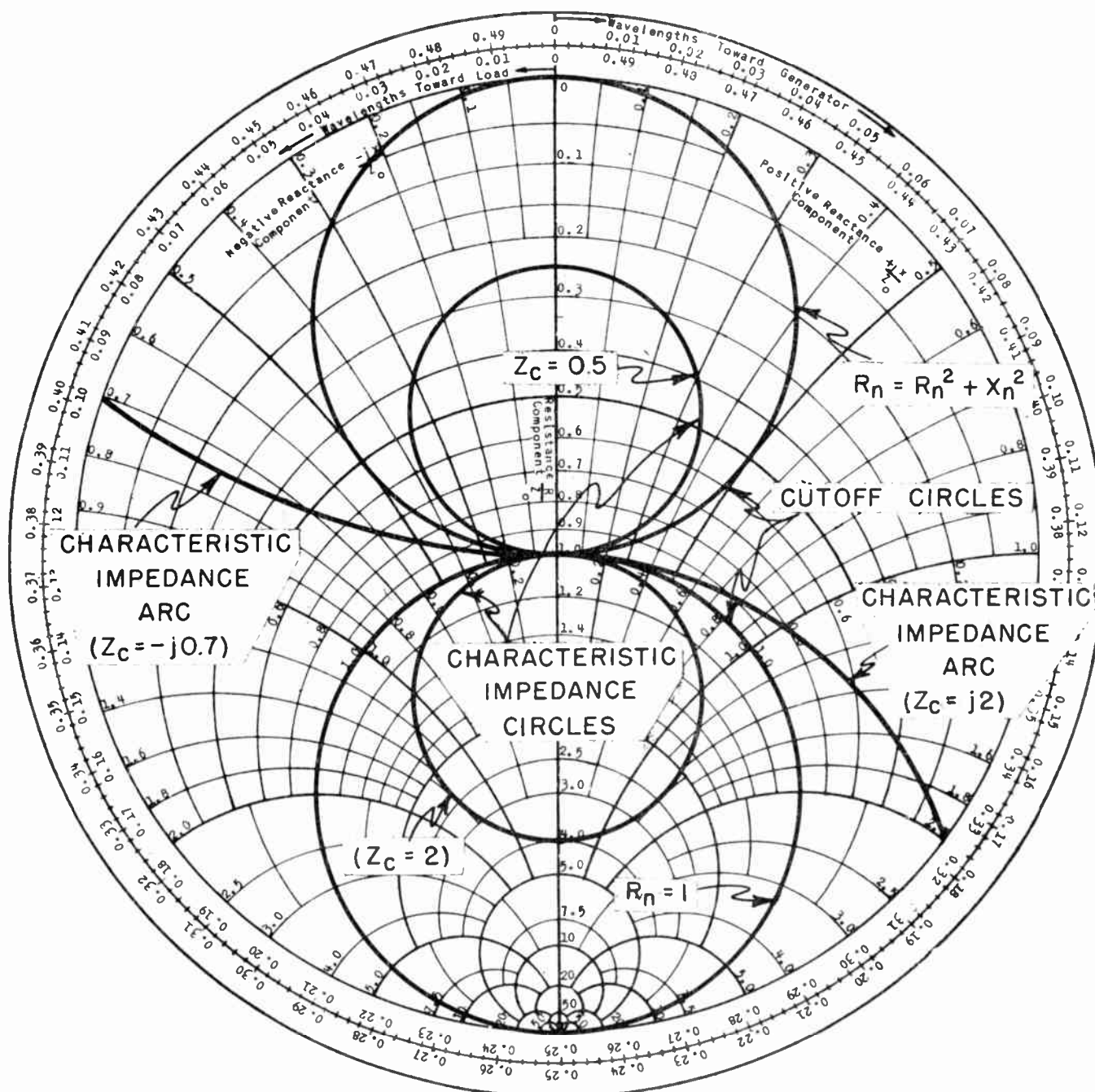


Fig. 1

Further consideration of (6) shows that Z_c is real whenever Z_n lies within one of the cut-off circles and is imaginary for all Z_n outside of the cut-off circles. Thus the cut-off frequencies may be determined and the pass and rejection bands identified directly from the input impedance curve of a filter by means of the cut-off circles.

For a fixed value of Z_c in the pass band, (Z_c real and positive), (6) is the equation of a circle on the Smith chart, which is tangent to the cut-off circles at the cen-

ter of the chart and lies wholly inside one of them as shown in Fig. 1. There is one of these circles, which will be called a "CHARACTERISTIC IMPEDANCE CIRCLE," associated with each positive real value of Z_c . This circle will intersect the real axis at the point

$$R_n = Z_c^2, \quad (7)$$

as determined by setting $X_n = 0$ in (6). Eq. (6) indicates that the input impedance Z_n of a filter at any given frequency in the pass band must fall on the impedance

circle corresponding to the characteristic impedance of the filter at that frequency. Conversely, a characteristic impedance circle constructed on a Smith chart [this may be done quite easily using (7)], will intersect the input impedance curve of a filter at the frequencies for which the characteristic impedance of the filter is equal to the value associated with the intersecting circle.

Note that if the input impedance of a filter is known at any particular frequency, the characteristic impedance of the filter at that frequency may be found by constructing the unique characteristic impedance circle through the known impedance point and calculating Z_c [by (7)] from the point at which the circle intersects the real axis.

A number of typical characteristic impedance circles may be constructed as auxiliary co-ordinates on a Smith chart and used to determine characteristic impedances of a filter, in its pass bands, directly from an input impedance curve of the filter plotted on the same chart.

Now for a fixed value of Z_c in the rejection band (Z_c imaginary), (6) is the equation of an arc on the Smith chart which is tangent to the cut-off circles (and also to the characteristic impedance circles), at the center (see Fig. 1), and terminates on the rim of the chart. There is a unique arc, which we shall call a, "CHARACTERISTIC IMPEDANCE ARC," associated with each imaginary value of Z_c , terminating at the point

$$X_n = -jZ_c \quad (8)$$

on the rim of the chart (since $Z_c = \sqrt{-X_n^2} = jX_n$ when $R_n = 0$). Thus, (6) implies that the input impedance of a filter at any given frequency in the rejection band must fall on the impedance arc which corresponds to the characteristic impedance of the filter at that frequency. Hence, if the input impedance of the filter is known at any given frequency in the rejection band, the characteristic impedance of the filter at that frequency may be determined by constructing the unique characteristic impedance arc through the known impedance point (by means of a compass), and noting the point at which it intersects the chart rim. The value of Z_c follows by (8).

Conversely (6) implies that a characteristic impedance arc will intersect the input impedance curve of a filter at a frequency for which the characteristic impedance of the filter is equal to that associated with the intersecting arc. A number of typical characteristic impedance arcs may be constructed as auxiliary co-ordinates on a Smith chart [(8) may be conveniently used for this purpose], and used to determine the characteristic impedances of the filter at rejection band frequencies by means of the intersections of the arcs with the input impedance curve of the filter.

FILTER ANALYSIS CHART

Fig. 2, facing page, is a chart, based upon principles just considered, which may be used to determine many

important properties of filters directly from a plot of their input impedances. The co-ordinates of this chart are normalized characteristic impedance and propagation constant. The characteristic impedance co-ordinates consist of the cut-off circles, the characteristic impedance circles and the characteristic impedance arcs discussed previously. The propagation constant co-ordinates are obtained by calculating typical values [by means of (2)] and plotting these on the Smith chart. Note that the co-ordinates of the filter analysis chart are considered to be superimposed upon a Smith chart even though the usual impedance co-ordinates are not shown. Desired impedance points are located by means of a calibrated cursor and the scale around the outside rim.

In the pass band the propagation constant, which is imaginary and hence consists only of the phase constant, is scaled in wavelengths for convenience in use with transmission lines. In the rejection band the propagation constant, which is real and hence consists only of the attenuation constant, is scaled in decibels.

To make use of the chart in analyzing a filter, the filter is terminated in a real impedance z_L and the input impedance measured as a function of frequency over the range of interest. These impedances are then normalized to z_L and plotted on the analysis chart in the same manner as on a Smith chart. (See Fig. 3, page 440). Since these measurements are usually made and plotted in terms of phase shift and voltage standing-wave ratio (or voltage-reflection coefficient), they may be plotted directly on the analysis chart just as conveniently as on the Smith chart itself if z_L is chosen to be equal to the characteristic impedance of the slotted line (or directional coupler), used in making the measurements.

Now the cut-off frequencies may be determined and the rejection and pass bands identified by means of the cut-off circles as described previously. The normalized characteristic impedance and the propagation constant at any frequency may be read directly from the co-ordinates of the corresponding input impedance point. The characteristic impedance read from the co-ordinates is normalized to z_L . (That is $Z_c = z_c/z_L$.)

In addition the attenuation of the filter (which in this case is equal to its reflection loss), at any given frequency may be determined by reading the reflection loss of the corresponding input impedance point by means of a calibrated cursor,² or may be calculated by means of the formula²

$$\text{db} = 10 \log (1 - |\rho|^2), \quad (9)$$

where ρ is the reflection coefficient (either current or voltage) of the input impedance point.

Note that:

1. The input impedance curve is obtained by routine procedures, using standard equipment and techniques;
2. Only one measurement is required at any one fre-

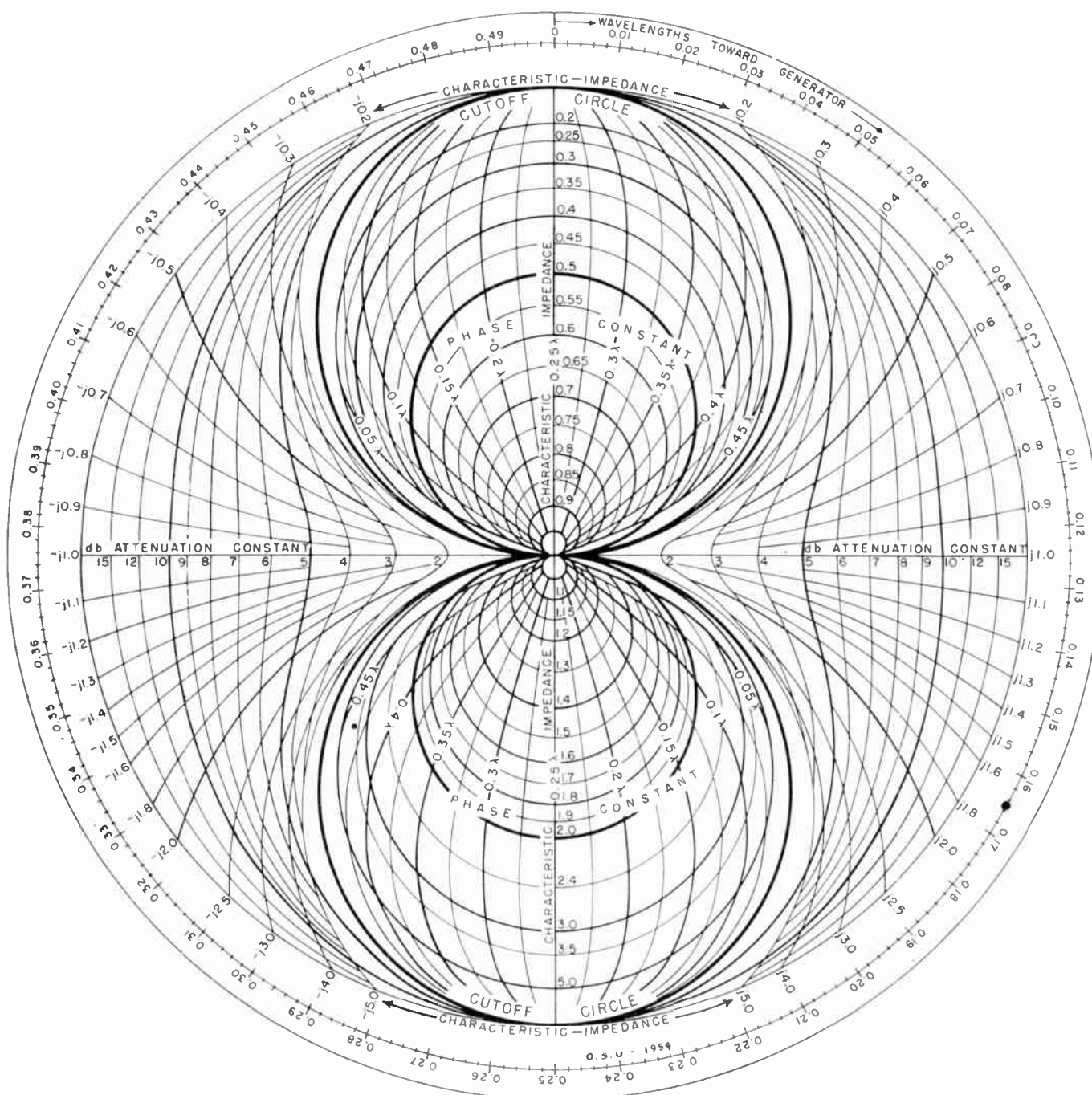


Fig. 2

quency (as against two required for the open- and short-circuit method); and

3. Calculations are eliminated by reading values directly from the chart co-ordinates (or a cursor).

The properties obtained by such procedures are:

1. Cut-off frequencies;
2. Pass bands;
3. Rejection bands;
4. Characteristic impedances;
5. Propagation constants; and
6. Attenuation.

FURTHER APPLICATIONS OF THE CHART⁴

If a proposed filter is to be analyzed from calculated data, the input impedance of only a single section is necessary, since the cut-off frequencies, the rejection and pass bands, and the characteristic impedances of a filter are the same as for the component sections. Hence all these properties may be determined directly from

⁴ H. N. Dawirs and F. K. Damon, "Application of the Ohio State University Filter Analysis Chart," presented at *NEC*, Chicago, Ill.; October 5, 1954.

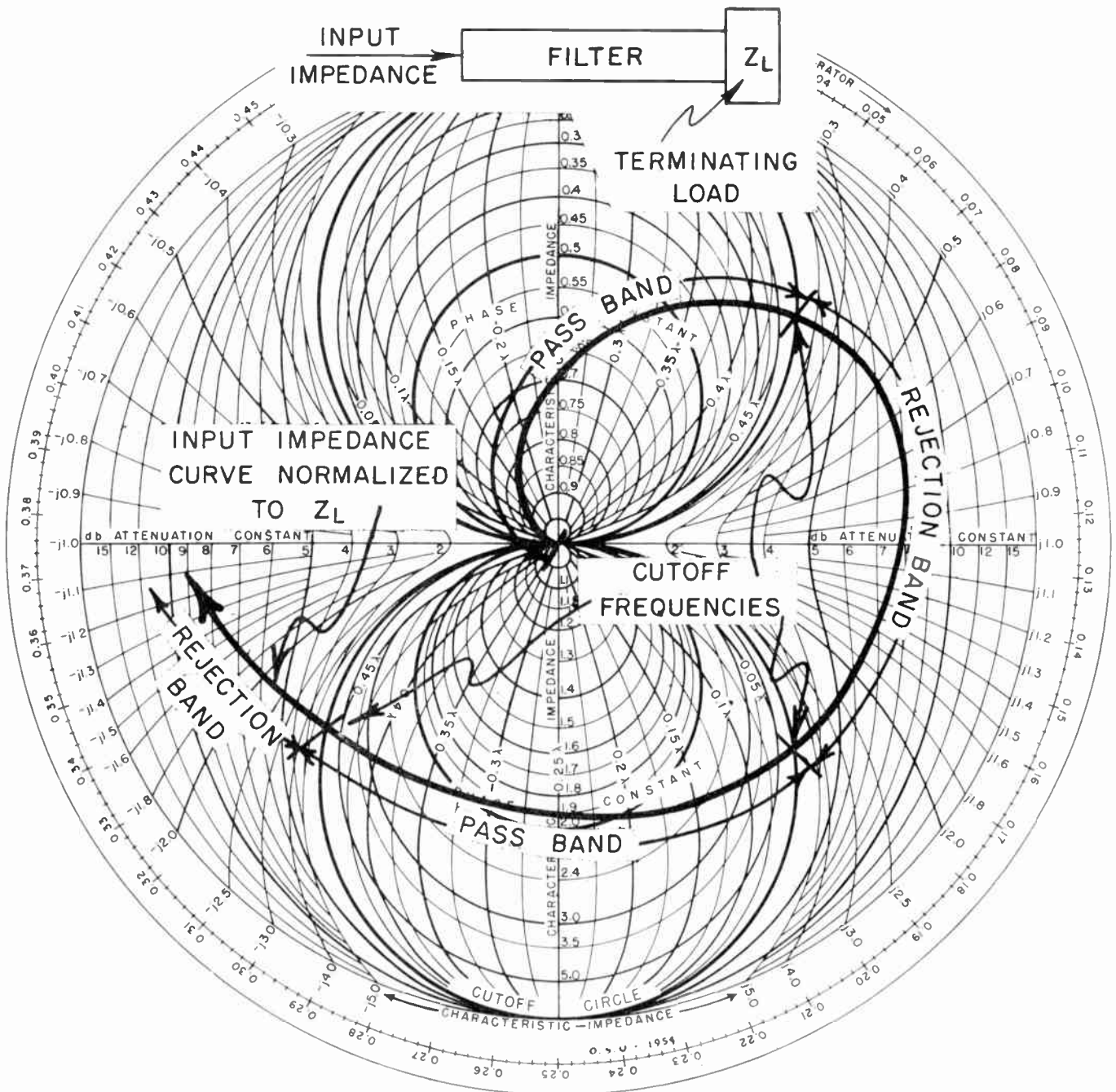


Fig. 3

the input impedance curve of a single section. The propagation constant $\gamma_n = n\gamma_1$, of an n -section filter is easily determined from the propagation constant γ_1 of a single section as read from the chart.

The value of the terminating impedance may often be chosen to simplify the calculation of the input impedance of the single section. Consider, for example, a filter section consisting of a reactance located at the midpoint of a transformer section of transmission line as shown in Fig. 4. If Z_L is chosen to be equal to Z_0 , the normalized impedance looking towards the reactance gap at the

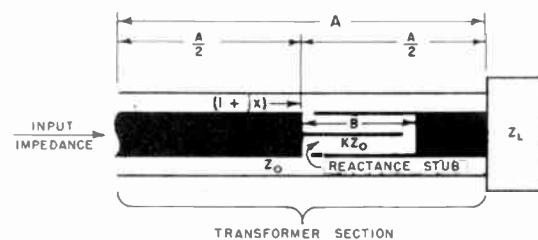


Fig. 4

center will be simply $1 + jX$, where the normalized reactance of the series stub is determined by usual Smith

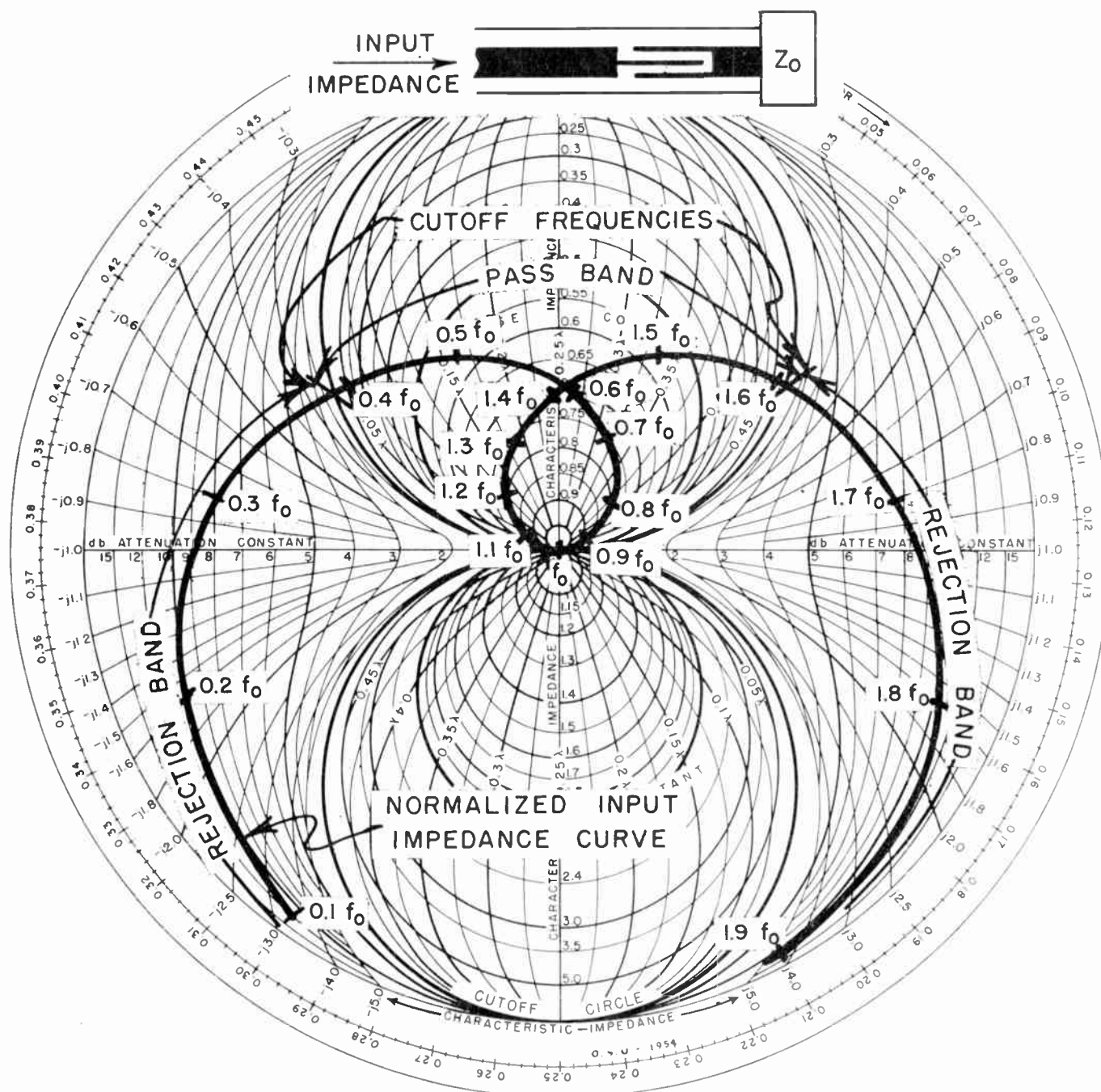


Fig. 5

chart methods. The impedance $1+jX$ is then transformed through the remaining half of the transformer section (a simple task on the Smith chart) to obtain the normalized input impedance curve of the section. Fig. 5 shows a typical input impedance curve for a filter section of this type.

It is interesting to observe the effect on the input impedance of a filter as a result of adding sections. Since, at any given frequency, the characteristic imped-

ance of a filter is independent of the number of sections, the input impedance point Z_n , must always lie on the impedance co-ordinate corresponding to the characteristic impedance of the filter at that frequency. However the propagation constant of the filter at a given frequency ($\gamma_n = n\gamma_1$), is a function of the number of sections.

At a particular frequency in the pass band, for example, the input impedance point is confined to a spe-

cific characteristic impedance circle, but progresses around and around this circle as sections are added. At other frequencies in the pass band the corresponding input impedance points progress around other impedance circles and at different rates as sections are added. Note that as an input impedance point progresses around its impedance circle the attenuation (or reflection loss), oscillates between zero when the impedance point is at the center of the chart, and a maximum when the point is at $R = \sqrt{Z_c}$. The maximum attenuation at each frequency in the pass band can be determined from the corresponding impedance circle by means of a cursor or (9) and plotted as shown in Fig. 6. The maximum attenuation curve in Fig. 6 was obtained from the input impedance curve in Fig. 5 for the filter section shown in Fig. 4. This curve is characteristic of a filter section and is a property of a filter composed of such sections, which is independent of the number of sections involved. The attenuation of a filter will oscillate between zero and the maximum attenuation curve in the pass band as a function of both the frequency and the number of sections, but can never exceed the values indicated by the curve.

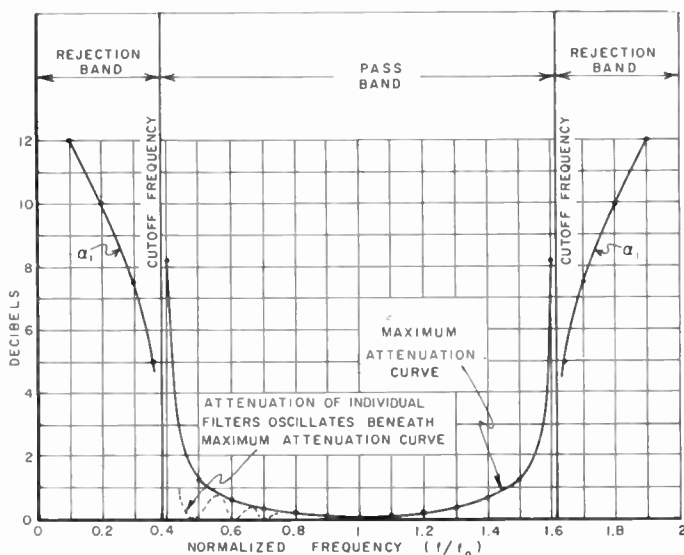


Fig. 6

At any particular frequency in the rejection band the input impedance point must remain on the corresponding characteristic impedance arc, but progresses outward along it as sections are added, approaching the rim as a limit. Thus the reflection loss, and hence the attenuation of the filter increases as sections are added. In fact the db attenuation of the filter at a given frequency in the rejection band becomes nearly a linear function of the number of sections. That is

$$\text{db} \cong |A + (n - k)\alpha_1|, \quad (10)$$

where A is the attenuation due to the first k sections. This becomes more nearly true as k becomes large but may often be used for making useful approximations,

even when k is small. For example

$$\text{db} \cong (n - 1)\alpha_1 \quad (11)$$

is usually a reasonable estimate of attenuation for practical purposes but does require a certain amount of judgment in its use.

A chart, such as is shown in Fig. 6, which may be used to evaluate the general performance of a filter section and to determine its suitability for given filter applications, may be constructed using the filter analysis chart. The cut-off frequencies of the section are indicated, the attenuation constant α_1 is plotted over the rejection band and the maximum attenuation curve discussed previously is plotted over the pass band. The cut-off frequencies and the rejection and pass band frequencies of a filter constructed of these sections are shown, the range over which the pass band attenuation of such a filter will be less than a specified value (regardless of the number of sections) may be determined from the maximum attenuation curve, and the number of sections required to obtain the desired attenuation at specified frequencies or over given ranges in the rejection band may be estimated from the curve of α_1 . Thus, such a chart is of considerable practical value in the design of transmission-line filters.

In addition to the above analysis and design application, the filter analysis chart may be used as an aid in the synthesis of some types of filters. As an example of this consider the filter section previously discussed and shown in Fig. 4. Examination of the input impedance curve shown in Fig. 5 indicates that the characteristic impedance of the filter will be equal to the characteristic impedance of the transformer section at a frequency which we shall call f_0 . This is the frequency for which the length, B , of the reactance stub is a quarter of a wavelength long. Hence the stub appears as a short circuit across the reactance gap and the filter section consists only of the transformer section of transmission line. Thus the characteristic impedance of the transformer section and the length of the reactance stub can be chosen so that the filter will match into a given load at a given frequency.

Further examination of the input impedance curve shown in Fig. 5 indicates that there exists a cut-off frequency, which we shall call f_c , which is below f_0 . Now the filter analysis chart may be used to synthesize a filter section of the type being considered which will match into a given load at a specified frequency and will have a given cut-off frequency f_c . To do so the characteristic impedance of the transformer section is chosen to be equal to the impedance of the given load and the length B of the reactance stub is chosen to be a quarter of a wavelength long at the specified frequency. Now to obtain the desired cut-off frequency it is necessary to determine the characteristic impedance of the reactance stub and the length of the transformer. Either of these may be chosen and the other determined. A choice of

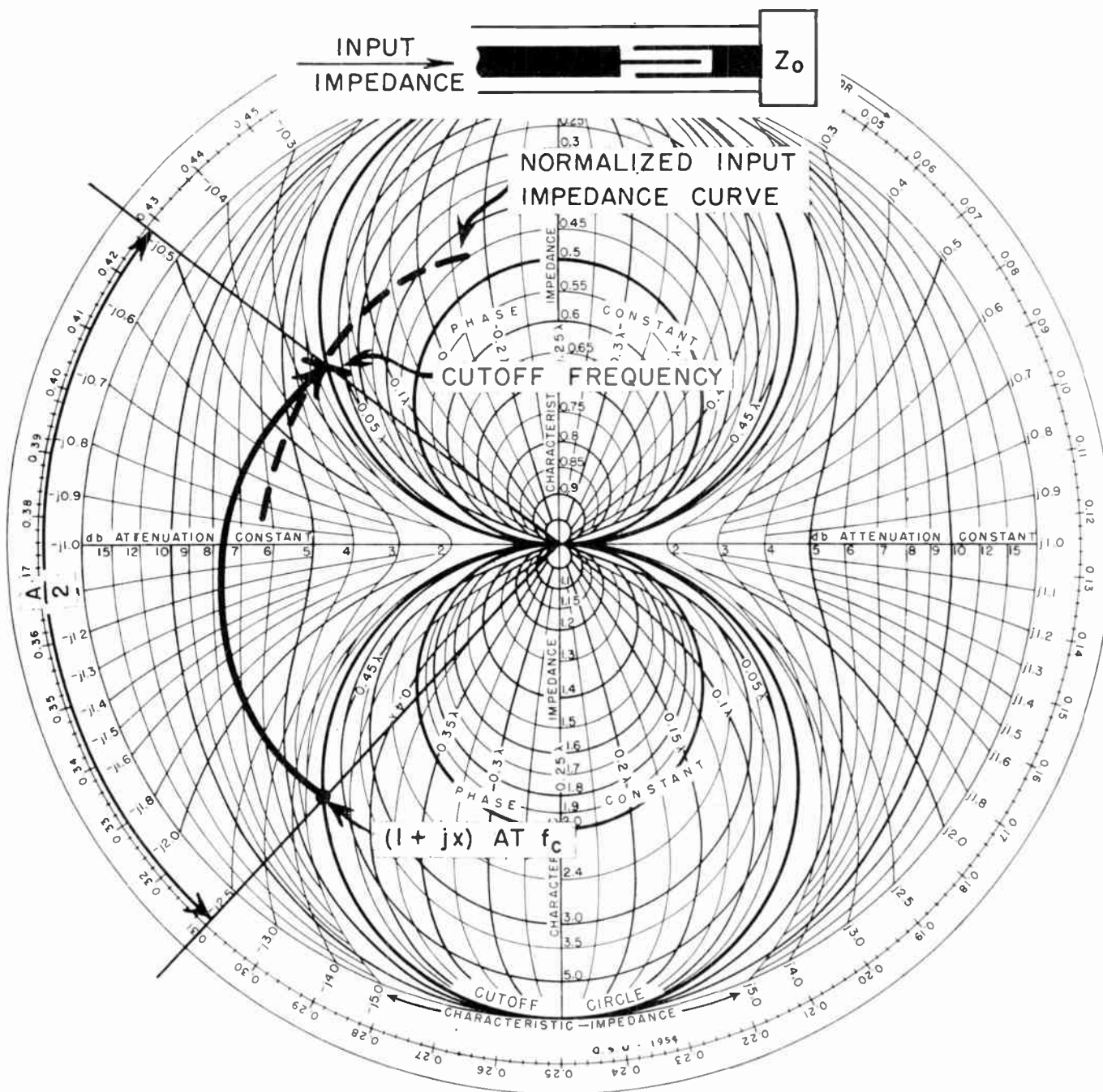


Fig. 7

a convenient value for the characteristic impedance of the reactance stub is best as an arbitrary choice of the transformer length may result in an impractical value for the characteristic impedance of the stub.

With the length and characteristic impedance of the series stub established, the input reactance, x , of the stub may be determined at the desired cut-off frequency, f_c , by ordinary Smith chart methods. It is now only necessary to determine the length of transmission line required to transform the resulting impedance point, $(1 + jX)$, to the nearest cut-off circle as shown in Fig. 7.

This is $A/2$ or one-half the total required length of the transformer section at f_c . Thus all of the values required to properly construct the filter section have been determined.

ACKNOWLEDGMENT

It would be difficult to name all of those who assisted in the preparation of this paper. However, of particular importance was the co-operation of L. A. Kail in drawing the filter analysis chart itself as well as a number of the other illustrations.

Concerning the Noise Figure of a Backward-Wave Amplifier*

T. E. EVERHART†

Summary—The noise figure of a traveling-wave amplifier has been derived as a function of circuit loss and space charge. The minimum-obtainable noise figure of the backward-wave amplifier is shown to be the same as the minimum-obtainable noise figure of the forward-wave amplifier, i.e., about 6 db. The noise figure of an ordinary backward-wave amplifier has been measured as a function of gain. The calculated noise figure checks well with the measured values.

NOISE, as related to thermionic vacuum tubes, is caused by random motion of the electrons. This random motion may be attributed to thermal kinetic energy, random emission from thermionic cathodes, random division among the various electrodes in a vacuum tube, or various other causes. The minimum input signal which can be detected by a given tube is determined by the noise of that tube; the higher the noise, the larger the minimum detectable input signal must be. For this reason, it is desirable to predict the noise one might expect from a certain type of amplifier, to devise means to reduce this noise, and to establish a minimum noise level which might be attained with a given type of tube. The noise properties of the forward-wave traveling-wave tube amplifier have been analyzed by several persons;¹⁻³ we shall examine the noise properties of a backward-wave traveling-wave tube amplifier, using the results of these previous analyses where they are applicable.

The backward-wave amplifier is a relatively new beam-type amplifier. It is closely related to the backward-wave oscillator,^{4,5} which is a voltage-tunable, regenerative traveling-wave tube oscillator. In both amplifier and oscillator, the electron beam interacts with a space-harmonic of the circuit whose group and phase velocities are oppositely directed. The phase velocity is toward the collector (in the same direction as the electron velocities), while the group velocity, and hence energy flow, is toward the electron gun end of the circuit. When the beam current is low, and the electron velocity is very nearly the circuit velocity at the frequency in question, a signal impressed at the collector

end of the circuit will appear amplified at the gun end of the circuit. If the beam current is raised, it will reach a discrete value at which the backward-wave amplifier becomes a backward-wave oscillator. At currents slightly below the start oscillation current, a very high, narrow-band electronic gain is obtained, the center frequency of which is voltage tunable. Such an amplifier has several applications; it could be used as a narrow-band detector, an electronically tuned filter, or in numerous microwave devices. A knowledge of the noise figure of this amplifier is both interesting and important. We shall derive a general expression for the noise figure, including the effects of space charge and circuit loss.

Noise figure is defined as the noise power output of an amplifier whose input is matched to its characteristic impedance divided by the output of a noiseless amplifier whose input is similarly matched. The output of the latter is simply power gain times the noise power available from the characteristic impedance, or, symbolically, $P_t = GkTB$. (Because the notation in this paper will be familiar to many workers in the field, the definition of symbols used is omitted from the text. These definitions are presented at the end of the paper.)

Fig. 1 gives a schematic picture of a backward-wave amplifier. In analyzing the interaction space, we shall assume a circuit input voltage at $z=l$, and an alternating velocity and current noise modulation in the beam at $z=0$. The electrons are constrained to move only in the z direction by an infinite axial magnetic field.

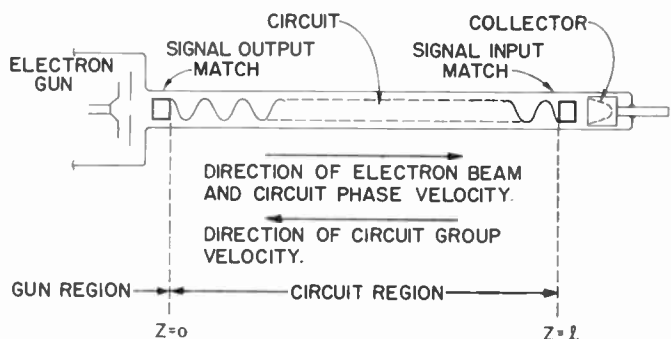


Fig. 1—Schematic diagram of a helix-type backward wave amplifier.

ANALYSIS OF CIRCUIT REGION

We assume that the reader is familiar with Pierce's analysis of the traveling-wave tube.⁶ Johnson has modified this analysis to describe backward-wave interac-

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¹ J. R. Pierce, "Traveling-Wave Tubes," D. Van Nostrand Co., Inc., New York, N. Y., Ch. 10; 1950.

² D. A. Watkins, "Noise Reduction in Beam Type Amplifiers," Electronics Research Laboratory, Stanford University, Technical Report No. 31, Stanford, California; March 15, 1951.

³ S. Bloom and R. W. Peter, "A minimum noise figure for the traveling-wave tube," *RCA Rev.*, vol. 15, pp. 252-267; June, 1954.

⁴ R. Kompfner and N. T. Williams, "Backward-wave tubes," *PROC. I.R.E.*, vol. 41, pp. 1602-1611; November, 1953.

⁵ H. R. Johnson, "Backward-Wave Oscillators," Hughes Aircraft Company Technical Memorandum No. 361; May, 1954.

⁶ Pierce, *op. cit.*, ch. 2, 7-9.

tion.⁷ He defines Pierce's impedance parameter K as follows:

$$K = -\frac{E_s^2}{2\beta^2 P}. \quad (1)$$

K is positive because P is negative for a backward-wave interaction. Pierce's loss parameter d is defined so that $d > 0$ corresponds to circuit attenuation for a backward wave. The other parameters of Pierce are defined as usual. This analysis yields a small C determinantal equation

$$-\delta^2 = \frac{1}{-b - jd + j\delta} + 4QC. \quad (2)$$

Realizing that the relation between the partial "total voltage" V_i , and the partial "circuit voltage" V_{ci} of Pierce is as follows,

$$\frac{V_{ci}}{V_i} = 1 + \frac{4QC}{\delta_i^2} \quad (i = 1, 2, 3), \quad (3)$$

and writing the equations relating partial circuit voltages to total circuit voltage, total alternating beam velocity, and total alternating current, we have the following expressions. (Unless otherwise indicated, all voltages are taken at axial position $z=0$.)

$$V_{c1} + V_{c2} + V_{c3} = V_c \quad (4)$$

$$\frac{V_{c1}}{\delta_1^2 + 4QC} + \frac{V_{c2}}{\delta_2^2 + 4QC} + \frac{V_{c3}}{\delta_3^2 + 4QC} = i \left(-\frac{2V_0 C^2}{I_0} \right) \quad (5)$$

$$\frac{\delta_1 V_{c1}}{\delta_1^2 + 4QC} + \frac{\delta_2 V_{c2}}{\delta_2^2 + 4QC} + \frac{\delta_3 V_{c3}}{\delta_3^2 + 4QC} = v \left(\frac{j u_0 C}{\eta} \right). \quad (6)$$

If (4), (5), and (6) are solved for V_{ci} , it is found that

$$V_{c1} = \frac{-V_c [(\delta_1^2 + 4QC)(\delta_2 - \delta_3)] + \left(j \frac{u_0 C}{\eta} \right) v [(\delta_1^2 + 4QC)(\delta_2^2 - \delta_3^2)]}{(\delta_1 - \delta_2)(\delta_2 - \delta_3)(\delta_3 - \delta_1)} + \frac{\left(-\frac{2V_0 C^2}{I_0} \right) i [\delta_2(\delta_3^2 + 4QC) - \delta_3(\delta_2^2 + 4QC)](\delta_1^2 + 4QC)}{(\delta_1 - \delta_2)(\delta_2 - \delta_3)(\delta_3 - \delta_1)}. \quad (7)$$

V_{c2} and V_{c3} are cyclic permutations of V_{c1} . Now $V_1(l) = V_1 \exp(2\pi CN \delta_1)$, and since V_1 and V_{c1} are related by a constant independent of axial position, z , we have $V_{c1}(l) = V_{c1} \exp(2\pi CN \delta_1)$. Adding $V_{c1}(l)$, $V_{c2}(l)$, and $V_{c3}(l)$, and solving for V_c results in the following expression:

$$V_c = -V_c(l) \frac{\xi(\delta)}{D(\delta, QC)} + \left(j \frac{u_0 C}{\eta} \right) v \frac{\tau(\delta, QC)}{D(\delta, QC)} + \left(-\frac{2V_0 C^2}{I_0} \right) i \frac{\phi(\delta, QC)}{D(\delta, QC)} \quad (8)$$

where

⁷ Johnson, *op. cit.*, pp. 2-3.

$$\xi(\delta) = (\delta_1 - \delta_2)(\delta_2 - \delta_3)(\delta_3 - \delta_1) \quad (8a)$$

$$\tau(\delta, QC) = (\delta_1^2 + 4QC)(\delta_2^2 - \delta_3^2) \exp 2\pi CN \delta_1 + \text{cyclic permutation} + \text{cyclic permutation} \quad (8b)$$

$$\phi(\delta, QC) = (\delta_1^2 + 4QC)[\delta_2(\delta_3^2 + 4QC) - \delta_3(\delta_2^2 + 4QC)] \cdot \exp 2\pi CN \delta_1 + \text{cyclic permutation} + \text{cyclic permutation} \quad (8c)$$

$$D(\delta, QC) = (\delta_2 - \delta_3)(\delta_1^2 + 4QC) \exp 2\pi CN \delta_1 + \text{cyclic permutation} + \text{cyclic permutation}. \quad (8d)$$

Eq. (8) expresses the voltage appearing on the circuit at $z=0$ (the circuit output) as a function of the voltage appearing on the circuit at $z=l$ (the circuit input) and the beam velocity and current modulation at $z=0$. The δ 's appearing in the above equation are functions of b , d , and QC . The latter two quantities, d and QC , do not vary if the loss and beam current of a given tube are held constant. On the other hand, b is varied merely by adjusting the beam voltage, holding the frequency constant. Johnson has calculated the δ 's and the CN for which oscillation begins for given values of d and QC .⁸ This is done by setting D above equal to zero, and solving for CN , using the expression relating δ to b , d , and QC . It is logical to assume that the b of maximum gain for the backward-wave amplifier is the same as the b of start oscillation. This is very nearly true if CN is within 5 to 10 per cent of $(CN)_s$. This assumption restricts the range of validity of the numerical results presented here. If exact results are desired, the roots of the determinantal equation may be found for various b 's, and these roots, together with the desired CN , may be sub-

stituted into (8). We therefore substitute the δ 's found by Johnson in (8). CN is quickly determined by the relation

$$CN = (CN)_s \left(\frac{I_0}{I_{0s}} \right)^{1/3} \quad (9)$$

where I_{0s} is the current at which the tube starts to oscillate and $(CN)_s$ is tabulated by Johnson.⁸

We must now transfer circuit voltage into circuit power. Since the thermal noise is not correlated to the beam noise, we must add the two as powers. To avoid

⁸ *Ibid.*, pp. 5, 7.

confusion, all voltages, currents, and velocities from this point will be taken as root-mean-square quantities. Keeping this in mind, and using the definition of C at the circuit entrance,

$$C^2 = \frac{I_0 V_c^2}{4 V_0 P}, \quad (10)$$

we find the noise power from the beam to be

$$P_b = \frac{I_0}{4 V_0 C^3} \left| \left(\frac{j u_0 C}{\eta} \right) v \frac{\tau}{D} + \left(- \frac{2 V_0 C^2}{I_0} \right) i \frac{\phi}{D} \right|^2. \quad (11)$$

The thermal noise power is

$$P_t = G k T B = k T B \left| \frac{\xi}{D} \right|^2. \quad (12)$$

Thus the noise figure is

$$F = \frac{P_t + P_b}{P_t} = 1 + \frac{I_0}{4 V_0 C^3 k T B} \left| \left(\frac{j u_0 C}{\eta} \right) v \left(\frac{\tau}{\xi} \right) + \left(- \frac{2 V_0 C^2}{I_0} \right) i \left(\frac{\phi}{\xi} \right) \right|^2. \quad (13)$$

The Gun Region

The gun of a beam-type amplifier consists of a thermionic cathode and various focusing and accelerating electrodes; it produces the electron beam with which the circuit interacts. There are several analyses of electron guns at microwave frequencies. We shall merely state a few general assumptions, and mention the procedure by which the results we use have been derived, referring the reader to the original source for a rigorous derivation. The expression for the noise figure of a backward-wave amplifier derived above contains two unknowns, the alternating velocity and current at the circuit entrance. Watkins,⁹ among others, analyzes the potential minimum to anode region of a space-charge limited diode with the Llewellyn-Peterson¹⁰ equations. Strictly speaking, these equations are valid only between parallel plane electrodes of infinite extent. The electrons may move only in a direction perpendicular to the electrodes, by the usual infinite axial magnetic field assumption. The Rack mean-square alternating velocity at the potential minimum is¹¹

$$v_a^2 = (4 - \pi) \frac{\eta k T_c B}{I_0}. \quad (14)$$

There is also another source of noise at the potential minimum, namely, the shot noise of the electrons:

$$\overline{i_a^2} = 2 e I_0 B. \quad (15)$$

Unless noise reduction schemes are incorporated in the gun, this noise is negligible. This shot noise will be neg-

lected until we find the minimum noise figure of a backward-wave amplifier.

Using the values of v and i found by Watkins, we obtain the following noise figures for the specified guns:

Case 1: Space-charge limited emission from cathode to circuit entrance:

$$F = 1 + \frac{(4 - \pi)}{2C} \frac{T_c}{T} \left| \frac{\tau}{\xi} - \frac{\phi}{\xi} \frac{\sqrt{2}}{\sqrt{4QC}} \right|^2. \quad (16)$$

Case 2: Space-charge limited emission from cathode to anode, anode followed by drift space followed by circuit; anode, drift space, and circuit at the same potential:

$$F = 1 + \frac{(4 - \pi)}{2C} \frac{T_c}{T} \left[1 + 2 \left(\frac{\omega_q}{\omega_p} \right)^2 \right] \cdot \left| \frac{\tau}{\xi} \cos \beta_{qz} + \frac{\phi}{\xi} \frac{\sin \beta_{qz}}{\sqrt{4QC}} \right|^2. \quad (17)$$

Case 3: Same as Case 2 except that the first drift space is followed by a second drift space at the helix potential. The two drift spaces are at different potentials, producing a velocity jump at the velocity maximum:

$$F = 1 + \frac{(4 - \pi)}{2C} \frac{T_c}{T} \left[1 + 2 \left(\frac{\omega_{q1}}{\omega_{p1}} \right)^2 \right] \frac{V_1}{V_2} \cdot \left| \frac{\tau}{\xi} \cos \beta_{q2z} + \frac{\phi}{\xi} \frac{\sin \beta_{q2z}}{\sqrt{4QC}} \right|^2. \quad (18)$$

Subscript 1 refers to the first drift space, subscript 2 to the second drift space. By making V_2 large with respect to V_1 , the noise figure can be markedly reduced. However, when V_2 becomes larger with respect to V_1 , the shot noise, (15), is amplified and may no longer be neglected; in short, the noise figure cannot be reduced without limit by velocity jumps.

The Minimum Noise Figure

A recent publication by Peter and Bloom³ concerns a theory on the minimum noise figure of a forward-wave traveling-wave tube amplifier. They start at the potential minimum, assuming full shot noise and the Rack fluctuating velocity; they consider the potential minimum to circuit region as an electron beam transmission line of variable impedance. They adjust the expression they find for the noise figure to its minimum value, which is

$$F_{\min} = 1 + 2\sqrt{4 - \pi} \frac{T_c}{T} \sqrt{4QC} \sqrt{f_{\max} f_{\min}}, \quad (19)$$

where

$$\begin{aligned} f(\psi) &= |\alpha \cos \psi - \beta \sin \psi| \\ 2f_{\max} &= |\alpha|^2 + |\beta|^2 + |\alpha + \beta^2| \\ 2f_{\min} &= |\alpha|^2 + |\beta|^2 - |\alpha^2 + \beta^2|. \end{aligned} \quad (20)$$

It can easily be shown that their parameter α may be

⁹ Watkins, *op. cit.*, pp. 11-19.

¹⁰ F. B. Llewellyn and L. C. Peterson, "Vacuum tube networks," *Proc. I.R.E.*, vol. 32, pp. 144-166; March, 1944.

¹¹ A. J. Rack, "Effect of space charge and transit time on the shot noise in diodes," *Bell Sys. Tech. Jour.*, vol. 17, pp. 592-619; 1938.

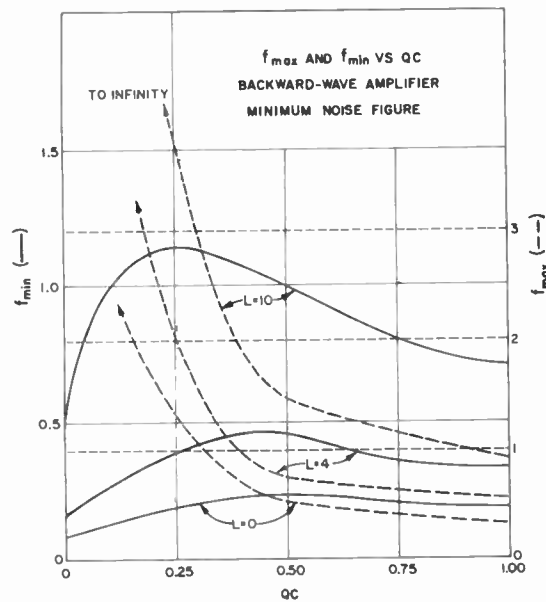


Fig. 2

replaced by our parameter τ/ξ , and their β may be replaced by our $[-\phi/\xi\sqrt{4QC}]$. The $f(\psi)$ of Peter and Bloom is taken directly from Watkins. The expressions derived here are the same as Watkins' if the mentioned replacements are made. We have calculated f_{\min} and f_{\max} for $L=0, 4$, and 10 ; and for $QC=0, 0.25, 0.50, 0.75, 1.00$.

In Fig. 2, f_{\min} and f_{\max} are plotted versus QC for the three values of loss. In Fig. 3, $(F_{\min}-1)T/T_c$ has been plotted versus QC for the three values of loss. These graphs are the backward-wave analogs of Fig. 2 and Fig. 4 in the above-mentioned paper of Peter and Bloom.³

It is interesting to note that the minimum noise figure for the zero loss case ($L=0$) is the same value for both the forward- and backward-wave amplifier, namely,

$$F_{\min} = 1 + .9265 \frac{T_c}{T} \quad (21)$$

Oxide cathodes operate about 1020 K; room temperature may be taken as 290 degrees K. The resulting minimum noise figure is about 6.5 db. Slightly different assumptions about the gun would lower this to about 6.0 db.¹²

In calculating the noise figure for the forward-wave amplifier, it is generally assumed that the circuit is long enough to justify dropping all but the growing wave term. The backward-wave theory described above assumes a CN near the CN of start oscillation, for it is only in this region that appreciable gain is found. Within the framework of these assumptions, it is extremely interesting to note that these two vastly different expressions lead to the same minimum noise figure,

¹² J. R. Pierce and W. E. Danielson, "Minimum noise figure of traveling-wave tubes with uniform helices," *Jour. Appl. Phys.*, vol. 25, pp. 1163-1165; September, 1954.

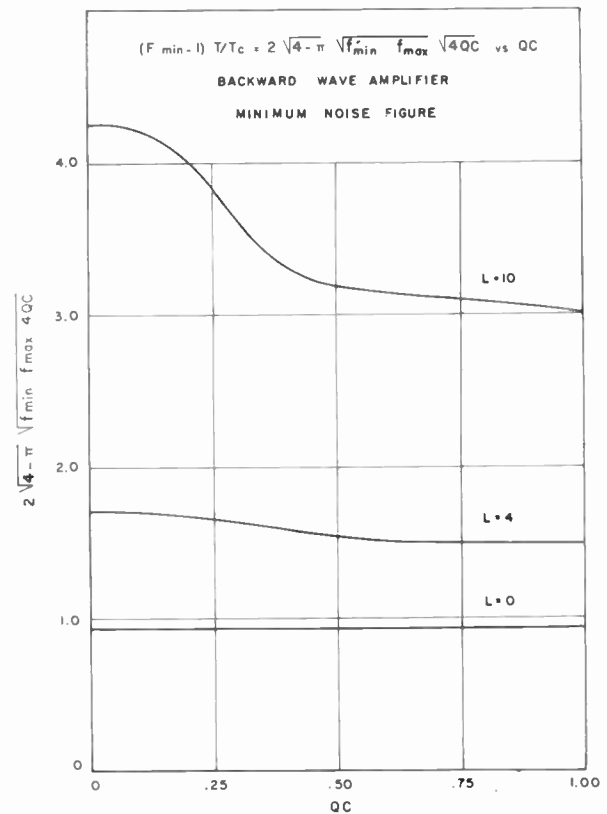


Fig. 3

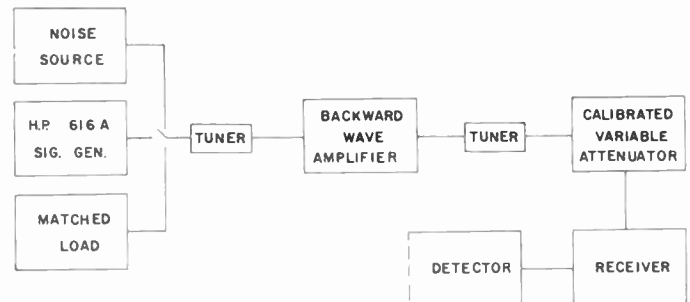


Fig. 4—Block diagram of noise figure measurement.

if the gun assumptions are the same in both cases. Watkins has shown that a klystron also has a theoretical minimum noise figure of about 6 db. This might indicate that all lossless beam-type amplifiers meeting the same assumptions have the same theoretical minimum noise figure.

If the effects of loss are included in the analysis, the minimum noise figure of a backward-wave amplifier differs from that of the forward-wave amplifier. At $QC=0$, and $L \neq 0$, the backward-wave amplifier has a higher minimum noise figure than the forward-wave amplifier; however, the slope of $(F_{\min}-1)T/T_c$ is negative in the case of the backward-wave amplifier, and positive in the forward-wave amplifier case. Consequently these curves cross, and for QC greater than a

certain value, the backward-wave amplifier will have a lower noise figure than the forward-wave amplifier. These curves can be used to predict the minimum noise figure of a backward-wave amplifier, if the circuit loss, L , and the space-charge parameter, QC , are known.

Because we wanted to compare our results with the work of Peter and Bloom, we have assumed full shot noise at the potential minimum. It should be mentioned that a recent analysis by Watkins¹³ predicts that there is a reduced shot noise at the potential minimum. This reduction lowers the minimum noise figure about 30 per cent for zero loss and space charge, or about 1.5 db.

EXPERIMENTAL RESULTS

The noise figure of a backward-wave amplifier has been measured as a function of gain. Care was taken to match the tube exceedingly well at the frequency of measurement, 3000 mc. The experimental setup is shown in Fig. 4. The experimental procedure was (1) measure the bandwidth of the receiver, (2) calibrate a Hewlett-Packard 616-A signal generator with a standard noise source (which in our case was a fluorescent lamp in waveguide that produced a "white" spectrum 15.8 db above kTB). This calibration gave us a setting, x , on the attenuator of the signal which corresponded to 15.8 db above kTB . Next, the matched load was connected to the tube, the calibrated attenuator was set at 0 db, and the detector reading was noted. Then the H-P 616A was connected to the tube, the calibrated attenuator set at 3 db, and the signal generator's attenuator adjusted to y , where the detector read the same as before. Thus the noise output of the tube, N_{out} , and the signal output, S_{out} , were identical, and both equaled the gain, G , times the signal input S_{in} . Since noise figure is defined

$$F = \frac{N_{out}}{GkTB} = \frac{S_{in}}{kTB}, \quad (22)$$

and the signal input is merely the attenuator setting of the signal generator, y ,

$$F = y - x + 15.8 \text{ db.} \quad (23)$$

This method of measurement eliminates any errors introduced by the receiver and detector, because the detector reading, and therefore the receiver input, remain constant during each experimental trial. Thus a non-linear, non-square-law detector may be used in this experimental setup.

It should be stressed that the tube used in this experiment was a laboratory model backward-wave oscillator which was operated below starting current, i.e., as an amplifier. No attempt was made to reduce the noise figure. The gun was quite similar to gun pictured in Fig. 1. After three points were measured the tube was

accidentally broken. For this reason no checks were made and no measure of experimental error was established. However, because of the method of measurement used, and because of the pains taken to match the tube well, we feel that these points are reasonably accurate. The parameters of this tube for these measurements were $K=3.5$ ohms, $V=990$ volts, $ka=.392$, $\gamma a=6.3$, mean beam radius $b=.202$ inch, mean helix radius $a=.245$ inch, $b/a=.825$, and the start oscillation current, $I_s=1.2$ ma. A longitudinal magnetic field focused the hollow electron beam. Less than 2 per cent of the beam current was intercepted by the helix. The noise figure was calculated assuming space-charge-limited emission from cathode to helix, which is a good assumption for the voltages used. The results of our measurement are shown in Table I, together with the calculated gain of the tube and the calculated noise figures for losses of zero and 4 db.

TABLE I

Loss—db	Beam current—ma	1.1	1.0	0.9
2.4	Gain—db	19.2	13.4	9.6
2.4	F (measured)—db	23.3	23.1	23.8
0	F (calculated)—db	18.0	18.5	18.9
4	F (calculated)—db	22.4	22.7	23.0

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The author gratefully acknowledges the help of each of the following persons, whose contributions have made this report possible: Dr. H. R. Johnson, who suggested the project and gave valuable aid during its progress; Dr. C. K. Birdsall, who gave advice on the subject of noise; Mrs. Kazi Higa, who performed the necessary computation; and A. M. Anderson and other members of the Engineering Staff, who built the experimental model.

LIST OF SYMBOLS

(in order of appearance)

P_t	Thermal noise power at the tube output, $z=0$.
G	Power gain.
k	Boltzmann constant (1.38×10^{-23} joule/degree Kelvin).
T	Temperature of characteristic impedance matched at the tube input, $z=l$.
l	Tube length.
z	Axial distance measured from circuit entrance.
B	Bandwidth of amplifier.
K	Pierce's impedance parameter, defined for the backward interaction in (1).
E_z	Axial electric field.
β	Circuit propagation constant, ω/v_p , where v_p is cold circuit velocity.
ω	2π (frequency).
P	Total power flow on circuit.
d	Circuit loss parameter of Pierce.

¹³ D. A. Watkins, "Noise at the Potential Minimum in the High-Frequency Diode," *Jour. Appl. Phys.* (to be published).

C	Pierce's gain parameter, $(I_0 K/4V_0)^{1/3}$.	L	mass, m , 1.76×10^{-11} coulomb/kg.
I_0	Average electron convection current.	N	Total circuit loss, 54.6 Cnd decibels.
V_0	Average beam voltage.	N	Number of circuit wavelengths.
δ	Normalized propagation constant of Pierce.	T_c	Cathode temperature in degrees Kelvin.
b	$(u_0 - v_p)/Cu_0$, where u_0 is the average electron velocity, $\sqrt{2\eta V_0}$.	e	Electronic charge, coulomb.
QC	Space charge parameter of Pierce.	j	Symbolizes an imaginary number $\sqrt{-1}$.
V_c	Circuit voltage of Pierce (rms).	ω_q	Reduced plasma frequency, $p\omega_p$ or $p\sqrt{\eta I_0/\epsilon_0 u_0 \sigma}$, where
V	Total voltage of Pierce (rms).		σ = beam cross-sectional area,
i	Alternating electron convection current, positive if electrons flow toward positive z .		p = plasma reduction factor.
v	Alternating beam velocity, positive toward positive z .	β_q	ω_q/u_0 .
N	l/λ , circuit length in wavelengths.	$\xi(\delta)$	parameters defined for convenience after (8).
η	Ratio of electronic charge, e , to electronic	$\tau(\delta, QC)$	
		$\phi(\delta, QC)$	
		$D(\delta, QC)$	

The Nature of the Uncorrelated Component of Induced Grid Noise*

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Summary—An investigation of induced grid noise in vacuum tubes has been made. It was found that the uncorrelated component of grid noise can be explained in terms of electrons elastically reflected from the plate of the tube. Experimental and theoretical justifications of this explanation are presented. The accuracy of methods for predicting grid noise from measurements of input admittance is affected because of a component of input admittance arising from reflected electrons. A table is included showing typical (measured) values of the induced grid noise of eleven modern receiving tubes.

INTRODUCTION

TWO TYPES of vacuum tube noise, shot noise and induced grid noise, are of importance in the design of low-noise, high-frequency amplifiers. Shot noise is the fluctuating component of plate current caused by random variations in the cathode emission rate. Induced grid noise is generated by fluctuations in the number of current pulses induced in the grid circuit by the passage of electrons between grid wires. At the higher operating frequencies, induced grid noise is the limiting factor in low-noise amplifier design.^{1,2}

For many years the theory of induced grid noise has been in a rather unsatisfactory state. Calculations based

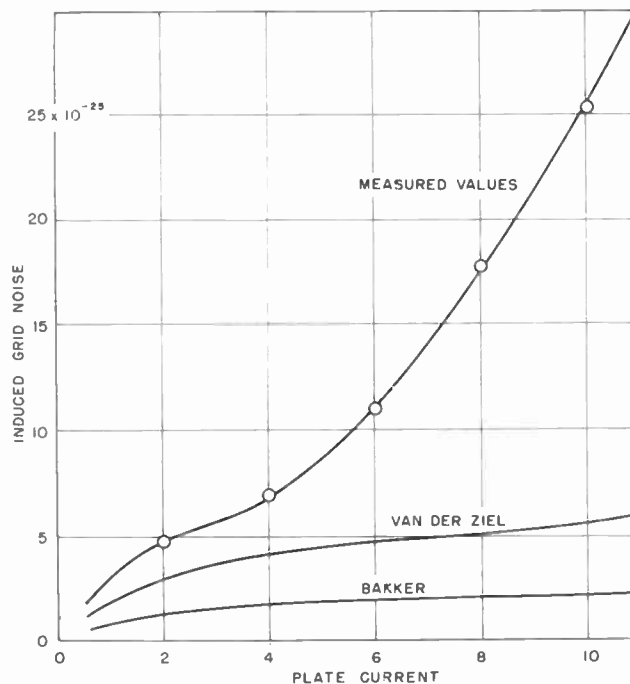


Fig. 1—Comparison between measured and calculated values of mean square induced grid noise current as a function of plate current for one section of a 6J6 double triode at 30 mc. Grid noise is expressed in amperes squared per unit bandwidth, and the plate current in ma.

directly on electron transit times yield induced grid noise magnitudes which are consistently low, often by a factor of two or three, (see Fig. 1), while predictions based on measurements of the input admittance are

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¹ Frequencies above about 15mc for modern miniature tubes.

² H. Wallman, A. B. Macnee, and C. P. Gadsden, "A low-noise amplifier," *Proc. I.R.E.*, vol. 36, pp. 700-708; June, 1948.

found to provide a fair amount of agreement with measured grid noise values.³⁻⁵

It is evident from experimental studies that a large component of the induced grid noise is uncorrelated with the shot noise in the plate current stream.^{6,7} Previous writers have suggested two explanations for the origin of this uncorrelated component:

1. Total emission noise⁸—fluctuating currents induced in the grid circuit by electrons returned to the cathode before reaching the potential minimum.
2. Induced partition noise⁹—fluctuations arising because of electron-trajectory variations and inhomogeneities in the electrode structure. It can be thought of as arising from fluctuations in average transit angle.

Consideration of the position of the potential minimum relative to electrode spacings indicates that total emission noise is negligible with respect to the total induced grid noise in tubes such as the 6AK5. Bell⁹ has estimated the magnitude of noise to be expected from electron-trajectory variations and concludes that the effect is small under normal operating conditions. The validity of this conclusion has not been completely substantiated by experiment, but it appears that these two explanations alone are not adequate to account for the observed excess grid noise.

This paper presents the results of theoretical and experimental studies which indicate that a major portion of the uncorrelated component of induced grid noise is caused by fluctuations in a small number of electrons which are elastically reflected by the plate. These electrons are reflected with sufficient energy to enable them to return through the grid, inducing additional current pulses in the grid circuit. This current increases the input admittance and the induced grid noise. A brief discussion concerning the accuracy of grid noise predictions based on measurements of input conductance or susceptance is included.

For the benefit of the design engineer a table has been included showing typical measured magnitudes of induced grid noise and plate noise for a number of modern miniature receiving tubes.

ANALYSIS OF INDUCED GRID NOISE

Theoretical studies, published first by Bakker³ and recently extended by van der Ziel,¹⁰ show that fluctua-

tions in the cathode emission of a planar triode should produce a mean square induced grid-noise current

$$\overline{i_g^2} = \overline{i_k^2} \left(\frac{\omega \tau_1}{3} \right)^2 \left[1 + 2 \left(\frac{\tau_2}{\tau_1} \right) \right]^2. \quad (1)$$

In this equation

- $\overline{i_k^2}$ = mean square space-charge-reduced shot-noise component of cathode current, I_k
- $= 2eI_k\Gamma^2\Delta f$
- τ_1 = transit time from potential minimum to grid-plane
- τ_2 = transit time from grid-plane to plate
- Γ^2 = space-charge reduction factor
- e = charge on an electron = 1.6×10^{-19} coulomb
- Δf = bandwidth in cps.

Eq. (1) is plotted for a typical case in Fig. 1. By a method similar to that described by Goldman¹¹ it has been shown that to a first approximation the induced grid noise can be expressed as¹²

$$\overline{i_g^2} = \frac{2I_b\Delta f\Gamma^2}{e} \cdot |S(\omega)|^2, \quad (2)$$

where

$$S(\omega) = 2\pi G(\omega) = \int_{-\infty}^{+\infty} F(t)e^{-j\omega t}dt. \quad (3)$$

$G(\omega)$ is the Fourier transform of $F(t)$, the current pulse induced in the grid by the passage of a single electron from cathode to plate [see Fig. 2(a)]. Comparison of (1) and (2) reveals the form of $|S(\omega)|^2$. Since the area of the grid current pulses must be zero, their power spectra and the mean square induced grid-noise current are both proportional to the square of frequency at small transit angles. This observation has been verified experimentally numerous times.³⁻⁶

The lack of agreement between measured and calculated values of induced grid noise as exemplified by Fig. 1 can be explained in terms of electrons reflected by the plate. It has long been known^{13,14} that electrons which are elastically reflected at the plate of a diode cause an increase in the shot noise. Strangely enough, no study has been published showing the effect of reflected electrons on grid noise.

An electron which is elastically reflected from the plate has sufficient energy to penetrate the retarding field that it meets between grid and plate. It will very likely succeed in passing back between the grid wires

³ C. J. Bakker, "Fluctuations and electron inertia," *Physica*, vol. 8, pp. 23-43; January, 1941.

⁴ D. O. North and W. R. Ferris, "Fluctuations induced in vacuum tube grids at high frequencies," *Proc. I.R.E.*, vol. 29, pp. 49-50; February, 1941.

⁵ R. L. Bell, "Induced grid noise," *Wireless Eng.*, vol. 27, pp. 86-94; March, 1950.

⁶ R. Q. Twiss and Y. Beers, "Minimal Noise Circuits," *Vacuum Tube Amplifiers*, vol. 18, Ch. 13, M.I.T. Rad. Lab. Series, McGraw-Hill Book Company, New York N.Y., 1948.

⁷ A. van der Ziel, "Noise suppression in triode amplifiers," *Canad. Jour. Tech.*, vol. 29, pp. 540-553; December, 1951.

⁸ A. van der Ziel and A. Versnel, "Induced grid noise and total emission noise," *Philips Res. Rep'ts*, vol. 3, pp. 13-23; February, 1948.

⁹ R. L. Bell, "Negative grid partition noise," *Wireless Eng.*, vol. 25, pp. 294-297; September, 1948.

¹⁰ A. van der Ziel, "Induced grid noise in triodes," *Wireless Eng.*, vol. 28, pp. 226-227; July, 1951.

¹¹ S. Goldman, "Frequency Analysis, Modulation and Noise," McGraw-Hill Book Company, New York, N.Y., 356 ff; 1948.

¹² T. E. Talpey, "A Study of Induced Grid Noise," Doctoral Thesis, July 1953, available on microfilm from University Microfilms, Ann Arbor, Michigan.

¹³ D. O. North, "Fluctuations in space-charge-limited currents at moderately high frequencies, Part II, diodes and negative grid triodes," *RCA Rev.*, vol. 4, pp. 441-472; April, 1940; vol. 5, pp. 106-124; July, 1940.

¹⁴ G. E. Duvall, "The Effects of Transit Angle on Shot Noise in Vacuum Tubes," M.I.T. Res. Lab. of Electronics, Tech. Rep. No. 82, Sept. 8, 1948.

before it loses its cathode-directed energy and is finally drawn back to the plate again. The pulse of current induced in the grid circuit by such a reflected electron will be approximately three times as long as the pulse produced by an ordinary electron, as indicated in Fig. 2(b).

By the application of two theorems from the study of Fourier integrals¹⁵ it is easily shown that the power spectrum of the complex pulse shown in Fig. 2(b) is given by the following expression:

$$|S_r(\omega)|^2 = |S(\omega) - S(-\omega)e^{j2\omega(\tau_1+\tau_2)} + S(\omega)e^{-j2\omega(\tau_1+\tau_2)}|^2 \quad (4)$$

where $|S(\omega)|^2$ is the power spectrum of the first third of the pulse, up to the point where the electron first arrives at the plate.

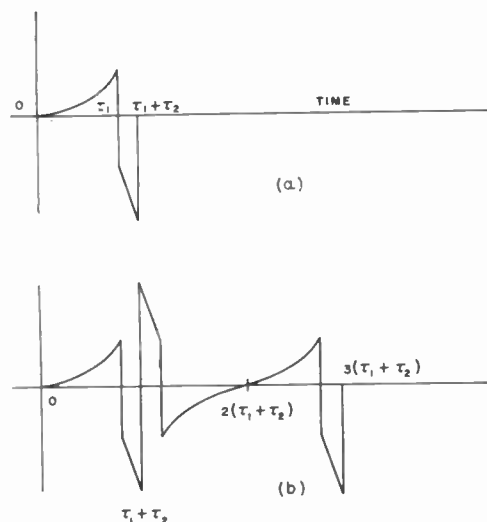


Fig. 2—(a) Current pulse induced in the grid circuit of an ideal space-charge-limited triode by the passage of a single electron, (b) Current pulse induced in the grid circuit by an electron which traverses the tube, is elastically reflected at the plate, succeeds in getting back close to the potential minimum, and then returns to the plate.

Now $S(\omega)$ can be expanded in a power series in terms of $\omega\tau_1$ as follows:

$$S(\omega) = a_1\omega\tau_1 + a_2(\omega\tau_1)^2 + a_3(\omega\tau_1)^3 + \dots \quad (5)$$

where the factors a_1, a_2 , etc. involve terms in (τ_2/τ_1) , $(\tau_2/\tau_1)^2$, etc., as in (1). Assuming that the transit angle $\omega\tau_1$ is considerably smaller than one radian,¹⁶ we can neglect all terms except the first in (5) and write

$$S(\omega) \cong -S(-\omega) \cong a_1\omega\tau_1. \quad (6)$$

Eq. (4) then becomes

$$|S_r(\omega)|^2 \cong |S(\omega)|^2 [1 + 2 \cos 2\omega(\tau_1 + \tau_2)]^2. \quad (7)$$

If $2\omega(\tau_1 + \tau_2)$ is also considerably less than one radian, the cosine is approximately unity and we obtain the relationship

¹⁵ See, for example, E. A. Guillemin, "The Mathematics of Circuit Analysis," John Wiley & Sons, New York, N.Y., Ch. VII, Article 22; 1949.

¹⁶ At 30 mc, the transit angles of all the miniature tubes studied are well below 0.1 radian.

$$|S_r(\omega)|^2 \cong 9|S(\omega)|^2. \quad (8)$$

The reflected electrons thus produce an induced grid noise component given approximately by

$$\overline{i_g^2} = \frac{2\Delta f}{e} (rI_b)(9)|S(\omega)|^2, \quad (9)$$

where r is the reflection coefficient of the plate, that is, the fraction of incident electrons which are elastically reflected.

Since fluctuations in the number of reflected electrons are independent of fluctuations in cathode emission, the induced grid noise components given by (2) and (9) add quadratically, giving as an approximate expression for the total induced grid noise at small transit angles

$$\overline{i_g^2} = \frac{2I_b\Delta f}{e} |S(\omega)|^2 [1^2 + 9r]. \quad (10)$$

Logarithmic extrapolation of experimental data reported by Farnsworth¹⁷ indicate that $r=0.03$ is a reasonable estimate of the reflection coefficient for plate voltages of 100 to 150 volts. Nominal values of 1^2 lie near 0.1, so that the bracket in (10) becomes

$$[1^2 + 9r] = [0.1 + .27] = 0.37.$$

The reflected electrons in this case have caused an approximately four-fold increase in induced grid noise. We thus conclude that the reflected electrons are entirely capable of producing the observed excess of measured grid noise over values predicted by earlier theories.

EXPERIMENTAL VERIFICATION

The effect of reflected electrons on induced grid noise was verified experimentally by measuring the induced grid noise of a type 6AS6 pentode as a function of suppressor voltage; the results of these measurements are shown in Fig. 3, on the following page.

The induced grid noise increases by a factor of six or so as the suppressor voltage varies from +20 to -20 volts. When the suppressor is negative, it creates a retarding field and some electrons are reflected before they reach the plate. As the suppressor is made more negative, more electrons are reflected until at about -10 volts they are all reflected and the plate current drops to zero. Those electrons which are not captured by the screen travel on toward the grid and induce additional current pulses in the grid circuit.

The correlation between induced grid noise and reflected electrons is even more striking if the data of Fig. 3 are plotted in a different manner. The deficiency in plate current with respect to its asymptotic value at positive suppressor voltages is a measure of the number of electrons which are artificially reflected by the field of the suppressor before they can reach the plate. The grid

¹⁷ H. E. Farnsworth, "Energy distribution of secondary electrons from copper, iron, nickel and silver," *Phys. Rev.*, vol. 31, pp. 405-422; March, 1928

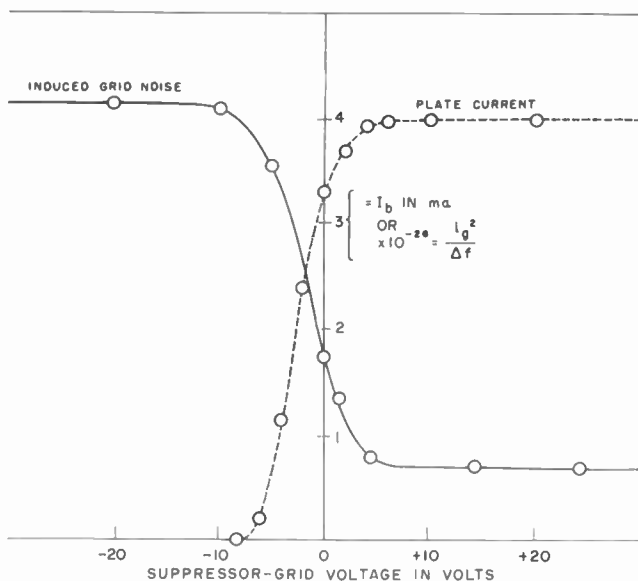


Fig. 3—Mean square induced grid-noise current vs suppressor-grid voltage for a 6AS6 at 30 mc with a fixed control grid voltage and a constant cathode current. The variation in plate current is also shown (dashed curve).

noise in excess of its asymptotic value at positive suppressor voltages should be directly proportional to the number of artificially reflected electrons according to (10). When the deficiency of plate current and the excess grid noise are both plotted as a function of suppressor voltage on the same set of coordinates the correlation is excellent, as Fig. 4 clearly shows.

It is significant that the grid noise is appreciably larger for zero suppressor voltage than for positive voltages. When the suppressor is at the same potential as the cathode, the resulting electric field is able to deflect a few electrons sufficiently to prevent them from reaching the plate.¹⁸ These electrons are returned to the vicinity of the control grid and thus cause an increase in the grid noise. If the suppressor connection is brought out to a separate pin, it should be connected to the plate and screen. This eliminates the possibility of reflected electrons being produced by deflection in the screen-suppressor region yet preserves the obstacles presented by the suppressor and screen wires to the return of reflected electrons from the plate. This connection was tried in a 30-mc cascode amplifier² employing a 6AS6 input stage. It was found that the noise factor could be reduced from 2.95 to 2.25 db by changing the suppressor connection from the cathode to the plate.

REFLECTED ELECTRONS AND INPUT ADMITTANCE

There is a direct relationship between the electrons which are elastically reflected at the plate of a vacuum

¹⁸ The deflection and subsequent reflection are caused by the combined field of the screen and suppressor grids. The reflection takes place just in front of the suppressor grid. The mechanism is similar to that described by W. G. Dow, "Fundamentals of Engineering Electronics," 2nd Ed., John Wiley & Sons, New York, N. Y., pp. 28-29; 1952.

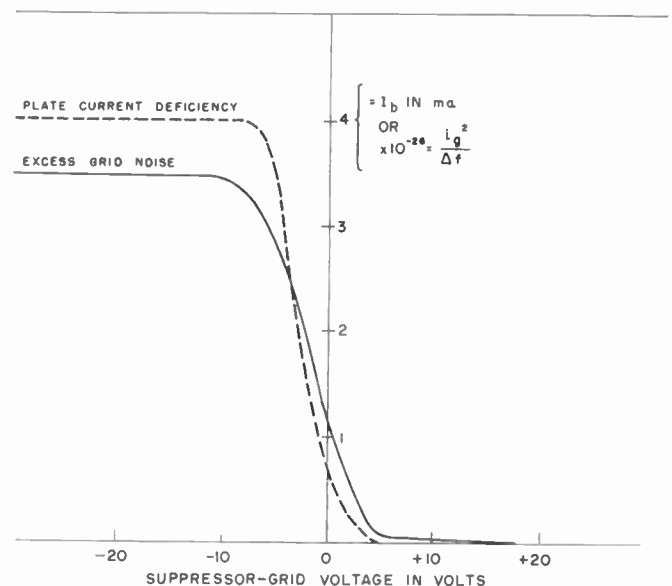


Fig. 4—Excess grid noise and deficiency in plate current (dashed curve) vs suppressor-grid voltage.

tube and the transit-time component of input admittance. This can be demonstrated qualitatively in the following manner. A signal applied to the grid of a tube causes variations in the control-grid voltage which are accompanied by variations in the plate current. For small signals it can be assumed that the reflection coefficient of the plate is constant, so that variations in the plate current will produce proportional variations in the number of reflected electrons.¹⁹ It follows that these reflected electrons must produce, in the grid circuit, a varying induced current which is proportional to the grid voltage. The component of this induced current which is in phase with the grid voltage produces additional input conductance; while the quadrature component produces additional input susceptance. When a measurement is made of the input admittance of a tube, the value obtained includes a component due to reflected electrons as well as the component due to the main electron stream.²⁰

Two methods have been advanced for predicting grid noise from measurements of the input admittance. One of these methods^{3,4} expresses the grid noise in terms of the transit-time component, G_o , of input conductance:

$$\overline{i_g^2} \cong 4kT\beta G_o \Delta f. \quad (11)$$

The value of the quantity β is usually taken as 5.0, although it is a function of the cathode temperature and certain geometrical factors. Unfortunately, because of lead inductance effects it is frequently more difficult to measure G_o than it is to measure the grid noise directly.

¹⁹ Sizable variations are being considered here; they should not be confused with the minute random fluctuations which give rise to induced noise. The random fluctuations are superimposed on the variations in plate current.

²⁰ There is also a component due to feedback from the cathode lead inductance. This feedback usually produces considerable input loading but negligible additional noise.

A second method of predicting grid noise makes use of the following expression:⁵

$$\overline{i_g^2} \cong \overline{i_p^2} \left(\frac{\omega C_e}{g_m} \right)^2, \quad (12)$$

where C_e is the space-charge component of input capacity—i.e., the difference between the input capacity when the tube is operating under normal conditions and the input capacity with the tube biased beyond "cut-off."

Both (11) and (12) imply complete correlation between grid noise and plate noise. Both of these equations were derived under the assumption that both grid noise and plate noise owe their origin to fluctuations in the primary electron stream passing the grid. These fluctuations are affected by space charge in the cathode-grid region and are often said to be "space-charge reduced." On the other hand, fluctuations in the number of reflected electrons are not influenced by space charge in the input region. We are thus led to the conclusion that the grid noise produced by reflected electrons, while linearly related to a component of input admittance, must be related in a slightly different manner than indicated by (11) or (12). The use of these equations for predicting the total induced grid noise consequently involves a certain amount of inaccuracy. Further study of the connection between input admittance and reflected electrons is necessary before the magnitude of the error can be ascertained.

THE INDUCED GRID NOISE OF MINIATURE RECEIVING TUBES

During the course of the research leading to the formulation of the above theory many measurements were made of the induced grid noise of a variety of commercially available receiving tubes. Table I presents a summary of these measurements. Because of its usefulness in network calculations, the induced grid noise values are expressed in terms of an equivalent grid noise conductance (βG_g), based on the representation defined by

(11). (No claim for the measurement of G_g alone is intended.) Measurements were also made of the plate noise of these tubes, and the results are presented in Table I in terms of an equivalent shot noise resistance.¹³

The method employed for the measurement of induced grid noise was essentially the same as that described by Bakker.³ A resonant capacitor was connected from plate to cathode of the tube under test to short-circuit the plate noise and prevent feedback, and the grid was connected directly to the input of a high-gain low-noise amplifier. Noise currents induced in the grid circuit of the tube under test were compared with noise from a temperature-limited diode. The change which occurred in the impedance level of the input circuit when the tube under test was turned on (loading due to transit-time and lead inductance effects) required that a correction be applied to the noise diode reading. The correction was determined from the effect of this loading on the amplifier output level. At 30 mc the effect of lead inductance on *grid noise* is negligible.

The values given in Table I represent the averages of measurements taken on a few tubes of each type, with particular values of voltage and current. Values for any given tube may differ considerably (20 to 30 per cent) from these data, even under the same operating conditions.

The use of a lower grid bias to obtain more plate current was found to cause an increase in grid noise in approximately the same ratio as the g_m was increased. Raising the plate voltage (leaving bias fixed) increases g_m without any appreciable rise in grid noise. It follows that for a given plate-current value, a high plate voltage and high negative grid bias are desirable for the attainment of a low noise figure.

SUMMARY

Experimental and theoretical evidence indicate that the origin of a major portion of the uncorrelated component of induced grid noise is that small fraction of the electron stream which is elastically reflected at the plate of a vacuum tube. It has been shown qualitatively that

TABLE I
MEASURED VALUES OF INDUCED GRID NOISE AT 30MC

Tube type	Number examined	Induced grid noise: Equivalent noise conductance βG_g in micromhos			Plate current I_b in ma	Equivalent shot noise resistance Req.	Transconductance g_m in micromhos
		Average	Lowest	Highest			
6AG5	21	142	112	156	7	480	6,000
6AK5	30	46	36	58	10	460	5,500
6AS6*	4	46	42	52	10	450	5,800
6AU6	17	212	184	254	12	420	6,600
6BC5	6	130	120	142	8	590	5,800
6BH6*	3	224	—	—	10	—	—
6BC6*	10	174	144	194	12	410	7,300
6J4	2	200	—	—	15	321	12,000
6J6†	10	60	42	82	8	720	4,400
2C51†	8	40	38	46	8	550	5,400
404A	9	72	68	36	15	240	16,000

* Suppressor connected to plate and screen.

† Values are for a single section.

the presence of reflected electrons will affect the accuracy of induced grid noise predictions based on measured values of input admittance.

It is conceivable that the effect of reflected electrons could be eliminated or materially reduced by the use of a specially constructed tube. If this could be done, the remaining induced grid noise would be more completely correlated with shot noise. By properly detuning the input circuit of a suitable amplifier, it should then be

possible to use this correlation to cause a partial cancellation of the effects of induced grid noise and thereby obtain substantially lower noise figures at high frequencies.

ACKNOWLEDGMENT

The authors would like to thank Professors W. G. Dow and G. Hok of the University of Michigan for helpful discussions during the course of this study.

On the Possibility of Amplification in Space-Charge-Potential-Depressed Electron Streams*

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Summary—Hahn-Ramo theory is used to derive a characteristic wave equation for an electron stream whose single-valued velocity is a function of spatial co-ordinates. A means of solving this equation is found, for the particular case of two-dimensional Cartesian co-ordinates. A specific, but practical, linear velocity distribution is assumed. It is shown that for several types of boundary conditions, the only waves which can be set up in such a beam are purely propagational and not growing, bearing out the result derived by an approximate method by G. Kent.

Numerical analysis for a cylindrical beam with potential depression was performed by a digital computer. As before, the results showed absence of any growing waves.

In order to check early results of Haeff,¹ which appeared to show the possibility of gain in single beam devices, an experiment was set up whereby a movable pickup cavity measured the amplitude of space-charge waves at a number of points along a drift tube.

Outputs were compared at different drift lengths for pulsed and continuous operation. No evidence of growing waves was observed, verifying the analytical results. It was found, however, that operation of the collector electrode at very low potentials created secondary electrons which returned to the gun region, were reflected, and then they flowed back with the primary beam. This double-stream action produced electronic gains up to 30 db. It is believed that either a similar effect, or else a space-charge wave gain produced as a direct result of nonuniform beam flow, can explain any signal gains found in conventional single-stream tubes.

INTRODUCTION

SEVERAL years ago, the double-stream amplifier, a mechanism for obtaining amplification in electron streams, was described by several authors.¹⁻⁴ It

has been shown analytically and experimentally that the mixture of two homogeneous electron streams of slightly different velocity should give rise to exponentially increasing space-charge waves, the result of the perturbation of the two original sets of space-charge waves by one another. Other authors⁵⁻⁸ have expanded this theory to general multiple-stream amplifiers, with finite boundaries. These theories indicate an optimum gain when only two distinct velocities are present. For the case of homogeneous mixtures of beams of different velocity, it has been shown that velocities in a Gaussian distribution cannot give rise to amplification.⁸ A general proof given by Walker⁹ indicates that a homogeneous mixture of electrons whose velocity distribution is flat or monotonic, or has a single maximum, cannot produce gain.

Some of the experimental data presented in Haeff's paper¹ indicated large amplification in the case of an electron stream coming from a single cathode. It was proposed that the observed gain was a consequence of the spatial velocity distribution caused by space-charge depression of potential. This had not been taken into account in the existing analyses, which dealt with a one-dimensional problem. A considerably more complicated problem results when the electron velocity is made a spatial function. Kent¹⁰ has used a series approximation to show that gain cannot be self-consistent in a ribbon beam having space-charge depression of potential. The

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¹ A. V. Haeff, "The electron-wave tube," *Proc. I.R.E.*, vol. 37, pp. 4-10; January, 1949.

² L. S. Nergaard, "Analysis of a simple model of a two-beam growing wave tube," *RCA Rev.*, vol. 9, pp. 585-601; December, 1948.

³ J. R. Pierce and W. B. Hebenstreit, "A new type of high frequency amplifier," *Bell Sys. Tech. Jour.*, vol. 28, pp. 33-51; January, 1949.

⁴ A. V. Hollenberg, "Experimental observation of amplification by interaction between two electron streams," *Bell Sys. Tech. Jour.*, vol. 28, pp. 52-58; January, 1949.

⁵ P. Parzen, "Theory of space-charge waves in cylindrical waveguides with many beams," *Elect. Commun.*, vol. 28, pp. 217-219; September, 1951.

⁶ J. R. Pierce, "Double-stream amplifier," *Proc. I.R.E.*, vol. 37, pp. 980-985; September, 1949.

⁷ C. K. Birdsall, "Interaction Between Two Electron Streams for Microwave Amplifications," Tech. Rep. No. 36, Elec. Res. Lab., Stanford Univ., Palo Alto, Calif.

⁸ H. Haus, "A Multivelocity Electron Stream in a Cylindrical Drift Tube," unpublished report, Res. Lab. Elec., MIT; June 5, 1952.

⁹ L. R. Walker, "The dispersion formula for plasma waves," *Jour. Appl. Phys.*, vol. 25, p. 131; January, 1954.

¹⁰ G. Kent, "Space charge waves in inhomogeneous electron beams," *Jour. Appl. Phys.*, vol. 25, pp. 32-41; January, 1954.

present paper contains a solution of the same problem, in a more rigorous manner. In addition, the space-charge wave modes for the more practical case of a cylindrical beam have been calculated numerically. The results in both cases strengthen the general conclusions reached earlier in the one-dimensional cases.

FORMULATION OF THE PROBLEM

The basic method used in problems involving space-charge interaction is the Hahn-Ramo method.¹¹⁻¹² This assumes: (1) a linear interaction in which the perturbations of space-charge density and electron velocity are small fractions of their average values, (2) that all-time varying quantities are of the form $f_T e^{j\omega t - \gamma z}$, where: f_T indicates a function of the transverse co-ordinates only, ω is the angular frequency of the perturbation, γ is the propagation constant (in general complex), z is the direction of electron velocity, t is time; and (3) electron velocities are single-valued functions of position.

The particular situation to be discussed involves an electron beam of space-charge density ρ and velocity v (the average values of which, ρ_0 and v_0 , are functions only of the co-ordinates transverse to z), moving in free space or bounded at or beyond its edge by perfectly conducting metal walls. An infinitely strong axial magnetic field assures zero transverse velocity.¹³

To obtain a solution for the waves in an electron beam one employs Maxwell's equations:

$$\begin{aligned}\nabla \times \bar{H} &= \rho \bar{v} + \epsilon_0 \frac{\partial \bar{E}}{\partial t} \\ \nabla \times \bar{E} &= -\mu_0 \frac{\partial \bar{H}}{\partial t} \\ \nabla \cdot \bar{E} &= \rho / \epsilon_0 \\ \nabla \cdot \bar{H} &= 0\end{aligned}\quad (1)$$

and the equation of motion of an electron

$$m \left[\frac{\partial \bar{v}}{\partial t} + v_x \frac{\partial \bar{v}}{\partial x} + v_y \frac{\partial \bar{v}}{\partial y} + v_z \frac{\partial \bar{v}}{\partial z} \right] = e [\bar{E} + \bar{v} \times \bar{B}]. \quad (2)$$

The velocity and space charge are separated into static part and perturbation. All cross-products of perturbations are neglected. The equation involving the axial component of electric field can be simplified to the following form:

$$\Delta_z^2 E_z + \left[\gamma^2 + \frac{\omega^2}{c^2} \right] \left[1 + \frac{\rho_0 \eta / \epsilon_0}{[j\omega - \gamma v_0]^2} \right] E_z = 0. \quad (3)$$

Δ_z^2 is the Laplacian operator of the transverse axes,

c the velocity of light, η the charge-mass ratio of the electron, and ϵ_0 the permittivity of free space.

Eq. (3), if solved under particular boundary conditions, is a characteristic equation specifying characteristic values of γ . Values of γ in the second quadrant of the complex plane are necessary for gain.

The solutions of (3) are real for purely imaginary γ , and complex for complex γ . In cases where there is no gain, the values of γ correspond to solutions having none, one, two, three, etc., zeros interior to the range of solution. The solutions having no zeros (except, perhaps, at the boundaries) are the familiar space-charge waves of the low-level klystron.

Velocity and Space-Charge Distribution

The space-charge density distribution chosen will greatly influence the values of γ . For practical reasons, we shall choose ρ_0 to be constant over the beam, and zero outside the beam proper. This distribution is not too far from that which can be obtained experimentally.

The velocity distribution will be chosen to approximate that found in an electron beam whose potential is depressed by space-charge forces. For an axially cylindrical, drifting beam, the solution of Poisson's equation

$$\nabla^2 V = -\rho_0 / \epsilon_0$$

gives:

$$v_0 \sim v_{00} + \frac{\rho_0 \eta}{4 \epsilon_0 v_{00}} r^2, \quad (4)$$

where v_{00} is the value of v_0 on the axis and r is distance from the axis.

For a two-dimensional "ribbon" beam of infinite extent in the y direction, the solution of Poisson's equation will depend on the potential of the electrodes bounding the beam in the x -direction. One possible solution is linear in x . This is particularly interesting, for it is the only distribution leading to a transformation enabling analytical solution.

Boundary Conditions

In any experimental space-charge amplifier the beam must be enclosed in metal walls. This is essential in order to prevent excessive potential depression due to space charge and to eliminate the effects of charge on insulating walls. Space-charge wave solutions for multiple-stream amplifiers show that the presence of metal walls very near the beam tends to inhibit the interaction of electrons near the wall. Furthermore, if the wall is removed by a distance greater than the order of the distance between bunches, its effect becomes negligible. There is no evidence that intermediate positions of the enclosing walls can produce any new effects.

The boundary conditions of (3) at metal walls are $E_z = 0$. At the center of an axially symmetrical system not containing a line charge on the axis, $\partial E_z / \partial r = 0$.

When the electron beam does not completely fill the space enclosed by the metal walls, an additional bound-

¹¹ W. C. Hahn, "Small signal theory of velocity-modulated electron beams," *Gen. Elec. Rev.*, vol. 42, pp. 258-270; June, 1939.

¹² S. Ramo, "The electronic-wave theory of velocity modulation tubes," *PROC. I.R.E.*, vol. 27, pp. 757-763; December, 1939.

¹³ Of course, relaxing the restraint on transverse velocities allows other types of interaction, some of which can give rise to gain. This is beyond the scope of this paper.

any condition must be satisfied at the boundary between the electron beam and free space. Outside of the beam $\rho_0 = 0$, making the solution much simpler in that region. It may be expressed as the sum of two linearly independent solutions, whose coefficients are determined by requiring that E_z and its transverse gradient be equal on each side of the boundary.

Cartesian Case

Ribbon Beam: If the electron beam in question is infinite in extent in the y direction, and has the linear velocity distribution described above, (3) will become

$$\frac{d^2 E_z}{dx^2} + \left(\gamma^2 + \frac{\omega^2}{c^2} \right) \cdot \left[1 + \frac{\rho_0 \eta / \epsilon_0}{\left[j\omega - \gamma \left(v_{00} + \frac{dv_0}{dx} x \right) \right]^2} \right] E_z = 0. \quad (5)$$

This assumes no y -components of field or velocity. The further assumption will be made that the wave velocity of a desired solution is much less than the velocity of light c . In that case:

$$\frac{\omega^2}{c^2} \ll |\gamma^2|,$$

and the ω^2/c^2 term may be dropped from the equation. A transformation, used first by MacFarlane and Hay¹⁴ in the solution of the crossed-field amplifier, may be applied. Let:

$$u = \frac{\omega + j\gamma v_0}{\frac{dv_0}{dx}}; \quad (6)$$

then (5) may be reduced to:

$$\frac{d^2 E_z}{du^2} - \left[1 - \frac{\omega_0^2}{\left(\frac{dv_0}{dx} \right)^2 u^2} \right] E_z = 0. \quad (7)$$

The ω^2/c^2 term has been dropped, and the notation $\omega_0 = \sqrt{\rho_0 \eta / \epsilon_0}$ used. ω_0 is commonly denoted the "plasma frequency."

It is a characteristic of the linear velocity distribution that $dv_0/dx = \omega_0$. Eq. (7) then reduces to the simple form:

$$\frac{d^2 E_z}{du^2} - \left(1 - \frac{1}{u^2} \right) E_z = 0 \quad (8)$$

free of explicit dependence on γ . The method of solution will be to first determine a solution satisfying the desired boundary conditions in the u variable, and then transform from u back to x , obtaining γ in the process.

The boundary conditions at a metal wall are $E_z = 0$.

For a beam not enclosed by conducting walls at its edges, the solution outside the beam must satisfy

$$\frac{d^2 E_z}{dx^2} + \gamma^2 E_z = 0.$$

Such solutions are $e^{\pm i\gamma x}$.

The solution for a beam in free space must vanish at infinity. Hence only the $e^{i\gamma x}$ term is allowable for large positive x , and $e^{-i\gamma x}$ for negative x . For positive x the matching condition

$$\frac{1}{E_z} \frac{dE_z}{dx} = +j\gamma$$

is replaced by

$$\frac{1}{E_z} \frac{dE_z}{du} = -1$$

in the u plane. On the negative x side of the beam,

$$\frac{1}{E_z} \frac{dE_z}{du} = +1$$

is the matching condition.

Since γ is not known, the mapping of x into the complex u plane is not known. Because (8) does not contain γ , it can be solved, two independent solutions being found. These solutions, combined linearly, encompass an entire class of physical problems, and by proper combination of the two solutions, each allowed γ for a particular set of parameters can be located.

Two values of u (u_A and u_B) must be found for which the boundary conditions at the beam edges (x_A and x_B) are satisfied; and for some complex γ and for given $v(x_A)$ and $v(x_B) < v(x_A)$, they must transform into real values of x . For given $v(x_A)$ and $v(x_B)$ there is a countable set of such u_A and u_B . The requirements on u_A and u_B are shown in Fig. 1 (opposite). Since u is linear in x and, more directly, in $v_0(x)$, the line joining u_A and u_B must pass through $u = \omega/\omega_0$, which for practical tubes is of the order of 10 or more. u_A and u_B must also be the proper distances from $u = \omega/\omega_0$ so that the ratio of these distances equals the ratio of v_A to v_B . A requirement for gain modes is that u_A to u_B fall in the first and second quadrants of the u plane, or, alternatively, in the fourth and third quadrants, so that the phase velocity of the wave will lie somewhere between v_A and v_B . This is a general requirement on growing waves produced by systems of different velocities.¹⁵

Under the boundary conditions required by conducting walls at x_A and x_B , we require the solution $E_z(u)$ to have zeros at u_A and u_B . If u_A and u_B satisfy the other stated conditions, $E_z(u)$ along the line joining u_A and u_B gives the solution along the corresponding segment of the x -axis. We now proceed to show that no such u_A and u_B giving rise to gain do exist.

¹⁴ G. G. Macfarlane and H. G. Hay, "Wave propagation in a slipping stream of electrons: small amplitude theory," *Proc. Phys. Soc.*, B. 63, pp. 409-426; June 1, 1950.

¹⁵ C.f., J. R. Pierce, "Coupling of mode of propagation," *Jour. Appl. Phys.*, vol. 25, pp. 179-183; February, 1954.

In order to find such u_A and u_B , $E_z(u)$ must be determined over a large part of the complex u plane. Let a particular pair of independent solutions of (8) be called $f_1(u)$ and $f_2(u)$. The general solution is $c[f_1(u) - \phi f_2(u)]$, in which c and ϕ are arbitrary complex numbers. We are only interested in the zeros of solutions, since u_A and u_B must be among them. All zeros of all linear combinations can be found by plotting f_1/f_2 . Choosing a point where $\phi = \phi_0$, we search for other points where $f_1/f_2 = \phi_0$. These points are all zeros of some particular solution of (8). Curves may be drawn (in the u plane) along which f_1/f_2 varies recurrently through the same set of values. These we call the loci of zeros of the solutions, and on them the next argument will be based.

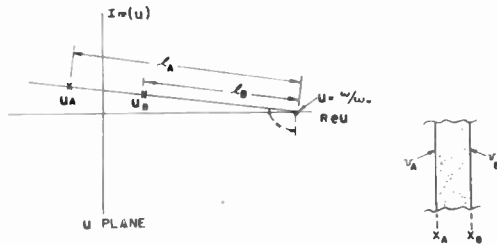


Fig. 1—Illustrating the relation two points u_A and u_B must satisfy to define a solution to (4) with zero boundary conditions. the relation $l_A/l_B = v_A/v_B$ must hold, and a line through u_A and u_B must intersect the real u axis at $u = \omega/\omega_0$. Then $|\gamma| = |l_A|/v_A = |l_B|/v_B$; $\arg \gamma = \alpha$.

Consider the equation

$$\frac{d^2 E_z}{du^2} + \frac{1}{u^2} E_z = 0. \quad (9)$$

For u sufficiently less than unity in absolute value, (8) approaches (9). Similarly, solutions of (8) become asymptotic to those of (9) in the same region.

Two solutions of (9) are

$$\sqrt{u} \sin\left(\frac{\sqrt{3}}{2} \ln u\right) \quad \text{and} \quad \sqrt{u} \cos\left(\frac{\sqrt{3}}{2} \ln u\right).$$

The zeros of these functions lie on the positive real u axis, and loci of zeros of arbitrary solutions of (9) in the u plane are radial lines.

If ju is substituted for u in (8), one obtains:

$$\frac{d^2 E_z}{d(ju)^2} + \left(1 + \frac{1}{(ju)^2}\right) E_z = 0. \quad (10)$$

For real ju , that is, imaginary u , the solutions of (10) will be oscillatory, with zeros lying on the imaginary u axis. The origin is a branch point and essential singularity, as in the asymptotic functions. The loci of zeros of arbitrary solutions of (10), which also qualify as zeros of solutions of (8), will lie as illustrated in Fig. 2. Since $f(-u)$, $f(u^*)$ and $f(u)$ all satisfy (8), all loci have counterparts in each quadrant.

The search for points which satisfy all requirements on u_A and u_B for gain is fruitless, for none of the loci illustrated in Fig. 2 could pass through two such points as are shown in Fig. 1. We conclude that no gain modes are possible.

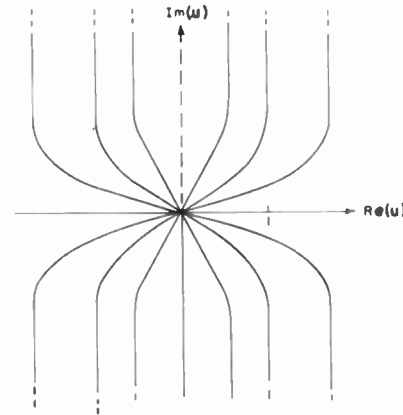


Fig. 2—Shapes of the loci of zeros of solutions to (4). Each radial line represents the locus of zeros of a set of independent solutions.

For the propagating modes which do exist, u_A and u_B lie on the real u axis, and must be spaced appropriately with respect to the ω/ω_0 point. Within the limitations of u_A and u_B can be found one infinite set of zeros lying on the positive real u axis, and one on the negative real u axis. These give rise to two sets of characteristic values of γ , one of which gives waves with phase velocity $> v_A$ and the other with phase velocity $< v_B$. No waves are found which have the velocity of any of the electrons. In the two lowest order modes E_z has no zeros between x_A to x_B . The next two modes, with velocities closer to electron velocities, have a zero of E_z between x_A and x_B . Such modes would be hard to detect, for most detectors (cavities, helices) indicate an integrated value of the field over the beam. The other modes are of successively higher order, and their only importance in practice is the power which they make unavailable for output. The modes of this "ribbon" beam have velocities similar to those calculated for a cylindrical beam in Fig. 3, on the following page.

It is interesting to postulate $(dv_0/dx) \neq \omega_0$, to determine whether a large artificially produced linear velocity distribution could give amplification. For smaller velocity gradients

$$\left(\frac{dv_0}{dx} < \omega_0\right),$$

the form of the solutions will be similar to those discussed above, except that the characteristic values of the propagation constant will be closer to one another.

If the gradient $(dv_0/dx) \geq 2\omega_0$, the solutions show only two possible values of γ , whose velocities correspond to the beam velocity at the two edges. For such high velocity gradients, the assumption of a plane wave solution is somewhat questionable. Particularly simple solutions are derived when $v_0' = 2\omega_0$, these being reducible to Bessel functions of zero order. Such solutions were examined in detail and found to produce no gain modes.

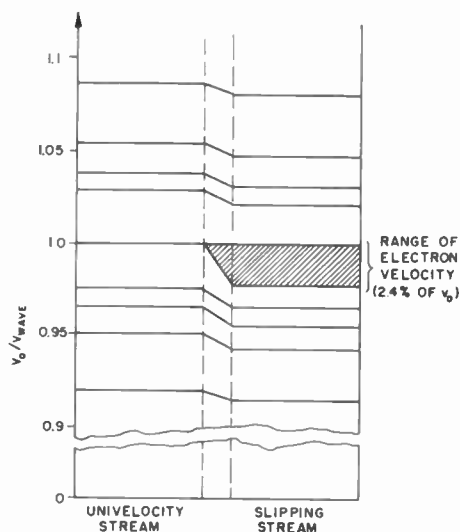


Fig. 3—Computed propagation constants of eight lowest modes of an electron beam with and without slip.

To obtain values of γ for a beam in free space, consider an arbitrary solution of (8), characterized by its locus of zeros. Its derivative dE_z/du also has zeros along approximately the same line (required by the approximate symmetry of the function near its locus of zeros). The function $(dE_z/du)/E_z$ has alternating zeros and poles along and near the locus.

Since E_z is real along its locus of zeros, the points at which the function $(dE_z/du)/E_z$ has values of ± 1 lies along a line which is very near a locus of zeros.

The same argument as before may be used to show the absence of gain modes.

The solution for the case of a beam enclosed by conducting walls at a distance is not readily visualized, and will not be discussed. It may be expected to yield the same results.

The case of an electron beam in which the (linear) velocity distribution has even symmetry about the center may be easily investigated, for the symmetry of velocity requires even or odd symmetry in the solution. For odd symmetry, the solution must be zero in the beam center, a simple boundary condition. For solutions with even symmetry, the first derivative must be zero at the center. Since the locus of zeros of dE_z/dx lies almost upon that of E_z , no new results would be expected.

Thus we have found that in the case of a "ribbon" beam, the only difference between a "slipping" beam and a beam with uniform velocity is the altering of

propagation constants of the space-charge waves. This effect is small, and no attempt was made to measure it in the experiments. It is perhaps harmful in the operation of very high current traveling-wave tubes, where bandwidth is already limited by high space-charge density, for the effect is similar to that of a higher space-charge density.

Cylindrical Beam—Computed Solutions: The more practical problem of a cylindrical, axially symmetrical beam with velocity distribution described by (4) was also attacked. All attempts at analytical solution having met with no success, the problem was presented to the Institute for Advanced Study Electronic Digital Computer.

The method chosen for preliminary analysis was to solve the appropriate form of (3),

$$\frac{d^2 E_z}{dr^2} + \frac{1}{r} \frac{dE_z}{dr} + \left(\gamma^2 + \frac{\omega^2}{c^2} \right) E_z = 0 \quad (11)$$

$$\left(1 + \frac{\omega_0^2}{[j\omega - \gamma v_0]^2} \right) E_z = 0$$

for chosen values of γ and attempt to determine values of γ satisfying the boundary conditions. A complete mesh survey of the γ plane required excessive time, while iteration and successive approximations might not converge correctly when handled by the computer. The chosen method consisted of forward integration of (11) satisfying the boundary conditions $E_z' = 0$ at $r = 0$, to the final value of r , at the conducting boundary. The value $E(\gamma, r_{\text{boundary}})$ was called $f(\gamma)$. It is zero only for γ satisfying the boundary conditions, and since the coefficients of (11) are regular except at $\gamma = j\omega/v_0$, the function of $f(\gamma)$ will be analytic everywhere in the finite plane except on portions of the imaginary axis.¹⁶ We are thereby led to the method of counting zeros of a complex function inside a closed contour by measuring the increase in $\arg f(\gamma)$ as γ traverses a chosen contour (Nyquist's criterion).

The net number of times $f(\gamma)$ encircles the origin, in the absence of singularities of the function within the contour, equals the number of zeros in the contour of γ . For the present problem, contours of γ are chosen such that all gain modes of interest would lie inside. These contours do not contain the imaginary axis of γ , where the singularities of the equation are known to lie. γ is traversed point-by-point, with a sufficiently fine mesh so that $f(\gamma)$ does not proceed more than 90 degrees between two successive points. A record is made when $f(\gamma)$ crosses the co-ordinate axes, with due regard to direction. When the contour is closed, four quantities are obtained, being the net number of times each of the four major axes was crossed in a specified direction. These should all be equal to the number of zeros in the contour. This provides a useful check of the method.

¹⁶ This quality of the solution results from a theorem of Fuchs; see, for example, E. T. Copson, "Introduction to the Theory of Functions of a Complex Variable," Oxford Univ. Press, Oxford, Eng., pp. 233-34, 1935.

In the light of the negative analytical and experimental results, the root-counting code was used only in broad sweeps of the γ plane; for practical constants it was confirmed that no roots were present. By confining the operation to purely imaginary values of γ , the propagational modes were computed. In Fig. 3 are shown some mode velocities calculated for a slipping cylindrical stream, as compared with mode velocities of an equivalent univelocity beam. The beam used in this calculation was 0.080 inch in diameter and completely filled its drift tube. Beam-center voltage was 100 volts and current 6 milliamperes.

EXPERIMENTAL INVESTIGATION

Although the analytical and computed results did not make the chance of finding gain experimentally seem optimistic, it was still felt that some gain phenomenon might arise from the complicated electron flow conditions arising in a practical tube. A demountable tube was constructed, the main features of which are illustrated in Figs. 4 and 5. The tube consists of an evacuated glass envelope, inside of which is a cylindrical brass frame. This frame acts as an aligning jig for the electron gun, the two resonant cavities, the collector, and a beam diameter measuring iris. The electron gun used throughout the investigation has an 0.050-inch diameter oxide cathode and produces a beam which can be readily confined to 0.060-inch diameter through the length of

rent on the diaphragm. Tuning and adjustment of the beam aperture is performed by means of a rod having a half-gear attached, which meshes with other gears in the tube to perform these functions. The liquid-air cooled vacuum system can maintain a pressure of 5×10^{-8} mm Hg.

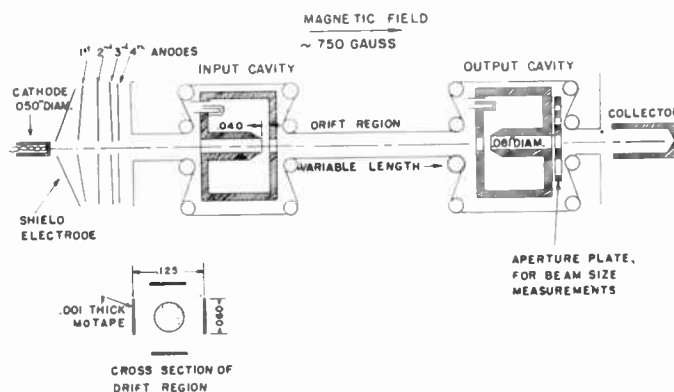


Fig. 5—Diagram of experimental demountable tube (not to scale).

The radio frequency system is sketched in Fig. 6. The loop in each resonant cavity is fed from the coaxial line through a nonmatched hermetic seal. The coaxial line is rigid, and emerges through an O-ring gland seal. Cavities are moved longitudinally by pushing and pulling on this coaxial line; coupling is adjusted by rotating the line, thereby rotating the loop inside the cavity. This has proved to be a satisfactory type of coupling for this type of work, in which little signal power is involved, and fairly large losses are tolerable.

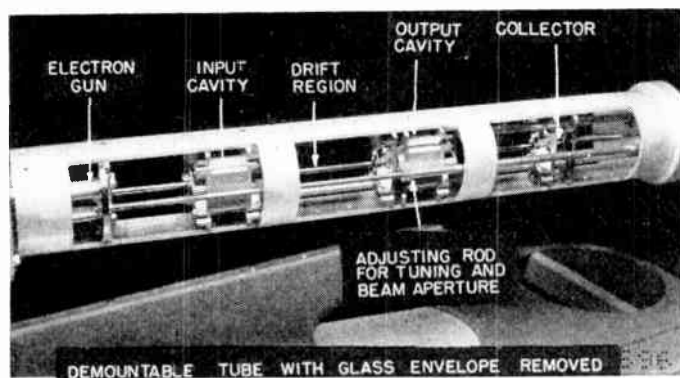


Fig. 4—Experimental demountable tube.

the tube, about fifteen inches from gun to collector. In most of the experiments, the electron beam was surrounded by a set of four molybdenum tapes, which formed a drift space of square cross section, 0.125-inch on a side. The tapes are 0.001-inch thick, and can be made to roll on rollers around the cavities. In this way cavity movement is possible without varying the drift region cross section.

Among the refinements in this demountable tube assembly are: cavities which can be tuned and matched from outside the vacuum, interchangeable collector assembly, and a diaphragm with apertures of graded diameter for measuring beam size by interception of cur-

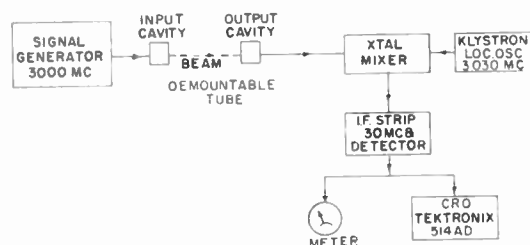


Fig. 6—Block diagram of radio frequency system.

The initial experiments were designed to determine if increasing waves could be set up in a slipping cylindrical stream of electrons. In order to insure that positive ion neutralization of the beam would not destroy the velocity difference, experiments with a continuous beam were compared with experiments in which the beam current was pulsed by application of one microsecond pulses to a gun electrode. In both these cases the measurement consisted of applying signal to the input cavity and observing the output voltage while the voltage of the entire cavity system and drift region was swept with

respect to that of the gun electrodes. Typical of the observed waveforms are those shown in Fig. 7. The curves on the right differ from those on the left only in that they contain the pulse edges; it is the envelope of the curve which is of consequence. Fig. 7 illustrates that there is no difference in the space-charge waves under pulsed or continuous operation.

COMPARISON OF SPACE CHARGE WAVES IN PULSED AND DC BEAMS. PULSE LENGTH 1μ SECOND; 5000 CYCLE RATE; CAVITY SPACING $11 \frac{3}{4}$ INCHES

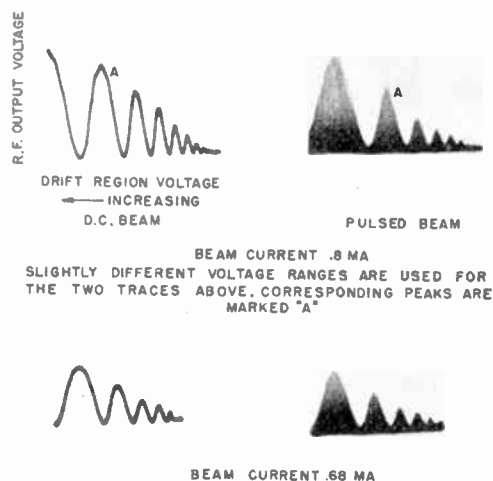


Fig. 7—Output voltage vs drift potential, for pulsed and for continuous beam. Pulse length, one microsecond; cavity spacing $11 \frac{3}{4}$ inches; beam current 0.8 ma. Corresponding peaks marked "A."

The curves obtained by sweeping cavity and drift-tube voltage, for a tube with infinitesimally short cavity gaps, are expected to be of the form $|\sin 2\pi z/\lambda_p|$, where z is the cavity spacing and λ_p the plasma wavelength of the space-charge waves. λ_p is a monotonically decreasing function of drift voltage. The curves shown have a decided decrease of amplitude at low voltages. This may be explained by the effect of transit time in the cavity gap. The curves of Fig. 7 are typical of results obtained with a two-cavity tube having no exponential gain; it is simply a low-level klystron.

To make a sensitive determination of the presence or absence of exponential gain, it is only necessary to make several such records with different cavity spacings. Any exponential gain will show up as a higher output at larger cavity spacing. Naturally, the number of maxima will increase as the spacing is increased, for each maximum is a measure of one plasma half-wavelength. If gain occurs as a result of space-charge depression of potential, the effect should be greater at low-drift voltages. This should show up as an increase in the height of the curve at the low-voltage end. In Fig. 8, showing curves taken at three-cavity spacings, the curve envelopes are all the same, indicating absence of exponential gain. The tube was operated over a wide

range of beam currents, up to four milliamperes, with similar results.

From the known beam diameter and velocity the theoretical velocity variation from beam center to edge was determined. The amount of gain produced by the

CAVITY OUTPUT vs. DRIFT REGION VOLTAGE FOR 3 DIFFERENT DRIFT LENGTHS. BEAM CURRENT 0.8 MA., FIXED R.F. INPUT

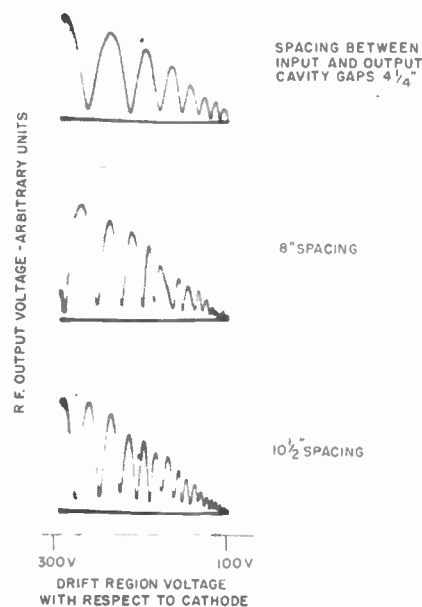


Fig. 8—Output voltage vs drift potential as a function of drift length. Beam current 0.8 am. Rf input fixed.

equivalent double-stream amplifier with half its electrons having the velocity of the beam center and the other half at the edge velocity would be, for a tube of the length used here, about 30 db, under typical voltage and current conditions. Gain in the electron beams observed, due to slipping, was definitely nonexistent.

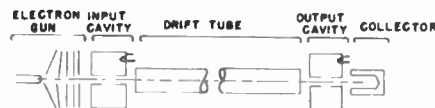


Fig. 9—Diagram of tube of the type used by Haeff.

In a second set of experiments, the drift region of the tube was altered so that it was somewhat similar to that described by Haeff¹ (see Fig. 9). The tape assembly was replaced by a metal cylinder of greater diameter which could be operated at a potential different from the cavity potential. Again, the potential of the drift region

was swept. Results were very similar to those of Figs. 7 and 8, except that the transit-time effect of the gaps could be eliminated by operating the cavities at a high, fixed potential.

It was found that with this configuration, curves such as shown in Fig. 10(b) could be obtained at very low-collector potential (50 to 150 volts). This may be interpreted as exponential gain, since the curve rises at low voltage, superimposing the gain on the usual oscillatory space-charge wave solution. The mechanism operating here could not have been observed by Haeff¹ because it requires a lower collector voltage than he employed. The gain observed at low-collector voltage was finally traced to secondary electrons emitted at low velocity from the collector, focused back through the cavities and drift tube, reflected from the region in front of the cathode, and refocused through the tube. These secondary electrons, having energy approximately equal to that of the primary beam less the collector potential, could not return to the cathode. With carefully adjusted gun voltages and collector voltage, the primary and secondary beams interacted as a double stream to produce an output above 30 db higher than the normal space-charge waves. The net gain of the device was never positive, but since the space-charge wave gain (klystron gain) of a tube can be greater than unity, the mechanism could produce net gain if used in a tube with better beam-to-cavity coupling.

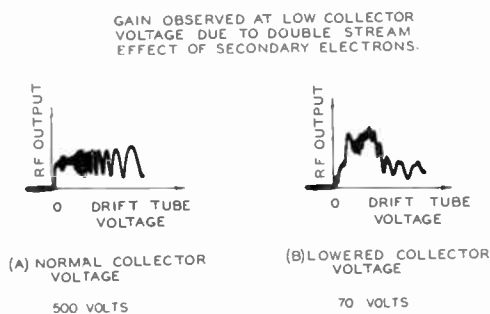


Fig. 10—Output voltage vs drift potential, illustrating gain obtained at lowered collector voltage.

Analysis was carried out to determine if the collector voltage, drift-tube voltage, and beam current were consistent with double-stream operation. Double-stream theory gives as the condition for maximum gain $(\omega_0/\omega)(v/\Delta v) = \text{a certain constant}$, where Δv is the difference of stream velocities. Here Δv would be approximately the collector voltage. This condition, written in terms of current and voltages, may be expressed at constant frequency as $IV^{3/2}_{\text{Dr. tube}}/V^2_{\text{collector}} = \text{constant}$. Data taken at maximum gain are presented in Fig. 11. The curves should be a family of parabolas. Anomalies in the data were due to multiple peaks in the gain maximum, which were, in turn, due to improper focusing at certain voltages.

The large drift tube was replaced once more by the tapes, and measurements proved that while some double-stream interaction was present, at the same voltage as before, the smaller cross section did not allow secondaries to return as readily and gain was reduced to the order of several decibels.

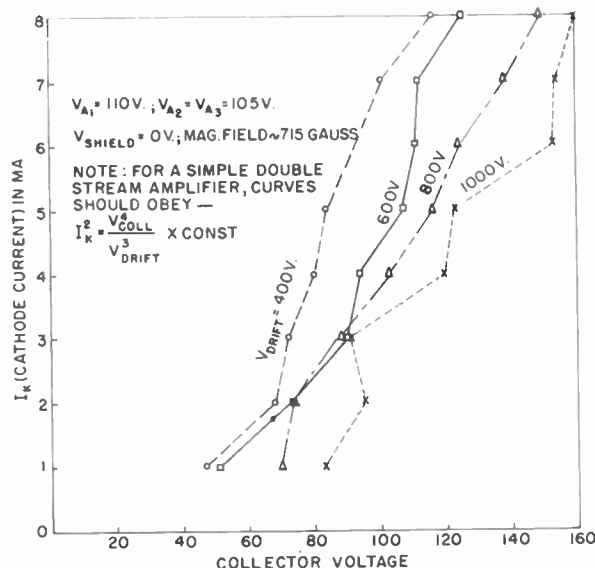


Fig. 11—Illustrating voltage and current conditions for maximum gain at lowered collector potential. Curves should be parabolic, to fit simple double-stream theory. Deviations near 3 and 5 ma are due to poor focusing.

While the experimental tube was operated in a manner which is typical of many beam-type tubes (confined flow, long-drift region, various drift-tube diameters), there are some established sources of gain which were not and could not have been observed. One of these is the growth of space-charge waves in a beam in perturbed Brillouin flow, as described by Bloom and Peter.¹⁷ This phenomenon could not have occurred because the beam was always confined by a strong magnetic field, with flux through the cathode. Another mechanism for single-stream gain, first disclosed by Field, Tien, and Watkins^{18,19} is the velocity-jump effect. This requires velocity changes of a certain type between the modulating and output cavities. Conditions in the measurements described never allowed any measurable gain of this type. A subsequent review of data taken by Haeff indicated that, in certain measurements, each of these phenomena could have resulted in the measured gain. There were also instances where it was possible that secondary electrons could have given double-stream gain similar to that produced by collector secondaries.

¹⁷ R. S. Peter, S. Bloom, and J. A. Ruetz, "Space-charge wave amplification along an electron beam by periodic change of the beam impedance," *RCA Rev.*, vol. 15, pp. 113-120; March, 1954.

¹⁸ L. M. Field, P. K. Tien, and D. A. Watkins, "Amplification by acceleration and deceleration of a single-velocity stream," *Proc. I.R.E.*, vol. 39, p. 194; February, 1951.

¹⁹ P. K. Tien and L. M. Field, "Space-charge waves in an accelerated electron stream for amplification of microwave signals," *Proc. I.R.E.*, vol. 40, pp. 688-695; June, 1952.

SUMMARY AND CONCLUSIONS

Analysis of a ribbon-shaped electron beam, and numerical computation of the space-charge waves on a cylindrical beam, have proved that the spatial distribution of electrons of different velocity is not sufficient to alter the criterion for gain found for a homogeneous mixture of electrons of different velocity. While the theory is admittedly limited to certain simple types of velocity distribution, these seem representative of practical situations. The space-charge wave solutions for beams with space-charge depression of potential were found to vary little from those of a univelocity beam. Measurements made on a tube similar to that described by the numerical computations yielded no evidence of gain, supporting the theoretical findings.

Under particular conditions of beam-focusing, it has been found that other mechanisms may give rise to gain: the presence of stray secondary electrons at appropriate velocity, the variations in beam diameter, or the existence of velocity changes in the interaction region.

ACKNOWLEDGMENT

The author wishes to thank Prof. J. Weber, University of Maryland, and Dr. L. Malter and many others at RCA Laboratories, whose advice and encouragement were of great aid. The numerical calculations were made possible by the co-operation of Dr. H. Goldstine of the Institute for Advanced Studies, Princeton, N. J., and were in part supported by a U. S. Army Ordnance Corps contract, held by the Institute.

RLC Lattice Transfer Functions*

A. D. FIALKOW† AND IRVING GERST‡

Summary—Necessary and sufficient conditions that a real rational function be the transfer function of an RLC symmetric lattice are derived. The range of allowable values of the multiplicative factor (which determines the gain level) in the transfer function is determined and an algorithm for its calculation is given. While the zeros of the transfer function may be anywhere in the complex number plane, the poles must lie in the left-half plane or its boundary excluding the values 0 and ∞ . However, at a pole of the transfer function on the pure imaginary axis, the function must satisfy certain further properties which are derived. In virtue of these latter conditions, it is found that there are realizable transfer functions which cannot be synthesized (even up to a multiplicative factor) by means of a symmetric lattice. These realizable transfer functions, which lie outside the symmetric lattice structure, always have some pure imaginary poles which do not satisfy our special conditions for lattice realizability. However, when all the conditions are met, a synthesis procedure for obtaining a corresponding lattice is given. The results are illustrated by an example.

1. INTRODUCTION

IN AN EARLIER paper [5],¹ we have obtained a complete theory for the transfer function of the general two terminal-pair network, either grounded or ungrounded, containing resistance, capacitance and self-inductance but no mutual coupling. If one considers two terminal-pair networks with these elements having a particular configuration, then the transfer function has further distinctive properties which are charac-

teristic of that structure. Because of the widespread use of the symmetric lattice in network theory, it is of interest to determine these characteristic properties for it. This is done in the present paper for the case of the open-circuit symmetric lattice driven by a zero impedance generator. However, the necessary conditions for lattice realizability developed below hold also in the case of a resistive source or a resistive load. Also, by a modification of the synthesis method given, it is usually possible to provide for either of these terminations.

We proceed to a statement of our principal result. As usual the transfer function $A(p)$ is defined as the ratio of steady-state output voltage to input voltage in the domain of the complex frequency variable p . Write

$$A(p) = KN/D \quad (1)$$

where $N = p^n + a_1 p^{n-1} + \dots + a_n$, $D = p^m + b_1 p^{m-1} + \dots + b_m$ are polynomials with real coefficients having no common factors, and K is a constant. It is known [5] that in order for $A(p)$ to be the transfer function of a general two terminal-pair network it is necessary that the poles of $A(p)$ be in the left-half plane² (l.h.p) or on its boundary excluding $p=0$ and $p=\infty$. Poles on the imaginary axis must be simple and have a pure imaginary residue. These then are also necessary conditions on the transfer function of a symmetric lattice. However, the

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¹ Numbers in brackets refer to the bibliography at end of paper.

² In this paper the abbreviations l.h.p. and r.h.p. will be used to denote respectively the interior of the left-half and the right-half planes. Reference to the boundary of these half-planes, whenever required, will always be made explicitly.

following further condition on the zeros and poles is also necessary in the case of a symmetric lattice:

(A) Define

$$X_h = \operatorname{Re} \left\{ j^h \frac{d}{dp} \left[\frac{1}{A} \right] \cdot \frac{d^h}{dp^h} \left[\frac{1}{A} \right] \right\}, \quad h = 2, 3, \dots$$

Then at each pure imaginary pole of $A(p)$ either all the values of X_h are zero or the first nonzero value of X_h occurs for h even and is negative.

As for the multiplicative constant K ,

(B) it is necessary that $-K_0 \leq K \leq K_0$ where the maximum gain constant K_0 is the smallest value of κ , $\kappa > 0$, such that at least one of the equations $D - (\kappa + \epsilon)N = 0$, $D + (\kappa + \epsilon)N = 0$ has a zero in the right-half plane for all small positive ϵ . An algorithm for the practical determination of K_0 is given in the proof. Conversely, a function $A(p)$ satisfying all of the above conditions may be realized as the transfer function of a symmetric lattice.

We find that our results are at variance in essential points with results obtained in papers by Kahal [6] and Weinberg [7]³ on the symmetric lattice. Reserving the detailed consideration of these matters for appropriate portions of the sequel, the following remarks may be made now. Kahal [6, p. 131] states that the transfer function of the general two terminal-pair network can always be realized to within a constant factor by the symmetrical lattice structure, while Weinberg [7, p. 427] states that he has realized any physically realizable RLC transfer function within a multiplicative constant by means of a symmetric lattice. Both of these statements are incorrect. (The same criticism applies to a recent abstract by Weinberg [8].)

It is found that there are physically realizable RLC transfer functions having some pure imaginary poles which cannot be realized by any symmetric lattice. As a simple counter-example, the physically realizable transfer function⁴ $A(p) = K(p^2 - 0.5p + 0.5)/(p^2 + 1) \cdot (p + 1)$ cannot be synthesized as a symmetric lattice for any $K \neq 0$. For applying (A) above to this function, we find that at the pole $p = j$ of $A(p)$, $X_2(j) = 64/K^2 > 0$. It therefore follows from (A) that this function is unrealizable by means of a symmetric lattice. Thus in so far as the transfer function is concerned, the symmetric lattice is limited as compared to the general two terminal-pair network not only in the maximum gain constant attainable (as was already noted in the RC-case [2, pp. 64-65; 4]) but also in the possible zero-pole combinations.

Coming to another matter, the lattice realization of a transfer function is often used as an intermediary in the practically desirable realization of this transfer function as a grounded network. We do not consider this conversion problem since it appears more natural and more

general to synthesize the transfer function directly by means of a grounded network whenever this is theoretically possible as was done in [5]. Further, it does not seem to be generally realized that in order for the conversion process to succeed, it is necessary, at the very least, that the transfer function satisfy the conditions for a grounded network [5, p. 118] e.g., the zeros cannot be positive real. It is thus apparent that there are cases in which the conversion of the lattice into a grounded structure is impossible.

2. THE BASIC THEOREM

Let $A(p)$ be given by (1). Without loss of generality, (by interchanging two terminals if necessary) we may assume $K > 0$. The following theorem enables us to characterize $A(p)$ completely.

Theorem 1: Necessary and sufficient conditions for $A(p) = KN/D$ to be the transfer function of a symmetric lattice are that the equations $D - \kappa N = 0$, $D + \kappa N = 0$ have no roots in the r.h.p. for all $0 < \kappa \leq K$.

Proof: (a) *Necessity.* Suppose $A(p)$ as given by (1) is the transfer function of a symmetric lattice whose constituent impedances are Z_a and Z_b . Then we have

$$A(p) = \frac{KN}{D} = \frac{Z_b - Z_a}{Z_b + Z_a}. \quad (2)$$

If we replace Z_a and Z_b in the lattice by new impedances Z_a' , Z_b' where

$$Z_a' = \left(\frac{1 + \lambda}{2} \right) Z_a + \left(\frac{1 - \lambda}{2} \right) Z_b$$

$$Z_b' = \left(\frac{1 - \lambda}{2} \right) Z_a + \left(\frac{1 + \lambda}{2} \right) Z_b,$$

with $0 < \lambda \leq 1$, we get a new transfer function $A_1(p)$ such that

$$A_1(p) = \frac{Z_b' - Z_a'}{Z_b' + Z_a'} = \frac{\lambda(Z_b - Z_a)}{Z_b + Z_a} = \lambda K \frac{N}{D}. \quad (3)$$

Write $\kappa = \lambda K$ so that $0 < \kappa \leq K$. Then it follows from (3) that

$$\frac{D - \kappa N}{D + \kappa N} = \frac{Z_a'}{Z_b'}. \quad (4)$$

Since the polynomials $D - \kappa N$ and $D + \kappa N$ can have no common zeros, we conclude from (4) that their zeros must lie in the l.h.p. or on its boundary.

(b) *Sufficiency.* Suppose now that both $D - \kappa N = 0$ and $D + \kappa N = 0$ have no zeros in the r.h.p. for $0 < \kappa \leq K$. We shall find constituent impedances, Z_a and Z_b of a lattice having KN/D as its transfer function. First we prove that the rational function

$$\psi(p) = \frac{D - KN}{D + KN} \quad (5)$$

cannot assume a negative real value in the r.h.p. For suppose a value $p = p_1$ exists with $\operatorname{Re} [p_1] > 0$ such that

³ This paper is based on a chapter of [9] and was published on the recommendation of the I.R.E. Professional Group on Circuit Theory.

⁴ Cf. [5, pp. 125-126] where this function with $K=2$ is synthesized as a grounded, two terminal-pair network. Using an alternative technique [5, p. 125] which is valid for sufficiently small K , this function with $K=2/3$ may be synthesized as a two terminal-pair network containing 8 elements.

$$\operatorname{Re} [\psi(p_1)] < 0, \quad \operatorname{Im} [\psi(p_1)] = 0. \quad (6)$$

Now⁵

$$\operatorname{Re} [\psi(p_1)] = \frac{|D(p_1)|^2 - K^2 |N(p_1)|^2}{|D(p_1) + KN(p_1)|^2}, \quad (7)$$

$$\operatorname{Im} [\psi(p_1)] = \frac{K}{j} \frac{[\overline{N}(p_1)D(p_1) - N(p_1)\overline{D}(p_1)]}{|D(p_1) + KN(p_1)|^2}. \quad (8)$$

Since $D(p_1) \neq 0$, it follows from (6) and (8) that $N(p_1)/D(p_1) = \overline{N}(p_1)/\overline{D}(p_1)$, which shows that $N(p_1)/D(p_1)$ is real. Hence from (6) and (7)

$$1 - K^2 \left[\frac{N(p_1)}{D(p_1)} \right]^2 < 0.$$

But then for $\kappa = |D(p_1)/N(p_1)|$, the inequalities $0 < \kappa < K$ hold, but we have $D^2(p_1) - \kappa^2 N^2(p_1) = 0$ which contradicts the first sentence of this paragraph. This proves that $\psi(p)$ cannot be negative real in the r.h.p.

We now make use of a theorem which we state here but whose proof is deferred until Appendix I.

Theorem 2: Let $F(p)$ be a real rational function which does not assume any negative real values in the r.h.p. and let $Z(p)$ be a positive real rational function such that $\operatorname{Re}[ZF] \geq 0$ on $p = j\omega$. Then ZF is a positive real function.

In our application of this theorem we take F as ψ defined by (5) and we form a particular Z as follows: Let the imaginary part⁶ of ψ on $p = j\omega$ be denoted by $V(\omega)$. Since $V(\omega)$ is an odd function of ω , it may be written in the form

$$V(\omega) = \epsilon \omega (\delta_1^2 - \omega^2) \cdots (\delta_r^2 - \omega^2) V'(\omega^2),$$

where $0, \pm\delta_1, \pm\delta_2, \dots, \pm\delta_r$ are all the real zeros and poles of odd multiplicity of $V(\omega)$, arranged so that $0 < \delta_1^2 < \delta_2^2 < \dots < \delta_r^2$. Then ϵ may be chosen as either $+1$ or -1 such that $V'(\omega^2) \geq 0$ for all ω . Depending on whether $\epsilon = +1$ or -1 , take Z respectively as $(p^2 + \delta_1^2) \cdot (p^2 + \delta_3^2) \cdots / p(p^2 + \delta_2^2) \cdots$ or the reciprocal of this function. One may readily verify that $\operatorname{Re}[Z\psi] \geq 0$ when $p = j\omega$ so that $Z\psi$ is a positive real function by Theorem 2. If the impedances of the lattice are now chosen as $Z_a = Z\psi$, $Z_b = Z$ then according to (2) and (5), the required transfer function KN/D is obtained. This completes the proof of Theorem 1.

The reactance Z which appears above was first defined by Kahal [6]. It is used here to wind up the proof of Theorem 1 in economical fashion. However its use may result in complicating the network realization by introducing certain redundant factors in Z_a . It may be shown (but we omit the proof here) that in general we may find Z_a and Z_b containing no redundant factors and neither one restricted to be a reactance such that $\psi = Z_a/Z_b$. By this means it is frequently possible to provide for a resistive source or a resistive load.

⁵ A bar denotes the conjugate complex number.

⁶ We assume that $V(\omega) \neq 0$. The case in which $V(\omega) \equiv 0$ is considered in Appendix II.

3. THE ZEROS AND POLES OF LATTICE TRANSFER FUNCTIONS

All the properties characterizing the transfer function of the symmetric lattice which are listed in the introduction may be derived from Theorem 1. However, as some of these have already been established for general two terminal-pair networks in [5], we limit ourselves to deriving the remaining ones, (A) and (B) of the introduction, which apply specifically to the symmetric lattice. In this section, we prove (A) reserving the proof of (B) for the following section.

It follows from Theorem 1 that if $A(p) = KN/D$ is to be the transfer function of a lattice for some K (i.e. up to a multiplicative constant) then for all small κ , $\kappa > 0$, no zero of either $D - \kappa N$ or $D + \kappa N$ can be in the r.h.p. If D is properly Hurwitz, then this follows automatically by continuity considerations; for any zero of D in the l.h.p. will vary into a zero of $D \pm \kappa N$ which is also in the l.h.p. if κ is small enough. However, if D has zeros on the imaginary axis a further investigation is required. This case is not considered by Weinberg [7], leading him to the error mentioned in the introduction.

Coming to the proof of (A), we consider a pure imaginary zero of D , $p = j\omega_0$, $\omega_0 \neq 0$. Consider the equation $D(p) - \kappa N(p) = 0$ which defines p implicitly as a function of κ . Since $p = j\omega_0$ is a simple root of this equation when $\kappa = 0$, we may expand p as a power series in κ for sufficient small κ , of the form

$$p = j\omega_0 + \alpha_1 \kappa + \alpha_2 \kappa^2 + \cdots \quad (9)$$

The corresponding root of $D(p) + \kappa N(p) = 0$ has an expansion which is obtained from (9) by replacing κ by $-\kappa$. Hence we require that for all small real κ the right member of (9) have a nonpositive real part. Necessary and sufficient for this to occur are that either⁷

- (i) $\operatorname{Re}(\alpha_h) = 0$ for all h .
- (ii) If α_h is the first coefficient in (9) whose real part is not zero, then h must be even and $\operatorname{Re}(\alpha_h) < 0$.

It will be shown in Appendix III that these conditions (i), (ii) on the coefficients of (9) are equivalent to condition (A) of the Introduction. (See also the last sentence of the next section.)

In drawing the erroneous conclusion that $A(p)$ may always be realized by a lattice for sufficiently small K , Kahal [6, p. 132] incorrectly assumes that $\operatorname{Im}[A(p)]$ in the neighborhood of $p = j\omega_0$ may always be approximated by the imaginary part of the first term in the Laurent expansion of $A(p)$. This error is equivalent to the supposition that the terms $\alpha_2 \kappa^2 + \alpha_3 \kappa^3 + \cdots$ of (9) may be ignored in our present discussion.

4. THE MAXIMUM GAIN CONSTANT

In this section, it is assumed that D and N meet the conditions for realizability with small K . We now consider the maximum gain constant K_0 which may be

⁷ It may be shown that (i) obtains if and only if $A(p)$ is an even function of p , i.e. we are in the LC-case.

realized. This question did not arise in previous RLC lattice synthesis techniques which realize the transfer function up to a multiplicative constant.⁸ While it simplifies the problem, this commonly used but artificial restriction sidesteps a basic question; for with transfer functions, unlike impedances, the level of response is an intrinsic quantity if transformers are excluded. An amplifier to boost the level may be unnecessary practically and in any case is a device which is alien to passive network theory.

According to Theorem 1, K_0 must have the property that $D - \kappa N$ and $D + \kappa N$ have no zeros in the r.h.p. for $0 < \kappa \leq K_0$, while at least one of $D - (K_0 + \epsilon)N$, $D + (K_0 + \epsilon)N$ has a zero in the r.h.p. for any ϵ sufficiently small and positive. This description of K_0 immediately suggests a way of determining it. For in order to pass from l.h.p. to r.h.p., the zeros of $D \pm \kappa N$ as κ increases positively, must first cross the imaginary axis. Hence K_0 must be one of the positive values of κ for which $D \pm \kappa N = 0$ has some pure imaginary roots.

We assume now that we are not in the LC-case,⁹ i.e., N/D is not an even function of p . If and only if this is so, there are just a finite number of values κ for which $D \pm \kappa N = 0$ has pure imaginary roots. If $m = n$ in (1), then $p = \infty$ is a root of $D - \kappa N = 0$ corresponding to $\kappa = 1$. The values of κ which correspond to finite pure imaginary roots of $D \pm \kappa N = 0$ are obtained by solving the simultaneous system

$$\begin{aligned} D_e \pm \kappa N_e &= 0 \\ D_o \pm \kappa N_o &= 0, \end{aligned} \quad (10)$$

where N_e , D_e and N_o , D_o are respectively the even and odd power terms of N and D . One root of (10) is always $p = 0$, which corresponds to the value¹⁰ $\kappa = \lfloor b_m/a_n \rfloor$. From [5] we know that $K_0 \leq \text{Min}(1, \lfloor b_m/a_n \rfloor)$ if $m = n$ and $K_0 \leq \lfloor b_m/a_n \rfloor$ if $m > n$. Let $0 < \kappa_1 < \kappa_2 < \dots < \kappa_s$ be those solutions of the system (10) (if any such exist) for which $\kappa_s < \text{Min}(1, \lfloor b_m/a_n \rfloor)$ if $m = n$ and $\kappa_s < \lfloor b_m/a_n \rfloor$ if $m > n$. Starting with κ_1 and proceeding to each κ_i ($i = 1, 2, \dots, s$) in turn, we must test to see whether any zeros of $D \pm (\kappa_i + \epsilon)N$ are in the r.h.p. for small positive ϵ . If κ_M is the first κ_i for which this occurs, then $K_0 = \kappa_M$. If there are no κ_i with this property then $K_0 = \text{Min}(1, \lfloor b_m/a_n \rfloor)$ if $m = n$ and $K_0 = \lfloor b_m/a_n \rfloor$ if $n < m$.

To carry out the above test we need examine only the variation of the pure imaginary zeros of $D \pm \kappa_i N$ when κ_i is replaced by $\kappa_i + \epsilon$. For example, let $p = j\omega_a$ be a zero of $D - \kappa_a N$ where κ_a is one of the above κ_i , and where for $0 < \kappa < \kappa_a$, all the zeros of $D \pm \kappa N$ have been in the l.h.p. (or its boundary). It follows using Lemma 1(ii) of

Appendix I (applied to ψ of (5) with $K = \kappa_a$) that $p = j\omega_a$ can be either a simple zero or a double zero of $D - \kappa_a N$. Using (vi) of the same lemma it can be proved that if $p = j\omega_a$ is a double zero of $D \pm \kappa_a N$, then κ_a must be κ_M defined above.

Now consider the remaining case in which $p = j\omega_a$ is a simple zero of $D - \kappa_a N$. Then in a procedure similar to that used in the preceding section, we may expand the solution p of $D(p) - \kappa N(p) = 0$ in the neighborhood of $p = j\omega_a$ as a Taylor series in $\kappa - \kappa_a$ of the form

$$p = j\omega_a + \alpha_1(\kappa - \kappa_a) + \alpha_2(\kappa - \kappa_a)^2 + \dots \quad (11)$$

Recalling that p is in the l.h.p. for $\kappa - \kappa_a$ small and negative, it follows that p will be in the r.h.p. for $\kappa - \kappa_a$ small and positive if and only if the first coefficient α_h whose real part is not zero has h odd.¹¹ The results of Appendix III may now be applied to express this condition directly in terms of the transfer function. It follows from (13) of this Appendix and the argument preceding (13) there, that if we define a sequence W_h ($h = 1, 2, \dots$) where

$$\begin{aligned} W_h &= \text{Re} \left\{ \frac{d^h}{dp^h} \left[\frac{1}{A} \right] \right\}, \quad h \text{ odd}; \\ W_h &= \text{Im} \left\{ \frac{d^h}{dp^h} \left[\frac{1}{A} \right] \right\}, \quad h \text{ even}, \end{aligned}$$

then at least one root of $D - (\kappa_a + \epsilon)N = 0$ will be in the r.h.p. if and only if the first nonzero value of $W_h(j\omega_a)$ occurs for h odd. All of the above results apply equally well if $p = j\omega_a$ is a zero of $D + \kappa_a N$.

The preceding conditions for determining κ_M when $p = j\omega_a$ is a simple zero of $D \pm \kappa_a N$ are stated directly in terms of the transfer function. We now derive an alternate set of conditions in this case which are given indirectly, but whose use requires a minimum of computation. Let the series

$$\kappa = \kappa_a + \beta_1(p - j\omega_a) + \beta_2(p - j\omega_a)^2 + \dots$$

be the inverse of (11), i.e., the expansion of $\kappa = D/N$ at $p = j\omega_a$. Then, when $p = j\omega$, we have

$$\text{Im} \left[\frac{D}{N} \right] = \text{Re} (\beta_1)(\omega - \omega_a) - \text{Im} (\beta_2)(\omega - \omega_a)^2 + \dots$$

Since

$$\beta_i = \frac{1}{i!} \left[\frac{d^i \kappa}{dp^i} \right]_{\kappa = \kappa_a},$$

(13) of Appendix III together with the above italicized conditions on the α_h may be used to show that if $\kappa_a \neq \kappa_M$ then ω_a is an even order zero of $\text{Im} [D/N]$. Since

$$\text{Im} \left[\frac{D}{N} \right]_{p=j\omega} = \frac{1}{i} \left[\frac{N_e D_o - N_o D_e}{|N|^2} \right]_{p=j\omega},$$

⁸ The maximum gain constant for RC lattice networks has been determined (Cf. [1; 3, p. 58]). On the other hand Weinberg's generalization [7, §3] of [1] for RLC lattice networks yields a constant which is not the true maximum gain constant K_0 . Thus, for example, if $A(p) = K/(p^2 + 0.1p + 1)$, then $K_0 = 1$, but the maximum value possible by Weinberg's method of §3 is $K = 1/800$.

⁹ For the determination of K_0 in the LC-case, see Appendix II.

¹⁰ We assume here that a_n is different than zero. If $a_n = 0$, the ratio b_m/a_n can be interpreted as ∞ in our results.

¹¹ The existence of a coefficient α_q with $\text{Re} (\alpha_q) \neq 0$ is assured here. For if $\text{Re} (\alpha_i) = 0$ for all i , then $A(p)$ is again an LC-transfer function which we have excluded from this discussion.

it follows that for $j\omega_a$ a simple zero of $D \pm \kappa_a N$, p of (11) will be in the r.h.p. for $\kappa - \kappa_a$ small and positive if and only if $j\omega_a$ is an odd order zero of the polynomial $N_e D_0 - N_0 D_e$.

Summarizing the preceding discussion we have the following procedure for determining K_0 : (a) Solve the system (10) for all pairs ($p=j\omega_i$, $\kappa=\kappa_i$), ω_i real and $\neq 0$, $\kappa_i > 0$. By eliminating κ in (10), the values $p=j\omega_i$ may be determined first as the pure imaginary zeros of the polynomial $Q = N_e D_0 - N_0 D_e$ and the κ_i obtained subsequently by substituting in (10). (b) Denote by κ'_i those values of κ_i for which either $p=j\omega_i$ is a double root of $D \pm \kappa N = 0$ or $p=j\omega_i$ is an odd order zero of Q . (c) Then $K_0 = \text{Min} [1, \lfloor b_m/a_n \rfloor, \kappa'_i]$ if $n=m$ or $K_0 = \text{Min} [\lfloor b_m/a_n \rfloor, \kappa'_i]$ if $n < m$.

By employing an argument similar to the one just used, we can also show that condition (A) of the Introduction which was proved in the preceding section, may be replaced by the following condition which is simpler to apply computationally: Let $p=j\omega_0$ be a pole of N/D with residue $j\gamma$, γ real, $\neq 0$. Then for realizability we must have $[N_e(j\omega)D_0(j\omega) - N_0(j\omega)D_e(j\omega)]/j = a(\omega - \omega_0)^r + \dots$, $a \neq 0$ with r even and $a\gamma < 0$.

5. EXAMPLE

The following example will serve to illustrate the foregoing procedure. Consider the realizable transfer function

$$A(p) = \frac{KN}{D} = K \left(p^3 + \frac{5}{3}p^2 + \frac{5}{3}p + 1 \right) / \left(p^4 + \frac{5}{4}p^3 + \frac{11}{4}p^2 + \frac{7}{4}p + \frac{5}{4} \right).$$

We first determine its maximum gain constant K_0 , when it is synthesized as a symmetric lattice. The final criterion of the previous section for determining K_0 will be applied. Since here

$$Q = N_e D_0 - N_0 D_e = -p(p^2 + \frac{1}{3})(p^2 + 1)^2,$$

we have the pure imaginary zeros $p = \pm j$ and $p = \pm j\sqrt{3}/3$ which may be verified to be simple zeros of $D - \kappa N$ for $\kappa = 3/4$, $\kappa = 1$ respectively. As just $p = \pm j\sqrt{3}/3$ are odd order zeros of Q (for $p \neq 0$), it follows that the only κ' is $\kappa' = 1$. Hence, since $m = 4 > 3 = n$, $b_m = 5/4$, $a_n = 1$, we have $K_0 = \text{Min} [5/4, 1] = 1$.

The synthesis of $A(p)$ when $K = 1$ may now be accomplished by the method following Theorem 2. For the function ψ given by (5) is here

$$\psi = \left(p^4 + \frac{1}{4}p^3 + \frac{13}{12}p^2 + \frac{1}{12}p + \frac{1}{4} \right) / \left(p^4 + \frac{9}{4}p^3 + \frac{53}{12}p^2 + \frac{41}{12}p + \frac{9}{4} \right) = N_1/D_1.$$

Then we find that $V = \text{Im} [\psi(j\omega)]$ is

$$V = \frac{-2\omega(\frac{1}{3} - \omega^2)(1 - \omega^2)^2}{|D_1(j\omega)|^2} = \epsilon\omega(\frac{1}{3} - \omega^2)V'(\omega^2),$$

where $\epsilon = -1$ and $V'(\omega^2) = (1 - \omega^2)^2 / |D_1(j\omega)|^2$. Hence

$$Z_b = Z = \frac{p}{p^2 + \frac{1}{3}};$$

$$Z_a = Z\psi = \frac{p(p^2 + \frac{1}{4}p + \frac{3}{4})}{p^4 + \frac{9}{4}p^3 + \frac{53}{12}p^2 + \frac{41}{12}p + \frac{9}{4}}.$$

APPENDIX I

Proof of Theorem 2

We require two preliminary lemmas for the proof of Theorem 2.

Lemma 1: Let $F(p)$ be a real rational function which does not assume any negative real value in the r.h.p. Further let $U(\omega)$ and $V(\omega)$ denote respectively the real and imaginary parts of F on the imaginary axis $p=j\omega$. Then F , U and V enjoy the following properties:

- (i) F is analytic in the r.h.p. and has no zeros there.
- (ii) On the imaginary axis $p=j\omega$, F can have at most second order zeros and poles.

The remaining properties all concern the behavior of F , U and V at a point $p=p_0=j\omega_0$ on the imaginary axis, the indicated expansions being the Laurent expansions at this point.

- (iii) If $F = a + \dots$, $V = a_2(\omega - \omega_0)^{2q+1} + \dots$, where $aa_2 \neq 0$, $q \geq 0$, then $a > 0$.
- (iv) If $F = a(p - p_0) + \dots$, $V = a_2(\omega - \omega_0)^{2q+1} + \dots$, where $aa_2 \neq 0$, $q \geq 0$, then $a_2 > 0$.
- (iv)' If $F = a/(p - p_0) + \dots$, $V = a_2(\omega - \omega_0)^{2q-1} + \dots$, where $aa_2 \neq 0$, $q \geq 0$, then $a_2 < 0$.
- (v) If $F = a(p - p_0) + \dots$, $V = a_2(\omega - \omega_0)^{2q} + \dots$, where $aa_2 \neq 0$, $q \geq 1$, then $U = a_1(\omega - \omega_0) + \dots$ with $a_1 a_2 < 0$.
- (v)' If $F = a/(p - p_0) + \dots$, $V = a_2(\omega - \omega_0)^{2q} + \dots$, where $aa_2 \neq 0$, $q \geq 0$, then $U = a_1/(\omega - \omega_0) + \dots$ with $a_1 a_2 > 0$.
- (vi) If $F = a(p - p_0)^2 + \dots$, where $a \neq 0$ and $V(\omega) \neq 0$, then $a > 0$ and $V = a_2(\omega - \omega_0)^{2q+1} + \dots$, with $a_2 > 0$, $q \geq 1$.
- (vi)' If $F = a/(p - p_0)^2 + \dots$, where $a \neq 0$ and $V(\omega) \neq 0$, then $a > 0$ and $V = a_2(\omega - \omega_0)^{2q-1} + \dots$, with $a_2 < 0$, $q \geq 0$.

Remarks: This same subject has already been studied by Kahal [6, Appendix I]. In his discussion, (i) is incorporated as part of the definition of F , (ii) is stated and proved, (iii) and part of (vi) are stated without complete proof. In place of (iv) and (v), he attempts to prove a result [6, p. 132] which we may paraphrase as follows: "If $F = a(p - p_0) + \dots$, $V = a_2(\omega - \omega_0)^q + \dots$, where $aa_2 \neq 0$, then either $q = 1$ and $a_2 > 0$, or else q is even." The latter conclusion is false as may be seen by taking F equal to the product $Z_1 Z_2$ where Z_1 is a reactance function such that $Z_1(j\omega) = jX_1(\omega)$, $X_1(\omega_0) = 0$, and Z_2 is an impedance function such that $Z_2(j\omega) = R_2(\omega) + jX_2(\omega)$, $R_2(\omega_0) = 0$, $X_2(\omega_0) \neq 0$. Then $V = R_2(\omega)X_1(\omega)$ has an odd order zero of at least third order at $\omega = \omega_0$. For example, let $Z_1 = (p^2 + 1)/p$, $Z_2 = (2p^2 + p + 1)/(p^2 + p + 2)$

and consider $p=j$. This mistake vitiates Kahal's proof for the synthesis of a lattice when $F=\psi(p)$ defined by (5) has simple zeros or poles on the boundary.

Proof: It suffices to prove the above properties as regards the zeros of F . The statements about the poles then follow by considering $1/F$ which again assumes no negative real value in the r.h.p. Our proof will depend upon the following well-known mapping property of analytic functions:

(M) Let $z=f(p)$ be analytic at $p=p_0$, and have the expansion

$$z = z_0 + c(p - p_0)^t + \dots, \quad c \neq 0, t \geq 1$$

there. Then the map of the sector $S: \theta_1 \leq \arg(p - p_0) \leq \theta_2$ for p in the neighborhood of p_0 , will completely cover a curvilinear sector $S': \phi_1 \leq \arg(z - z_0) \leq \phi_2$ for $|z - z_0|$ sufficiently small. If α and β designate the rays $\arg z = \theta_1$, $\arg z = \theta_2$ and if α' and β' are respectively the maps of α and β , then ϕ_1 and ϕ_2 are the variable angles given by $\phi_1 = \arg \alpha'$, $\phi_2 = \arg \beta'$. Furthermore, the angle between β' and α' at $z = z_0$ is $t(\theta_2 - \theta_1)$.

We now proceed to prove the various statements of the lemma seriatim. If now F were zero at $p=p_i$ in the r.h.p. then by (M) with $f(p)=F(p)$, $\theta_1=0$, $\theta_2=2\pi$, the map of a neighborhood of p_i would cover a neighborhood of the origin in the z -plane. Thus for some point in the r.h.p. F would be negative real which is a contradiction. This proves (i).

In proving the remaining properties we shall apply (M) with $f=F$ and with S as a semi-circular neighborhood $-\pi/2 \leq \arg(p - p_0) \leq \pi/2$, so that α and β are respectively the lower and upper halves of the imaginary axis. Suppose now contrary to (ii) that F has a zero of order t , $t \geq 3$ at $p=j\omega_0$. Then the curvilinear sector S' will have an angle of $t\pi \geq 3\pi$ at $z=0$, and thus surely contains negative real points. This contradiction proves (ii).

To prove (iii) suppose $a < 0$. Then in view of the expansion of V one of α' and β' will be in the upper half z -plane and the other in the lower half z -plane. Thus S' certainly contains a part of the negative real axis which is a contradiction.

Similarly if a_2 were negative in (iv), then α' would be in the upper half-plane and β' in the lower half-plane, so that S' again would include negative real points. This contradiction proves (iv).

To prove (v) suppose that $a_1 a_2 > 0$. Then it follows from the expansions of U and V that either α' is in the second quadrant and β' is in the first quadrant; or α' is in the fourth quadrant and β' is in the third quadrant. Either of these choices results in S' containing a negative real segment which is a contradiction.

Finally suppose in (vi) that $V = a_2'(\omega - \omega_0)^{2q} + \dots$, $a_2' \neq 0$, $q \geq 1$. Then α' and β' would both be in the upper half-plane or the lower half-plane (depending on the sign of a_2'). Since t in (M) is 2 here, the angle of S' is 2π . Thus S' must include a segment of the negative real axis. Hence ω_0 cannot be an even order zero of V and we have $V = a_2(\omega - \omega_0)^{2q+1} + \dots$, $a_2 \neq 0$, $q \geq 1$, and a is

real. By using the argument of (iv), a_2 must be positive. If now $a < 0$, then $U = -a(\omega - \omega_0)^2 + \dots$, and it follows that α' is in the fourth quadrant while β' is in the first quadrant. This again leads to a contradiction and the proof of the lemma is complete.

Lemma 2: Let $G(p)$ be a real rational function such that $\text{Re}[G(p)] \geq 0$ on $p=j\omega$. Then at a pole $p=p_0=j\omega_0$ of $G(p)$ of the first or second order respectively the following expansions hold.

- (i) $G(p) = c/(p - p_0) + \dots$, c real;
(ii) $G(p) = (c + dj)/(p - p_0)^2 + \dots$, $c \leq 0$, d real.

Proof: Suppose $G(p) = (c + dj)/(p - p_0)^q + \dots$, $q=1, 2$, where c and d are real. Then for $p=j\omega$

$$\text{Re}[G(p)] = \frac{\text{Re}[j^{-q}(c + dj)]}{(\omega - \omega_0)^q} + \dots$$

If $q=1$ then $d=0$, otherwise the real part will assume both positive and negative values in the neighborhood of ω_0 . If $q=2$, $\text{Re}[j^{-2}(c + dj)] = -c \geq 0$, to have $\text{Re}[G(p)] \geq 0$. This establishes Lemma 2.

We now consider the proof of Theorem 2. Let u , U' and v , V denote respectively the real and imaginary parts of Z and F for $p=j\omega$. Since $\text{Re}[ZF] \geq 0$ on $p=j\omega$ and ZF is analytic in the r.h.p. by Lemma 1 (i), we must show that on $p=j\omega$, ZF has at most poles of first order with positive residues. By Lemma 1 (ii) and the fact that Z is positive real, the orders of ZF , Z and F at a possible singularity $p=p_0=j\omega_0$ of ZF may be listed as follows:

	ZF	Z	F
(α_1)	-1	+1	-2
(α_2)	-1	0	-1
(α_3)	-1	-1	0
(β_1)	-2	0	-2
(β_2)	-2	-1	-1
(γ)	-3	-1	-2

To establish Theorem 2, we will show that cases β_1 , β_2 , γ are impossible while in cases α_1 , α_2 , α_3 , the residue of the pole of ZF is positive.

Case (α_1) : Here

$$Z = b(p - p_0) + \dots, \quad b > 0; \quad F = a/(p - p_0)^2 + \dots, \quad a > 0$$

by Lemma 1 (vi)' and the result is immediate.

Case (α_2) : Write

$$Z = b_1 + b_2 j + \dots, \quad b_1 \geq 0, \quad b_2 \text{ real}, \quad b_1^2 + b_2^2 \neq 0;$$

$$F = \frac{a_2 + a_1 j}{p - p_0} + \dots, \quad a_2 \geq 0, \quad a_1 \text{ real}, \quad a_1^2 + a_2^2 \neq 0,$$

where Lemma 1 (iv)' has been used to establish the sign of a_2 . Then

$$ZF = \frac{(a_2 b_1 - a_1 b_2) + (a_1 b_1 + a_2 b_2)j}{p - p_0} + \dots$$

Hence by Lemma 2 (i) with $G = ZF$,

$$a_1 b_1 + a_2 b_2 = 0. \quad (12)$$

If $b_1 a_2 \neq 0$ then we may use (12) to write the residue

$$a_2 b_1 - a_1 b_2 = a_2 b_1 + \frac{a_1^2 b_1}{a_2},$$

which shows the residue at $p = p_0$ is positive. If $b_1 = 0$ then by (12) $a_2 = 0$ and conversely. But then $a_1 b_2 \neq 0$. Now $\text{Re}_{p=j\omega}[ZF] = uU - vV$. We have $u = b_3(\omega - \omega_0)^{2k} + \dots$, $b_3 > 0$, $k \geq 1$, $V = a_3(\omega - \omega_0)^s + \dots$, $a_3 \neq 0$, $s \geq 0$. Hence

$$\begin{aligned} \text{Re}_{(p=j\omega)}[ZF] &= a_1 b_3 (\omega - \omega_0)^{2k-1} + \dots \\ &\quad - a_3 b_2 (\omega - \omega_0)^s + \dots \end{aligned}$$

If s is odd then $\text{Re}[ZF] \geq 0$ requires that $s = 2k - 1$ and $a_1 b_3 - a_3 b_2 = 0$. But then by Lemma 1 (iv)' $a_3 < 0$, and the residue $-a_1 b_2 = -a_3 b_2^2 / b_3 > 0$. If s is even then $s < 2k - 1$ and we must have $-a_3 b_2 > 0$. By Lemma 1 (v)', $a_1 a_3 > 0$. Hence $-a_1 b_2 > 0$. Thus in every case ZF has a positive residue.

Case (α_3): Let

$$Z = b/(p - p_0) + \dots, \quad b > 0; \quad F = a_1 + a_2 j + \dots$$

Then

$$ZF = (a_1 b + a_2 b j)/(p - p_0) + \dots,$$

and as in the preceding case we conclude that $a_2 b = 0$, which implies that $a_2 = 0$. If V has an odd order zero at $\omega = \omega_0$ then by Lemma 1 (iii)', $a_1 > 0$ and the residue $a_1 b > 0$. If $V = a_3(\omega - \omega_0)^{2q} + \dots$, $a_3 \neq 0$, $q \geq 0$ and writing $u = b_1(\omega - \omega_0)^{2k} + \dots$, $b_1 > 0$, $k \geq 0$ then

$$\begin{aligned} \text{Re}_{(p=j\omega)}[ZF] &= a_1 b_1 (\omega - \omega_0)^{2k} + \dots \\ &\quad + a_3 b (\omega - \omega_0)^{2q-1} + \dots \end{aligned}$$

We must have $2k < 2q - 1$ and $a_1 b_1 > 0$ otherwise $\text{Re}[ZF] \not\geq 0$. Thus $a_1 > 0$ and again the residue $a_1 b > 0$.

Case (β_1): We can write, using Lemma 1 (vi)'

$$Z = b_1 + b_2 j + \dots, \quad b_1 \geq 0, \quad b_2 \text{ real}, \quad b_1^2 + b_2^2 \neq 0;$$

$$F = \frac{a}{(p - p_0)^2} + \dots, \quad a > 0.$$

Then

$$ZF = (ab_1 + ab_2 j)/(p - p_0)^2 + \dots,$$

and the use of Lemma 2(ii) leads to $ab_1 \leq 0$. This implies $b_1 = 0, b_2 \neq 0$. By Lemma 1 (vi)' we have $V = a_2(\omega - \omega_0)^{2q-1} + \dots$, $q \geq 0$, $a_2 < 0$. If $u = b_3(\omega - \omega_0)^{2k} + \dots$, $b_3 > 0$, $k \geq 1$ then

$$\begin{aligned} \text{Re}_{(p=j\omega)}[ZF] &= -ab_3(\omega - \omega_0)^{2k-2} + \dots \\ &\quad - a_2 b_2 (\omega - \omega_0)^{2q-1} + \dots \end{aligned}$$

We must have $2k - 2 < 2q - 1$ and $-ab_3 > 0$. Thus case (β_1) is impossible.

Case (β_2): Here, using Lemma 1 (iv)'

$$Z = \frac{b}{p - p_0} + \dots, \quad b > 0;$$

$$F = \frac{a_2 + a_1 j}{p - p_0} + \dots, \quad a_2 \geq 0, \quad a_1 \text{ real}, \quad a_1^2 + a_2^2 \neq 0;$$

$$ZF = \frac{ba_2 + ba_1 j}{(p - p_0)^2} + \dots$$

As in the preceding case, $ba_2 \leq 0$ which implies $a_2 = 0$, $a_1 \neq 0$. Let

$$\begin{aligned} V &= a_3(\omega - \omega_0)^s + \dots, \quad a_3 \neq 0, \quad s \geq 0; \\ u &= b_1(\omega - \omega_0)^{2k} + \dots, \quad b_1 > 0, \quad k \geq 0. \end{aligned}$$

Then

$$\begin{aligned} \text{Re}_{(p=j\omega)}[ZF] &= a_1 b_1 (\omega - \omega_0)^{2k-1} + \dots \\ &\quad + a_3 b (\omega - \omega_0)^{s-1} + \dots \end{aligned}$$

If s is odd then $s - 1 < 2k - 1$ and $a_3 b > 0$. This is impossible, for by Lemma 1 (iv)', $a_3 < 0$. If s is even, then we must have $s - 1 = 2k - 1$ and $a_1 b_1 + a_3 b = 0$. Again this is impossible, for by Lemma 1 (v)', $a_1 a_3 > 0$.

Case (γ): By Lemma 1 (vi)', $Z = b/(p - p_0) + \dots$, $b > 0$; $F = a/(p - p_0)^2 + \dots$, $a > 0$.

Let $u = b_1(\omega - \omega_0)^{2k} + \dots$, $b_1 > 0$, $k \geq 0$. By Lemma 1 (vi)', $V = a_2(\omega - \omega_0)^{2q-1} + \dots$, $a_2 < 0$, $q \geq 0$ so that

$$\begin{aligned} \text{Re}_{(p=j\omega)}[ZF] &= -ab_1(\omega - \omega_0)^{2k-2} + \dots \\ &\quad + a_2 b (\omega - \omega_0)^{2q-2} + \dots \end{aligned}$$

This is impossible since both $-ab_1$ and $a_2 b$ are negative. Thus all the pure imaginary poles of ZF are simple and have positive residues. This completes the proof of Theorem 2.

APPENDIX II

LC Lattice Transfer Functions

Suppose $\text{Im}[\psi(j\omega)] \equiv 0$. Then using (8) with p_1 replaced by $j\omega$ we have $N(j\omega)/D(j\omega) \equiv \bar{N}(j\omega)/\bar{D}(j\omega)$ which implies that N/D is an even function of p . Thus $A = KN/D$ with N and D even polynomials, and where the zeros of D are pure imaginary and distinct. Hence A is an LC transfer function. The results for this case follow by paraphrasing the discussion of the RC case given in [3]. By using an argument similar to that in [3, pp. 56-58] with p replaced by p^2 , it follows that if $D \pm \kappa N$ has no zeros in the r.h.p. for $0 < \kappa \leq K$ then

$$\psi = \frac{D - \kappa N}{D + \kappa N} = \eta \frac{(p^2 + \xi_1^2)(p^2 + \xi_2^2) \dots}{(p^2 + \xi_1^2)(p^2 + \xi_2^2) \dots}, \quad \eta > 0,$$

where the ξ_i^2 and ζ_i^2 are non-negative, and the combined sequence¹² of ξ_i^2 and ζ_i^2 arranged in ascending order

¹² Kahl's statement [6, p. 124] of the order relations of the zeros and poles of ψ is weaker than the one given here and is insufficient to accomplish the factorization of ψ into reactance functions Z_a and Z_b despite his statement to the contrary. For example the function $(p^2 + 1)(p^2 + 2)/(p^2 + 3)(p^2 + 4)$ satisfies his order conditions but may not be factored as Z_a/Z_b .

of magnitude consists of consecutive pairs (ξ_i^2, ζ_i^2) or (ζ_i^2, ξ_i^2) which may occur in either order, except that the last term in the sequence may be an unpaired ζ^2 or ξ^2 . We may now choose¹³

$$Z_a = \lambda_1 p \frac{(p^2 + \zeta_2^2)}{(p^2 + \xi_1^2)} \cdot \frac{(p^2 + \zeta_4^2)}{(p^2 + \xi_3^2)} \cdots, \quad \lambda_1 > 0;$$

$$Z_b = \lambda_2 p \frac{(p^2 + \xi_2^2)}{(p^2 + \zeta_1^2)} \cdot \frac{(p^2 + \xi_4^2)}{(p^2 + \zeta_3^2)} \cdots, \quad \lambda_2 > 0,$$

with $\lambda_1 \lambda_2 = \eta$, as the required lattice impedances in (2).

The maximum gain K_0 is given [3, pp. 58, 72-73] as

$$K_0 = \text{Min} [1, |b_m/a_n|, K_d], \quad n = m$$

$$= \text{Min} [|b_m/a_n|, K_d], \quad n < m.$$

Here K_d is the smallest numerical value of κ for which the equation $D - \kappa N = 0$ has a multiple root. This means that K_d is the real root of smallest absolute value (if it exists) of the equation in κ obtained by equating the discriminant of $D - \kappa N$ to zero.

APPENDIX III

The Pure Imaginary Roots of $D(p) - \kappa N(p) = 0$

In §3 and §4 we were led to the consideration of the pure imaginary roots of $D(p) - \kappa N(p) = 0$ with $\kappa = 0$ and $\kappa = \kappa_a$ respectively. In this Appendix, we investigate these roots in detail and obtain the results which were used in these earlier sections. Let

$$p = j\omega_0 + \alpha_1(\kappa - \kappa_0) + \alpha_2(\kappa - \kappa_0)^2 + \cdots, \quad \alpha_1 \neq 0,$$

be a solution of the equation $D(p) - \kappa N(p) = 0$ for κ in the neighborhood of κ_0 . The above equation corresponds to (9) if κ_0 is chosen as $\kappa_0 = 0$ or to (11) if $\kappa_0 = \kappa_a$. The coefficients α_h are given by

$$\alpha_h = \frac{1}{h!} \left[\frac{d^h p}{d\kappa^h} \right]_{\kappa = \kappa_0}.$$

Write κ_h for $d^h \kappa / d p^h$ ($h = 1, 2, \cdots$). Then we have

$$\frac{d p}{d \kappa} = \frac{1}{\kappa_1}, \quad \frac{d^2 p}{d \kappa^2} = -\frac{\kappa_2}{\kappa_1^3},$$

$$\frac{d^3 p}{d \kappa^3} = -\frac{(\kappa_3 \kappa_1 - 3 \kappa_2^2)}{\kappa_1^5}, \quad \cdots$$

Thus $\text{Re} [\alpha_1] = \text{Re} [1/\kappa_1(j\omega_0)]$. If $\kappa_0 = 0$ (as in §3) then it follows from the fact that the residue of N/D at $p = j\omega_0$ is pure imaginary that $\kappa_1(j\omega_0)$ is pure imaginary

¹³ There are other choices for Z_a and Z_b as follows from the fact that the discussion in [3, p. 55] for the RC case may be made to apply here also.

and hence $\text{Re} [\alpha_1] = 0$. For other values of κ_0 this need not be so. However, in the case that $\text{Re} [\alpha_1] = 0$, we suppose now that $\text{Re} [\alpha_1] = \text{Re} [\alpha_2] = \cdots = \text{Re} [\alpha_{r-1}] = 0$, $\text{Re} [\alpha_r] \neq 0$, $r \geq 2$. Then it follows from the above equations that if $r = 2$, $\text{Re} [\alpha_2] = \text{Re} [-\kappa_2(j\omega_0)/\kappa_1^3(j\omega_0)]/2!$; if $r = 3$, then $\text{Im} [\kappa_2(j\omega_0)] = 0$, $\text{Re} [\alpha_3] = \text{Re} [-\kappa_3(j\omega_0)/\kappa_1^4(j\omega_0)]/3!$, \cdots .

In general, it may be shown using mathematical induction that if $1 \leq h < r$ then $\text{Re} [\kappa_h(j\omega_0)] = 0$ for h odd, $\text{Im} [\kappa_h(j\omega_0)] = 0$ for h even, and

$$\text{Re} [\alpha_r] = \text{Re} [-\kappa_r(j\omega_0)/\kappa_1^{r+1}(j\omega_0)]/r!.$$

Hence writing

$$X_h' = \text{Re} [j^h \kappa_1(j\omega_0) \kappa_h(j\omega_0)], \quad (h = 2, 3, \cdots)$$

$$\kappa_1(j\omega_0) = j\beta \neq 0,$$

β real, we conclude that

$$\text{Re} [\alpha_1] = 0,$$

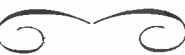
$$\text{Re} [\alpha_h] = X_h' = 0 \quad (h = 2, 3, \cdots, r-1)$$

$$\text{Re} [\alpha_r] = \frac{(-1)^r}{\beta^{r+2} r!} X_r'. \quad (13)$$

We may use these equations to complete the proof of (A) of the introduction. Since $\text{Re} [\alpha_r] \neq 0$, this is also true of X_r' . Furthermore if r is even, then $\text{Re} [\alpha_r]$ and X_r' have the same sign. Thus we may use X_h' in place of $\text{Re} [\alpha_h]$ in (i) and (ii) of the italicized sentence of §3. Also, since $\kappa = D/N = K/A$, we may replace κ in X_h' by $\kappa/K = 1/A$ to get X_h defined in the introduction, as this merely multiplies X_h' by the positive constant $1/K^2$. This completes the proof of (A) in the introduction.

BIBLIOGRAPHY

1. Bower, J. L., and Ordung, P. F., "The Synthesis of Resistor-Capacitor Networks." *PROC. I.R.E.*, Vol. 38 (March, 1950), pp. 263-269.
2. Fialkow, A., "Two Terminal-Pair Networks Containing Two Kinds of Elements Only." *Proc. Symposium on Modern Network Synthesis*, Polytechnic Institute of Brooklyn, (1952), pp. 50-65.
3. Fialkow, A., and Gerst, I., "The Transfer Function of an RC Ladder Network." *Jour. Math. Phys.*, Vol. 30 (July, 1951), pp. 49-72.
4. Fialkow, A., and Gerst, I., "The Transfer Function of General Two Terminal-Pair RC Networks." *Quart. Appl. Math.*, Vol. 10, July, 1952, pp. 113-127.
5. Fialkow, A., and Gerst, I., "The Transfer Function of Networks Without Mutual Reactance." *Quart. Appl. Math.*, Vol. 12 (July, 1954), pp. 117-131.
6. Kahal, R., "Synthesis of the Transfer Function of Two Terminal-Pair Networks." *Trans. IEE*, Vol. 71, Part I (1952), pp. 129-134.
7. Weinberg, L., "A General RLC Synthesis Procedure." *PROC. I.R.E.*, vol. 42 (February, 1954), pp. 427-437.
8. Weinberg, L., "Networks Terminated in Resistance at both Input and Output." *Proc. I.R.E.*, Vol. 42 (March, 1954), p. 625.
9. Weinberg, L., *New Synthesis Procedure for Realizing Transfer Functions of RLC and RC Networks*. Tech. Rep. No. 201, Res. Lab. of Electronics, Mass. Inst. Tech., 1951.



Modes and Operating Voltages of Interdigital Magnetrons*

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Summary—Methods have been discussed for obtaining a desirable frequency spectrum of the modes of an interdigital resonator, so that it may be possible to get useful operation in more than one mode. The consequences of phase reversal at certain locations in the anode, in the case of nonzero order modes, have been analyzed, together with the effects of phase-shifting fingers. Experimental results have been given, which are seen to be in substantial agreement with theory.

INTRODUCTION

THE RESONANCE frequencies as well as Q values of modes of various orders in interdigital magnetrons have been studied theoretically as well as experimentally.¹⁻⁴ The frequency spectrum is characterized by the fact that the modes of various orders are well separated from one another. In particular, the zero-order mode can be operated without the use of any additional mode control devices, such as straps. However, efficient operation in this mode requires the use of decoupling chokes.⁵ Such chokes are not needed in the case of nonzero-order modes, as the latter do not couple strongly with a symmetrically located cathode. On the other hand, these modes occur in degenerate pairs, and also have points of phase reversal around the anode.

Because of these disadvantages of nonzero-order modes, most of the work on operating interdigital magnetrons has been done in the zero-order mode.⁶⁻⁹ Typical results which have been reported in this mode include a cw power output of 500 watts or a peak power output of 300 kilowatts at frequencies of the order of 2,500 mc/sec. and over-all efficiencies up to 70 per cent. This mode has also been used in an inverted mag-

netron,¹⁰ giving 1,500 watts at 5 per cent duty ratio, with an over-all efficiency up to 50 per cent. However if the problems of degeneracy and phase reversal in the case of nonzero-order modes can be solved satisfactorily, then an interdigital magnetron offers the possibility of useful operation at more than one frequency. With tunable tubes having overlapping tuning ranges for various modes, wide tuning ranges would also be possible. Apart from this, nonzero-order modes can give higher frequencies for the same dimensions of the resonator. Interest in a nonzero-order mode has been shown in another paper¹¹ also, dealing with the design of a waveguide loaded interdigital resonator for operation in the second-order mode.

In order to remove the degeneracy between the two first-order modes, Crawford and Hare used two capacitive fingers behind the main set of fingers. Also, in order to make the "polarity of the fingers alternate regularly all the way around the anode," they used the "phase reversing anode."¹² At each position of zero E -field, two adjacent fingers were joined to the same side of the cavity. The two could be made as one broad finger, called a phase-shifting finger. Crawford and Hare obtained a cw power output of 50 watts at an efficiency of 40 to 50 per cent in the wavelength range of 6 to 12 centimeters.

In the work reported here, problems relating to the frequency spectrum of the interdigital magnetron, and to its operation in nonzero-order modes, have been investigated further. The results of an experimental study for correlating the frequencies of various modes with resonator parameters have been given. Methods have been discussed for removing the degeneracy of nonzero-order modes, and making all the resultant modes equally spaced in frequency.

Experimental evidence and theoretical justification for the excitation of each of the nonzero-order modes at more than one voltage have been given. It has been found that phase-shifting fingers do not ensure operation at only one voltage for one mode. An alternative approach to this problem has been suggested.

MODES OF A SIMPLE INTERDIGITAL RESONATOR

In order to control the frequency spectrum of an interdigital resonator, it is first of all desirable to know the general manner in which the modes of various or-

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† National Physical Laboratory of India, New Delhi, India.

¹ F. H. Crawford and N. D. Hare, "A tunable squirrel-cage magnetron—the donutron," *Proc. I.R.E.*, vol. 35, pp. 361-369; April, 1947.

² J. F. Hull and L. W. Greenwald, "Modes in interdigital magnetron," *Proc. I.R.E.*, vol. 37, pp. 1258-1263; November, 1949.

³ W. S. Lucke, "Obstacle Loaded Cylindrical Cavities with Application to the Interdigital Magnetron," *Cruft Laboratory Technical Report No. 60*, Harvard University; November, 1948.

⁴ A. Leblond, "Study of an interdigital line used as an anode circuit of a magnetron oscillator for U.I.F.," *Ann. Radioélect.*, vol. 8, pp. 194-210; July, 1953.

⁵ J. F. Hull and A. W. Randalls, "High-power interdigital magnetrons," *Proc. I.R.E.*, vol. 36, pp. 1357-1363; November, 1948.

⁶ *Ibid.*

⁷ H. W. Welch, Jr. and G. R. Brewer, "Operation of Interdigital Magnetrons in Zero Order Mode," *Technical Report No. 2*, Electron Tube Laboratory, University of Michigan.

⁸ F. Ludi, "Single cavity magnetron," *Tijdschr. ned. Radiogenoot.*, vol. 18, pp. 89-103; March 1953. Also F. Ludi, *Proc. I.R.E.*, vol. 41, p. 799; June, 1953.

⁹ A. Leblond; O. Doehler and R. Warnecke, "A new magnetron oscillator with interdigital circuit," *C. R. Acad. Sci. Paris*, vol. 236, pp. 55-57; 5th Jan., 1953.

¹⁰ J. F. Hull, "Inverted magnetron," *Proc. I.R.E.*, vol. 40, pp. 1038-1041; September, 1952.

¹¹ G. Hok, "Design of waveguide loaded resonator for magnetron with interdigital circuit," *Proc. I.R.E.*, vol. 41, pp. 763-769; June, 1953.

¹² Crawford and Hare, *op. cit.*

ders are influenced by the parameters of the fingers and the cavity. With this in view, a demountable resonator was designed, so that several combinations of anode and cavity parameters were easily obtainable. Fig. 1 shows a

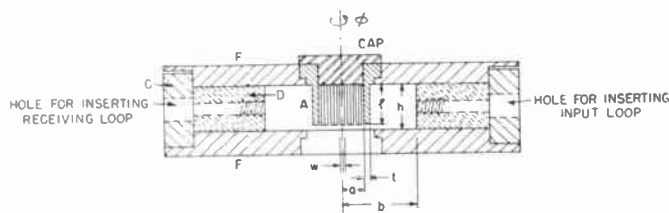


Fig. 1—Cross section of demountable resonator.

cross section of the resonator, the lower set of fingers being omitted for the sake of clarity. The cavity was formed by clamping two face plates *F* against a thick annular disc *D*. Fingers attached to two cylinders were inserted through holes in the face plates. The squirrel cage formed by the fingers was made coaxial with the annular disc by fitting the face plates and the disc into an outer cylinder *C*. Two caps completed the resonator. The resonance frequencies were determined by inserting two coupling loops into the cavity, one for feeding in power from a test oscillator, and the other for detecting the amplitude of oscillations. The modes were identified by plotting the field patterns with the help of a rotating probe. The rectified and amplified output of the probe was fed to a recording milliammeter, while the probe was slowly rotated by a motor with a step-down gear-box.

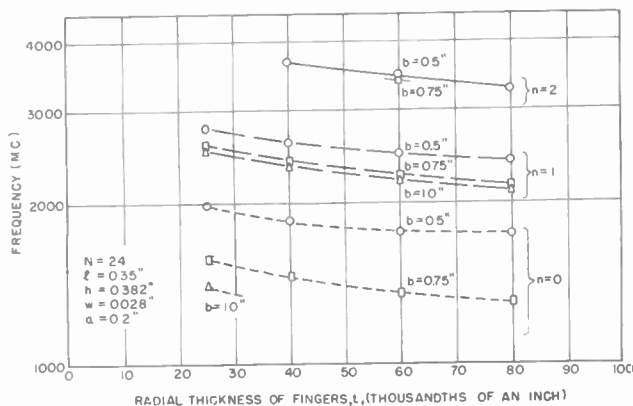


Fig. 2—Variation of frequency with capacity at the anode.

Graphs of resonance frequency of the different modes as a function of resonator parameters are given in Figs. 2 and 3. There *N* stands for the total number of fingers and *n* stands for the order of the mode. The symbols for other parameters are explained in Fig. 1. The following general conclusions can be drawn from the data:

1. The ratio by which the frequency decreases for a given increase of cavity radius is smaller for higher-order modes. Consequently, as the ratio of cavity radius

to anode radius (denoted by b/a) increases, the separation of adjacent modes also increases. In a typical case, the separation between the zero- and first-order modes increases from 41 per cent to 68 per cent as b/a increases from 2.5 to 3.75. Under the same conditions the separation between first- and second-order modes increases from 41 per cent to 52 per cent.

2. Increase of radial thickness of the fingers, or decrease of separation between adjacent fingers reduces the resonance frequencies in a ratio which is nearly the same for the various orders. Thus the mode separations do not depend critically on these two parameters.

These conclusions would be of assistance in obtaining desired intervals between the modes of various orders.

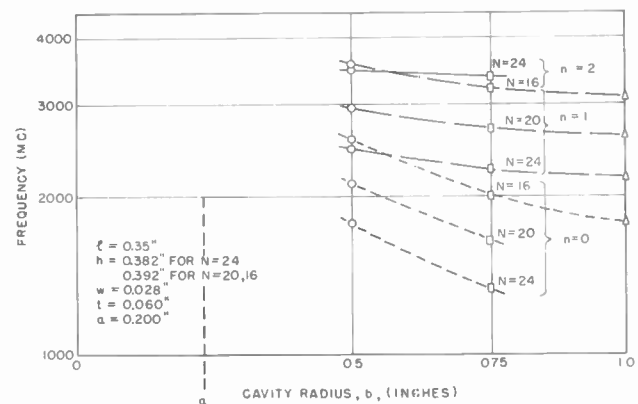


Fig. 3—Variation of frequency with cavity radius.

MODES OF INTERDIGITAL RESONATOR WITH SHORTING WIRES AT FINGERS

In the early stages of this work shorting wires were used to solve the problem of degeneracy in nonzero-order modes. At that time the primary interest was in getting only a second-order mode to give steady operation. The resonator had 24 ordinary fingers, and four phase-shifting fingers located at intervals of 90 degrees. The tips of these four fingers were short-circuited to the opposite face of the cavity. It was expected that all the modes except the second-order mode with nodes of *E*-field at the shorting wires would become inoperable.

However, it was found that the other second-order mode and the two first-order modes were still operable. The shorting wires had greatly distorted their field patterns, and raised their frequencies. The higher second-order mode was separated from the undisturbed second-order mode by an interval of 32 per cent, and the two first-order modes were separated from the latter by intervals of 4 and 2 per cent respectively. Fig. 4 (next page) shows the field patterns of the 3 lowest modes. Locations of phase-shifting fingers are represented by *P*, and that of the coupling loop by *L*.) It is seen that the *E*-field of the first-order modes was nearly zero in opposite quadrants. When cold tests were performed with the

shorting wires put between the faces of the cavity at successively increasing distances behind the fingers, the frequencies of the first-order modes were found to decrease. At a certain stage, the frequency of the higher second-order mode came within the range of the test oscillator, and was found to approach that of the undisturbed second-order mode. The above experiments showed that shorting wires at fingers are not suitable for frequency control. However, they suggested the possibility of using shorting wires or radial vanes in the cavity.

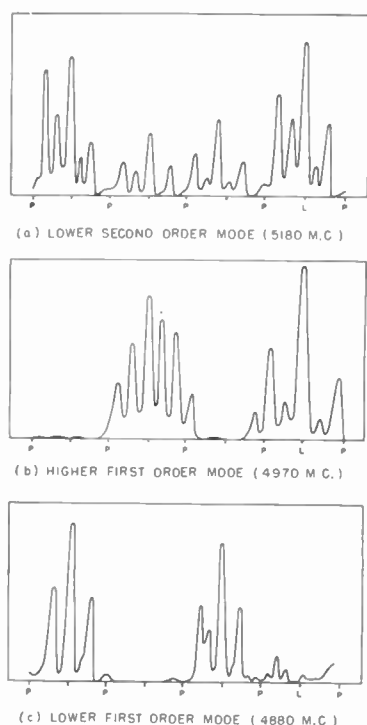


Fig. 4—Field patterns with shorting wires. (a) Lower second-order mode (5,180 mc). (b) Higher first-order mode (4,970 mc). (c) Lower first-order mode (4,880 mc).

MODES OF INTERDIGITAL RESONATOR WITH VANES IN THE CAVITY

Further cold tests were performed using radial vanes in the cavity. A cross section of the resonator with vanes is shown in Fig. 5. Four radial slots were cut in one of the face plates, so as to lie behind the four phase shifting fingers. Vanes were inserted into these slots and clamped in place. The radial penetrations of the vanes could be varied independently. However, radially opposite vanes were always set at symmetrical locations. The resonance frequencies were determined for all the combinations of a set of values of d_a and d_b , where d_a was the radial length of one pair of vanes and d_b was that of the other pair.

Fig. 6 (page 473) shows the results in graphs of resonance frequencies versus d_a with various values of d_b as parameter. It is seen that the frequencies of the zero-order mode and of the second-order mode, which ordinarily has maxima of E -field at the position of the

vanes, are increased by an increase of d_a as well as d_b . The frequency of each of the first-order modes is independent of one pair of vanes, and rises with increase in penetration of the other pair. The second-order mode having zeros of E -field at the vanes is practically unaffected by both d_a and d_b . In general, (1) insertion of vanes leaves the resonance frequency of a mode unaltered if the E -field for the undisturbed mode is zero at the positions of the vanes; (2) the resonance frequency is increased when the above condition is not satisfied; (3) the rate of increase of frequency with increase of d becomes larger as d increases.

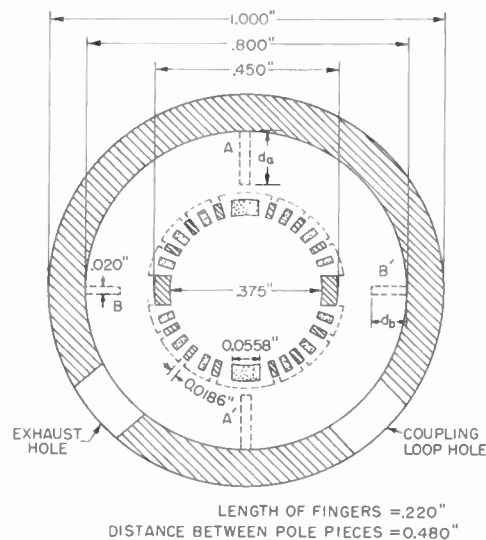


Fig. 5—Cross section of resonator.

The graphs also show how five different modes can be located at convenient intervals, with the help of two pairs of vanes in the cavity. The five modes are the zero-order mode, the two first-order modes, and the two second-order modes. By adjusting the difference between d_a and d_b the separation between the first-order modes can be adjusted. By adjusting the actual magnitudes of d_a and d_b the separation between the two second-order modes can be adjusted. In this way, intervals of the order of 9 per cent were obtained in operating tubes having a ratio of cavity radius to anode radius equal to 2:1. The intervals can be increased by increasing this ratio, as discussed in the first section.

The field patterns of the two second-order and two first-order modes obtained when vanes were used are shown in Fig. 7 (page 473). Since phase-shifting fingers were present, a regular field configuration was obtained only with the lower second-order mode, but the distortion was much less than when shorting wires were used.

It is seen that a desirable frequency spectrum can be obtained without undue distortion of the field patterns, by using vanes in the cavity. Other problems to be considered in connection with controlled operation in more than one mode relate to the Q and to the admittance presented by the fingers to the electron stream for the

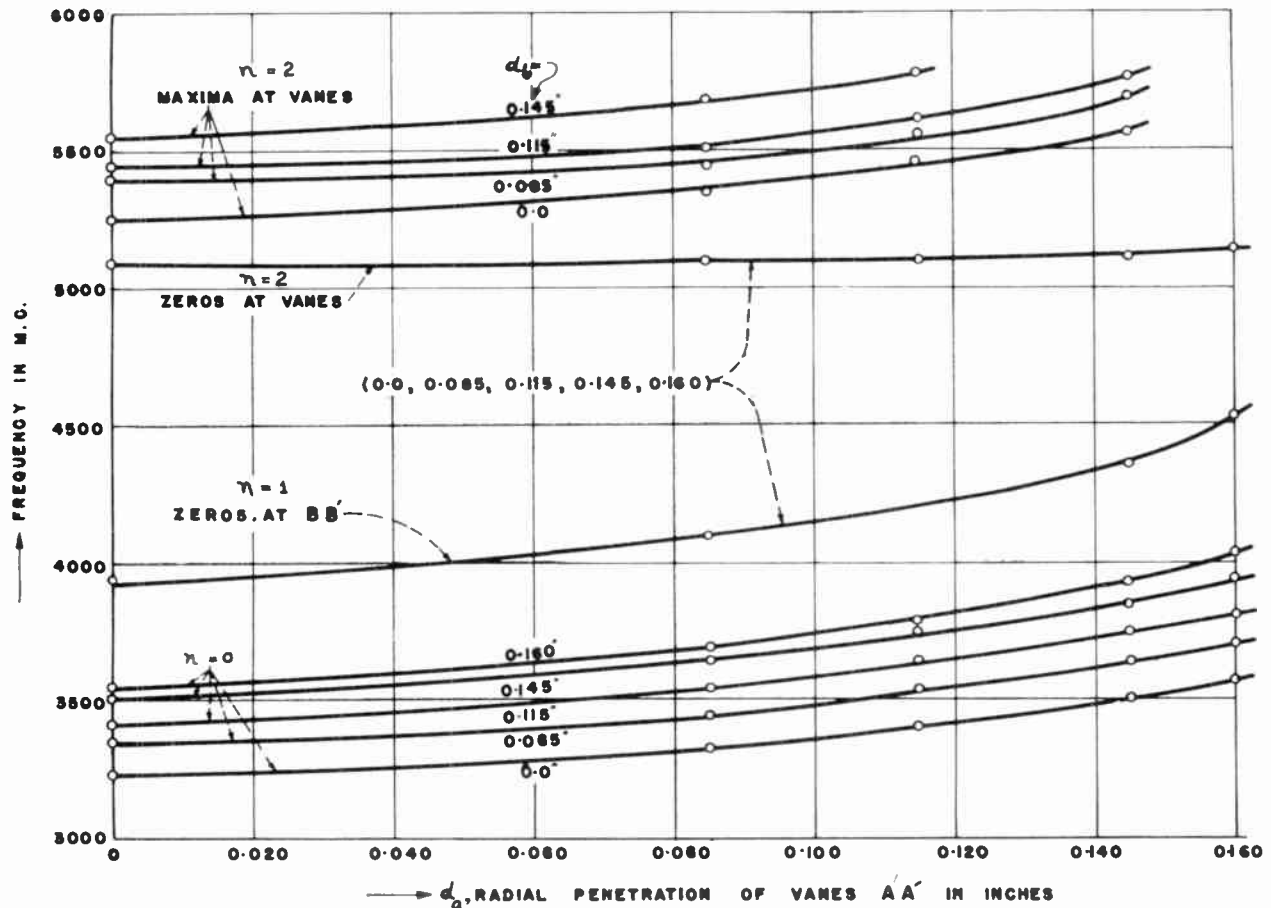


Fig. 6—Variation of resonance frequencies with penetration of vanes.

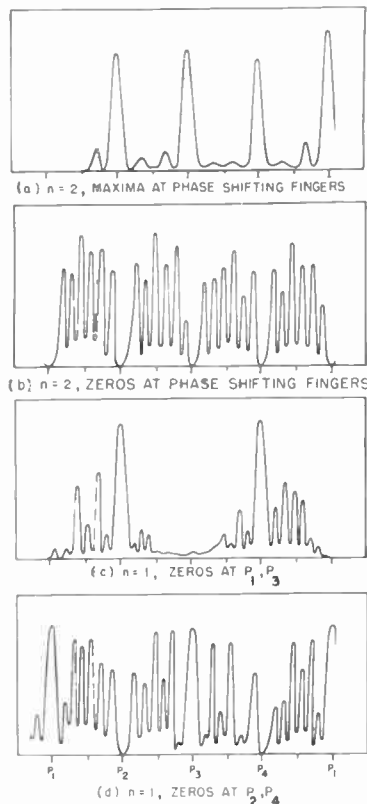


Fig. 7—Field patterns with vanes.

various modes. Some further work has been done by the author along these lines at the National Physical Laboratory of India.¹³ A cavity with rectangular cross section has been used to separate the two first-order modes, keeping the surface losses small. Also the radial width of the fingers has been made to vary, so that the widest fingers are located at the axial plane parallel to the length of the cavity. In this way, further separation of the two first-order modes has been obtained, and at the same time the admittances presented to the electron stream by the two modes can be made nearly equal.

OPERATING VOLTAGES OF NONZERO-ORDER MODES

The fact that electric field in the cavity changes sign at certain locations, in the case of nonzero-order modes, results in operation being possible for each at more than one voltage. Fig. 8 (next page) shows a current vs voltage curve, at constant magnetic field, for a tube having four phase-shifting fingers. The load was matched to the transmission line at the frequency of the lower second-order mode. The duty ratio was 0.1 per cent. The curve shows that operation in the lower second-order mode and lower first-order mode was obtained in two distinct ranges of voltage in each case. The same was found to

¹³ A. Singh and N. C. Vaidya, "A new method of mode control of interdigital magnetron," *Jour. Sci. Ind. Res.*, vol. 13, pp. 512-515; November, 1954.

be true for the higher first-order mode also, in other cases. Operation in more than two ranges of voltage for a given mode was observed in some cases. Such observations were made with tubes having shorting wires as well as with those having vanes. It is clear that even in the lower second-order mode more than one Fourier component could be excited, in spite of the presence of phase-shifting fingers at suitable locations. Thus it is of interest to study the Fourier components of the field configurations of different modes, with and without phase shifting fingers.

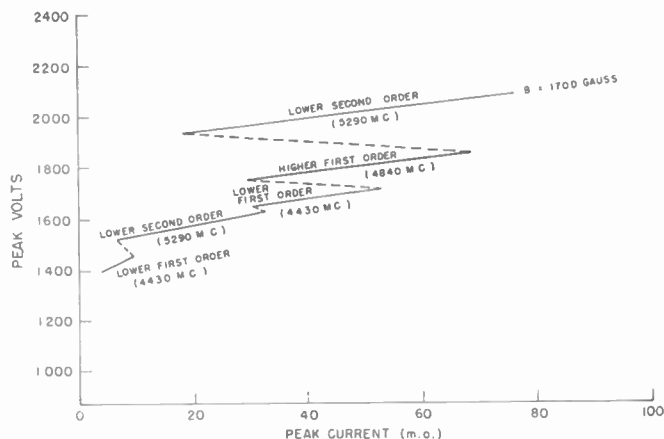


Fig. 8—Modes of operation of Tube 29.

FOURIER COMPONENTS OF THE FIELD CONFIGURATIONS

The azimuthal component $E_\phi(a, \phi)$ of the E -field at the anode in the median plane can, in the general case, be written

$$E_\phi(a, \phi) = Z(\phi) \cos n\phi P(\phi), \quad \text{or} \quad Z(\phi) \sin n\phi P(\phi),$$

where $Z(\phi)$ depends upon the total number of fingers and the ratio of gap width to finger width, and is of the nature shown in Fig. 9; n is the order of the mode; and $P(\phi)$ is a function depending upon the number and location of the phase-shifting fingers, as shown in Fig. 9. $Z(\phi)$ can be analyzed into its Fourier components as follows:

$$Z(\phi) = \frac{4}{\pi} \left\{ \sin \frac{\pi}{3} \cos M\phi + \frac{1}{3} \sin \frac{3\pi}{2\rho} \cos 3M\phi + \frac{1}{5} \sin \frac{5\pi}{2\rho} \cos 5M\phi + \dots \right\},$$

where the ratio of gap width to finger width is 1: ($\rho - 1$), and M is half the total number of fingers.

When no phase-shifting fingers are present, then for higher order modes,

$$E_\phi(a, \phi) = Z(\phi) \cos n\phi$$

$$= \frac{2}{\pi} \left[\sin \frac{\pi}{2\rho} \{ \cos (M + n)\phi + \cos (M - n)\phi \} + \frac{1}{3} \sin \frac{3\pi}{2\rho} \{ \cos (3M + n)\phi + \cos (3M - n)\phi \} + \dots \right].$$

Alternatively,

$$E_\phi(a, \phi) = Z(\phi) \sin n\phi = \frac{2}{\pi} \left\{ \sin \frac{\pi}{2\rho} \{ \sin (M + n)\phi - \sin (M - n)\phi \} + \frac{1}{3} \sin \frac{3\pi}{2\rho} \{ \sin (3M + n)\phi - \sin (3M - n)\phi \} + \dots \right\}.$$

Let γ represent the number of complete cycles around the anode, for a given Fourier component. It is seen that when no phase-shifting fingers are present, γ assumes the values $M \pm n$, $3M \pm n$, $5M \pm n$, etc.

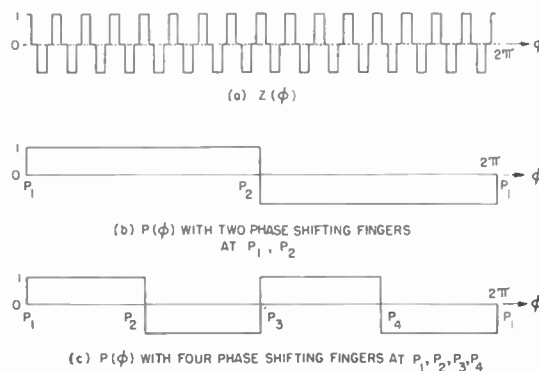


Fig. 9—Functions used for obtaining Fourier components of field configurations.

The case where four phase-shifting fingers are present is considered next. If in that case $P(\phi)$ is denoted by $P_4(\phi)$, then by Fourier analysis,

$$P_4(\phi) = \frac{4}{\pi} \sin 2\phi + \frac{4}{3\pi} \sin 6\phi + \frac{4}{5\pi} \sin 10\phi + \dots$$

$$P_4(\phi) \sin \phi = \frac{2}{\pi} \cos \phi - \frac{2}{\pi} \cos 3\phi + \frac{2}{3\pi} \cos 5\phi - \frac{2}{3\pi} \cos 7\phi + \dots$$

$$P_4(\phi) \cos \phi = \frac{2}{\pi} \sin \phi + \frac{2}{\pi} \sin 3\phi + \frac{2}{3\pi} \sin 5\phi$$

$$+ \frac{2}{3\pi} \sin 7\phi + \dots$$

$$P_4(\phi) \sin 2\phi = \frac{2}{\pi} - \frac{4}{3\pi} \cos 4\phi - \frac{4}{15\pi} \cos 8\phi - \dots$$

$$P_4(\phi) \cos 2\phi = \frac{8}{3\pi} \sin 4\phi + \frac{16}{15\pi} \sin 8\phi + \frac{24}{35\pi} \sin 12\phi + \dots$$

The most significant Fourier components of $E_\phi(a, \phi)$ would be obtained from the product of the first term in the expansion of $Z(\phi)$ and the function $\cos n\phi P_4(\phi)$ or $\sin n\phi P_4(\phi)$ appropriate to the given mode. The components given by the subsequent terms in the $Z(\phi)$ expansion would have their γ and excitation voltage far removed from those corresponding to the first term. Taking unity as coefficient of this first term, and expressing products as sums and differences, one obtains

$$P_4(\phi) \cos M\phi$$

$$= \frac{2}{\pi} \{ \sin (M+2)\phi - \sin (M-2)\phi \} + \frac{2}{3\pi} \{ \sin (M+6)\phi - \sin (M-6)\phi \} + \dots$$

$$P_4(\phi) \sin \phi \cos M\phi$$

$$= \frac{1}{\pi} \{ \cos (M+1)\phi + \cos (M-1)\phi \} - \frac{1}{\pi} \{ \cos (M+3)\phi + \cos (M-3)\phi \} + \frac{1}{3\pi} \{ \cos (M+5)\phi + \cos (M-5)\phi \} + \dots$$

$$P_4(\phi) \cos \phi \cos M\phi$$

$$= \frac{1}{\pi} \{ \sin (M+1)\phi - \sin (M-1)\phi \} + \frac{1}{\pi} \{ \sin (M+3)\phi - \sin (M-3)\phi \} + \frac{1}{3\pi} \{ \sin (M+5)\phi - \sin (M-5)\phi \} + \dots$$

$$P_4(\phi) \sin 2\phi \cos M\phi$$

$$= \frac{2}{\pi} \cos M\phi - \frac{2}{3\pi} \{ \cos (M+4)\phi + \cos (M-4)\phi \} - \frac{2}{15\pi} \{ \cos (M+8)\phi + \cos (M-8)\phi \} + \dots$$

$$P_4(\phi) \cos 2\phi \cos M\phi$$

$$= \frac{4}{3\pi} \{ \sin (M+4)\phi - \sin (M-4)\phi \} + \frac{8}{15\pi} \{ \sin (M+8)\phi - \sin (M-8)\phi \} + \dots$$

The values of γ thus obtained from the first term are $M \pm 2$, $M \pm 6$, etc. for the zero-order mode; $M \pm 1$, $M \pm 3$, $M \pm 5$, etc. for the first-order modes; M , $M \pm 4$, $M \pm 8$, etc. for one second-order mode; and $M \pm 4$, $M \pm 8$, etc. for the other second-order mode.

The effect of four phase-shifting fingers has been to make the value M available for γ in one of the second-order modes. But, at the same time, in all the nonzero-order modes one pair of components has been replaced by a number, whose amplitude does not fall off so rapidly as to make their excitation improbable.

The foregoing study is helpful in explaining the observed voltages of operation. For a given mode and magnetic field, the voltage is inversely proportional to γ , to a first approximation. This approximation was considered adequate because there was uncertainty in choosing values of voltage from the data for comparison with theory, as each mode of excitation was obtained over a range of voltage. The starting point of each range was chosen for comparison with theory.

The following limitations of the above simple theory have to be recognized. The presence of any irregularities in the geometry of the resonator would introduce Fourier components not given by the above analysis. Examples of such irregularities are the coupling loop, an imperfectly aligned cathode, and irregular spacing of fingers. Yet, notice that for Fourier analysis the range of azimuth over which the function $E_\phi(a, \phi)$ was defined was 0 to 2π . However, in exciting a given field configuration, the electrons need not be in synchronism with the field over this whole range. The following shows the kind of difference between simple theory and experiment expected on the above basis. In the first-order mode, when four phase-shifting fingers are present, a value of $\gamma = M$ is not given by Fourier analysis, but an excitation voltage corresponding to $\gamma = M$ would be possible, if an electron, while giving energy to the rf field, could reach the anode without crossing the two phase-shifting fingers at the maxima of the E -field.

COMPARISON OF EXPERIMENTAL RESULTS WITH THEORY

In these tubes M was equal to 16, as the tube had 24 ordinary fingers and four phase-shifting fingers, each one of the latter being equivalent to two ordinary ones. Data were taken on the two first-order modes, and the second-order mode whose E -field was zero at the phase-shifting fingers. The zero-order mode could not be operated since cathode decoupling chokes were not used. Operation in the other second-order mode was erratic, due to points of phase reversal around anode.

TABLE I

Tube No.	Mode	Magnetic field in Gauss	Proportions of reciprocals of voltages
22	Lower Second	1370	16.0:17.7:20.2
22	Lower Second	1920	16.0:18.3:19.9
23	Higher First	1920	16.0:17.2
23	Lower Second	1920	16.0:18.0
28	Lower Second	1700	16.0:19.9:20.8:22.0
28	Lower Second	2260	16.0:17.3:20.1:21.3
29	Lower First	1480	13.3:16.0:19.2
29	Lower First	1700	16.0:19.1:21.3
29	Lower First	2140	16.0:19.1
29	Lower First	2610	16.0:19.0
29	Lower Second	1480	16.0:20.5
29	Lower Second	1700	16.0:20.0
29	Lower Second	2140	16.0:20.5
29	Lower Second	2610	16.0:20.5

The data given in Table I above may be discussed under the following headings:

1. Within experimental error, the reciprocals of voltages for each mode are proportional to numbers in the series, $16, 16 \pm 1, 16 \pm 2$, etc. Table I shows the actual numbers obtained in the series, 16 being taken as the reference number.

2. The presence of voltages corresponding to $\gamma = 16$ and 20 in the second-order mode and to $\gamma = 13, 17, 19$ and 21 in the first-order modes is in conformity with values obtained from Fourier analysis.

3. The presence of a voltage corresponding to $\gamma = 16$ in the first-order modes is due to the fact that electrons can reach the anode by covering only a small range in azimuth, as already discussed. The assumption that the voltage corresponds to $\gamma = 16$ was checked by using the voltage for $\gamma = 16$ in the second-order mode as reference. The voltages were found to be directly proportional to the frequencies to within 1 per cent, as would be expected when γ is the same for the two cases.

4. The occasional presence of voltages corresponding to other values of γ may be ascribed to irregularities in the structure, which would modify the field and thus introduce additional Fourier components.

5. The consistent absence of a voltage corresponding to $\gamma = 17$ from the lower first-order mode is understandable, since the same order of voltage can excite the

$\gamma = 20$ component of the lower second-order mode; and the latter appears to be more easily excited.

Before the advantage of a number of well separated modes can be exploited, the difficulty of more than one voltage of operation for the nonzero order modes will have to be removed. Phase-shifting fingers are of doubtful advantage. A different approach would be to increase the number of fingers, without using phase-shifting fingers. The separation (in percentage) of the two components will thus be reduced. Considering the fact that operation at a voltage corresponding to $\gamma = M$ is also likely in addition to those corresponding to $\gamma = M \pm n$, it appears possible that all the three voltage ranges may merge into one continuous range. This may be expected particularly for small values of n . The problem invites further work.

CONCLUSIONS

It has been shown that the degeneracy of nonzero order modes can be removed and the resonance frequencies accurately controlled, without undue distortion of the field patterns in the interaction space, by using radial vanes in the cavity. The dimensions of the fingers, the cavity, and the vanes, can be so chosen as to obtain a number of modes at desired intervals.

It was found that each nonzero-order mode normally operates at more than one voltage. The way to get around this difficulty appears to be to use a large number of ordinary fingers, without introducing any phase-reversing ones. This alternative is also to be preferred if operation in more than one mode is desired.

ACKNOWLEDGMENT

The guidance and encouragement received from Prof. E. L. Chaffee, Dr. D. L. Benedict and Prof. R. W. P. King are gratefully acknowledged. The work of Dr. C. Yeh and Mr. J. P. Jasionis on operating tubes using shorting wires provided the stimulus for further work on mode control. The observation that the interdigital magnetron operated at more than one voltage for a nonzero-order mode was also first made by them. The data on operating voltages of Tubes 22 and 23 were taken by Dr. Yeh.



Measurement of Minority Carrier Lifetime and Surface Effects in Junction Devices*

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Summary—The characteristics of junction devices are influenced to a considerable degree by the lifetime of the minority carriers. Accordingly, methods for the measurement of this quantity are of considerable importance. Methods have been described for the measurement of the lifetime of minority carriers when these carriers are produced within the volume of a semiconductor. When the minority carriers are introduced near the surface of a semiconductor the resulting effective lifetime may be determined to a large extent by the nature of the surface. For most junction devices, it is the effective lifetime that is of primary importance.

This paper describes a simple method for the measurement of effective lifetimes of injected minority carriers. The measurements may be applied to practical junction structures as, for example, an alloyed junction transistor. Measurements may be made on either completed or partially completed devices. The resulting data are potentially of value as quality controls during the fabrication of transistors and similar devices.

In many cases, the effective lifetime is a good indication of the surface conditions, and immediate evaluation of these conditions may be obtained at various stages of device processing. With selected geometries, the measurement method may be applied to determine absolute values of surface recombination velocities and should therefore be in studying surface conditions and treatments.

The measurement method is described in terms of junction devices using germanium as the semiconductor. However, the method is equally applicable to junction devices made with other semiconductor materials.

INTRODUCTION

AN IMPORTANT material property which affects the performance of transistor devices is the lifetime of minority carriers in the semiconductor. This lifetime depends on the nature of the material and on the various treatments to which the material has been subjected. Electrically, it is a direct factor in many transistor parameters such as saturation current, current amplification factor, and others. It is, therefore, of considerable practical importance to be able to evaluate this factor directly on junction devices.

Earlier studies of minority carrier lifetimes have been mainly directed to evaluations as a property of the material¹ (volume lifetime) or as a property of a surface² (surface lifetime). As a result, the methods developed in these studies have not used the geometries of practical junction devices nor have they generally involved a p - n junction. This paper will describe a simple method³

which is directly applicable to junction devices. Indeed, this method uses a p - n junction to inject minority carriers by means of a current pulse applied to the junction in the forward direction. The decay of the injected carriers is observed by open-circuiting the p - n junction and observing the junction voltage on an oscilloscope. A particular advantage for investigative work is that this measurement can be made on a single junction, thereby avoiding the more complex construction of a complete transistor. Furthermore, immediate evaluation can be made at various stages in the processing of junction units as a control in the fabrication of transistors or as a measurement in the study of process variations.

EXPERIMENTAL METHOD

The circuit of Fig. 1 shows an experimental arrangement for applying a constant current pulse in the forward direction through a p - n junction and, by means of a thermionic diode, open-circuiting the p - n junction at the termination of the current pulse. The open-circuited junction voltage is observed on an oscilloscope. Minority carriers are injected into the base region during the

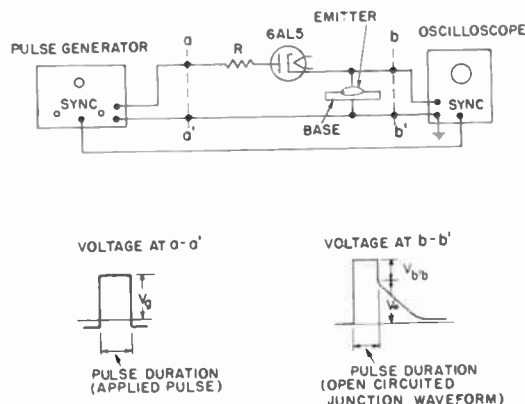


Fig. 1—Circuit illustration for applying constant current pulse to emitter-base junction and observing an open-circuited junction voltage upon termination of pulse.

time the pulse is applied to the junction in the forward direction. Upon completion of the pulse, the thermionic diode effectively opens the circuit between the generator and the emitter. As a result, the junction voltage is a direct measure of what happens to the injected carriers. A typical open-circuited voltage wave form is also illustrated in Fig. 1 (voltage at b - b').

It is observed that, after an initial drop due to an internal series resistance, the open-circuited junction voltage decays approximately linearly with time, and this linear decay is followed by an approximately exponen-

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‡ RCA Labs., Princeton, N. J.

¹ L. B. Valdes, "Measurement of minority carrier lifetime in germanium," *Proc. I.R.E.*, vol. 40, pp. 1420-1423; November, 1952.

² D. Navon, R. Bray, and H. Y. Fan, "Lifetime of injected carriers in germanium," *Proc. I.R.E.*, vol. 40, pp. 1342-1347; November, 1952.

³ A related development has been described by B. R. Gossick, "Post-injection barrier electromotive force of p - n junction," *Phys. Rev.*, vol. 91, pp. 1012-1013; August 15, 1953.

tial decay. As will be shown below, this linear portion of the voltage variation lends itself very readily to computation of a minority carrier lifetime, which is here designated as an *effective lifetime* since it results from the combined effect of volume and surface lifetimes.

In Fig. 2 there is shown a flexible circuit for use in connection with a suitable pulse generator and an oscilloscope for observing either the reverse bias (to be described subsequently) or the open-circuited junction

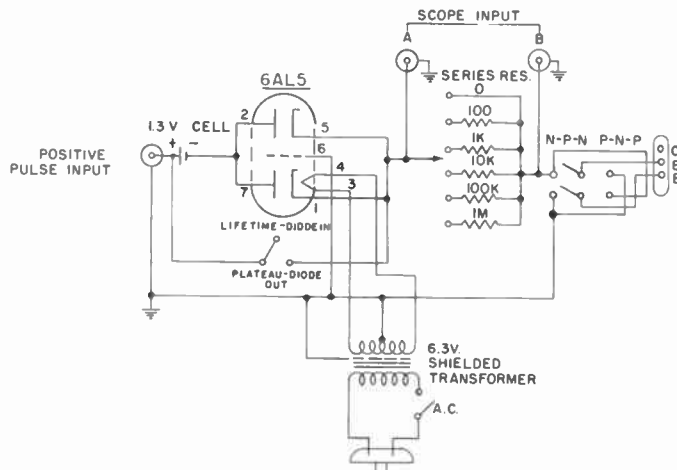


Fig. 2—Circuit used for the measurement of effective lifetime and related characteristics.

voltage. The pulse rise time and more important, the pulse decay time of the pulse generator, should be reasonably small—1/10 of the effective lifetimes to be measured should be adequate. Pulse length and repetition frequency usually used are 10 μ sec and 3,000 p/sec, but the exact values employed are not important. The pulse amplitude and generator output impedance are also of no great importance. The oscilloscope response should be at least comparable to the pulse generator decay time mentioned above. It is important that the vertical amplification and horizontal trace speeds be calibrated. A differential oscilloscope connected as shown in Fig. 2 is a convenient means for measuring the pulse current flowing through the junction device. The 1.3-volt battery is inserted in series with the diode to eliminate a spurious voltage arising from the thermal velocities of the cathode-emitted electrons. A reversing switch is provided to accommodate both *n*-type and *p*-type devices with a single socket arrangement. A transparent alignment device with radial lines engraved thereon can be used for measuring effective lifetimes easily and quickly from oscilloscope displays of open-circuited junction voltages. When the vertical deflection sensitivity and horizontal sweep time are suitably adjusted, the effective lifetime is read directly by aligning one of the radial lines with the linear portion of the open-circuit junction-voltage waveform.

THEORETICAL DEVELOPMENT

A theoretical interpretation of the observed junction voltage can be made on the basis of simple but approximate junction theory. A *p-n* junction in which the conductivity of the *p*-region is much greater than that of the *n*-region, as in an alloyed junction of indium on germanium, will be considered. In such a junction, the current flow across the transition region of the junction is predominantly a hole flow, and holes are injected into the *n*-type germanium. The results, however, will apply with equal validity to a junction in which the conductivity of the *n*-region is much greater than that of the *p*-region. In this case, electrons would be injected into the *p*-type region.

Let p_n be the hole density present in the *n*-region under thermal equilibrium conditions, and Δp be the additional injected hole density in the *n*-region at the boundary of the junction transition region. The total hole density at the junction boundary will be

$$p = p_n + \Delta p. \quad (1)$$

From the theory of the *p-n* junction,⁴ the hole density in the *n*-region at the junction boundary is given by

$$p = p_n e^{qV/kT}, \quad (2)$$

where V is the junction voltage. Combining (1) and (2) the solving for the voltage,

$$V = \frac{kT}{q} \ln \left(1 + \frac{\Delta p}{p_n} \right). \quad (3)$$

If the assumption is made that the excess carrier concentration, Δp , decays exponentially according to a single effective lifetime, τ_e , then

$$\Delta p = \Delta p_0 e^{-t/\tau_e}, \quad (4)$$

where Δp_0 is excess carrier concentration at the termination of the forward current pulse. Eq. (4) can be placed in (3). The constant $(1 + \Delta p_0/p_n)$ can be readily evaluated in terms of the junction voltage, V_0 , at $t=0$ (this is the junction voltage immediately before and immediately after the removal of the forward pulse—see Fig. 1), since

$$V_0 = \frac{kT}{q} \ln \left(1 + \frac{\Delta p_0}{p_n} \right). \quad (5)$$

The open-circuited junction voltage as a function of time is then

$$V = \frac{kT}{q} \ln [1 + (e^{qV_0/kT} - 1)e^{-t/\tau_e}]. \quad (6)$$

For t/τ_e very small, and if, as usual, $V_0 \gg kT/q$ (6) may be simplified to

⁴ William Shockley, "Electrons and Holes in Semiconductors," D. Van Nostrand Company, Inc., New York, p. 312; 1950.

$$V \cong V_0 - \frac{kT}{q} t/\tau_e. \quad (7)$$

The initial voltage variation is linear with time. The slope of the linear variation is a measure of the effective lifetime.

$$\tau_e = - \frac{\Delta t}{\frac{q}{kT} \Delta V} = - \frac{kT}{q} \times \frac{1}{\text{Slope of Linear Decay}}. \quad (8)$$

The values of Δt and ΔV may be read directly with the use of a calibrated oscilloscope. Fig. 3 shows some typical voltage wave shapes and some typical calculations for effective lifetimes.

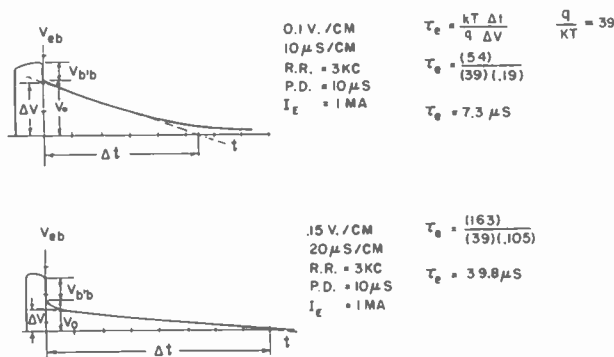


Fig. 3—Measurement of effective lifetime emitter-to-base open-circuited junction voltage.

The basic p - n junction theory applied above and based on (2) assumes that the injected minority carrier density is small compared with the majority carrier density. Accordingly, accurate measurements of τ_e should be made using small enough currents so that this assumption is valid. However, if the current is too small, a well-defined linear region is not obtained. For most of the devices that have been measured, a junction current of about 2 ma has been appropriate. When the junction current is increased so as to invalidate the assumption mentioned above, calculations similar to those above can be carried out, but the results are considerably more complex. The voltage decay for this case, as viewed on the oscilloscope, will exhibit a "hump" separating two regions of approximate linear decay. The latter decay corresponding to the region where the minority carriers are again small compared with the majority carriers can be used for measurement.

As is implied above, the preceding analysis does not possess a high degree of rigor. Only the life history of the holes has been considered, and the manner of their decay has been assumed without consideration of accompanying diffusion effects. In Appendix I this problem is examined in a more rigorous fashion. Both hole and electron carriers with independent lifetimes as well as diffusion effects are included. It is again assumed that the

minority carrier density is small compared with the majority carrier density. A study of the solution indicates that as long as the portion of the junction current due to minority carriers (holes) is approximately equal to the total junction current (injection efficiency, γ , = 1), then the resulting junction voltage decay will be that due to holes irrespective of the lifetime of the electrons. Further, it appears that the method of measuring lifetime discussed above should give results that are adequate for engineering purposes. As an additional check, the method of measuring lifetime discussed herein has been compared with another more involved method of measurement, and good agreement between the two methods of measurement has been obtained.⁵

LIFETIME MEASUREMENTS

Typical Measurements

Measurement of effective lifetimes for p - n - p junction transistors will give results generally ranging from 1 to 10 μ sec.⁶ Sample diodes made with materials having volume lifetimes of 1, 4, and 700 μ sec gave effective lifetimes of 0.5, 3.4, and 39.8 μ sec, respectively. Effective lifetimes as small as 0.01 μ sec have been measured.

It is important to observe that the effective lifetime of the units made from material having 700 μ sec volume lifetime was measured as only 39.8 μ sec. On the other hand, the effective lifetime measured on units which were made from low volume lifetime material was quite close to the volume lifetime. This seems reasonable assuming effective lifetime to be a measure of the combined effects of volume recombination and surface recombination. In accordance with calculations for simple geometries,⁷ effective lifetime, τ_e , volume lifetime, τ_v , and surface lifetime, τ_s , are related as

$$\frac{1}{\tau_e} = \frac{1}{\tau_v} + \frac{1}{\tau_s}. \quad (9)$$

The surface lifetime, τ_s , will be dependent upon the geometry and upon the surface recombination velocity. For a fixed geometry, the effective lifetime together with the volume lifetime (measured by conventional methods on the bulk material) can be used for determining a surface lifetime which is directly related to the surface treatment. When the volume lifetime is much larger than the measured effective lifetime (as is usually the case in practical device geometries), effective lifetime is very nearly a measure of surface lifetime and can accordingly be used as an index of surface treatment.

⁵ These measurements were carried out by Dr. A. R. Moore, RCA Laboratories and utilize the decay of photoconductivity following illumination with a pulsed light source. This technique has been described by D. T. Stevenson and R. J. Keyes, "Measurement of lifetimes and diffusion constants in germanium," *Phys. Rev.*, vol. 94, p. 1416; June 1, 1954.

⁶ R. R. Law, C. W. Mueller, J. I. Pankove, and L. Armstrong, "A developmental germanium p - n - p junction transistor," *PROC. I.R.E.*, vol. 40, pp. 1352-1357; November, 1952.

⁷ W. Shockley, *op. cit.*, pp. 318-325.

TABLE I
CHANGE IN EFFECTIVE LIFETIME VALUES AS A RESULT OF ETCHING

Specimen	τ_v Volume Lifetime	τ_e After Fabrication and chemical etch $s = 400$ cm/sec	τ_e Dipped in etch containing $\text{Cu}(\text{NO}_3)_2$ $s = 7400$ cm/sec	τ_e Electrolytic etch 2 min 2 ma $s = 250$ cm/sec	τ_e Electrolytic etch 5 min 3 ma $s = 250$ cm/sec	τ_e Electrolytic etch 5 min 3 ma $s = 250$ cm/sec
T-6 T-61	4 μsec 700 μsec	3.4 μsec 37 μsec	1.8 μsec 4.1 μsec	40 μsec	58 μsec	58 μsec

Effects of Etching on Effective Lifetime

To observe the effect of surface treatment on effective lifetime, two germanium alloy junctions having 0.045-inch diameter emitter dots, base wafer-thickness of 0.005 inch, and volume lifetimes that were substantially different were first chemically etched and measured and then dipped in an etch containing copper nitrate for 15 seconds. This etch was chosen because of its ability to produce a high surface recombination velocity which has been reported to be approximately 7,400 cm/sec.⁸ It was noticed upon removing the junction from the etch that copper was deposited on the dot and on the germanium surface. Following the etch treatment, the effective life was measured and indicated a substantial lower lifetime than before etching. The copper was next removed by an ammonia and hydrogen-peroxide solution, and the unit was washed in distilled water. The junction was then electrolytically etched in 1 per cent sodium hydroxide for two minutes at 2 ma current. Subsequent measurement of effective lifetime indicated a decided increase from its previous value. The effective lifetime was further increased by additional electrolytic etching; subsequent etching produced no further increase in τ_e . The measured data for the sequence of etching together with reported values of surface recombination velocities produced by these etching solutions on germanium are shown in Table I.⁸ The data in Table I indicate that there is a close correlation between surface recombination velocity and effective lifetimes when the volume lifetime is large. With the aid of (9), $\tau_e = 39.5$, 4.12, and 63.3 μsec are obtained for the chemical etch, containing copper nitrate, and electrolytic etches, respectively. If these surface lifetimes are proportionally related to surface recombination velocities as

$$\frac{1}{\tau_e} = Ks \quad (10)$$

values of the geometrical factor, K , can be computed as 63.2, 32.8, and 63.2 cm^{-1} . If the second value is discarded, a geometrical factor of 63.2 cm^{-1} is applicable for the units described above. Data similar to that shown in Table I can be used to obtain geometrical factors for different junction devices. After the geometrical factor of the unit has been determined, measurements of

effective lifetime and volume lifetime can be used for determining the surface recombination velocity of completed or partially completed units. Often only a relative comparison of surface treatments is desired. In this case, the geometrical factor need not be determined. The surface lifetime serves as an index of comparison.

Absolute Determination of the Surface Recombination Velocity

It is sometimes necessary to make a direct determination of surface recombination velocity. Thus, the efficacy of the etching solution may be in question, or a new solution may need calibration.

The absolute calibration can be made by using a junction geometry amenable to analysis as carried out by Shockley.⁹ Thus, for the geometry as shown in Fig. 4,

$$s = \frac{1}{\tau_e \left[\frac{1}{B} + \frac{1}{C} \right]} \quad (11)$$

The dimensions of the sample are not critical. It has been convenient to use wafers whose dimensions are $2A = 0.215$ inch, $2C = 0.125$ inch and $2B = 0.005$ inch.

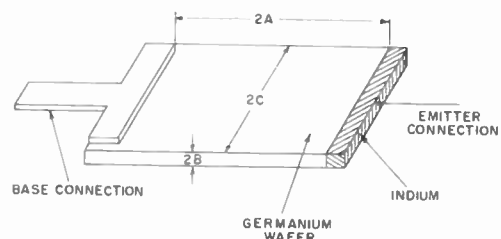


Fig. 4—Device for direct measurement of surface recombination velocity.

The $2A$ dimension should be chosen several times larger than the volume diffusion length. The $2B$ dimension should preferably be chosen so as to be the dominant term in (11). In this case then, (11) is valid as long as $s(2B/D) \leq 1$ (D is the diffusion constant for the minority carriers under consideration).

As a typical example of the application of this method of direct determination, several specimens were made with germanium whose volume lifetime was $\tau_v = 700$

⁸ A. R. Moore and J. I. Pankove, "The effect of junction shape and surface recombination on transistor current gain," *Proc. I.R.E.*, vol. 42, pp. 907-913; June, 1954.

⁹ W. Shockley, "The theory of $p-n$ junctions in semiconductors and $p-n$ junction transistors," *Bell Sys. Tech. Jour.*, vol. 28, pp. 435-489; July, 1949. See also W. Shockley, *op. cit.*, pp. 318-325.

μsec . The effective lifetime for these specimens averaged $36.5 \mu\text{sec}$. Accordingly, using (9), $\tau_s = 38.5 \mu\text{sec}$ is computed. Finally, with the aid of (11), $s = 320 \text{ cm/sec}$ is computed. The surface treatment in question was an electrolytic etch so that this value of s is in good agreement with $s = 250 \text{ cm/sec}$ that has been previously used (see Table 1).

MEASUREMENT OF BASE-LEAD RESISTANCE

In an alloyed junction device the base-lead resistance, $r_{bb'}$, is the majority carrier (ohmic) resistance of the semiconductor between the metallic contact to the semiconductor and the region near the actual p - n junction. It is an important factor in the performance of many junction devices.¹⁰

The method of measurement to be described below measures a diode base-lead resistance which is generally different from that of the corresponding device operating as a transistor. This difference is due to the dissimilar current distribution within the body of the semiconductor for diode and transistor operation.

The initial drop in the open-circuited junction voltage upon termination of the pulse can be used for the measurement of the resistance. The voltage, $V_{b'b}$, corresponding to the drop across $r_{bb'}$ is shown in Figs. 1 and 3. If the positive amplitude of the generator pulse, V_g , is measured, then $r_{bb'} = V_{b'b} R / V_g$, where R is the current limiting resistance in series with the pulse generator. This assumes that the voltage drop across the 6AL5 tube, the voltage across the junction, and the voltage of the series battery are negligible in comparison with the voltage drop across the current limiting resistor. If this assumption is not valid, the junction current just before the pulse is removed can be determined by measuring the appropriate voltage across R with the aid of the differential input to the oscilloscope (see circuit of Fig. 2).

OBSERVATIONS OF REVERSE BIAS WAVEFORM

During the course of this work experimental observations were made of the junction recovery voltage under conditions of applied reverse bias. In this case, a reverse bias is applied to the junction immediately after the termination of the forward pulse. The junction waveform under these conditions is observed on an oscilloscope. This type of switched junction operation has been investigated.¹¹⁻¹⁴ Since the interpretation of the observed waveform is somewhat more complex than that of the open-circuited case discussed above, this observation

was not developed into a system for the determination of minority carrier lifetimes. However, qualitative observations made under these conditions may be quite valuable, and in some cases this method of operation is a more sensitive indication of whether or not minority carriers are being injected. A switch is included in the circuit of Fig. 2 to enable this observation to be made. This switch shorts out the 6AL5 diode and bias battery and is labeled "plateau-diode out." The reverse bias is supplied by a blocking condenser in the output of the generator. This condenser becomes charged during the forward pulse. After the forward pulse is terminated, the charged blocking condenser applies a reverse bias to the p - n junction.

Observations of the junction waveform under the conditions of a reverse bias following a forward pulse (see Fig. 5) show first an immediate drop in voltage after the termination of the pulse due to the base-lead resistance.

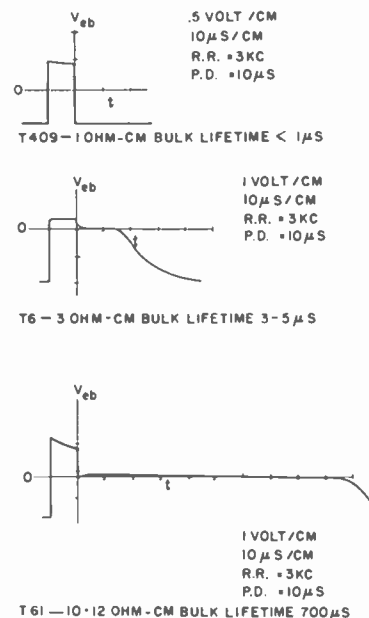


Fig. 5—Variation in plateau length of emitter-base voltage for different base wafer lifetimes.

This is similar to that discussed above in the case of the open-circuited junction voltage. This immediate drop in voltage is generally followed by an extended period of approximately zero voltage after which the reverse voltage across the junction gradually increases in magnitude as the injected carriers recombine and permit the junction to be biased in the reverse direction. The existence of the zero-voltage plateau indicates that minority carrier injection has taken place. These observations can be utilized in a qualitative manner to check for minority carrier injection and as a qualitative observation of the effective lifetime. These effects are illustrated by the experimental observations shown in Fig. 5. This figure shows the experimental waveforms under the reverse bias conditions observed on junction diodes made from

¹⁰ L. J. Giacoletto, "Study of p - n - p alloy junction transistor from dc through medium frequencies," *RCA Rev.*, vol. 15, pp. 506-562; December, 1954.

¹¹ E. M. Pell, "Recombination rate in germanium by observation of pulsed reversed characteristics," *Phys. Rev.*, vol. 90, pp. 278-279; April 15, 1953.

¹² R. G. Shulman and M. E. McMahon, "Recovery currents in germanium p - n junction diodes," *Jour. Appl. Phys.*, vol. 24, pp. 1267-1272; October, 1953.

¹³ R. H. Kingston, "Switching time in junction diodes and junction transistors," *Proc. I.R.E.*, vol. 42, pp. 829-834; May, 1954.

¹⁴ B. Lax and S. F. Neustadter, "Transient response of a p - n junction," *Jour. Appl. Phys.*, vol. 25, pp. 1148-1154; September, 1954.

germanium having different volume lifetimes. It is seen that, for the unit made from germanium having a volume lifetime of less than 1 microsecond, there is essentially no plateau region. An appreciable plateau region is observed in the second case for the unit having a volume lifetime between 3 and 5 microseconds. Finally, a rather extended plateau is observed in the third case for a unit made from material having a volume lifetime of 700 microseconds.

APPENDIX I: OPEN-CIRCUITED JUNCTION VOLTAGE

This appendix contains the solution for the open-circuited junction voltage following operation in the forward direction when both holes and electrons with independent lifetimes are considered, and when the movement of these carriers is governed by the one-dimensional continuity equation. The material in this appendix is the work of Dr. D. O. North, RCA Laboratories, Princeton, New Jersey.

The p - n junction is operated in a forward direction until a steady-state condition is reached, and at time $t=0$, the forward bias is removed and the open-circuited junction voltage determined as a function of time. The solution for the open-circuited junction voltage when displacement currents are neglected and when the minority carrier density is small compared with the majority carrier density is

$$\frac{e^{\Delta V(t)} - 1}{e^{\Delta V_0} - 1} = \frac{J_p}{J_p - J_n} \left[1 - \operatorname{erf} \sqrt{\frac{t}{\tau_p}} \right] - \frac{J_n}{J_p - J_n} \left[1 - \operatorname{erf} \sqrt{\frac{t}{\tau_n}} \right] + \frac{\sqrt{J_n J_p}}{J_p - J_n} \sqrt{A} e^{-Bt} \left[\operatorname{erf} \sqrt{\frac{J_n}{J_p} A \frac{t}{\tau_p}} - \operatorname{erf} \sqrt{\frac{J_p}{J_n} A \frac{t}{\tau_n}} \right], \quad (12)$$

where

$$A = \frac{J_p J_n (\tau_p - \tau_n)}{J_p^2 \tau_p - J_n^2 \tau_n}, \quad (13)$$

$$B = \frac{J_p^2 - J_n^2}{J_p^2 \tau_p - J_n^2 \tau_n}, \quad (14)$$

and the various quantities have the following meaning:

$A = \frac{q}{kt}$ of suitable sign so that ΔV_0 is a positive quantity,

$V(t)$ = open-circuited junction voltage following $t=0$,

V_0 = forward junction voltage at $t=0$.

τ_n, τ_p = electron and hole lifetimes in p -type and n -type semiconductors, respectively.

$J_n = n_p \sqrt{\frac{D_p}{\tau_n}}$ = thermally generated electron current density in p -type semiconductor.

$J_p = p_n \sqrt{\frac{D_n}{\tau_p}}$ = thermally generated hole current density in n -type semiconductor.

n_p, p_n = electron and hole density present in p -type and n -type semiconductors respectively under equilibrium condition.

D_n, D_p = electron and hole diffusion constant in p and n semiconductors respectively.

$$\operatorname{erf} y = \frac{2}{\sqrt{\pi}} \int_0^y e^{-x^2} dx.$$

The solution given above is applicable to the general case where the n -type and p -type semiconductors have arbitrary characteristics. Certain special cases can now be considered.

1. If neither $\tau_n \rightarrow 0$ or $\tau_p \rightarrow 0$ and n_p and p_n remain finite, then respectively $J_n \rightarrow \infty$ or $J_p \rightarrow \infty$ and $V(t) = 0$. This is the case when the minority carrier lifetime in either semiconductor approaches zero.

2. If $J_n \rightarrow 0$ by $n_p = 0$, then

$$\frac{e^{\Delta V(t)} - 1}{e^{\Delta V_0} - 1} = 1 - \operatorname{erf} \sqrt{\frac{t}{\tau_p}}.$$

Since $n_p p_p = n_i^2 = a$ constant, $n_p \rightarrow 0$ is the same as $p_p \rightarrow \infty$. This is the case of the conductivity of the p -type semiconductor being infinitely large. In this event the minority carrier lifetime, τ_n , can be arbitrarily small provided only that $J_n \rightarrow 0$. The same limit solution is obtained if $J_n \rightarrow 0$ by $\tau_n \rightarrow \infty$. Due to the symmetry of (12), the solution for $J_p \rightarrow 0$ is obtained by interchanging τ_n for τ_p .

3. If $\tau_p = \tau_n = \tau$, then

$$\frac{e^{\Delta V(t)} - 1}{e^{\Delta V_0} - 1} = 1 - \operatorname{erf} \sqrt{\frac{t}{\tau}},$$

irrespective of the values of J_n and J_p .

4. If $J_n = J_p$, then

$$\begin{aligned} \frac{e^{\Delta V(t)} - 1}{e^{\Delta V_0} - 1} &= 1 - \frac{1}{2} \left[\operatorname{erf} \sqrt{\frac{t}{\tau_p}} + \operatorname{erf} \sqrt{\frac{t}{\tau_n}} \right] \\ &\quad - \frac{1}{2} \left[\operatorname{erf} \sqrt{\frac{t}{\tau_p}} - \operatorname{erf} \sqrt{\frac{t}{\tau_n}} \right] \\ &\quad \cdot \left[\frac{\tau_p + \tau_n}{\tau_p - \tau_n} + \frac{4t}{\tau_p - \tau_n} \right] \\ &\quad + \frac{2}{\sqrt{\pi}(\tau_p - \tau_n)} \left[\tau_n \sqrt{\frac{t}{\tau_n}} e^{-t/\tau_n} \right. \\ &\quad \left. - \tau_p \sqrt{\frac{t}{\tau_p}} e^{-t/\tau_p} \right] \end{aligned}$$

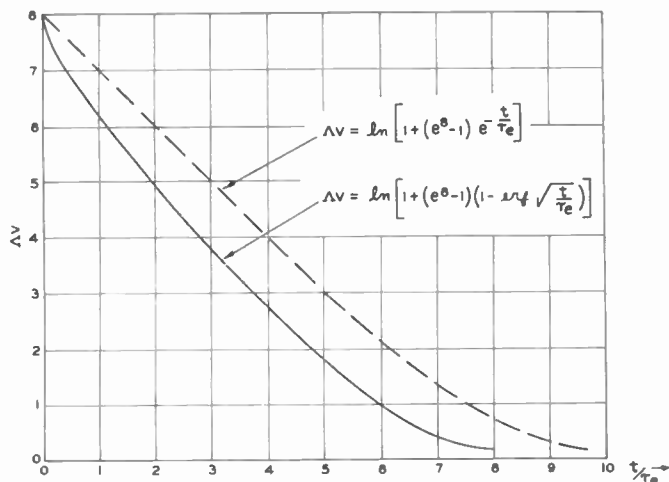


Fig. 6—Comparison of voltage decay for exponential and error function time dependency.

Case 2 is the solution applicable to the operation considered herein. This solution differs from the solution given in (6) which states that

$$\frac{e^{\Delta V(t)} - 1}{e^{\Delta V_0} - 1} = e^{-t/\tau_p}.$$

When the injected minority carrier density is small, the voltage decay does have a form similar to that given by the error function solution as shown in Fig. 6 for an arbitrary case of $\Delta V_0 = 8$. For a somewhat larger minority carrier injection level, the voltage decay has more nearly the form of that given by the exponential solution also shown in Fig. 6 for comparison. At still larger minority carrier injection the voltage decay exhibits a "hump" as described in the text.



Correspondence

Understanding the Gyrator*

The gyrator, postulated by Tellegen¹ as a new nonreciprocal network element, is attracting the attention of network theorists nowadays. Shekel² has shown that a four-pole network with nonreciprocal admittance matrix $\|Y_{ij}\|$ can be separated into the parallel combination of a reciprocal network and a gyrator (Fig. 1) of gyrating admittance $\gamma = (Y_{12} - Y_{21})/2$, i.e.

$$\|Y_{ij}\| = \begin{vmatrix} Y_{11} & Y_{12} - \gamma \\ Y_{21} + \gamma & Y_{22} \end{vmatrix} + \begin{vmatrix} 0 & \gamma \\ -\gamma & 0 \end{vmatrix}.$$

Carlin³ has found the necessary and sufficient conditions for the synthesis of nonreciprocal networks by means of reciprocal networks and real gyrators.

The gyrator's physical significance can be seen from the equivalent circuit of Fig. 2 with admittance matrix $\|Y_{ij}\|$; for an arbitrary value of the admittance Y_2 there follows:

$$V_1 = Y_{11} - Y_2, \quad Y_2 = Y_{22} - Y_2 \\ I' = (Y_{12} + Y_2)V_2, \quad I'' = (Y_{21} + Y_2)V_1.$$

* Received by the IRE, January 6, 1955.
¹ B. D. H. Tellegen, "The Gyrator, a New Electric Network Element," Philips Res. Rep. 3, pp. 81-101; 1948.

² J. Shekel, "The gyrator as a 3-terminal element," Proc. I.R.E., vol. 42, pp. 1014-1016; August, 1953.

³ H. J. Carlin, "Theory and Application of Gyrator Networks," Polytechnic Inst. of Bklyn., Res. Rep. 289; March, 1954.

Considering separately the system of two current generators it is seen that its total input power is $\text{Re}(Y_{12} + Y_2 + Y_{21}^* + Y_2^*)V_1^*V_2$.

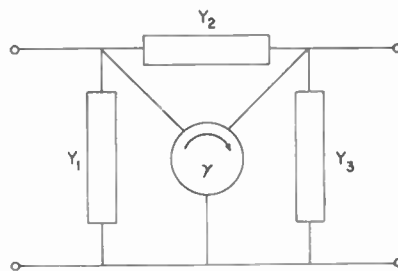


Fig. 1—Separation of a gyrator from a nonreciprocal network.

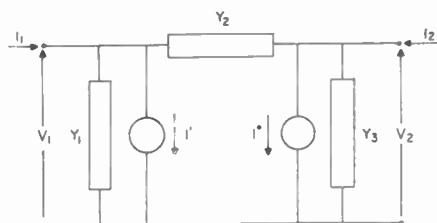


Fig. 2— π -equivalent circuit of a nonreciprocal network.

If in particular $Y_2 = -(Y_{12} + Y_{21})/2$, this power is zero and the system (I' , I'') reduces to the gyrator.

Similarly, starting from the network's impedance matrix $\|Z_{ij}\|$ and assuming an equivalent circuit of the type of Fig. 3, there follows for an arbitrary value of the impedance Z_2

$$Z_1 = Z_{11} - Z_2, \quad Z = Z_{22} - Z_2 \\ V' = (Z_{12} - Z_2)I_2, \quad V'' = (Z_{21} - Z_2)I_1.$$

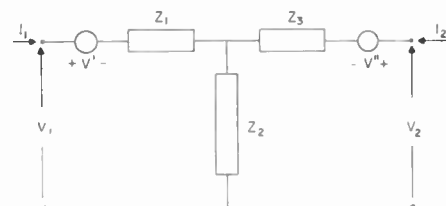


Fig. 3—T-equivalent circuit of a nonreciprocal network.

In particular, if $Z_2 = (Z_{12} + Z_{21})/2$, the system of two voltage generators reduces to a gyrator of gyrating impedance $\zeta = (Z_{21} - Z_{12})/2$.

These considerations suggest methods of simple realization of gyrators by means of current or voltage generators.

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Effect of Heisenberg's Principle on Channel Capacity*

The limitations imposed by thermodynamics on the amount of energy necessary to transmit one bit of information, has been discussed by Felker and Pierce.¹ A minimum of $kT \log_2 2$ ergs per bit was obtained. k is Boltzmann's constant, and T is absolute temperature.

Professor Fano has suggested that Heisenberg's principle may affect channel capacity.

The following analysis shows that the energy necessary to transmit one bit is not appreciably increased by quantum mechanical considerations, providing that

$$w \ll \frac{2}{3\pi} \frac{kT}{h}$$

w is the channel bandwidth in cycles per second and h is Planck's constant.

Consider a simple channel of bandwidth w , with a signal power per cycle of s/w ergs, and a noise power per cycle of N/w ergs. In order to optimize efficiency, we shall assume $N \gg s$.¹

A single measurement of the signal by the receiver will, according to quantum mechanics, involve some uncertainty in its energy. Let this uncertainty in energy be denoted by ϵ . It will contribute the equivalent of less than 2ϵ ergs of additional noise power per cycle.

Associated with this energy uncertainty is an uncertainty of time of measurement, which we will call τ . Heisenberg's principle states that

$$\tau = \frac{h}{\epsilon}$$

is about the most accurate in time-of-measurement we can obtain.

To find the power of the noise equivalent to the time-of-measurement uncertainty, consider that one is observing an equivalent of signal-plus-noise of flat spectrum

$$G(f) = \frac{s}{w} + \frac{N}{w} + 2\epsilon \text{ ergs,}$$

extending from $f=0$ to $f=w$.

Measuring the signal at time $t-\tau$, and using this measurement as an estimate of the value of the signal at time t , will almost always result in some error. The size of this error is the same as the amplitude of the difference between the original signal and the output of a hypothetical delay circuit of delay time τ , into which the signal could be fed.

The delay circuit is of frequency response

$$e^{2\pi i f \tau}$$

The error signal can be obtained from the original signal, by subjecting it to a filter of frequency response

$$1 - e^{2\pi i f \tau}$$

The spectrum of the error will be

$$\left(\frac{s}{w} + \frac{N}{w} + 2\epsilon \right) |1 - e^{2\pi i f \tau}|^2$$

To find the error power per cycle, we integrate this spectrum over all frequencies at which the signal exists, and divide by w , obtaining

$$\frac{1}{w} \int_0^w |1 - e^{2\pi i f \tau}|^2 \left(\frac{s}{w} + \frac{N}{w} + 2\epsilon \right) df$$

additional ergs of "noise" power per cycle.

Although the error in time of measurement will not always be τ , but will have a probability distribution of zero mean and width τ , a more exact treatment results only in an unimportant scale factor of the order of unity.

We may approximate this integral rather well by

$$\frac{4\pi^2}{3} N\tau^2,$$

if

$$s \ll N, \quad 2\pi w\tau \ll 1$$

and

$$2\epsilon \ll \frac{N}{w}$$

The total additional noise contribution due to both energy and time uncertainties is

$$\frac{4\pi^2}{3} Nw\tau^2 + 2\epsilon = \frac{4\pi^2}{3} Nw \frac{h^2}{\epsilon^2} + \epsilon$$

Since we may make ϵ arbitrary, let us choose it so that this total additional noise is minimized. We obtain

$$\epsilon = \left(\frac{4\pi^2}{3} Nw h^2 \right)^{1/3} \text{ ergs.}$$

The total equivalent increase in noise power then becomes

$$3 \left(\frac{4}{3} \pi^2 Nw h^2 \right)^{1/3} \text{ ergs.}$$

If this noise is to contribute negligibly to the channel equivocation, it must be much less than N/w , the ordinary thermal noise power per cycle, that is

$$3 \left(\frac{4}{3} \pi^2 Nw h^2 \right)^{1/3} \ll \frac{N}{w}$$

or

$$6\pi w h \ll \frac{N}{w}$$

Since $N/w = 4kT$, we obtain

$$w \ll \frac{2}{3\pi} \frac{kT}{h}$$

From purely dimensional considerations, it can also be shown that additional channel equivocation approaches zero, as wh/kT approaches zero, but no clear indication could be obtained as to the relative rates of approach.

Using $T = 300$ degrees absolute, we find $w < 1.6 \times 10^{13}$ cycles per second at room temperature.

This limitation on bandwidth is not serious from a practical standpoint, but even if one did want to transmit information faster than this, it would be possible to use

several independent channels in parallel, keeping the bandwidth of each below the limit, and still obtain an over-all channel capacity in excess of that suggested by the formula.

From the foregoing, it appears that Heisenberg's principle imposes no additional efficiency limitations on information channels.

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On Entropy Equivalence in the Time- and Frequency-Domains*

Shannon^{1,2} has found two different expressions for the entropy of a discrete, stationary, gaussian time series, by analysis in the time- and frequency-domains, respectively. It is interesting to note that by equating these two results a relationship is obtained which is of use in evaluating certain high-order determinants. This relation was first found by Szegő,³ and has recently been derived independently by Whittle^{4,5} using a different procedure. Thus we have the pleasing example of a rather obscure identity which can now be explained heuristically by an information-theoretic argument. Further, a result of Kolmogoroff⁶ and Wiener⁷ on the extrapolation of a discrete stationary time series can be seen to be a natural consequence of entropy considerations.

Using Shannon's expression for the entropy of an n -dimensional gaussian distribution, we have for the entropy per term, or per degree of freedom, H , of the discrete time series $\dots, x_{-1}, x_0, x_1, \dots$,

$$H = \lim_{n \rightarrow \infty} \frac{1}{n} \log [(2\pi e)^{n/2} |a_{ij}^{(n)}|^{1/2}], \quad (1)$$

where $|a_{ij}^{(n)}|$ is the determinant whose elements are a_{ij} :

$$a_{ij} = \overline{x_i x_j} = \phi(|i-j|); \quad i, j = 1, \dots, n. \quad (2)$$

Elias⁸ gives an expression similar to (1):

$$H = \lim_{n \rightarrow \infty} \frac{1}{n} \log [(2\pi e) |a_{ij}^{(n)}| / |a_{ij}^{(n-1)}|]. \quad (3)$$

Now let

$$F(f) = \begin{cases} \sum_{k=-\infty}^{+\infty} \phi(k) e_k \cos 2\pi f k; & 0 \leq f \leq 1, \\ \epsilon_k = \begin{cases} 1; & k = 0 \\ 2; & k \neq 0 \end{cases} \\ 0; & \text{elsewhere.} \end{cases} \quad (4)$$

* Received by the IRE, December 27, 1954. The research in this paper was supported jointly by the Army, Navy and Air Force under contract with Mass. Inst. of Tech.

¹ C. E. Shannon, "A mathematical theory of communication," *Bell Sys. Tech. Jour.*, vol. 27, pp. 379 and 623; October, 1948.

² G. Szegő, "Beiträge zur Theorie der Toeplitzchen Formeln," *Math. Zeit.*, vol. 6, p. 167; 1920, and vol. 9, p. 167; 1921.

³ P. Whittle, "Hypothesis Testing in Time Series Analysis," Almquist & Wiksells AB, Uppsala, Sweden; 1951.

⁴ P. Whittle, "Some results in time series analysis," *Skandinavisk Aktuarietidskrift*, vol. 1, p. 48; 1952.

⁵ A. H. Kolmogoroff, "Sur l'interpolation et extrapolation des suites stationnaires," *Compt. Rend. (Paris)*, vol. 208, p. 2043; 1939.

⁶ See also *Bull. Acad. Sci. (URSS)*, vol. 5, p. 3; 1941.
⁷ N. Wiener, "The Extrapolation, Interpolation and Smoothing of Stationary Time Series with Engineering Applications," Technology Press; 1949.

* Received by the IRE, October 4, 1954; revised manuscript received, November 26, 1954.

¹ J. H. Felker, "A link between information and energy," *Proc. I.R.E.*, vol. 40, pp. 728-729; June, 1952.

Then

$$\phi(k) = \int_0^1 F(f) \cos 2\pi f k df. \quad (5)$$

Applying Shannon's expression² for the entropy per degree of freedom of a gaussian process with limited spectrum, we find

$$H = \frac{1}{2} \int_0^1 \log [2\pi e F(f)] df. \quad (6)$$

Equating (1) and (6),

$$\lim_{n \rightarrow \infty} |a_{ij}^{(n)}| = \exp \left\{ n \int_0^1 \log F(f) df \right\}, \quad (7)$$

a result obtained by Szegő³ and Whittle.^{4,5} They have also obtained the result of equating (3) and (6),

$$\lim_{n \rightarrow \infty} \left[|a_{ij}^{(n)}| / |a_{ij}^{(n-1)}| \right] = \exp \left\{ \int_0^1 \log F(f) df \right\}, \quad (8)$$

a limit postulated by Polya in 1915.

Finally, using a suggestion of Elias,⁶ we may employ the Kolmogoroff-Wiener^{6,7} prediction theory to determine the entropy of a time series term when all preceding terms are known. This entropy, for a gaussian process, is given by (3), but is also given by

$$H = \frac{1}{2} \log 2\pi e \sigma^2, \quad (9)$$

where σ^2 is the variance of the irreducible error of the Kolmogoroff-Wiener procedure. This variance is found by Kolmogoroff⁶ and Wiener⁷ to be

$$\sigma^2 = \exp \left\{ \int_0^1 \log F(f) df \right\}. \quad (10)$$

Substituting (10) in (9), we obtain (6). Thus expression (10) may now be understood intuitively. Further, we have a simple demonstration, on information-theoretic grounds, of the well-known result that linear prediction is optimal for a gaussian time series.

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⁶ P. Elias, "A note on autocorrelation and entropy," *Proc. I.R.E.*, vol. 39, p. 839; July, 1951.

Beam-Hugging Plates for Unlimited Cathode Ray Deflection*

The need to present rapid single events with adequate brightness on a cathode-ray oscilloscope has driven tube makers to adopt signal deflecting plates which limit the picture height to a 2-inch on a 5-inch screen. Even such close-spaced plates require about 100 volts for a 2-inch deflection, and wide-band amplifiers for much larger undistorted signal output become quite unwieldy.

It is widely believed that increased deflection sensitivity by means of long and close-spaced plates must inevitably be paid for by limiting maximum deflection. But this is not the case if lateral predeflection and twisted beam-hugging plates are used.

How to design such a system can be learned from Fig. 1, showing after the second anode

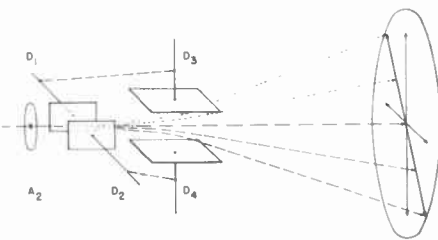


Fig. 1—All possible beam trajectories after subsidiary deflection.

A₂ two crossed pairs D_1, D_2 and D_3, D_4 of parallel plates. Omitting sweep deflection for the moment, and with the signal applied to both pairs connected together, the diagonal deflection is, of course, linearly proportional to signal voltage. The interesting point of the picture is that all possible paths—of which five are shown—of the electron beam form one twisted sheet, nowhere thicker than the beam. Therefore one may proceed in an imagined experiment in the following manner:

1. Leave the subsidiary plates D_1, D_2 unchanged (D_s).
2. Keep the main deflecting plates D_3, D_4 parallel (D_m).
3. Move D_3 and D_4 closer together, but stop wherever a surface point touches the beam; while
4. Simultaneously reducing the signal voltage applied to D_3 and D_4 in the same proportion as they are closer spaced.

With this procedure the deflecting field strength and all paths of the beam remain unchanged. At the end of the experiment both plates D_3, D_4 will just touch the sheet of possible trajectories on either surface and be twisted like it, parallel to it and to each other. But their close spacing requires much less signal voltage, although no limit is imposed on maximum angle of deflection.

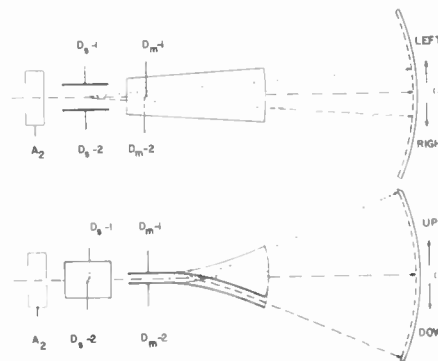


Fig. 2—Top and side view of beam-hugging deflecting plate system.

Fig. 2 shows top and side view of the so developed system D_1, D_2 combined from D_s and D_m . It is now clear that the subsidiary predeflecting plates D_s were introduced to spread the beam trajectories through the main deflecting plates. Beyond this, their contribution to the total signal deflection is of no importance. The vector diagram Fig.

3(a) shows how the total "vertical" deflection D_V is composed of D_s and D_m . A third pair of plates would provide scanning "horizontal" deflection D_H at right angle to D_V . Actually, the deflection due to the twisted main plates is more accurately represented by the—otherwise similar—diagram Fig. 3(b). The resultant deflecting angle of D_V

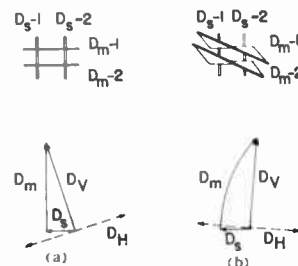


Fig. 3—End-on views of deflecting plates and resulting deflections.

depends on the relative effectiveness of the two pairs D_s and D_m but, like the shape of the plates, it is permanent for a given design. Fig. 4 shows that only one pair of signal terminals is brought out and that the location of the new sweep deflection plates D_3, D_4 is as usual.

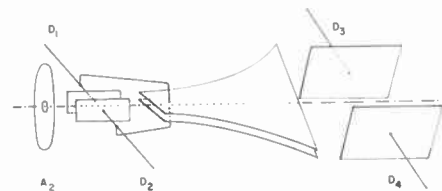


Fig. 4—Complete deflecting plate system.

The surfaces of such close-spaced plates must be accurate and smooth, or inhomogeneous fields will de-focus the beam. Defocusing by image charges due to the proximity of the plates need not be feared. A recent study¹ indicates instead that space charge depression due to such plates should rather aid the focus.

A model of a 5-inch tube, 17-inches long, has been tested. The main plates were spaced approx. 0.1-inch for a length of 2.5-inches. They had less than 5 μf capacity and a sensitivity of about 6v per inch per kv acceleration without limiting full screen deflection. Transit time in such plates should reduce, at 2 kv, the sensitivity by about 4 per cent at 100 mc.

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¹ J. S. Hickey, Jr. and T. G. Mihan, "The spreading of an electron beam," *Proc. I.R.E.*, vol. 40, p. 994; August, 1952.

Single-Sideband Transmission without Transient Distortion*

It has long been recognized that each sideband of an amplitude-modulated carrier contributes two components to the detector output signal, an undistorted in-phase com-

* Received by the IRE, December 22, 1954.

* Received by the IRE, December 22, 1954.

ponent and a 90-degree phase-displaced distortion. In double-sideband reception the former contributions add, the latter, being of opposite phase, cancel each other. If only one sideband is transmitted, failure of this cancellation shows up as a distorted transient response.¹ The vestige of the suppressed sideband in vestigial-sideband television systems serves to reduce this fault. It had also been understood² that the distorting out-of-phase component could be entirely eliminated in a synchronous detector whose beat oscillator is locked to the carrier frequency; but this solution seemed impractical at the time.

It would now appear that a very similar problem is being solved in the demodulator for the coloration subcarrier in color television. The essential step is to control the frequency and phase of a synchronous local oscillator by periodically comparing it with the received carrier at a time when that is unmodulated (or perhaps modulated by a known signal).

To suppress single-sideband distortion in television (monochrome or color) then requires the following steps, none of them at the transmitter:

1. Use a synchronous second detector, with a stable local oscillator at the nominal video IF carrier frequency.
2. Provide a gate circuit that opens when there is no modulation or a well-defined modulation, for instance during the color "burst."
3. During gate time compare the second LO in frequency and phase with the received IF carrier as it reaches the second detector.
4. The output of the comparison circuit then controls either the frequency of the second I.O., or perhaps that of the first LO, so as to minimize the error.

Failure of the system, for instance during warm-up, will merely mean that single-sideband distortion remains as now.

The key to the control of the synchronous oscillator is, of course, the periodic comparison during the synchronizing periods, and any other transmission system that provides such periods of clean carrier can thus be made a single-sideband system without penalty of transient distortion.

Regarding transmission systems that do not provide regular synchronizing intervals, the question remains whether double-sideband transmission with its redundant waste of half the bandwidth is the only possible way to avoid single-sideband distortion. This is not the case. For instance:

1. In a single-sideband system, let there be transmitted two pilot frequencies so chosen and locked that their difference after demodulation is an exact measure of the transmitter carrier frequency; or
2. Transmit one pilot frequency so controlled by the whole modulation at the transmitter that after demodulation it yields an exact measure of the true transmitter carrier frequency; or

3. In systems where intervals without modulation are sufficiently frequent, even if irregularly spaced, let the gate select such periods for comparison.

Such methods would not now seem very attractive, except for particular applications. On the other hand, phase distortion in a sharply cutting single-sideband filter need not be feared; suitable filters are known.³

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* H. E. Kallmann, "Transversal filters," *Proc. I.R.E.*, vol. 28, pp. 302-310; July, 1940, (see Fig. 7).

Quasi-Fraunhofer Gain of Parabolic Antennas*

The gain of a parabolic antenna over an isotropic radiator may be defined as

$$G = \frac{4\pi P_m}{W}, \quad (1)$$

where W is the total radiated power and P_m is the maximum radiated power per unit solid angle in the axial direction of the paraboloidal reflector.

Let us assume that the wave is everywhere in-phase at the aperture plane of the considered parabolic antenna (Fig. 1). Since all the wavelets from various parts of the aperture plane do not arrive simultaneously at point P , a phase error exists which will effect the measured value of gain. However, for $R \geq 2D^2/\lambda$, D being the aperture diameter, this phase error is small and the gain measured under this condition may approximate the true Fraunhofer gain, which is the value measured at $R \rightarrow \infty$.

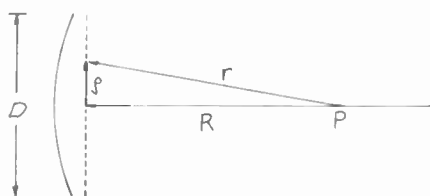


Fig. 1—Paraboloidal reflector.

In the practice of parabolic antenna design, tapered aperture illumination is employed for the reasons of optimum over-all efficiency and sidelobe reduction. Wavelets from regions near the edge of the aperture, which are largely responsible for the phase error, are now less weighted due to the tapering in the primary illumination toward the edge. It is the purpose of this note to show how the measured gain varies with R for several assumed tapered aperture illuminations. The transition region between the Fresnel and Fraunhofer regions, $D^2/4\lambda < R < 2D^2/\lambda$, is called Quasi-Fraunhofer region.

Assume that aperture illumination is:

$$E = E_0 \left[1 - \left(\frac{\rho}{a} \right)^2 \right]^n, \quad (2)$$

where a is the aperture radius and n is the tapering constant. If $(\rho/R)^2 < 1$, then we find $r \approx R + \rho^2/2R$. Thus the field¹ at P is

$$E_p = j \frac{E_0}{\lambda R} e^{-jkR} \int_0^{2\pi} \int_0^a \left[1 - \left(\frac{\rho}{a} \right)^2 \right]^n \cdot e^{-jk\rho^2/2R} \rho d\rho d\phi, \quad (3)$$

where $k = 2\pi/\lambda$. The corresponding radiation intensity at P is

$$P_m = \frac{1}{2} \eta R^2 |E_p|^2. \quad (4)$$

The total power radiated by the aperture is

$$W = \frac{1}{2} \eta E_0^2 \int_0^{2\pi} \int_0^a \left[1 - \left(\frac{\rho}{a} \right)^2 \right]^n \rho d\rho d\phi. \quad (5)$$

The above integrations may be readily carried out for $n=0, 1$ and 2 . On substituting (3), (4), and (5) into (1), we obtain

$$G = g_n 4\pi^2 a^2 / \lambda^2,$$

where g_n is the gain factor with the following expressions for $n=0, 1$ and 2 .

$$g_0 = \frac{1}{K^2} [\sin K]^2$$

$$g_1 = \frac{3}{K^4} [2 + K^2 - 2 \cos K - 2K \sin K]$$

$$g_2 = \frac{20}{K^6} \left[\left(1 - \frac{K^2}{2} - \cos K \right)^2 + (K - \sin K)^2 \right],$$

in which $K = ka^2/2R$.

It is not difficult to show that, as $R \rightarrow \infty$, we have the true Fraunhofer gain

$$G_\infty = g_\infty \cdot \frac{4\pi^2 a^2}{\lambda^2} = \frac{2n+1}{(n+1)^2} \frac{4\pi^2 a^2}{\lambda^2}.$$

The normalized quantities g_0/g_∞ , g_1/g_∞ and g_2/g_∞ are plotted in Fig. 2. It is noted that $n=0$ is the special case of uniform aperture illumination and the result is the same as that given by Silver.¹

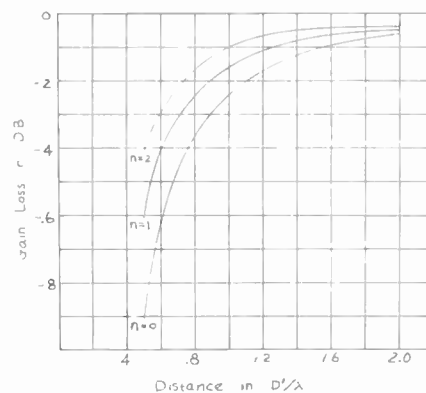


Fig. 2—Grain loss vs. distance for various tapered dish illuminations.

It is seen from Fig. 2 that, at a given distance, the error in the measured gain value is less for the tapered aperture illumination. If the primary feed pattern may be fitted into one of the family curves given by expression (2) for a large part of its major lobe, then Fig. 2 may be used as a correction curve to find the true Fraunhofer gain. This is permissible because the effects of the deviations between the hypothetical illumination [eq. (2)], and the practical primary feed pattern tend to average out in the process of integration.

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H. E. Kallmann, R. E. Spencer, and C. P. Singer, "Transient response of single-sideband systems," *Proc. I.R.E.*, vol. 28, pp. 557-563; December, 1940.

¹ *Loc. cit.*, p. 560.

* Received by the IRE, September 27, 1954.
¹ S. Silver, "Microwave Antenna Theory and Design," *Rad. Lab. Ser.*, vol. 12, pp. 198-199, McGraw-Hill Book Co., New York, N. Y.; 1949.

On Fourier Transforms in the Theory of Cathode-Ray Tubes*

It has been shown how Fourier Transforms can be used in a theory of determining the dynamic sensitivity of cathode-ray tubes at VHF.¹ Using the notations stated in the paper, the relative dynamic sensitivity of a cathode-ray tube is

$$\frac{A_d}{A_0} = \left| \frac{\int_{-b/2}^{b/2} \phi(x) e^{-i(2\pi x/\lambda_e)} dx}{\int_{-b/2}^{b/2} \phi(x) dx} \right|. \quad (1)$$

The determination of the relative dynamic sensitivity at VHF may be divided into two steps:

1. Determination of the static field strength distribution $E_y(x)$ along the x axis by either calculation or measurement when the forms of the plates are prescribed.

2. Determination of the sensitivity curve by means of the Fourier transform theory. As the electron velocity in the x direction is very nearly constant, $\phi(x)$ is proportional to $E_y(x)$, according to the equation

$$\phi(x) = \frac{d\varphi}{dx} \approx \frac{e}{mv_x^2} E_y(x). \quad (2)$$

Knowing $\phi(x)$, the normalized Fourier transform immediately gives the relative dynamic sensitivity A_d/A_0 , choosing a convenient length l of the field strength distribution.

Fig. 1 illustrates the well-known case of parallel plates neglecting stray fields and exit displacement.

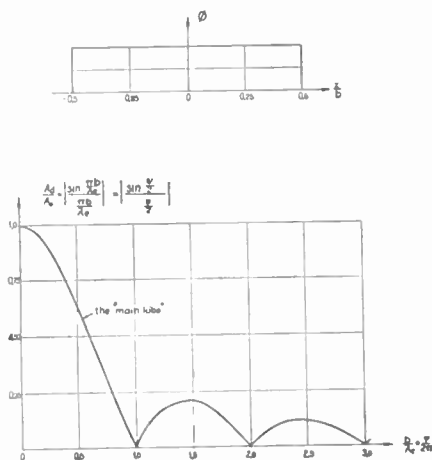


Fig. 1—The distributions of $\phi(x/b)$ and $A_d A_0(b/\lambda_e)$ in the parallel plates case.

In the same way as in the theories of antennas, inhomogeneous lines, time pulses, etc., a law equivalent to the uncertainty relation of Heisenberg determines the limits within which the distributions of the two Fourier transforms may vary. As is well known, the uncertainty relation has the following appearance for a particle with the momentum p in the x direction:

* Received by the IRE, October 28, 1954.

¹ E. Folke Bolinder, "A theory of determining the dynamic sensitivity of cathode-ray tubes at very high frequencies by means of Fourier transforms," *Trans. I.R.E. PGED*, scheduled for early publication.

$$\Delta p \cdot \Delta x \approx h, \quad (3)$$

where h = Planck's constant and the sign \approx means "is proportional to and is of the order of . . ."

Using the connection

$$= \frac{h}{\lambda}, \quad (4)$$

we obtain the corresponding formula for waves:

$$\Delta \frac{1}{\lambda} \cdot \Delta x \approx 1. \quad (5)$$

In the theory of cathode-ray tubes we get

$$\Delta \frac{1}{\lambda_e} \cdot \Delta x \approx \text{constant}, \quad (6)$$

or

$$\Delta \frac{l}{\lambda_e} \cdot \Delta \frac{x}{l} \approx \text{constant}. \quad (7)$$

Here $\Delta(l/\lambda_e)$ = the width of the A_d/A_0 distribution, and $\Delta(x/l)$ = the width of the ϕ distribution. l must be constant in all cases to be compared.

As the electron velocity v_x is constant through the deflection field, (6) may be written

$$\Delta \frac{v_x}{\lambda_e} \cdot \Delta \frac{x}{v_x} \approx \text{constant}. \quad (8)$$

or

$$\Delta f \cdot \Delta \tau \approx \text{constant}, \quad (9)$$

where τ is the electron transit time.

It is thus evident that the relative dynamic sensitivity is increased if the lengths of the plates are decreased, or if the electron velocity is increased. In the limiting case, when $\Delta \tau \rightarrow 0$, we obtain theoretically a field strength distribution in the form of a mathematical Dirac-pulse which corresponds to a horizontal A_d/A_0 distribution, i.e., a cathode-ray tube having the same relative sensitivity for all frequencies. However, at the same time the absolute sensitivity is approaching zero. This has been counteracted in modern cathode-ray tubes having small plates by using large distances between the plates and the screen and accelerating the electrons after deflection.

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Intrinsic Barrier Transistor*

A new junction transistor, the p - n - i - p has recently been described.¹ Included in its structure is a thick collector-depletion layer of intrinsic (i -type) semiconductor. Theory predicted the extension of the useful frequency range of junction transistors by greatly reducing collector capacitance, while maintaining low ohmic base resistance and high collector breakdown voltage.

Transistors have recently been constructed using laboratory techniques which

verify details of the theory. Alpha cutoff frequencies ($f_\alpha - 3\text{db}$) in the 50- to 100-mc range, collector capacitances in the 0.3 to 0.7 mmf range, ohmic base resistances between 100 and 200 ohms, and collector breakdown voltages greater than 100 volts have been obtained. The best unit produced thus far oscillates stably at 465 mc.

As shown in Fig. 1, collector capacitance

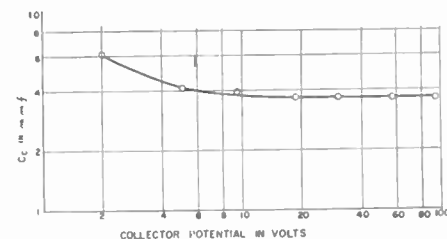


Fig. 1. p - n - i - p collector capacitance.

characteristically decreases with initial increase of collector bias and then becomes constant at higher voltage, as expected from theory. Collector reverse cutoff current (I_{co}), nominally below $20\mu\text{A}$ at 100v bias, increases 4.5 per cent per degree C. temperature rise. A concurrent study of small area p - i - n diodes with series ohmic resistances of less than 1 ohm showed a dynamic forward impedance of $2(kT/qI)$ (52 ohms at 1 ma). Again, this result agrees with theory.²

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² R. N. Hall, "Power rectifiers and transistors," *Proc. I.R.E.*, vol. 40, pp. 1512-1518; November, 1952.

Checking Codes for Digital Computers*

There has been considerable interest recently in the representation of decimal digits in digital computers by binary expressions. The most general binary code consists of ten arbitrary binary expressions assigned to the decimal digits in some order. The most obvious code is the so-called 8421 code, which represents the decimal digit n by the number n written in the binary scale. A more useful code is the excess 3, which represents the decimal digit n by the binary number $n+3$. Its advantage is the property of nines—complementing by interchange of zeros and ones.

While the $n+3$ code is simple and useful, requiring only four binary digits, and having a simple addition rule in addition to the complementing property, it does not allow of a check. This letter reports the results of a study of checking codes made at the Moore School, University of Pennsylvania, in 1950-1951, on contract with the Burroughs Adding Machine Co., by Morris Plotkin and myself.

A "check," in the sense used here, is an examination of the expressions at a given point in the machine to determine whether

* Received by the IRE, November 12, 1954.

* Received by the IRE, October 29, 1954.

¹ J. M. Early, "P-N-I-P and N-P-I-N junction transistor triodes," *Bell Sys. Tech. Jour.*, vol. 33, p. 517; May, 1954.

or not they belong to the code. If such an examination is to detect even a single error in a binary digit, the code must be so constructed that no code expression can be changed into another by a single error. It is convenient to define the "distance" between two binary expressions as the number of binary digits in which they differ. Thus, the expressions 1101 and 0011 are at distance three. A code is then said to be of distance d if the minimum distance between any two of its ten expressions is d . A code must, therefore, be of distance 2 or greater to qualify as a checking code. A code of distance d will give an alarm if $d-1$ or fewer errors occur simultaneously in a single expression.

The question then arises of constructing checking codes of distance 2, 3, etc., which retain the desirable addition and complementing properties of the $n+3$ code. It can be shown that a code must be of the form $an+b$ if addition is to be realizable through binary addition of the code expressions, with at most a constant additive correction (an additional correction is allowed in case of decimal carry). This requirement can be pictured as uniform spacing (with spacing n) of the code expressions in the list of consecutive binary numbers. The complementing property simply amounts to symmetrical spacing of the code expressions about the center line, for a given number of binary digits. Both points are illustrated in Table I, below, by the $n+3$ code, which is of distance 1, with four binary digits:

TABLE I

Decimal Digit n	$n+3$	Binary Expression
	0	0 000
	1	0 001
	2	0 010
0	3	0 011
1	4	0 100
2	5	0 101
3	6	0 110
4	7	0 111
5	8	1 000
6	9	1 001
7	10	1 010
8	11	1 011
9	12	1 100
	13	1 101
	14	1 110
	15	1 111

Naturally, it is desirable to use as few binary digits as possible in constructing a code. It is known that a code of distance 2 requires at least five binary digits, and a code of distance 3 at least seven. For the case of distance 2, the only five binary digit code with all properties is the $3n+2$ code. This code has the additional useful property of using all expressions of the form $3n+2$ in five binary digits, thus making for a simple checking process. Unfortunately, no comparable situation exists for distance 3 codes in seven binary digits—it is necessary to go to eight digits, making admissible the $27n+6$ code which has all properties, including the property of using all expressions of the form $27n+6$ in eight digits. But before dismissing the possibility of an acceptable distance 3 code in seven digits, several variations on the $an+b$ form were investi-

gated, including codes whose expressions fall into two blocks of five each, with the same spacing in each block, and codes consisting of the 8421 , excess 3, or 2421 nonchecking codes plus a three-digit check, which is added separately (though it is permitted to receive a carry correction). It was shown that no such possibility exists. Therefore, unless more complicated addition laws than the ones considered are found to be usable, the $27n+6$ code is the most desirable code of distance 3. It and the $3n+2$ code are believed to be new.

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"Valve Noise Produced by Electrode Movement"

With reference to the above paper,¹ a formula for cathode resonance following eq. (30) was given assuming 19.9×10^{11} dynes per square centimeter as Young's Modulus. This is the value at 20 degrees C. Since the cathode operating temperature is about 825 degrees C., the Young's Modulus corresponding to this temperature should be used.

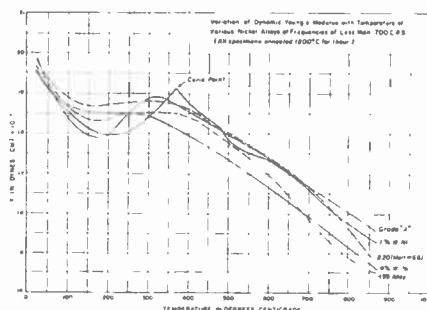


Fig. 1

As shown in Fig. 1, a considerable change in the Modulus occurs with increasing temperature. The curves shown are for grade "A" nickel, no. 220 alloy (melt #66), no. 499 alloy, 1 per cent and 4 per cent tungsten-nickel alloy.

Assuming a value of 15.25×10^{11} for no. 220 alloy at 825 degrees C., (30) reduces to:

$$f_{220} = \frac{1.24 \times 10^6 (r_1^2 + r_2^2)^{1/2}}{l^2}$$

The specimens plotted on the curves were measured by the transverse vibration method. The specimens were made with a cross section and length for resonance at 500–700 cps. The effect of higher frequencies is to lower the dynamic modulus at elevated temperatures. As lower frequencies are used, the modulus approaches that obtained by static testing. The highest value of modulus at cathode-operating temperatures will be that of a single crystal. Fine-grained nickel appears to have a somewhat lower modulus.

I am indebted to Mr. Richard L. Hoff, Assistant Development Metallurgist, Super-

* Received by the IRE, June 6, 1954.
¹ P. A. Handley and P. Welch, "Valve noise produced by electrode movement," *Proc. I.R.E.*, vol. 42, pp. 565–563; March, 1954.

rior Tube Company, Norristown, Pa., for this data and information.

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Rebuttal²

In the derivation of the formula for calculating the resonant frequency of a cathode, reference was made to the original empirical relationship for the resonant frequency of nickel rods supported by mica insulators and the value of Young's Modulus at 20 degrees C. was used in the absence of further information.

We are indebted to Mr. Glauber for his graph showing the change in Young's Modulus with temperature. It is of interest that this change depends upon the type of cathode nickel used and that therefore the resonant frequency must be modified by the ratio of the square root of this value to that of the square root of 19.9×10^{11} dynes.

During our investigational work we have not observed the effects of cathode resonance and you will note that the noise frequency diagrams do not show them. For this reason we have concluded that the actual resonant frequency of cathodes in miniature valves is generally so high that it does not affect the valve performance in the same way as the lower frequency grid resonances and low frequency rattle noises. However, the fact that cathode resonance will be lower than that given by the empirical relation by about 10–15 per cent shows that this could be a source of trouble if long thin cathodes were used.

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Russian Vacuum-Tube Terminology*

In a note under the above heading, on page 1023, vol. 42, of the June, 1954 PROCEEDINGS OF THE I.R.E., G. F. Schultz makes two statements which, to me, born and educated in Russia, appear not to be strictly correct.

The first statement is that "Russia has no 'h,' and this letter is commonly transliterated as 'g' in words of non-Slavic origin." The Russian language has a letter corresponding to "h." It is "x," and can be seen as the first letter of the Russian equivalent for the word "characteristics," as given by Mr. Schultz; it is a soft "h," as in "home," and in this way different from the hard "x" or "chi" of the Greek.

The second is that "Russian has no equivalent to the English 'plate'." There is a Russian equivalent to the term "plate." It is "Пластика," phonetically spelled "plastinka," and not infrequently found in Russian technical literature.

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* Received by the IRE, December 8, 1954.

Continuous Radar Echoes from Meteor Ionization Trails*

It is now established that there is a scattering medium in the *E*-region of the ionosphere which supports extended-range vhf radio propagation.^{1,2} That is, when sufficient power is used, a substantially continuous signal can be propagated over medium distances (500 to 2000 km) at frequencies (30 to 100 mc) which are above the "maximum-usable-frequencies" of the regular ionospheric layers. The exact nature of the scattering process has never been adequately demonstrated. The principal theories offered in explanation of *E*-region scatter are: (1) scattering from "blobs" of ionization which may be created by turbulence;^{1,3} and (2) overlapping of reflections from numerous meteor ionization trails.^{2,4,5}

We are here proposing a simple experimental method of determining the relative contributions of (1) and (2) to *E*-region scatter. On the basis of measurements taken in a preliminary application of this method, we conclude that extended-range vhf propagation may be almost entirely supported by reflections from meteor ionization trails.

If the scattering medium consists of a horizontal layer of blobs formed by isotropic turbulence, the number of scatterers in a narrow radar antenna beam varies as R^3 , where R is the range to the illuminated scatter region. (As used here, the term radar implies that the transmitter and receiver are at the same location.) The power scattered from each blob would vary as R^{-4} , so that the integrated power received from the illuminated region would be proportional to R^{-1} . Thus, the maximum radar response from blobs formed by isotropic turbulence would be obtained when the antenna beam is pointed vertically.

If the scattering medium consists of numerous meteor trails, the antenna elevation angle giving the strongest integrated radar echo is different from that required for blobs. The number of meteor trails illuminated by the narrow beam varies as R^3 . These trails, being long and thin, produce strong echoes only when they are normal to a ray from the radar. The power reflected from an individual trail which is so oriented is proportional to R^{-3} . It has been shown that the number of properly-oriented trails in a limited volume of the *E*-region is dependent upon the location of this volume relative to the radar site.⁴ In particular, very few meteor reflections can be obtained

from an area directly over the radar, since there are very few horizontal meteor trails. The maximum number of properly-oriented trails per unit volume occurs in those areas of the *E*-region which are about 100 km away from the point over the radar site. It follows that the radar antenna beam should be elevated about 45 degrees from the horizontal to obtain the maximum integrated response from meteor ionization trails.

Weak radar echoes have been observed from what appears to be a scatter region by groups at Ottawa⁶ and Saskatoon,⁷ Canada. The relative effect of the elevation angle of the antenna beam was not studied by either group. The experimental results obtained at these locations, when considered in terms of the antenna effects outlined above, appear to provide conflicting evidence on the cause of the echoes. The possibility of auroral effects also makes it difficult to use these results to differentiate between meteoric and other types of scatter.

A number of attempts were made at Stanford University during the summer of 1953 to detect *E*-region scatter with a radar system. The frequency used in the Stanford tests was 23 mc, and the average radiated power was about 2 kw. Range gating, coherent detection, and narrow bandwidth were used in the receiver for increased sensitivity.⁸ On those occasions when a high-gain, vertically-directed antenna was used, continuous scatter signals could not be detected at any range. On the other hand, continuous signals were readily observable when a fairly-broad antenna beam having maximum gain at an elevation angle of about 45 degrees was used. In this latter instance, the range gate was set to approximately 140 km.

In the experiments where the ever-present scatter echoes were recorded, strong individual meteor-bursts were very much in evidence. The remainder of the signal fluctuated randomly, as would be expected if it were due to the integrated effect of many small-amplitude, short-duration echoes occurring at random times. The characteristics of the continuous signal observed between the individually-discernible, larger meteor bursts is not of much help in explaining the nature of the scattering process, owing to the random nature of the integrated resultant whatever the causative agency. The fact that a continuous signal could not be obtained from overhead, whereas it could easily be obtained from more remote *E*-region areas, is regarded as evidence that overlapping meteor echoes were responsible for the total signal in the second case. From the wavelength and distance dependence of the signal amplitude measured in existing vhf propagation circuits,⁹ and the dependence of meteoric reflections on wavelength and path length,⁴ the conclusion is drawn that these preliminary radar results

provide support for the view that meteoric ionization plays the dominant role in extended-range vhf propagation.

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"A Mathematical Technique for the Analysis of Linear Systems"*

Ragazzini and Bergen¹ have shown how the *z*-transformation can be applied to the analysis of linear systems. In their method, the time response of a feedback control system can be obtained fairly readily as the coefficients of the infinite series that results when the numerator of the system pulse transfer function is divided by the denominator. In obtaining the over-all pulse transfer function, the Laplace Transfer Function of the individual component blocks in the feedback loop must be known in factored form.

A relation exists, which has not, to the writer's knowledge, appeared in the literature, that permits one to check the derivation of a particular *z*-transform when the continuous transform is known. The relation is:

$$\lim_{T \rightarrow 0} TF^*(z) = F(S) \quad (1)$$

where:

T = Sampling Interval

$F^*(z)$ = *Z*-transform

$F(S)$ = Laplace-transform.

This relation can be derived by noting that the Polygonal Approximation utilized¹ to the true time function approaches the true time function as the sampling interval approaches zero. The factor, T , is required in the equation to allow for the fact that the *z*-transform is based upon impulses of infinitesimal time duration rather than upon the generating triangles whose sum yields the Polygonal approximation.

The following example illustrates the use of (1). Given a *z*-transform that has been obtained by operation upon a time function:

$$F^*(z) = \frac{z}{z - \exp(atT)} \quad (2)$$

In this particular case, the function of time is $f(t) = \exp(at)$. Substituting (2) into (1), and utilizing the relation $z = \exp(ST)$ one has:

$$\lim_{T \rightarrow 0} \frac{T \exp(ST)}{\exp(ST) - \exp(aT)} = \frac{0}{0} \quad (3)$$

Applying L'Hospital's Rule:

$$\lim_{T \rightarrow 0} \frac{TS \exp(ST) + \exp(ST)}{\exp(ST) - a \exp(aT)} = \frac{1}{S - a} \quad (4)$$

The result of (4) is known to be the Laplace-transform of $f(t) = \exp(at)$.

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¹ Proc. I.R.E., vol. 42, pp. 1645-1651; November, 1954.

* Received by the IRE, January 10, 1955. This work was supported by the U.S. Navy (Office of Naval Research), the U.S. Army Signal Corps, and the U.S. Air Force, Contract N6onr-251 Task 7.

¹ D. K. Bailey, R. Bateman, L. V. Berkner, H. G. Booker, G. F. Montgomery, E. M. Purcell, W. W. Salisbury, and J. B. Wiesner, "A new kind of radio propagation at very high frequencies observable over long distances," *Phys. Rev.*, vol. 86, pp. 141-145; April, 1952.

² O. G. Villard, Jr., A. M. Peterson, L. A. Manning, and V. R. Eshleman, "Extended range radio transmission by oblique reflections from meteoric ionization," *Jour. Geophys. Res.*, vol. 58, pp. 83-93; March, 1953.

³ H. G. Booker and W. E. Gordon, "A theory of radio scattering in the troposphere," *Proc. I.R.E.*, vol. 38, pp. 401-412; April, 1950.

⁴ V. R. Eshleman and L. A. Manning, "Radio communication by scattering from meteoric ionization," *Proc. I.R.E.*, vol. 42, pp. 530-536; March, 1954.

⁵ D. W. R. McKinley, "Dependence of integrated duration of meteor echoes on wavelength and sensitivity," *Can. Jour. Phys.*, vol. 32, pp. 450-467; July, 1954.

⁶ D. W. R. McKinley and P. M. Millman, "Long duration echoes from aurora, meteors, and ionospheric back-scatter," *Can. Jour. Phys.*, vol. 31, pp. 171-181; February, 1953.

⁷ P. A. Forsyth, B. W. Currie, and F. E. Vawter, "Scattering of 56-mc/s radio waves from the lower ionosphere," *Nature*, vol. 171, pp. 352-353; February, 1953.

⁸ P. B. Gallagher and A. M. Peterson, "Ionosphere sounding by cross-correlation techniques," paper presented at the Western Electronic Show and Convention, San Francisco, Calif.; August 21, 1953.

⁹ D. K. Bailey, talk presented at the 11th General Assembly of URSI, The Hague, Netherlands; August 30, 1954.

High-Voltage Silicon Diodes*

Breakdown of a semiconductor junction diode is characterized by a rapid increase in back current as the applied reverse voltage increases slightly beyond some critical value. This value is such that the electric field across the junction, or parts of it, is sufficient to cause either Zener field emission or avalanche breakdown. The particular breakdown mechanism involved, as well as the magnitude of the critical voltage, depends on the semiconductor diode material, its resistivities, and the nature of the junction (step, linear gradient, etc.).

Although this paper is not concerned with the nature of the breakdown mechanism it should be noted that recent experimental evidence¹ indicates that the fields necessary for Zener emission are attainable only in narrow germanium junctions. Breakdown of broad germanium junctions and of silicon junctions in general is apparently of the avalanche type.

The high back resistance attainable in a properly made grown junction silicon diode, together with the superior temperature characteristics of silicon, make it an obvious choice for a "high voltage" diode. We shall arbitrarily list diodes with breakdown voltages in excess of one kilovolt in this class. Since the impurity distribution of a grown junction diode determines the magnitude of the breakdown voltage, the majority of the effort in such a project must necessarily be concentrated in the crystal growing stage where the impurity distribution is originally built in. For a given type of junction the breakdown voltage will, of course, be larger, the smaller the concentration of impurities (the higher the resistivity). However, it is important to realize that for an operational diode the forward resistance not only of the junction but of the entire unit must be kept as low as possible. Thus, the desired forward characteristics of the diode may set the upper limit on the resistivity.

When growing a junction crystal from the melt it is impossible to obtain a true "step" junction, at least as compared with that resulting from an alloy fusion process. However, when one considers the width of the space charge region at high reverse voltages, it becomes evident that this region is considerably wider than the junction itself, so that the abruptness of the original transition region (perhaps several tenths of a mil wide) becomes relatively unimportant. Calculations for one of the units discussed below indicate a space charge region width of 7 mils at a reverse voltage of 2.3 kilovolts. This is computed from the following expression:

$$d = \sqrt{2\epsilon(\mu_n\rho_n + \mu_p\rho_p)V} \text{ in M.K.S. units}$$

where

d = space charge width

ϵ = dielectric constant

$\mu_n\mu_p$ = drift mobilities of electrons and holes respectively

$\rho_n\rho_p$ = resistivities of the n and p regions respectively

V = applied reverse voltage.

* Received by the IRE, December 29, 1954. Supported in part by the Bureau of Ships, Department of the Navy, and the Signal Corps.

¹ K. G. McKay, "Avalanche breakdown in silicon," *Phys. Rev.*, vol. 94, pp. 877; May 15, 1954.

This is obtained directly from solution of Poisson's equation for a step junction.

From the above argument it might seem that the distinction between the "step" and "graded" junctions was entirely meaningless as regards high-voltage diodes. This, however, is not true, as the concentration gradients across the junction regions can be varied over such wide limits that the breakdown behavior differs considerably between the two types of junctions, even at high reverse voltages.

With these design criteria as a guide, an attempt was made to assess the practical possibilities of high-voltage grown-junction silicon diodes. Several high-resistivity junction crystals were grown and sliced into bars of suitable dimensions. Low resistance ohmic contacts were made to the ends of these bars. Various etching and surface techniques were used, and the best ones selected. The following figures will show the characteristics of two of these experimental units, which were contained in glass envelopes filled with an inert atmosphere.

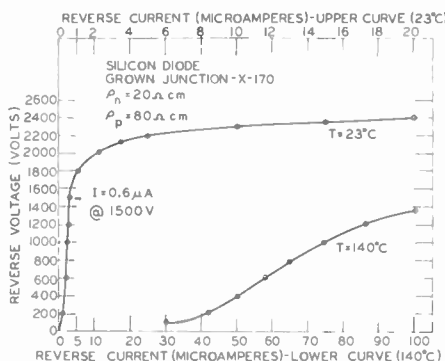


Fig. 1

Fig. 1 shows the reverse characteristics of a silicon step junction diode with resistivities of 80-ohm/cm on the p side and 20-ohm/cm on the n side. Data were taken at room temperature and at 140 degrees C., and were plotted on a linear scale to show detail in the high-voltage region. Points of interest on the room temperature curve (upper abscissa) include the 2,500 megohm resistance at 1,500 volts and breakdown at 2,300 volts. At 140 degrees C., the unit still exhibits better than 20 megohms of back resistance at 1,400 volts. The diode was about 50 mils square in cross section, so that the abscissas should be multiplied by about 50 to get the current density. (1 ma/cm² full scale for the upper abscissas, 5 ma/cm² full scale for the lower.)

Fig. 2 shows the forward characteristics of the same diode. The forward current was the same at 140 degrees C. as at room temperature. The forward resistance above 2 volts is lower than that of the 5V4G high vacuum rectifier, which is an indirectly heated cathode type. Of course, in terms of replacement possibilities, it must be pointed out that the 5V4G is a full wave rectifier and will pass 175 ma of forward current at a dissipation of 4 to 5 watts. The silicon laboratory diode discussed here is outclassed in this respect at this point.

Fig. 3 shows the result of measurements of an interesting diode. This unit was cut from a crystal of 40 ohm/cm resistivity for both the p and n side. Its concentration

gradient differs from that of the first unit shown in that it is a "graded" junction. With this unit, even though the resistivities are of the same order of magnitude as those of the step junction, one would expect a higher breakdown voltage. Various surface treatments were necessary before it became possible to obtain data at the highest voltages.

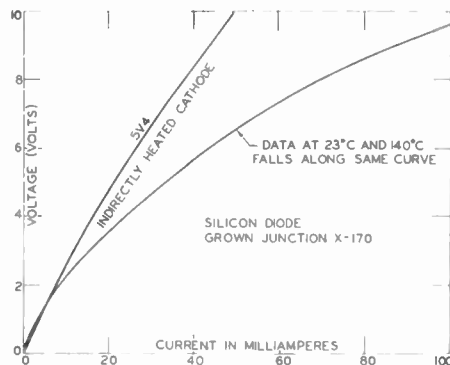


Fig. 2

The highest such point corresponds to less than 6 microamps at 5,300 volts. Above this voltage, a region of what was believed to have been surface instability was found. In this region, the readings were not reproducible, i.e., the currents of 12 and 16 microamps corresponding to 5,500 and 5,700 volts respectively could change by as much as 100 per cent from one instant to the next. The location of this unstable region was very much a function of the surface treatment. It is assumed that instability of surface prevented observation of a junction breakdown.

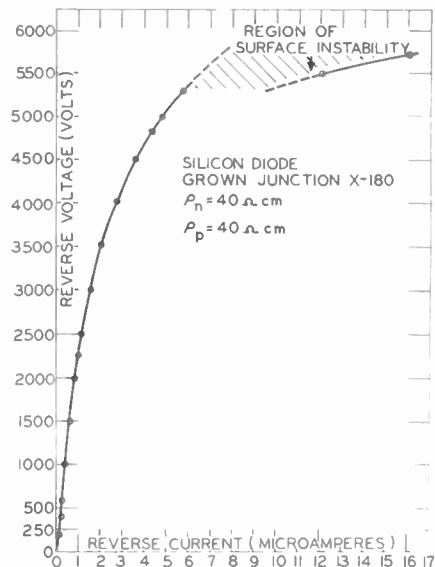


Fig. 3

It may thus be seen that an extension of the useful range of junction diodes has been made possible through the use of high-resistivity silicon junctions. For the first time, high-voltage characteristics are obtainable in a single unit solid-state rectifier such that the device becomes comparable to high-vacuum rectifier tubes.

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For a photograph and biography of A. B. Macnee, see page 1026 of the June, 1954 issue of the PROCEEDINGS OF THE I.R.E.

For a photograph and biography of A. R. Moore, see page 1026 of the June, 1954 issue of the PROCEEDINGS OF THE I.R.E.

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J. R. PIERCE

Since 1936 he has been a member of the technical staff of the Bell Telephone Laboratories, Inc., where he has been concerned with various vacuum-tube problems. In January, 1952, he became director of Electronics Research at Bell Laboratories.

In 1948 Dr. Pierce received the IRE Fellow Award for his "many contributions to the theory and design of vacuum tubes."

Dr. Pierce is the recipient of the Eta Kappa Nu "Outstanding Young Electrical Engineer" award for 1942, and the IRE Morris Liebmann Memorial Prize for 1947. Dr. Pierce is the author of two widely known books in his field, *Theory and Design of Electron Beams*, and *Traveling Wave Tubes*, and in addition has written many popular science articles for various magazines.

He is a Fellow of the American Physical Society, and a member of the American Institute of Electrical Engineers, the British Interplanetary Society, Tau Beta Pi and Sigma Xi. He is Editor of the PROCEEDINGS OF THE I.R.E. and a Director of the I.R.E., 1954-1955.

A. Singh received his M.S. degree in physics from Punjab University in 1945. Following this he was sent to the United States on a Government of India Scholarship. He received a Master of Engineering Science degree in 1947, and a Ph.D. in 1949, both from Harvard University.



A. SINGH

Dr. Singh was a lecturer in radio physics at the University of Delhi from 1949 to 1953. Since 1953 he has been a scientific officer with the National Physical Laboratory of India.

T. E. Talpey (S'47-A'50-M'53) was born on March 20, 1925, in Auburn, N. Y. He received the B.E.E. degree from Cornell University in 1946 and the M.S. and Ph.D. degrees in electrical engineering from the University of Michigan in 1948 and 1954. In 1951 he was awarded a Fulbright grant to study at the University of Grenoble, France, where he obtained the Doctorat d'Université in 1952.



T. E. TALPEY

From 1946 to 1953, he was an instructor in the Department of Electrical Engineering at the University of Michigan. In July, 1953 he joined the technical staff of the Bell Telephone Laboratories, Murray Hill, N. J., where, as a member of the Electron Tube Development Department, he has been engaged in the study of noise in grid-control tubes.

Dr. Talpey is an associate member of the AIEE and a member of Sigma Xi, Tau Beta Pi, Eta Kappa Nu and Phi Kappa Phi.

R. R. Warnecke (SM'48-F'50) was born on November 16, 1906, at Tours, France. He received the degree of Docteur de l'Université in Paris in 1933. Following this he became chief of the vacuum tube laboratory of the Société Française Radioélectrique; in 1940 he was head of the electronic tube research laboratory of the Compagnie Générale de Télégraphie Sans Fil and, in 1946, chief engineer at the research center of this company. He is now technical director of the Electronics Department in the same organization.



R. R. WARNECKE

Dr. Warnecke is a member of the Société Française de Physique, the Société Française des Electriciens, the Société des Radioélectriciens, and the Société des Ingénieurs et Techniciens du Vide. He received the IRE Fellow award in 1950 for "his engineering and research contributions to vacuum-tube theory and design in France."

Dr. Warnecke also received the Prix Ancel of the Société Française des Electriciens in 1943, the Prix H. Becquerel de l'Académie des Sciences de Paris in 1945, the Blondel Medal in 1951, and the Prix d'Aumale de l'Académie des Sciences de Paris in 1952. In 1954, he was the recipient of the Morris Liebmann Memorial Prize Award.

For a photograph and biography of W. M. Webster, see page 346 of the March, 1955 issue of the PROCEEDINGS OF THE I.R.E.

IRE Awards, 1955

Medal of Honor Award

Morris Liebmann Memorial Prize



ARTHUR V. LOUGHRIN

For his leadership and technical contributions in the formulation of the signal specification for compatible color television.



HARALD T. FRIIS

For his outstanding technical contributions in the expansion of the useful spectrum of radio frequencies, and for the inspiration and leadership he has given to young engineers.

Browder J. Thompson Memorial Prize



BLANCHARD D. SMITH, JR.

For his paper entitled, "Coding by Feedback Methods," which appeared in the August, 1953 issue of the PROCEEDINGS OF THE I.R.E.

Harry Diamond Memorial Award



BERNARD SALZBERG

For his contributions in the fields of electron tubes, circuits, and military electronics.

Vladimir K. Zworykin Television Prize



HAROLD B. LAW

For development of techniques and processes resulting in a practical form of shadow-mask tri-color kinescope.

New Fellows



V. J. ANDREW

For his contributions to radio antennas and transmission lines.



R. M. ASHBY

For his contributions to radar detection theory and integration of fire control-flight control systems for aircraft.



C. H. BACHMAN

For his contributions in the field of electron physics.



G. I. BACK

For his leadership in the field of military communications and communication systems.



B. G. BALLARD

For his direction of radar and electronic research in Canada.



G. S. BROWN

For his contributions to automatic control systems and to engineering education.



G. H. BROWNING

For his early contributions and his inspirational leadership in the electronics field.



KENNETH BULLINGTON

For his contributions to the field of radio propagation.



V. S. CARSON

For his contributions to the development and analysis of long-range aeronautical electronic navigation systems.

New Fellows



J. A. CHAMBERS

For his contributions to the development of high power broadcast transmitters and military electronic equipment.



R. D. CHIPP

For his contributions to the development of radar and television apparatus for the Navy.



C. E. CLEETON

For his contributions to microwave spectroscopy and electronic identification systems.



J. W. COLTMAN

For his contributions to the fields of microwave techniques, X-ray applications, and nuclear studies.



A. G. COOLEY

For his contributions to facsimile transmission methods.



F. A. COWAN

For his contributions to long-distance communication, particularly in the development of television network facilities.



C. C. CUTLER

For his research on microwave antennas and tubes.



HARRY DAVIS

For his contributions to the development of electronic aerial navigation systems.



J. W. DAWSON

For his contributions to the advancement of scientific and engineering knowledge.

New Fellows



R. L. DIETZOLD

For his application of mathematics to network design and military problems.



C. S. DRAPER

For his contributions to the theory and practical application of precise instrumentation and to engineering education.



O. M. DUNNING

For his contributions to the field of sound recording, and his effective organization of engineering effort.



J. B. FISK

For his contributions to the development of the magnetron and his leadership in basic electronic research.



J. W. FORRESTER

For his contributions to the development and engineering design of high speed digital computers.



G. L. FREDENDALL

For his applications of network analysis and synthesis to television system problems.



F. J. GAFFNEY

For contributions to the field of electrical measurements.



R. S. GLASGOW

For his contributions to the field of engineering education.



HAROLD GOLDBERG

For his contributions to the field of guided missile armament.

New Fellows



T. E. GOLDUP

For his pioneering achievements in the design and development of thermionic tubes and his contributions to the technical and administrative counsels of the British radio industry.



A. W. GRAF

For his contributions to the radio engineering profession.



C. E. GRANQVIST

For his contributions to air navigation systems and devices, and for his leadership in the engineering of electronic apparatus.



E. I. GREEN

For his contributions in the development of communication systems and apparatus components.



P. R. GUÉNARD

For his scientific and technical contributions in the field of microwave tubes.



W. A. HARRIS

For his contributions to the development of frequency converter tubes and to the understanding of fluctuation phenomena in electronic tubes.



A. E. HARRISON

For his contributions as a teacher, author and engineer, especially in the field of klystrons.



GERHARD HERZOG

For his contributions to radioactive instrumentation for geological survey and medical applications.



S. C. HIGHT

For his contribution to communication and weapon systems development.

New Fellows



G. W. O. HOWE

For his pioneering work in radio and his outstanding contributions to engineering education.



L. A. HYLAND

For contributions to aircraft radio direction finding, and his effective direction of research.



R. B. JANES

For his contributions to the development of improved camera tubes.



MARTIN KATZIN

For his contributions to the knowledge of microwave propagation.



V. R. LEARNED

For his contributions to research and development of microwave electron tubes.



E. A. LEDERER

For his contributions to the application of chemical and metallurgical science to electron tubes.



MEYER LEIFER

For his contributions to the fields of electronic navigation and information theory.



T. M. LIBBY

For his technical contributions and long service in the field of communications.



URNER LIDDEL

For his contributions to the establishment, promotion and integration of government sponsored nuclear research in academic institutions.

New Fellows



E. G. LINDER

For his contributions to microwave electronics.



B. D. LOUGHLIN

For contributions to color television, frequency modulation, and superregeneration.



C. J. MARSHALL

For his contributions to airborne television and radar research and development.



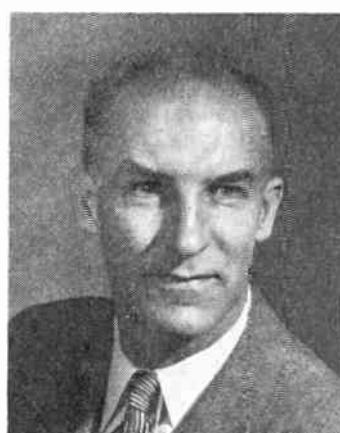
R. E. MOE

For his contributions to the field of electronics.



R. C. MOORE

For his contributions to television circuitry.



P. L. MORTON

For his contributions to the field of digital computers and the teaching of electronics.



W. A. NICHOLS

For his contributions to the construction of the national radio system in Canada.



R. S. OHL

For his contributions to the development of solid state point contact rectifiers.



W. H. PICKERING

For his contributions as teacher of electronics, and for his leadership in missile guidance, control and instrumentation.

New Fellows



J. R. RAGAZZINI

For his contributions in the fields of computers and control systems and as a teacher of these subjects.



E. G. RAMBERG

For his theoretical analyses of electronic devices.



W. G. RICHARDSON

For his contributions to the art of broadcasting, both sound and television, in Canada.



L. N. RIDENOUR

For his stimulating leadership in the field of electronic engineering.



H. E. ROYS

For his contributions to the improvement of disk and tape recording.



O. H. SCHMITT

For his contributions to the application of electronics to the study of living organisms.



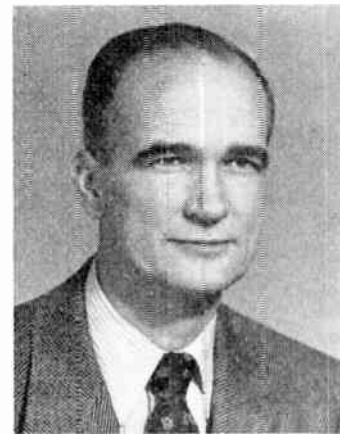
B. A. SCHWARZ

For his contributions to the development and production of automobile radios.



SAMUEL SEELY

For his contributions as an educator, author and as a director of research and development.



WILLIAM SHOCKLEY

For his contributions to development of the transistor.

New Fellows



C. M. SINNETT

For his contributions in the field of electronic circuitry.



C. E. SMITH

For his contributions to broadcast engineering and for his training activities.



J. E. SMITH

For his contributions to the art of radio communications.



P. L. SPENCER

For his contributions to the design and development of electron tubes.



G. C. SZIKLAI

For his contributions to television circuits and systems.



B. D. H. TELLEGEN

For his contributions and teachings in the field of vacuum tubes and communication networks.



J. R. TOLMIE

For his early contributions to radio.



W. G. TULLER (deceased)

For his contributions to the advance of theoretical analysis of information theory and its practical application.



C. H. VOLLUM

For his contribution to the development and manufacture of electronic laboratory instruments.

New Fellows

P. K. WEIMER

For his contributions to the development of television pickup tubes.



E. L. WHITE

For his leadership in advancing the use of radio in the interest of safety and efficiency in industry.



A. J. WILLIAMS, JR.

For his contribution to the field of self-balancing recorders of electrical quantities.



R. D. WYCKOFF

For his contributions to geophysical instrumentation, and the development of guided missiles.



IRE News and Radio Notes

IRE GRANTS MEMBERSHIP LEAVE TO MEMBERS IN ARMED SERVICE

At the January meeting, the Board of Directors voted on a policy to grant leave of absence from IRE membership to those in the Armed services. The leave will be granted, upon written request to IRE Headquarters, for a period up to three and one half years.

The member will have an inactive status during this period; he will not be required to pay dues, nor will he receive the PROCEEDINGS, or have other privileges of membership. When the member is discharged from service, he will be restored to the position he held preceding leave of absence. If Headquarters does not receive notice at the end of the three and one half year absence period, restoration to active membership will be made automatically.

NUCLEAR ENGINEERING CONFERENCE TO MEET IN CALIFORNIA

From April 27 to 29 a Conference on Nuclear Engineering will be held at University of California Los Angeles Campus.

A panel discussion will be featured each day of the conference: "Water and Liquid Metals as Primary Working Fluids," "Radiation Sources for Industrial Applications," and "Power Reactor Control During Load Changes." At an evening dinner session on April 28, John von Neumann, Institute for Advanced Studies, will speak.

The conference is sponsored by the UCLA Department of Engineering; Nuclear

Engineering Division, AICHE; Southern California Section, ASME; Southern California Section, AICHE; Golden Gate Chapter, ASM; San Francisco Section, AIEE; and Northern California Section, AICHE. Further information may be obtained from the University of California Extension, Los Angeles 24, California.

AUTOMATION SYMPOSIUM TO MEET AT MICHIGAN STATE IN MAY

The Engineering School of Michigan State College will sponsor a symposium called "Automation—Engineering for Tomorrow" on May 13. A part of the college's centennial year activities, the symposium will present general sessions, of interest to all branches of engineering, and special meetings, each presented by different departments and of interest to specific engineering fields.

Among the symposium speakers, will be W. R. G. Baker, Vice-President of the General Electric Company, and Eric A. Walker, Dean of Engineering at Pennsylvania State University. Dr. Baker will look at automation from a technical point of view, while Dean Walker will discuss it from the philosophical and sociological standpoint. Included for discussion at the general sessions are: Automatic Control Systems; Design of Systems; and Instrumentation of Automation.

Further details may be obtained from Prof. J. M. Apple, School of Engineering, Michigan State College, East Lansing, Michigan.

PGANE and Dayton Section Honor Dayton Univ.



Dayton University is honored for its support of the Dayton Section and PGANE of Dayton. From left to right, J. H. Parr and L. H. Rose, both of the university, receive scroll and set of MIT Radiation Laboratory Series from A. B. Henderson and P. G. Wiegert, the Chairmen of the Dayton Section and of Dayton PGANE respectively.

Calender of Coming Events

IRE-PIB Symposium on Modern Network Synthesis, Engineering Societies Building, New York, N. Y., April 12-15

Instrumentation Symposium, Proving Ground Instrumentation Committee of the American Ordnance Association, Patrick Air Force Base, Cocoa, Florida, April 14 and 15

IRE 9th Annual Spring Technical Conference, Cincinnati, Ohio, April 15-16

SMPTE 77th Semiannual Convention, Hotel Drake, Chicago, Ill., April 17-22

International Symposium on Electrical Discharges in Gases, Technical University, Delft, Netherlands, April 25-30

IRE Seventh Region Technical Conference and Trade Show, Hotel Westward Ho, Phoenix, Ariz., April 27-29

New England Radio Engineering Meeting, Sheraton Plaza Hotel, Boston, Mass., April 29-30

Semiconductor Symposium, Electrochemical Society, Cincinnati, Ohio, May 2-5

IRE URSI Spring Meeting, Washington, D. C., May 3-5

National Aeronautical Electronics Conference, Biltmore Hotel, Dayton, Ohio, May 9-11

IRE-AIEE-IAS-ISA National Telemetering Conference, Hotel Morrison, Chicago, Ill., May 18-20

AFCEA Global Communications Convention, New York City, May 19-21

IRE-AIEE-RETMA-WCEMA Electronic Components Conference, Hotel Ambassador, Los Angeles, Calif., May 26-27

American Society for Engineering Education Annual Meeting, Pennsylvania State University, State College, Pennsylvania, June 20-24

URSI-U. of Michigan International Symposium on Electromagnetic Wave Theory, University of Michigan, Ann Arbor, Mich., June 20-25

IRE-AIEE Conference on Industrial Electronics, Rackham Memorial Building, Detroit, Michigan, September 28-29

IRE PG on Electron Devices Annual Technical Meeting, Shoreham Hotel, Washington, D. C., Oct. 24-25

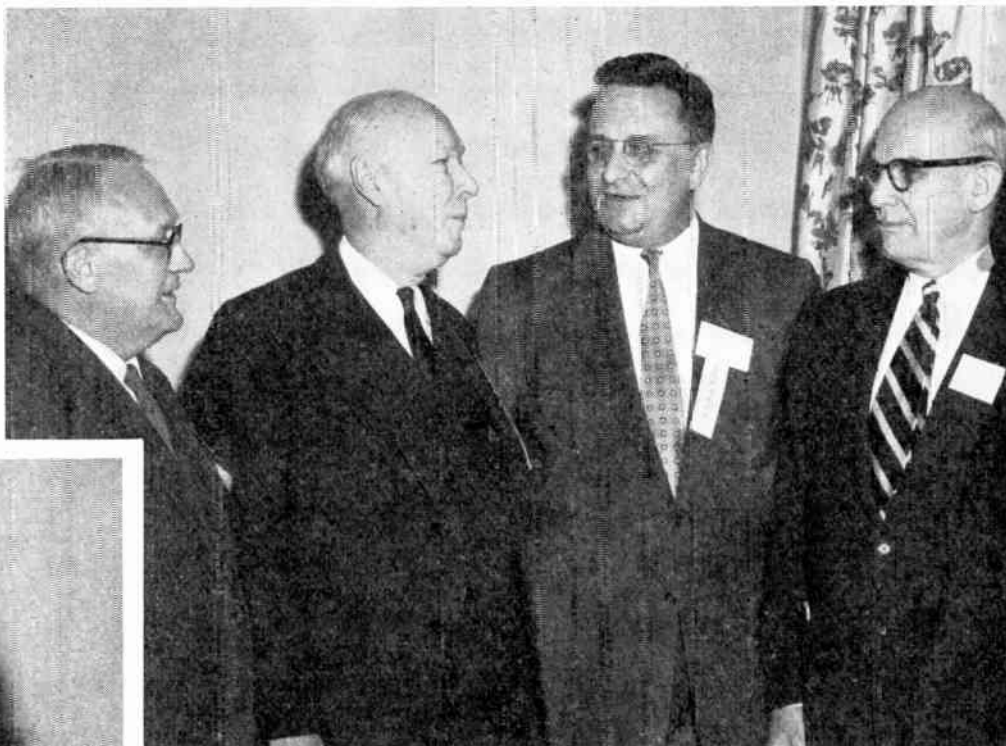
IRE-AIEE-ISA Eighth Annual Technical Conference on Electrical Techniques in Medicine and Biology, Washington, D. C., November

IRE Annual Electronic Conference, Towne House Hotel, Kansas City, Kansas, Nov. 3-4

IRE SOUTHWESTERN CONFERENCE

Sponsored by Dallas- Fort Worth Section

(Below) John A. Green, Chairman of the Dallas-Fort Worth Section which sponsored the Southwestern Conference. Mr. Green is president of John A. Green Company.



(Above) The opening day speakers at the Dallas meeting are T. A. Hunter, Editor of the IRE Student Quarterly and President of Hunter Manufacturing Company, W. V. Houston, President of Rice Institute, J. D. Ryder, President of the IRE and Dean of the Michigan State College School of Engineering, and George W. Bailey, Executive Secretary of IRE.

(Right) Conference leaders are R. A. Arnett of the Houston Technical Laboratories, and James G. Flynn, Jr., Vice-President of Collins Radio Company. They served as Vice-Chairman and Chairman, respectively of the Seventh Annual Meeting.



The Seventh Annual Southwestern IRE Conference and Electronics Show was held February 10 to 12 at the Baker Hotel in Dallas, Texas. The conference was sponsored by the Dallas-Fort Worth Section.

Special events included a reception for President John Ryder, an Executive Committee Meeting for Region Six, a ladies' program, a cocktail party and banquet, and a Directors meeting. John F. Jordan, IRE Fellow and Director of engineering and research for the Baldwin Piano Company, spoke at the banquet on "Musical Tone Color." Included in the special events were tours of seven plants in the Dallas area. At the Electronics Show there were exhibits by nearly eighty concerns.

Theme of the conference program was "Our Expanding Technology." The program included Airborne Electronics, Geophysics, Computers, Applied Electronics, Television, Automatic Control, Solid State Electronics, Microwaves, and Audio. Papers were presented by: Airborne Electronics—R. E. Lanier, B. L. Powell, B. H. Easter, E. H. Flath, Jr., and F. E. Schulte. Geophysics—

M. A. Arthur, G. W. Fordham, G. C. Summers, and H. J. Jones. Computers—L. E. Heizer, H. M. Martinez, J. F. Forrester, F. S. Preston, and P. A. Dennis. Applied Electronics—K. B. Bennett, F. W. Tatum, D. F. Sellers, W. G. Redmond, and C. J. Schultz. Television—J. W. Wentworth, I. C. Abrahams, D. A. Peterson, and J. P. Gallagher. Automatic Control—L. B. Wadel, E. J. Kompass, H. A. Spuhler, J. M. Salzer, R. G. Brown, and A. R. Teasdale, Jr. Solid State Electronics—W. J. Pietenpol, J. W. Englund, K. E. Loofbourrow, W. A. Adcock, and C. P. Dotson. Microwaves—N. L. Pappas, J. R. Lincio-come, V. Graziano, R. W. Haegle, and C. C. Campbell. Audio—W. Rudmose, J. D. Colvin, W. E. Stewart, and W. E. Seaman. Contributing to the success of the conference were: J. G. Flynn, Jr., *Chairman*; R. A.

Arnett, *Vice-Chairman*; J. H. Honsy, *Secretary*; J. S. McNeely, *Treasurer*; T. A. Wright, Jr., *Technical Program*; Mark Shepherd, Jr., *Registration*; Thomas B. Moseley, *Inspection Tours*; James O. Weldon, *Banquet*; John Albano, *Housing*; Theil Sharpe, *Exhibits*; Mrs. D. H. Clewell, *Ladies' Program*.

SEVENTH REGION TECHNICAL CONFERENCE WILL MEET MAY 27-29

From April 27 to 29 the Phoenix Section will be host to the Seventh Region Technical Conference and Trade Show at the Hotel Westward Ho in Phoenix, Arizona.

The three days of technical sessions will feature thirty-five papers on Engineering Management, Semiconductors, Specialized Components, Special Measuring Techniques and Testing Equipment, Missile Design Considerations and Missile Equipment, and Telemetering Problems. These papers will present interpretive rather than theoretical material, emphasizing information which can be applied to design and development problems.

The manufacturers will provide a show featuring the latest developments in the field of electronics and plans have been made for a women's program. Social activities will include an informal registration get-together on April 26, a conference luncheon on April 27, and a western barbecue dinner and cocktails, on April 28. Especially for the women, arrangements have been made for a fashion show featuring creations by Western designers, and a tour of the Valley of the Sun with a luncheon in Scottsdale, "the West's most Western town." On Saturday, April 30, special arrangements may be made for visits to Nogales, Mexico, Boulder Dam, Grand Canyon, and local resorts.

IRE MANUAL OF STANDARDIZATION NOW AVAILABLE TO MEMBERS

The IRE Manual of Standardization, which describes the procedure of standardization followed by Technical Committees, is now available to members. Copies may be obtained from the Office of the Technical Secretary at IRE Headquarters.

WILLIAM DUBILIER RECEIVES COOPER UNION ALUMNUS AWARD

William Dubilier, IRE Fellow and founder of the Cornell-Dubilier Electric Corporation, has received the Gano Dunn Medal for outstanding professional achievement by a Cooper Union alumnus. The annual award, given for the first time this year, was presented February 12 at the Founders Day Dinner in the Hotel Commodore, New York City.



Mr. Dubilier received the award for his contributions to the development of high voltage condensers. Among his other awards are the Chevalier Cross of the Legion of Honor, Officer of the French Academy, Grand Medal of the Association des Ingenieurs-Docteurs de France.

WILLIAM DUBILIER

CHICAGO TO BE HOST TO NATIONAL TELEMETERING CONFERENCE

The National Telemetering Conference, sponsored by the IRE, AIEE, IAS, and ISA, will meet May 18 to 20 at the Hotel Morrison in Chicago. The program will feature papers and exhibits in *Industrial Telemetering, Pickups and Transducers, Telemetering Components, Data Processing, Flight Testing, Multiplexing Techniques, Developments in Telemetry and Remote Control.*

Hugh L. Dryden, Director of the National Advisory Committee for Aeronautics will speak to the conference on "Problems in Ultra High Speed Flight." Luncheon speaker will be W. A. Wildhack of the National Bureau of Standards.

Inquiries regarding exhibits should be addressed to G. Brittain, Armour Research Foundation, Chicago, Illinois. For further information concerning the program, write to C. H. Hoepfner, Stavid Engineering, Inc., Plainfield, New Jersey.

SYRACUSE FORMS THREE CHAPTERS

On January 6 at the Syracuse Museum of Fine Arts three new professional group chapters were formed. These were the first chapters to be formed in the Syracuse Section and *pro tem* chairmen were selected to head each group. Donald E. Maxwell is Chairman of the Audio Chapter. Wilbur R. LePage of Syracuse University, heads the Circuit Theory Chapter, and C. Graydon Lloyd directs the Engineering Management Chapter.

Speakers on the program represented the various interests of the newly formed chapters. W. R. G. Baker, General Electric Vice-President and General Manager of the Electronics Division, was the first speaker. Dr. Baker gave a history of IRE professional groups, covering their initial conception and philosophy and their development up to the present. H. Brainard Fancher, General Electric General Manager of Germanium Products, reviewed the essential characteristics of a good manager. He defined the managerial function and explained how it was practically fulfilled. Norman Balabanian, Syracuse University, noted that the many branches of network theory arise from the application of various mathematical methods to network problems. The most used tool at present is complex function theory. Norman S. Cromwell, General Electric Radio and Television Department, discussed and demonstrated the application of low pass filters to prevent the energy in low frequency room and turntable vibrations from saturating and distorting the desired output of a high fidelity record player.



(Left to right) Past Chairman Richard Shea, Meeting Arrangement Coordinator; Speaker Norman Cromwell; Syracuse Section Chairman William Hall; Vice-Chairman Daniel Healy; Chairman Don Maxwell, Audio; Speaker H. Brainard Fancher; Speaker Norman Balabanian; Chairman Wilbur LePage, Circuit Theory; Chairman Graydon Lloyd, Engineering Management; Meetings and Papers Committee Chairman Don Arsen; Speaker Walter R. G. Baker, and the Secretary-Treasurer Major A. Johnson.

SUMMER COURSES IN NUCLEAR REACTORS AND AUTOMATIC CONTROL OFFERED BY U OF MICHIGAN

An intensive course on Nuclear Reactors and Radiations in Industry will be presented August 15 to 26 by the University of Michigan. Tuition, covering fee for a printed set of course notes, will be \$200 and the registration deadline will be June 1.

Sponsored by the Nuclear Engineering Committee of the College of Engineering, the course will be conducted by guest lecturers and staff members, including L. E. Brownell, H. J. Gomberg, W. Kerr, H. A. Ohlgren and J. R. Sellars. Further details may be received from Prof. William Kerr, Department of Electrical Engineering, University of Michigan, Ann Arbor, Michigan.

Two intensive courses in Automatic Control will also be offered by the university. Course I is scheduled for June 13 to 18 and Course II is scheduled for June 20 to 22. Closing date for registration is April 15.

The courses are built around the principles and application of measurement, communication, and control. Course I will consist of the fundamentals in each of these fields and will include work in nonlinear systems. Course II will consider applications of the fundamentals to more advanced problems. More information may be obtained by writing to Prof. L. L. Rauch, Room 1521, East Engineering Building, University of Michigan, Ann Arbor, Michigan.

TECHNICAL WRITERS WILL MEET

A two-day meeting of technical writers and editors will be held May 12 and 13 at the Statler Hotel in New York City. Ten papers will be presented by well-known representatives of industry, government, and education. The meeting will also discuss plans for a more formal organization of technical writers and editors.

Robert T. Hamlett of Sperry Gyroscope Company is Chairman of the group and Elsie Ray, Anaconda Copper Mining Company is Secretary.

TRANSISTOR BIBLIOGRAPHY READY

Through the courtesy of the Northwestern University Technological Institute, a comprehensive bibliography of transistors is now available. Reprints of the publication may be obtained without charge from A. R. Krull, Technological Institute Library, Northwestern University, Evanston, Illinois. Requests should designate "Transistors and Their Applications, A Bibliography, 1948-1953," by Alan R. Krull, IRE Trans. PGED, Vol. ED-1, No. 3, 1954, pp. 40-70.

SYMPOSIUM ON MODERN NETWORK SYNTHESIS TO BE HELD IN APRIL

An international Symposium on Modern Network Synthesis will be held April 13 to 15 at the Engineering Societies Building in New York. The fifth in a series by the Microwave Research Institute, it is part of the celebration commemorating the 100 anniversary of the Polytechnic Institute of Brooklyn.

The program will consider advances in the synthesis of passive networks in the frequency and time domains. It will include improved methods for designing RLC transducers and advances in the design of sampling filters. Developments in active and non-reciprocal circuits, such as unconventional applications of transistors, will also be presented. A roundtable discussion is planned on the significance of new network synthesis techniques to the solution of design problems in industry.

The cooperation of the PG On Circuit Theory and the co-sponsorship of the Office of Naval Research, Air Force Office of Scientific Research, and Signal Corps permits the symposium to be held without admission charge or registration fee. Volume V of the MRI Symposia Series, "Proceedings of the Symposium on Modern Network Synthesis, II" will be published by October 1955, at five dollars per copy. Copies of the program, hotel accommodation information and registration forms are available from: Polytechnic Institute of Brooklyn, Microwave Research Institute, 55 Johnson Street, Brooklyn 1, New York.

DIGITAL MACHINES FOR NATION-WIDE DIALING TO BE DISCUSSED

The Boston Section and the Boston Chapter of the Professional Group on Electronic Computers will hold a joint meeting on April 21 to discuss "Digital Machines for Nationwide Dialing."

John Meszar, Director of Switching Systems Development at Bell Laboratories, will outline the progress of the new automatic long distance switching system developed at the laboratory.

OBITUARIES

Charles Jackson Pannill, IRE Fellow, former President of the Radiomarine Corporation of America, and of RCA Institutes, Incorporated died recently.

Mr. Pannill held the first Certificate of Skill in Radio Communications and the first radio operator's license issued by the United States Government. He retired in 1947 after nearly half a century in radio communications.

His career in communications began in 1898 when he enlisted in the Navy as a telegrapher during the Spanish-American War. After tours of duty at the Norfolk Navy Yard and Coast Signal Service, he applied his telegraphic training to wireless communications under Professor Reginald A. Fessenden. It was at Brant Rock, Massachusetts in 1906 that Professor Fessenden conducted the first transatlantic demonstration of the spoken word. Mr. Pannill later assisted Professor Fessenden in the first installation of radio equipment aboard a United States battleship, and in the inauguration of overland wireless communications between New York, Philadelphia and Washington, D. C.

He joined the Marconi Wireless Telegraph Company of America, predecessor of RCA, in 1912 and served for two years as Superintendent of the Southern Division. Re-entering the Navy in 1914, he assisted in the establishment of the Naval Com-

munication System. He later became Assistant to the Director of Naval Communications, but resigned from the service in 1919.

During the following eight years, Mr. Pannill served as Vice-President and later President of the Independent Wireless Telegraph Company. In 1927, he was American delegate to the International Radio Conference. In 1928, he joined the Radiomarine Corporation of America, a service of RCA, as Vice-President and General Manager. He became Executive Vice-President in 1931, and was President from 1935 until his retirement in 1947. He served as President of RCA Institutes from 1932 until 1947.

Mr. Pannill received the Degree of Chevalier of the First Order of Leopold from the Belgian government for his work in international marine communications, and was awarded the Marconi Memorial Medal of Achievement by the Veteran Wireless Operators Association. He was a member of the Society of Naval Architects and Marine Engineers, the Cosmos Club, the New York Maritime Exchange, and the Board of Managers of Seaman's House of New York City. He was also a former Governor of the Propeller Club of New York.

PROFESSIONAL GROUP NEWS

THREE NEW CHAPTERS APPROVED

At a meeting on February 1, the Executive Committee approved petitions for the formation of the following chapters: PG on Audio, San Antonio Chapter; PG on Electronic Computers, Baltimore Chapter; PG on Information Theory, White Sands Proving Ground Chapter (El Paso Section).

TECHNICAL COMMITTEE NOTES

The **Antennas and Waveguides** Committee met at IRE Headquarters on January 12. Chairman P. H. Smith presided at the meeting. The possibility of standardizing impedance or reflection charts was discussed.

The committee discussed its published *Standards* which are in need of review (*Standards on Antennas, Modulation Systems, and Transmitters: Definitions of Terms, 1948 and Standards on Antennas: Methods of Testing, 1948*). It was announced that preparation of new definitions had been assigned to the West Coast Subcommittee and that H. Jasik's Subcommittee was preparing *Standards on Methods of Testing Waveguide Components*.

This committee met again on February 9 to discuss conflicts in the proposed *Definitions for Waveguide Components* with definitions approved by ASA Committee C42.

The **Facsimile** Committee met at the Times Annex in New York on January 7 with Vice-Chairman A. G. Cooley presiding and on February 11 with Chairman H. F. Burkhard presiding. The Committee's proposed definitions were discussed, and plans were made for reproducing the Facsimile Test Chart prepared by the committee.

The **Feedback Control Systems** Committee met at the MIT Faculty Club in Cambridge on January 11, with J. E. Ward presiding. The committee reviewed the pro-

posed *Standards on Graphical and Letter Symbols for Feedback Control Systems* and the proposed *Standards on Terminology for Feedback Control Systems*. The formation of a new subcommittee on Measurements was discussed. Terms prepared by G. Biernson and W. B. Williams were referred to the Definitions Subcommittee.

The **Radio Transmitters** Committee met on January 12 at IRE Headquarters with P. J. Herbst presiding. The committee formulated new terms in the field of spurious transmitter output. Three subcommittee chairmen made reports: Harold Goldberg reported that the proposed *Methods of Measurement of Pulse Quantities* was being revised and would be submitted to the main committee in two months. T. M. Ghayas, Jr. stated that work would be resumed on the proposed *Standards on Monochrome Television Transmitters*. Chairman B. Sheffield of Subcommittee 15.2 agreed to review with the subcommittee the suggestions of the main committee for the proposed *Standards on Methods of Testing Radio-Telegraph Transmitters (Below 50 MC)*.

The **Electron Devices** Committee met on January 21 at IRE Headquarters with Chairman W. J. Dodds presiding. The Subcommittee on Cathode-Ray and Television Tubes announced that they were beginning activity on methods of testing. The Subcommittee on Gas Tubes has prepared a list of Gas Tube Definitions. The Subcommittee on Camera Tubes, Phototubes, and Storage Tubes in Which Photoemission Is Essential reported that Camera Tube Definitions will soon be ready for considera-

tion by the committee and gave suggestions for liaison of the Electron Devices Committee and the new committee on nuclear science, soon to be formed. The Subcommittee on High-Vacuum Microwave Tubes reported that three task groups (Cold Test, Oscillators, and Amplifiers) were working on standards. The Physical Electronics Subcommittee reported that a review of definitions is almost completed. R. M. Ryder, Chairman of the Solid State Devices Subcommittee stated that proposed *Methods of Testing Transistors* are ready for submittal to the Standards Committee. He also reported on proposed standards in the following fields now in preparation in this subcommittee and its task groups: Large Signal Test, Point Contact Transistors; Letter Symbols for Transistor Qualities; Large Signal Tests for Junction Transistors; Single Tests on Power Transistors; Definitions of Hybrid Parameters. Recommendations for chairman of a Task Force on Ferroelectric and Ferromagnetic Devices were requested. Two subcommittees will make recommendations for assignment of work on temperature-sensitive resistors. It was noted that both the Tube Conference and the Semiconductor Conference were fully organized for 1955.

The **Measurements and Instrumentation** Committee met on January 31 at IRE Headquarters with P. S. Christaldi presiding. The following subcommittee reports were given: C. D. Owens (Magnetic Measurements Subcommittee) reported that the problem of testing ferrite core antennas for broadcast receivers would be considered by

this subcommittee. The written report of A. P. G. Peterson of the Subcommittee on Audio-Frequency Measurements stated that revision of proposed *Standards on Non-linear Distortion* was being continued in co-operation with two other IRE subcommittees. The Subcommittee on Video-Frequency Measurements reported that the work on angle and delay measurements should be completed by June, 1955. Dr. Showers, Chairman of the Interference Measurements Subcommittee, reported that his subcommittee was co-operating in the work of the Ad Hoc Committee on Spurious Radiation (Standards Committee). Definitions of terms compiled by the Subcommittee on Oscillography were described as nearly complete by the chairman, M. J. Ackerman. This subcommittee will also study measurements made with cathode-ray instruments. After hearing the subcommittee reports, the committee discussed the *Compilation of References to Methods of Measurement* which it plans to prepare.

The **Audio Techniques** Committee met at IRE Headquarters on February 9 with Chairman D. E. Maxwell presiding. L. D. Runkle, Chairman of the Definitions Subcommittee, reported on the action of his group on the proposed definitions. A report of Chairman R. C. Moody of the West Coast Subcommittee was read, describing the work on the proposed *Standard on Intermodulation Distortion*. The committee reviewed the proposed *Standards on Methods of Measurement of Gain, Loss, Amplification, Attenuation and Frequency Response* which is nearing completion.

Books

Electronics by A. T. Starr

Published (1954) by Pittman Publishing Corp., 2 West 45th St., New York City, 388 pages+7 page index+VIII pages, 352 figures, 8½×5½, \$7.50.

This book was written by Dr. Starr especially to cover the requirements of the electronics examination for the degree of Bachelor of Science in Engineering at the University of London. It makes no pretense at being a textbook. Within its scope it is an excellent reference volume; it should be a great help to engineers reviewing (not studying for the first time) the field of electronics in preparation for State Licensing or university doctoral examinations.

The emphasis of the book is on electronic circuitry rather than on the physics of electron tubes. Probably no two authors would agree on just what to include in such a book, but to this reviewer it seems that Dr. Starr has made a good selection. His presentation of topics is both clear and concise. At the end of each chapter there is a group of problems taken from recent London University examinations, and at the end of the book there is a set of answers to a few of these problems.

The six chapter titles are: "Physical Fundamentals"; "Valves"; "Rectification"; "Circuit Theory"; "Amplifiers, Oscillators and Detectors"; and "Electronic Applications."

There are, in addition, seven appendixes covering various mathematical topics.

This reviewer believes that there should be at least a few references to other literature on the various topics discussed, but, other than that, feels that Dr. Starr had done an excellent job.

F. T. McNAMARA

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Television, Second Edition by V. K. Zworykin and G. A. Morton

Published (1954) by John Wiley and Sons, Inc., 440 Fourth Ave., New York 16, N. Y. 1020 pages+XV pages+17 page index. Illustrated, 6½×9½, \$17.50.

What can be said in a love song that hasn't been said before? The second edition of this classic book is a complete reworking of the field and is even more of a gold mine than the first edition. Despite compression and omission of some of the topics which are better known now, the authors have been forced to expand the size of the book by approximately sixty per cent. Color fundamentals and practical color and industrial television systems account for about half of the expansion in four new chapters. The rest is just to take care of discussions of the new techniques evolved during the last fourteen years.

Every chapter has been revised and many completely rewritten.

The material covered ranges from the physics of electron emission and the fundamentals of color vision to discussions of the uses of television receivers in home management and interplanetary travel. The greatest stress is laid on those topics which are peculiar to television.

The treatment is thorough and restrainedly technical; mathematics is used where necessary. The prose is clear and readable. The book is well illustrated both with line drawings and half-tones. A knowledge of basic radio and circuit theory is assumed.

This is a must for every engineer interested in television.

KNOX MCILWAIN
Hazeltine Electronics Corporation
Little Neck, New York

REVIEW CREDIT CORRECTION NOTED

In the February issue H. J. Carlin was incorrectly credited with reviewing *Magnetic Amplifier Circuits* by W. A. Geyger. E. J. Smith, Polytechnic Institute of Brooklyn, should have been listed as the reviewer.

Electrical Transients by L. A. Ware and G. R. Towne

Published (1954) by Macmillan Co., 60 Fifth Ave., N. Y., N. Y. 219 pages + 2 page index + xi pages. Illustrated. 5½ × 8½. \$4.75.

This book is intended as a text for undergraduate electrical engineering students at the junior or senior level. It is primarily an introduction to the Laplace transform with applications to the solution of the differential equations of simple electric circuits. The position of the authors, that the teaching of the Laplace transform to undergraduate students is not only feasible but highly desirable, seems well taken. The limited aspects of the subject which are dealt with in this book require no mathematics beyond that normally given in undergraduate studies.

The title is perhaps inappropriate since there is very little discussion of transient phenomena as such. The direct Laplace transform is presented in Chapter Two and the remainder of the book is devoted to applications. The major emphasis is on the mechanics of the solution of circuit problems by Laplace transform. As an introduction to the transform method *Electrical Transients* is fairly good.

Unfortunately, the preoccupation with the manipulations required to solve various differential equations results in a complete neglect of the conceptual advantages afforded by the Laplace transform. The use of the impedance or admittance concept in obtaining transform relationships is ignored except for a footnote. Instead, the authors write down the differential equations for each problem and then transform them term by term. The link between the behavior of linear networks in the time and frequency domains, which is perhaps more important than any other single aspect of the Laplace transform, is also completely ignored. The treatment of initial conditions is not, as is strongly implied, the major advantage of the Laplace transform.

Many examples are worked out in the book and additional problems are given at the end of the chapters. The long involved treatment of the shunt-peaked video amplifier given in Chapter Nine seems of questionable value, but in general the problems are well chosen and should be helpful to the student. However, in spite of some good features and a worthwhile objective, the over-all treatment is not as good as it might be and leaves much to be desired.

LEONARD A. GOLDSTONE
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Brooklyn, N. Y.

Electromagnetic Theory by V. C. A. Ferraro

Published (1954) by John de Graff, Inc., 64 West 23 St., N. Y. 10, N. Y. 550 pages + viii pages + 5 page index. 161 figures. 8½ × 5½. \$7.00.

This book follows the historical approach set by Maxwell in his classic treatise, an approach which has been followed by many writers in electromagnetic theory. Ferraro's text differs from older treatises of this school (e.g. J. H. Jeans, *Electricity and Magnetism*, Cambridge 5th edition, 1933) in that vector notation is used, including an introductory chapter on vector analysis, and the book is modern in language and notation. It differs from other modern books of similar outline

(e.g. W. R. Smythe, *Static and Dynamic Electricity*, McGraw-Hill, second edition, 1950) in that classical cgs units are used rather than the rational mks system; there is slightly more emphasis on the mathematical proofs and somewhat less emphasis on specific applications. However, examples are worked out at the end of each chapter and there are extensive listings of problems, many taken from the Cambridge and University of London examinations.

The author states that the book is intended primarily for mathematicians and only secondarily for physicists and engineers. Thus there are several valuable proofs of uniqueness and existence for the stationary field. The major part of the text has, however, to do with the physical laws and their interpretation, and all of the book can be followed readily by graduates or advanced undergraduates in science and engineering.

Magnetic fields are introduced through the concept of magnetic charges. This approach has the advantage of stressing the similarities between electric and magnetic fields, but the disadvantage of masking the differences, and of requiring the difficult proof of equivalence between current flow and a magnetic shell. The author claims novelty for his proof of this point, and the proof seems direct.

The writing is concise and clear. The major fault is one which seems to come naturally to texts following the historical approach—treatment of the time-varying electromagnetic field is relegated to a secondary position out of keeping with its present importance. Less than one-fifth of this text is devoted to the time-varying field. Applications are seldom mentioned, and even the proofs of existence uniqueness, and convergence, which might be expected in a text for mathematics, are omitted.

J. R. WHINNERY
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Berkeley, California

Feedback Control Systems by Gilbert Howard Fett

Published (1954) by Prentice-Hall Inc., 70 Fifth Ave., N. Y., N. Y. 351 pages + 5 page index + ix pages + 4 page index + 4 page appendix. Illustrated. 8½ × 6½. \$10.00.

This book is a text covering the theory and analysis of feedback control systems. The presentation is pitched at a level suitable for senior undergraduate or graduate students or for practicing engineers who desire an introductory grounding in the field. The coverage of material ranges from a brief presentation of the components of feedback systems, time domain analysis of linear closed loop systems, a substantial discussion of frequency domain methods of analysis, to a chapter on non-linear systems.

The material is presented in a manner which reflects the fact that the author has had teaching experience. Explanations are clear and the order of presentation logical. On the other hand, there is a rather heavy emphasis on the more academic discussions of stability and stability diagrams at the expense of practical problems encountered in synthesis. For instance, considerable space is devoted to the Nyquist diagram and to modified transfer loci which result from the

mapping of paths other than the imaginary axis, such as a displaced imaginary axis, and of straight lines at a fixed angle with respect to the imaginary axis. While interesting from an academic viewpoint, these studies are not too important in practice. The low accuracy with which system constants are known and the difficulty of correlating these loci with time domain response do not always warrant such detailed and precise analyses of modified transfer loci.

The book would have profited from inclusion of more material on system design of feedback control systems. A unified analysis of the methods of compensation and their effect on over-all performance and a discussion of the practical advantages of one form or another is missed. In the treatment of multiple loop systems, techniques for the reduction of the system to an equivalent single loop system for study and design is not pointed out. Also the inverse (M^{-1}) plane, a powerful yet simple approach to the design of feedback systems with other than unity or proportional feedback, is not adequately exploited.

Despite some differences in viewpoint, this reviewer would like to reemphasize that the text is a careful and technically correct work and that a reader would benefit by studying it carefully. The book contains a comprehensive bibliography and a set of practice problems at the end of each chapter so that the reader is able to extend and apply some of the concepts developed in the text. Some of the more practical aspects of design and synthesis can be supplemented by use of other sources.

This text is well recommended for those engineers or students who wish to be introduced to the underlying theory of feedback control systems.

JOHN R. RAGAZZINI
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New York, N. Y.

RECENT BOOKS

Alsberg, Harold, Ed., *TV Field Service Manual with Tube Locations: Volume Three*. John F. Ryder Publisher, Inc., 480 Canal Street, New York 13, N. Y. \$2.10.

Courant, R. and Hilbert, D., *Methods of Mathematical Physics: Volume One*. Interscience Publishers, Inc., 250 Fifth Ave., New York 1, N. Y. \$9.50.

Gray, Truman S., *Applied Electronics*. John Wiley and Sons, Inc., 440 Fifth Ave., New York 15, N. Y. \$9.00.

La Joy, Millard H., *Industrial Automatic Controls*. Prentice-Hall Inc., 70 Fifth Ave., New York 11, N. Y. \$6.65.

MacLanachan, W., Ed., *Television and Radar Encyclopaedia*. Pitman Publishing Corp., 2 West 45 Street, New York, N. Y. \$6.00.

Noll, Edward M., *Television for Radiomen*. Macmillan Co., 60 Fifth Ave., New York 11, N. Y. \$10.00.

Snitzer, Milton S., *TV Manufacturers' Receiver Trouble Cures: Volume Six*. John F. Ryder Publisher, Inc., 480 Canal Street, New York 13, N. Y. \$2.10.

Zbar, Paul B., and Schildkraut, Sidney, *Advanced Television Servicing Techniques*. John F. Rider Publisher, Inc., 480 Canal Street, New York 13, N. Y. \$3.60.

THE 1955 NATIONAL CONFERENCE ON AERONAUTICAL ELECTRONICS

SPONSORED BY THE DAYTON SECTION AND
PG ON AERONAUTICAL AND NAVIGATIONAL ELECTRONICS

DAYTON OHIO, MAY 9-11

Monday Morning

SEMICONDUCTORS I—TRANSISTORS AND RECTIFIERS

Engineers Club, Auditorium

- "Medium Powered, Hermetically Sealed, Silicon Rectifiers for High Temperature Applications," A. Bergson and W. G. Mitchell, Raytheon Mfg. Co.
- "Semiconductor Power Rectifiers," J. W. Thornhill, Texas Instruments, Inc.
- "Some Practical Considerations Concerning the Limiting Operating Voltages of Junction Transistors," W. E. Sheehan, Raytheon Mfg. Co.
- "Transistor Pulse Characteristics," E. A. Hoskinson, North American Aviation, Inc.
- "Silicon Power Transistors and their Applications," J. W. Lacy and P. D. Davis, Jr., Texas Instruments, Inc.

ANTENNAS AND PROPAGATION I

Engineers Club, Italian Room

- "Investigation of the Electrical Characteristics of Low Frequency Transmitting Antenna Towers by Scale Model Measurements," Sidney Rosenberg and Paul Wilson, USAF Rome Air Development Center.
- "Aircraft Antenna System Lightning Protection," R. F. Huber, Joslyn Mfg. Co., M. M. Newman and J. D. Robb, Lightning & Transients Research Inst.
- "An Evaluation of Liaison Antennas for the Boeing Jet Transport," O. C. Boileau, Jr., Boeing Airplane Co.
- "Helicopter Antenna Design Considerations," A. R. Ellis, Stanford Research Institute.
- "The Antenna Crossover Problem in Conical Scan Radar," M. S. Wheeler, Westinghouse Electric Corp.
- "High Speed Sequential Lobing Antenna for Tracking Radar," J. T. McDonough, Westinghouse Electric Corp.

MANAGEMENT I—ENGINEERING AND PRODUCTION

Billmore Hotel, Main Ballroom

- "Management of a Study Program," N. V. Petrou and J. E. Darr, Westinghouse Electric Corp.
- "The Role of Electronics Research in Systems Engineering," Sidney Wald, The Glenn L. Martin Co.
- "Management and Production of Airborne Electronics in the Event of Atomic War," A. S. Brown, Stanford Research Inst.
- "A New Packaging Design Well Suited to Automation," D. H. Westwood, RCA.
- "An Approach to the Packaging of Sub-

- miniature Electronic Equipment," A. H. Stoney, Sylvania Electric Products, Inc.
- "Miniaturization and Unitization in Equipment Design," S. M. Stuhlberg, Crosley Div., AVCO Mfg. Co.

ELECTRONIC COMPONENTS I

Billmore Hotel, English Room

- "Design of Airborne Power Transformers from a Heat Transfer and Weight Point of View Using Forced Air Cooling and Metal Tape Windings," A. B. Cicero, Sylvania Electric Products, Inc.
- "Airborne High Temperature Transformer and Reactor Components," A. Lucic, North American Aviation, Inc.
- "Audio Frequency Selective Tunable Relay," Gerald Zomber, Avion Instrument Corp.
- "The Model 307 Photo-Electric DC Chopper," F. H. Davis, Avion Instrum. Corp.
- "Practical Design Criteria for High Order Mode Cavities," Amasa Pratt, Kearfott Co., Inc.

Monday Afternoon

SEMICONDUCTORS II—CIRCUITS

Engineers Club, Auditorium

- "Microwave Video Detection Characteristics of Crystals," R. E. Henning, Sperry Gyroscope Co.
- "Characteristics and Circuit Design for High Power Transistors," H. T. Mooers, Minneapolis-Honeywell Regulator Co.
- "Transistor DC-DC Converters," D. A. Paynter, General Electric Co.
- "Transistorized Time Encoder," J. C. Groce, Federal Telecommunication Labs., Inc.
- "A Silicon Transistor Resolver Amplifier," W. W. Wells, North American Aviation, Inc.
- "Transistor Application in a 2 to 8 MC Communications Receiver," H. J. Woll, RCA.

ANTENNAS AND PROPAGATION II

Engineers Club, Italian Room

- "Loop Antennas," Phyllis A. Kennedy and Thaddeus Kaliszewski, Harvard University.
- "Evaluation of Structural Dielectrics for Use in Flush Type Cap Antennas," Bruce M. Sifford and Henry J. Sang, Stanford Research Inst.
- "VSWR Circle Transformations," David A. Cope, Glenn L. Martin Co.
- "Obtaining a Uniform Field in the Diffraction Zone of a Large Aperture," J. O. Stenoien, Boeing Airplane Co.
- "Absolute Backscattering Measurements Employing the Synchrodyne Principle, Hybrid-T, and Image Plane in the

- K-Band," Capt. L. A. Yarbrough, USAF Institute of Technology.

- "Characteristics of Meteor Bursts on 15 MC Over a 608 KM Path," H. T. Castillo, Dayton, Ohio.

RELIABILITY

Billmore Hotel, Main Ballroom

- "Measuring, Assessing and Predicting Equipment Reliability," C. M. Ryerson, RCA.
- "System Function or Information Flow as a Measure of Reliability," A. Kohlenberg, Melpar, Inc.
- "Airborne Radar Reliability," A. M. Levine and A. J. Finocchi, Federal Telecommunication Lab.
- "Airborne Radar Reliability," L. A. Mayberry, Motorola, Inc.
- "Reliability in Complex Airborne Electronic Equipments," G. H. Scheer, USAF Wright Air Development Center.
- "Field Support of Complex Airborne Electronic Equipment," H. W. Brown, Jr., RCA.

MEASUREMENT AND TEST I

Billmore Hotel, English Room

- "Characteristics of X-Band Radar Test Set," Murray Kaye, Sperry Gyroscope Co.
- "A Calorimeter for Microwave Low Level Power Measurements," L. D. Strom, Texas Instruments, Inc.
- "Improvements in Calorimetric Wattmeters and Water Loads," Samuel Freedman, Chemalloy Electronics Corp.
- "Design Considerations for a New Type of Dummy Load," D. Self, Sperry Gyroscope Co.
- "Recent Developments on the National Bureau of Standards Microwave Refractometer," M. C. Thompson, Jr., National Bureau of Standards.
- "A Method of Wavelength Measurement for the Microwave and Millimeter Wave Regions," W. W. Balwanz, M. B. Rapport, and E. W. Ward, USN Naval Research Laboratory.

Tuesday Morning

FERROMAGNETICS AND PLASTICS

Engineers Club, Auditorium

- "Bimagnetic Applications in Airborne Control Systems," I. L. Auerback, Burroughs Corp.
- "A New Passive Magnetic Binary for Digital Applications," J. R. Horsch, General Electric Co.
- "Ferrite Duplexers for Microwave Radar Applications," T. N. Anderson, Airtron, Inc.
- "Plastics Material," J. H. DuBois, Mycalex Corp. of America.

"A New Class of Artificial Dielectrics for Microwave Applications," W. O. Puro, H. T. Ward, Jr., and D. M. Bowie, Melpar, Inc.

HUMAN ENGINEERING

Engineers Club, Italian Room

"A Miniature Airborne Pictorial Plotter," S. Romano, Avion Instrument Corp.

"A Preliminary Study of Operational Advantages of Pictorial Navigation Displays," F. S. McKnight, CAA Technical Development & Evaluation Center.

"Problems of Simulations with Human Subjects," M. Goetz, Westinghouse Electric Corp.

"Development of a Pilot Analog for the Single-Degree-of-Freedom Case," R. J. Mead and N. Diamantides, Goodyear Aircraft Corp.

"Some Human Engineering Problems in Fly by Wire Techniques," Arthur Kahn, Westinghouse Electric Corp.

"Human Engineering Analysis of Flight Director Systems," N. J. Cafarelli, Stavid Engineering, Inc.

COMPUTERS

Biltmore Hotel, Main Ballroom

"Gain Compensation for Airborne Analogue Computers," T. G. Nichols, Westinghouse Electric Corp.

"An Analogue Surface Function Generator," J. J. Earshen, Cornell Aeronautical Lab., Inc.

"Comparative Advantages of Airborne Digital Computers," D. L. Nettleton, RCA.

"Analysis of Systems Containing Digital Computers," E. Arthurs, RCA.

"The Flying Spot Scanner as a Digital Data Read Out Device," C. E. Jones, Federal Telecommunication Labs., Inc.

SERVOMECHANISMS

Biltmore Hotel, English Room

"Gain Equalization of Linear Servomechanisms which Solve Non-Linear Equations," G. E. Adams, Farnsworth Electronics Co.

"Feedback Control Systems Using Sampled Data," L. E. Mertens, RCA.

"Some Loading Effects on Servomechanism Performance," George Axelby, Westinghouse Electric Corp.

"Non-Linear Boost System Flow Characteristics and its Effect upon Autopilot Performance," A. M. Fuchs and F. J. Huddleston, Westinghouse Electric Corp.

Tuesday Afternoon

FORUM: THE WEAPONS SYSTEMS CONCEPT AND HOW IT AFFECTS AERONAUTICAL ELECTRONICS

Engineers Club, Auditorium

RADIO INTERFERENCE

Engineers Club, Italian Room

"Low-Impedance Gaskets for Radio-Frequency Applications," Verne Pulsifer and A. J. Hoehn, Armour Foundation.

"Measurement of Interference Fields about Aircraft," J. R. Stahmann, Lightning and Transients Research Institute.

"Radio Interference Control in Aircraft," A. L. Albin and J. E. McManus, Armour Research Foundation.

"A Study of Interference between Messages from Independent Multiple Sources on a Single Channel," Bobby Buchanan, USAF Cambridge Research Center.

"Interference Blankers," R. O. Engels, USAF Rome Air Development Center.

"Study of Noise Reduction by Feedback in Ultra-High Frequency Amplifiers," A. B. Glenn, RCA.

ENVIRONMENTAL CONDITIONS

Biltmore Hotel, English Room

"A Comparison of the Thermal Efficiencies of Subminiature Tube Shields Using a New Method of Measurement," L. C. Calhoun, Westinghouse Electric Corp.

"Relationship between Heat and Temperature or How Is a Dissipation in Watts Related to the Temperature of Parts," A. S. Gutman, Sylvania Electric Products, Inc.

"General Design Aspects for Cooling Electronic Equipment," M. Mark, Raytheon Mfg. Co.

"Reliable Tube Bulb Temperatures and Plate Operating Ratings," E. S. Mockus, Raytheon Mfg. Co.

Wednesday Morning

ELECTRON TUBES I

Engineers Club, Auditorium

"Some Results of a Comprehensive Program to Improve Tube Reliability," Arthur Kohlenberg, Melpar, Inc.

"Developmental Low Noise TW Tubes for L-, S-, and C-Band," P. R. Wakefield and A. G. Hogg, RCA.

"A Developmental High Power Tunable X-Band Pulse Magnetron for Airborne Applications," W. F. Beltz, RCA.

"A High Power X-Band Klystron," R. A. LaPlante, Philips Laboratories.

"Casting Waveguides Complete with Flanges by the Shell Molding Process," Samuel Freedman, Chemalloy Electronics Corp.

EQUIPMENT I

Engineers Club, Italian Room

"A Precise 60 CPS 6.5 KVA Power Source," F. A. Kahl, Bendix Radio.

"Ultra-Linear, Wide Range 400-Cycle to D-C Converter," Darwin Krucoff, Melpar, Inc.

"An Airborne Radio Sextant," R. M. Ringoen, Collins Radio Co.

"AVQ-10 Commercial Airlines Weather Radar," C. J. Monroe and Aubrey W. Vose, RCA.

"Precision Ranging with a Pulsed Optical Radar," Leonard Geller and John Lawton, Cornell Aeronautical Laboratory, Inc.

"Pod-Mounted Electronic Equipment has Advantages," H. A. Brelsford, RCA.

NAVIGATION I

Biltmore Hotel, Main Ballroom

"An All Weather Radio Sextant," D. O. McCoy, Collins Radio Co.

"Model Measurements of Rotor Modulation for VOR Antennas," W. E. Barrick, Electronics Research, Inc.

"Measurement of TACAN and VOR Bearing Errors," D. T. Latimer, Jr., USN Naval Air Test Center.

"Recent Developments in Distance Measuring Equipment (DME)," R. C. Borden, Civil Aeronautics Administration TDEC.

"Radio Beam Coupler System," Herbert Hecht and G. F. Jude, Sperry Gyroscope Co.

"The RHO in Navarho," Raymond Alexander, American Machine & Foundry Co.

MEASUREMENT AND TEST II

Biltmore Hotel, English Room

"Propeller Blade Angle—and Deflection—Measurement," J. C. Camm, Electronics Corp. of America.

"A Pulse System of Strain Recording," P. L. Toback, Armour Research Foundation.

"A Versatile 200 Channel Recorder for Static Stress Analysis," T. C. Fletcher, Beckman Instruments, Inc.

"A Miniaturized Telemetering System Resulting from Modern Design Techniques," I. P. Magasiny, Raymond Rosen Engineering Products, Inc.

"Flight Testing an Airborne Magnetic Tape Data Recorder," J. J. Dover, USAF Flight Test Center.

"Signal Generator," Norman Greenberg, Avion Instrument Corp.

Wednesday Afternoon

ELECTRON TUBES II

Engineers Club, Auditorium

"Magnetron Beam Switching Tube," Hilary Moss, Burroughs Corp.

"Recent Developments in the Raytheon Recording Tube," R. C. Hergenrother, A. L. Luftman, and C. S. Sawyer, Raytheon Mfg. Co.

"Stacked Ceramic Tubes," H. E. Sorg, Eitel-McCullough, Inc.

"Status of Stacked Tube Development," W. R. Wheeler, Sylvania Electric Products, Inc.

"Ceramic Techniques and Parts Fabrication for Vacuum Tube Applications," T. S. Stanislaw, Sylvania Electric Products, Inc.

INFORMATION THEORY

Engineers Club, Italian Room

"Z Transform for Multiple Sampled Systems," N. T. Simopoulos, U. of Dayton.

"Modern Network Theory Design of Crystal Filters for Communications and Navigation," M. Dishal, Federal Telecommunication Labs., Inc.

"Phase Detector for Pulsed I-F Signals," O. E. Linderman, General Electric Co.

"The Philosophy of Design of Data-Processing Systems," R. L. Whittle, Federal Telecommunication Labs., Inc.

- "A 30-Target Electronic Radar Simulator: Its Application to Human Engineering and Systems Research," Lowell Schipper, Ohio State University.
 "A Multiple Target Radar Simulator," Sidney Wald, Glenn L. Martin Co.

NAVIGATION II

Biltmore Hotel, Main Ballroom

- "The Magnetic Drum as an Aid for Air Traffic Control and Weather Reporting," G. E. Fenimore, CAA Technical Development & Evaluation Center.
 "A Novel Holding Pattern for Inbound Airplanes," C. E. Young, Cornell Aeronautical Laboratory, Inc.

- "Analytic Approach to the General Air Traffic Control Problem," L. J. Fogel and N. J. Cafarelli, Stavid Engineering, Inc.
 "Evaluation of the Rho/Theta Transponder System for Air Traffic Control," D. S. Crippen and J. E. Herrmann, CAA Technical Devt. & Eval. Ctr.
 "An Investigation of Ilas Beam Characteristics and Aircraft Tracks," Abe Tatz, Airborne Instruments Lab., and Capt. C. P. Thomas Wright Air Development Center.
 "A Broad Band Blue Lighting System for Radar Approach Control Centers," C. L. Kraft and P. M. Fitts, Ohio State University and Arthur Perong, USAF Wright Air Development Center.

CIRCUITS

Biltmore Hotel, English Room

- "A High Stability RF System for DME Interrogators," M. Feller, Federal Telecommunication Labs., Inc.
 "8.5-15 CM Plate Pulsed Reentrant Oscillator Circuits," W. E. Babcock, RCA.
 "TEM Mode Microwave Filters," D. V. Geppert and R. H. Koontz, Sylvania Electric Products, Inc.
 "A Wideband Low-Noise Amplifier for Millimicrosecond Pulses," Harry Kihn, RCA.
 "A Precision Omnibearing Selector for the Test and Adjustment of VOR Receivers," R. L. Olson, Collins Radio Co.

1955 IRE CONVENTION RECORD

All available papers presented at the 1955 IRE National Convention will appear in the IRE CONVENTION RECORD to be published in June. The CONVENTION RECORD will be issued in ten Parts, with each Part devoted to one general subject. The papers for each session are listed on pages 349-377 of the March issue.

Instructions on Ordering

1. If you are a member of a Professional Group and have paid the group assessment by April 30, you will automatically receive, free of charge, that Part of the CONVENTION RECORD pertaining to the field of interest of your group, as indicated in the chart below.

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CONVENTION RECORD

Part	Title	Free To Paid Members of Following Professional Groups	Prices for Members (M) College and Libraries (L) Non-Members (NM)		
			M	L	NM
1	Antennas & Propagation Sessions: 2, 10, 33, 40	Antennas & Propagation	\$1.00	\$2.40	\$3.00
2	Circuit Theory Sessions: 7, 32, 39	Circuit Theory	1.00	2.40	3.00
3	Electron Devices and Components Parts Sessions: 16, 23, 43, 44, 51, 52	Electron Devices Component Parts	1.50	3.60	4.50
4	Computers, Information Theory, Automatic Control Sessions: 8, 14, 24, 27, 35, 42, 48, 53	Electronic Computers Information Theory Automatic Control	2.25	5.40	6.75
5	Aeronautical and Navigational Electronics Sessions: 11, 19, 55	Aeronautical & Navigational Electronics	1.00	2.40	3.00
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8	Communications and Microwave Sessions: 3, 4, 28, 36, 47, 54	Communications Systems Vehicular Communications Microwave Theory and Techniques	2.00	4.80	6.00
9	Ultrasonics, Medical and Industrial Electronics Sessions: 5, 26, 34, 41, 49	Ultrasonics Engineering Medical Electronics Industrial Electronics	1.50	3.60	4.50
10	Instrumentation, Telemetry and Nuclear Science Sessions: 1, 6, 9, 15, 17, 22, 30, 45	Instrumentation Telemetry and Remote Control Nuclear Science	2.50	6.00	7.50
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NATIONAL TELEMETERING CONFERENCE

SPONSORED JOINTLY BY IRE, AIEE, IAS, AND ISA

CHICAGO, ILLINOIS, MAY 18-20

Wednesday Morning

COMPONENTS I

INDUSTRIAL TELEMETERING

Chairman, K. C. Black

- "A New High-Speed Telemeter Transmitter for dc Measurements," R. M. Stuart, General Electric.
- "A New Electronic Telemetering Transmitter for Pilot Wire Applications," T. Barabutes, Westinghouse Electric & Manufacturing.
- "An Incremental Remote Position Control System," Jonathan Mass, Kiryat Motzkin, P.O. Box 1, Haifa, Israel.
- "A New Time Interval Telemeter System," W. H. Howe, The Foxboro Co.

Wednesday Afternoon

SYSTEMS I

INDUSTRIAL TELEMETERING

Chairman, P. A. Borden

- "Pulse Telemetering for Industry," V. C. Kennedy, Jr., Streeter-Ames Co.
- "Automatic Teletype Transmitting System," J. R. Cunningham, Beckman Instruments, Inc.
- "Ultra Sonic Liquid Level Indicator System," R. L. Rod, Bogue Electric Mfg. Co.
- "Channels for Telemetering, Supervisory Control and Other Purposes," H. A. Rhodes, A. T. & T.
- "Telemetering System for Space Position Data," R. N. Nicola, The Newton Co.

COMPONENTS II

PICK-UPS AND TRANSDUCERS

Chairman, K. M. Uglow

- "Vibrotron Digital Telemetering System," J. Ohman, Southwest Research Inst.
- "A Commutatorless Direct Current Motor," H. D. Brailsford, Brailsford & Co.
- "Precision Data Recording and Repeating System (The Inductosyn)," J. L. Winget, Farrand Optical Co.
- "A Gravity Switch," P. Weaver, W. L. Maxson Corp.
- "A Phase Modulated Transistorized Pressure or Acceleration Telemetering Channel," A. I. Dranetz and J. L. Upham, Gulton Mfg. Corp.
- "New Developments in Miniature Telemetering Pick-Ups," L. A. G. TerVeen, Pacific Division, Bendix Aviation Corp.

Thursday Morning

SYSTEMS II

FLIGHT TESTING

Chairman, C. A. Taylor

- "New AKT-6 Flight Test," J. E. Spooner, Radiation, Inc.
- "Telemetry as a Flight Test Instrument,"

J. J. Dover, Air Force Flight Test Cntr., Edwards A. F. B.

- "A PDM-FM Telemetering System for Low Level DC Inputs," R. H. White, Natl. Advisory Committee for Aeronautics, Langley Aeronautical Lab., Langley Field.
- "An Analog Cross-Spectrum Analyzer for Telemetering," R. L. Kenimer, Natl. Advisory Committee for Aeronautics, Langley Aeronautical Lab., Langley Field.
- "Automatic Digital Recording of Flight Test Data," L. I. Goldfischer and S. G. Cohen, General Precision Labs.

Thursday Afternoon

SYSTEMS III

MULTIPLEXING TECHNIQUES

Chairman, E. L. Gruenberg

- "Mechanical Sampling Devices in Telemetering and Related Fields," J. F. Brinster, General Devices, Inc.
- "A New Subminiature Airborne FM Demultiplexer," L. Finkel, F. Shandelman, and J. Piontkowski, Raymond Rosen Engineering Products, Inc.
- "The Magnetron Beam Switching Tube," H. Moss, Burroughs Corporation Research Center
- "A Mercury Jet Commutating Switch," W. R. Davis, Detroit Controls Corp.
- "Miniaturized Airborne Electronic Commutator," R. O. DuBois, Electro-Mechanical Research, Inc.
- "Telemetry Filters and Their Effect on the Dynamic Accuracy of Multiplex FM Subcarrier Instrumentation Systems," G. S. Slough, R. A. Bunyan, W. H. Duerig, and G. E. Tisdale, Electro-Mechanical Research, Inc.

COMPONENTS III

NEW DEVELOPMENTS IN TELEMETRY AND REMOTE CONTROL

Chairman, J. T. Mengel

- "Mixing Airborne Telemetering Subcarriers for Maximum Isolation with Minimum Loss," W. F. Link, Pacific Division, Bendix Aviation Corp.
- "A New Ground Station Telemetering Receiver," M. S. Redden and H. W. Zancanata, Nems-Clarke, Inc.
- "The Use of AC Excited Gages in a PDM/PM Telemeter System," W. F. Carmody, Pilotless Aircraft Division, Boeing Airplane Co.
- "A New Instrumentation Direct Writing Recorder and its Application to Telemetry," G. E. Bower, Century Electronics Co.
- "Precision Multi-Channel Heads for Magnetic Tape Recording," A. V. Gangnes, Ampex Corp.
- "A Digital Approach to Telemeter Testing," C. R. Reid, Aerophysics Dept., Goodyear Aircraft Corp.

"An Automatic Landing System for Aircraft," M. H. Goldstein, Jr., and C. W. Merriam III, M.I.T.

Dinner

Speaker: Dr. Hugh L. Dryden, Director, National Advisory Committee for Aeronautics

"Problems in Ultra High Speed Flight"

Friday Morning

SYSTEMS IV

NEW DEVELOPMENTS IN TELEMETRY AND REMOTE CONTROL

Chairman, W. J. Mayo-Wells

- "Radar Beacon Telemeter," J. W. Poliseo, Stavids Engineering, Inc.
- "Flight Control Group AN/DRA-2," W. H. Eggerton, Melpar, Inc.
- "A Pulse Telemetering System for Use on Balloon-Launched Rockets," L. R. Davis, Naval Research Lab.
- "The AN/DKT-4 and AN/MKR-1 Telemetering System," T. B. Jackson, U. S. Naval Ordnance Lab.
- "Silicon Transistor Applications in Telemetering Equipment," C. E. Earhart and O. A. Becklund, Texas Instruments.
- "Coherent Pulse Telemetry," A. H. Cooper, E.M.I. Engineering Development, Ltd., Hayes, Middlesex, England

Luncheon

Speaker: Dr. William A. Wildhack, National Bureau of Standards
"In-Accurate Transmission of Mis-Information"

Friday Afternoon

COMPONENTS IV

DATA PROCESSING

Chairman, C. F. West

- "The Role of Magnetic Tape in Data Recording Processing and Analysis," G. L. Davies, The Davies Laboratories, Inc.
- "Talking to a Computer," R. F. Shaw, Electronic Computer Division, Underwood Corp.
- "A Precision Pressure Telemetering System with Digital Data Handling," J. Prast, Bell Aircraft Corp.
- "An Automatic Digital Data Reduction System Utilizing PDM Telemetering," R. F. Hummer, R. M. McClung, and D. J. Simmons, U. S. Naval Ordnance Test Station, Inyokern Aviation Ordnance Dept., China Lake, Calif.
- "Data Reduction Equipment Used with the FALCON Missile," H. D. Greif, Hughes Research & Development Laboratories, Hughes Aircraft Co.

Abstracts of Transactions of the I.R.E.

The following issues of Transactions have recently been published, and are available from the Institute of Radio Engineers, Inc., 1 East 79 Street, New York 21, N. Y., at the following prices. The contents of each issue and, where available, abstracts of technical papers are given below.

Sponsoring Group	Publication	Group Members	IRE Members	Non-Members*
Aeronautical and Navigational Electronics	Vol. ANE-1, No. 4	\$1.00	\$1.50	\$3.00
Audio	Vol. AU-3, No. 2	.95	1.40	2.85
Electronic Computers	Vol. EC-3, No. 1	1.10	1.65	3.30
Reliability and Quality Control	PGQC-4	1.20	1.80	3.60
Telemetry and Remote Control	PGRTRC-2	0.95	1.40	2.85
Proceedings of the WESCON Computer Sessions		1.45	2.15	4.35

* Public libraries and colleges may purchase copies at IRE Member rates.

AERONAUTICAL AND NAVIGATIONAL ELECTRONICS

VOL. ANE-1, No. 4, DECEMBER, 1954

Editorial—K. E. Black

Electronic Simulators for Study of Aircraft Flight Paths—S. L. McDonough. The simulation facilities described in this paper were developed at the Cornell Aeronautical Laboratory, Inc. in Buffalo, N. Y. under the sponsorship of the U. S. Navy Bureau of Ships. The simulator was designed for use as an aid in the study of the problems of coordinating many aircraft in a given geographic area. Correspondingly, the equipment is only as sophisticated as the requirements of the special problem dictated. The main significance of the simulator lies in its ability to develop with a small operating crew, a complex aircraft traffic situation as might be viewed by a ground based surveillance radar. Further, the simulator is capable of generating signals for use in testing and developing experimental data processing, computing, and recording equipment.

Several methods of simulating simplified aircraft tracks in three dimensions are included. Simulation equipments based on these methods are in use as data sources for track-while-scan radar repeater presentation and for direct use in ground installed computing and recording equipment. The methods involve both integration and multiplication techniques in generating voltages which represent tracks in rectangular coordinates. The voltages may be used directly in computers, or video circuits may be employed to convert aircraft position to polar form for PPI and RIII presentation. The simulation methods vary in the degree of aircraft maneuverability from non-maneuvering straight-line aircraft to that of fixed rate of turn aircraft. The complete installation has the capacity to simultaneously generate 24 aircraft paths.

High Voltage Problems in Flush and External HF Antennas—R. L. Tanner. At frequencies near the low end of the hf band the radiation resistance of aircraft antennas is almost always very small compared to antenna

reactance. To radiate appreciable power, therefore, large voltages must be applied to the antenna terminals, resulting in a severe voltage breakdown problem at the high operating altitudes of modern aircraft. This paper discusses the mechanisms of breakdown at radio frequencies and relates them to the problem of voltage protection of cap-type and fixed wire antennas and associated components. The use of models in the design of cap-type antennas for high breakdown voltage is considered, and experimental data are given on rf breakdown voltage as a function of altitude for typical geometries.

Development of the Ring Goniometer for Radio Direction Finders—Yoji Ito and Isokazu Tanaka. The radio direction finder of the Bellini-Tosi type requires a radiogoniometer to compare the direction of wave arrival to that of the rotating angle of the search coil. The newly designed Ring Goniometer described is the most efficient developed thus far due to high coupling coefficient (70–80%) and low coupling or octantal error (one degree without correction). It is simple to construct, consisting of two coils which are wound on ring-shaped concentric iron cores.

This paper presents the detailed theoretical analysis and method of measurement of such a ring goniometer.

A New Airborne DME Interrogator Designed for Stable Operation and Ease of Maintenance—A. R. Applegarth. This paper describes the technical design of the NARCO Model UDI-1 DME Interrogator. The transmitter section consists of a ten channel crystal controlled oscillator followed by a chain of amplifier-frequency multipliers to provide 500 watts pulse output in the band 960 to 990 mc. The receiver section is a ten channel superheterodyne using a crystal controlled local oscillator, which operates in the band 1185 to 1215 mc. The range unit contains electronic ranging circuitry along with improved magnetostriction delay code and decode systems, an extremely simple identity signal detector, and a novel means for causing intentionally irregular interrogations. The mechanical design which permits the set to be easily separated into its major components for maintenance is also described.

AUDIO

VOL. AU-3, No. 2, MARCH-APRIL, 1955

IRE-PGA News

A Versatile Audio Spectrometer—W. O. Essler. A new method of obtaining audio power spectra is described. Advantages of the new method are speed of use and great frequency of detail. The new method is based on the production of a "composite signal" which has a time continuous power spectrum. The composite signal method can be used with many types of signals such as speech, music, traffic noise, et cetera.

Triode Cathode-Followers for Impedance Matching to Transformers and Filters—T. J. Schultz. The material presented in this paper gives a selection of ready-made triode cathode follower circuits applicable to the problems of impedance matching to audio filters and transformers. Curves resulting from bread-board tests are given on five tube types, and an illustrative example is included.

Noise Analysis with a Heterodyne Type Sonic Analyzers—J. D. Richard, Jr., P. Smith, and F. H. Stephens. A technique for obtaining noise spectrum levels using a heterodyne type sonic analyzer is described. The analysis of a specific noise recording is shown as an example. A method is also described for obtaining spectrum levels directly from the cathode ray tube photographs.

Basic Principles of Stereophonic Sound—W. B. Snow. Stereophonic sound has become of vital importance to industry. The subject has been studied for many years, but the published material is scattered. This paper summarizes the fundamental theory underlying stereophonic sound so far as it has been published, and gives examples of how the theory is employed in representative practical situations. Fundamental differences between ordinary binaural listening and stereophony are pointed out, as well as similarities. It is shown that much qualitative but little quantitative information has been reported. Factors which aid some stereophonic effects are shown to be detrimental to others, and methods of minimizing the undesirable conditions are suggested. Applications to recording are discussed.

ELECTRONIC COMPUTERS

VOL. EC-4, No. 4, MARCH, 1955

News

Reviews

Engineering Description of the ElectroData Digital Computer—J. C. Aldrich. The operation of the ElectroData digital computer, control console, input-output equipment, and power units is described. The chief logical components are described in detail and the characteristics of some circuits are given, with a description of logical operation. Basic design decisions are stated.

Transistor Circuitry for Digital Computers—C. L. Wanlass. Transistor circuitry is presented that enables the construction of a digital computer which will operate at a clock frequency of 200 kc or less. The circuitry employs readily available germanium or silicon junction transistors of the type used in audio-frequency circuit work. A new system of diode gating is also presented as a necessary part of the circuit philosophy. No vacuum tubes are required or used with the computer.

A High Speed Permanent Storage Device—J. M. Weir. This paper describes a device useful for the permanent storage of digital information which ordinarily is not to be altered once it is stored. The device utilizes a large

magnetic-core matrix switch, of a type described by Rajchman, in conjunction with a storage system used with the Bell Computer, Model VI, to obtain permanent storage capacities up to about a million bits. The information is stored by suitability facing a set of drive leads from the output of the magnetic switch through an array of magnetic cores. This device is characterized by low-access time, large-operating tolerances, and a relatively small number of magnetic cores.

Control Features of a Magnetic-Drum Telephone Office—W. A. Malthaner and H. E. Vaughn. Several functional arrangements useful in conjunction with a parallel magnetic-drum memory are described with general reference to their application in an experimental telephone-switching system. The functions included are detection and registration of input information, counting, timing, transfer of information from one drum location to another, and translation of information from one form to another.

Stability of a Method of Smoothing in a Digital Control Computer—W. Karush. In a certain operation a digital computer was used as an element of a control system to smooth consecutive observational data. The method of smoothing consisted of predicting from past smoothed values and then combining the prediction with the next observation. In this paper an analysis of the stability of this useful method is made, and an explicit formula of the range of the parameters for which the method is stable is derived. Also, the statistical variance of the smoothed variable is calculated.

Review of Electronic Computer Progress During 1954—D. A. Brown.

RELIABILITY AND QUALITY CONTROL

PGQC-4, DECEMBER, 1954

Developments in Trustworthy-Value Techniques—E. G. Rowe and P. Welch. Quality-Control procedures in the manufacture of vacuum tubes, with particular emphasis on those designed for reliable or trustworthy service, are discussed. Histograms are used to distinguish between manufacturing variations and manufacturing errors to aid in their rapid detection and correction. Tests are made for short-circuits, disconnections, glass faults, premature heater and emission failures, noise, and short lives under high shock and vibration environment.

Reliability of Quantity Produced Transistors in Low Power Audio Applications—F. M. Dukat. Transistors have now been in quantity production for more than eighteen months and many hundreds of thousands of transistors have been put into daily use in hearing aids and other applications and are giving highly satisfactory and reliable service. A great deal of information has been accumulated relating to the performance of transistors at initial installation and during thousands of hours of operation.

Data will be presented summarizing this experience. The nature of defects that have been encountered will be discussed together with the relation of circuitry and other operating conditions to service performance and reliability.

Reliable Electronics Through Protective Coating Techniques—E. R. Gamson and H. Henesian. Stanford Research Institute has been engaged in a program involving studies and methods of protective coating materials. The techniques investigated, including thin coatings, cast-resin embedment, and foam-plastic encapsulation, were found to offer effective methods for improving the reliability of electronic systems. The protective medium surrounding such equipment could, in the ideal sense, eliminate failures due to vibration, moisture, or other environmental effects.

The results of these studies indicate the degree of protection now offered and the direction whereby the ideal may be obtained. Our research indicates that it is now possible to operate critical electronic devices in adverse environmental conditions for extended periods, by the application of the correct protective coating system.

Quality in Production—Dr. R. Weller. The quality of the product delivered by a manufacturer is largely determined by management policy. Where high quality is desired it is important to direct attention to ways and means of getting it.

The cost of an article from the user's point of view is a function of performance and reliability. The marginal value of improved performance is difficult to evaluate but the marginal value of increased reliability is readily stated.

Reliability can be discussed in terms of the reasons for failure, the identification and analysis of specific causes, the remedies necessary and especially the organizational steps necessary to ensure adequate flow of information and authority to secure a good product.

TELEMETRY AND REMOTE CONTROL

PGTRC-2, NOVEMBER, 1954

Delay Line Controlled Subcarrier Discriminator—K. A. Morgan and R. F. Blake. A new FM subcarrier discriminator utilizing a delay line as the frequency stable element has been developed. The design has exceptionally good linearity and stability without requiring parts selection. The output is essentially independent of tube characteristics since the sensitivity is primarily determined by a regulated reference voltage and the delay line. The linearity is theoretically perfect and in practice, linearities of plus or minus 0.1% have been obtained when used as a subcarrier telemetering discriminator. Input signal amplitude variations over a dynamic range of 100:1 have no noticeable effect on the output. This discriminator is particularly useful at high frequencies up to 1 mc (where other types of linear discriminators are difficult to construct) and as low as 3 kc. This low frequency limitation may be overcome by the design of miniature lumped constant or disturbed delay lines. Subminiaturization techniques are directly applicable to the design which requires only six tubes and a minimum number of electrical components.

Telemetering and Information Theory—Frank W. Lehan. Issue is taken with the reaction that seems to be in evidence among some radio engineers to the effect that information theory is an extremely complex and somewhat impractical subject fit only for mathematicians to amuse themselves with.

By way of illustration, a short engineering style discussion is given about the misbehavior of the audio frequency discriminator common in fm-fm telemetering at weak signal levels. An intuitive presentation is made as to why such misbehavior is not basic to the fm-fm system but is characteristic of the discriminator used. A different type of discriminator, suggested by information theory, is described and its performance outlined.

The discussion is extended to the fm transmitter, Receiver link and methods of improving the performance of this combination at poor signal to noise levels are suggested. It is speculated that some 10 db improvement may be possible here.

Finally, speculation concerning the possibilities of an integrated fm-fm transmission-reception system capable of operating at greatly reduced signal levels is indulged in. It is emphasized that such speculation is merely an indication of possible fruitful avenues of re-

search, not a proposal for an actual system.

A Temperature-Stable Transistor VCO—Fred M. Riddle. A transistor oscillator which is stable up to 150°F has been developed for obtaining electrical measurements. A reactance-modulation technique uses time rather than gain to control the frequency shift. Reactive current of constant amplitude is applied to an lc circuit for a controlled portion of each cycle. The portion of the cycle is controlled by a bias current injected into the modulating transistor. A differential converter circuit which functions also as a dc amplifier is used with the oscillator in making voltage measurements at high impedance level. Drift of the unit with variation in temperature, vibration, and supply voltage is comparable to that of vacuum-tube multi-vibrators and dc-amplifier combinations.

A Slope Modulator for FM Recording of Analog Data on Magnetic Tape—Louis W. Erath and Frank C. Smith, Jr. A new method of accomplishing frequency modulation enables this system to be used for the recording of analog data on magnetic tape with high signal-to-noise ratio and good linearity.

Conventional means of accomplishing frequency modulation, e.g. multivibrators, suffer from a number of limitations, especially in deviation ratio and linearity. Such circuits are much less elegant than the frequency-meter type of demodulator used in association with them.

The Slope Modulator described in this paper is a new circuit equal in refinement to the demodulator. Theoretically capable of 100 per cent deviation, it is operated in geophysical equipment at 75 per cent deviation for 100 per cent modulation, and in this application distortion is less than 1 per cent with noise down 80 db. Good carrier-frequency stability and freedom from tube characteristics are other features of the circuit.

The Slope Modulator shows promise for applications in telemetering, process control, vibration analysis and laboratory measurements where analog data must be recorded for reproduction. In its present form the system is capable of recording frequencies from zero to 500 cycles at a tape speed of 7½ inches per second. A maximum input signal to the modulator of 10 volts rms or 28 volts dc can be handled with less than 1 per cent distortion and an over-all signal-to-noise ratio (including tape drive mechanism) of approximately 56 db.

PROCEEDINGS OF THE WESCON COMPUTER SESSIONS

A Dependent Variable Analog Function Generator—C. J. Savant, Jr., and R. C. Howard. An all electronic, versatile, analog, arbitrary function generator which permits the rapid change of the functional form over a wide range is described. The basic components are explained with the aid of response oscillograms. The heart of the system is a linear-to-logarithmic converter which displays dependable performance when operating on the nonlinear portion of triode characteristics. After a slide-augmented description of the hardware, various nonlinear equations, including that of Duffing and Van de Pol, are mechanized and the solutions compared with analytical results.

Automatic Iteration on an Electronic Analog Computer—Louis B. Wadel. This paper discusses the employment of an electronic analog computer for automatic solutions of ordinary differential equations whose computer solution depends upon the application of an iterative process. Three types of equations falling in this category are noted, and a simple example of each is given. A computer solution procedure applicable to each example is outlined, with circuit diagrams included. Also described is the

use of a multipole stepping relay to effect the iteration procedures required. Illustrative results are presented.

A Logarithmic Voltage Quantizer—E. M. Glaser and H. Blasbalg. An analog to digital converter is described which converts a voltage into a chain of pulses whose number is proportional to the logarithm of the voltage. The device is automatic. It can handle input data at the rate of 10,000 voltage samples per second. Samples of amplitude between 3.3 and 100 volts and length greater than .3 microseconds can be quantized. Accuracy of conversion is adjustable to either 5 per cent or 10 per cent. A simple rc circuit performs the logarithmic conversion. The mathematical analysis of the conversion and quantization is given. Design equations are developed. A laboratory model of the quantizer is described and experimental results are shown.

A Digital Converter—J. B. Speller. A digital converter, which uses a unique disk pattern, has a shaft rotation input and unambiguous output in the natural binary code. The unit has 13 binary digits with an accuracy of one part in 2^{14} . By adding additional gear trains and disks it is possible to provide considerably higher count and accuracy. No transformations are required to obtain the natural binary code output. The input torque is about 0.2 in. oz. and is uniform throughout the range of the instrument. Sixty-four revolutions of the input shaft are needed for the complete count. The converter weighs less than 7 oz. and is similar in appearance to a small synchro. Brushes and commutator disks permit either dc or pulse voltage inputs and outputs.

Efficient Linkage of Graphical Data and Digital Computers—E. D. Lucas, Jr. This paper discusses the problem of transferring graphical data into digital form, and the converse problem of converting digital data into graphical form, for use with digital computers. Semi-automatic machines are described for analyzing and reading both oscillographs and film records, and associated machines for converting these readings into digital form. The latter includes conversion devices with digital output on punched tape, punched cards, or magnetic tapes. Several automatic plotting machines are described which present in graphical form the results of digitally computed data. These plotters will accept computer input in numerous digital forms, both serial and parallel, including punched tape and cards and electrical signals from contact closures. The plotters also accept input from analog devices.

Transistor Flip-Flops for High-Speed Digital Computers—Edmund U. Cohler. This talk sketches the design and operation of various point-contact transistor flip-flops for use in computers. The analysis of the various types are explained and the circuits discussed in terms of their limitations and capabilities. Schematics for these flip-flops are presented

and salient differences noted.

A system using various flip-flops and associated gates is described and evaluated in terms of extended use of transistors in computer circuitry.

Design Fundamentals of Photographic Data Storage—Gerhard L. Hollander. This paper is designed to give in one place sufficient background in the fundamentals of photography to permit engineers to evaluate film as a storage medium. The first section briefly presents the terminology and discusses available specifications which characterize films intended primarily for ordinary photography. With this background material the data needed to use film for data storage are described. Once these characteristics are determined, a design procedure can reply experimental development, and the memory transfer function of photographic media can be formulated. Finally, the separate problem of selecting the storage locations in photographic media is treated.

Pulse Response of Ferrite Memory Cores—James Robert Freeman. The response of magnetic-ferrite cores to current pulse in a two-to-one selection coincident-current magnetic-memory are classified as fourteen basic voltage outputs. These outputs are defined and described with relation to the hysteresis loops and pulse sequences involved. Photographs of the pulse responses are presented and certain distinctive differences compared. The concept of reversible and irreversible outputs is explained. Measurements of the various core voltage outputs for the General Ceramics body MF-1326B are given. Curves are included of the pulse voltage characteristics, and the switching and peaking times versus the driving current. An example is given of the use of pulse test data for evaluating the merit of a memory core. Disturb sensitivity is defined and its relationship to the driving pulse duration and overdriving is described. The effect driving pulse rise time also is considered.

Computer-Programmed Preventive Maintenance for Internal Memory Sections of the ERA 1103—S. R. Cray. Daily preventive maintenance routines for the electrostatic and magnetic drum storage systems of the ERA 1103 have greatly reduced the probability of storage failures during subsequent operation. Diagnostic test programs and marginal checking features are described which are used during preventive maintenance periods to insure that satisfactory operating margins prevail. Results obtained through the use of these procedures are presented. The general characteristics of the storage systems are discussed together with operational limitations.

An Input-Output System for a Digital Control Computer—L. P. Retzinger. A control system involving twelve input functions and fourteen output functions using a digital computer as the computation element is discussed. A method of converting shaft rotations to serial

binary information, time sharing the basic circuitry, is described. In going from serial digital control signals to output shaft positions a novel system is used to generate and store semi-proportional positioning signals which serve as inputs to magnetic amplifiers. Throughout the system, time sharing is utilized to a high degree, and independence of supply voltage, temperature, and other factors normally affecting analog systems, is stressed.

Characteristics of a Logistics Computer—Eugene Leonard. The ORDFIAC, a large-scale computer having punched-card input and output and a 10,000 word magnetic drum memory, was built to handle supply problems for the Ordnance Department of the Army. Its code includes an unusual and highly versatile vector instruction, which in addition to its original purpose of facilitating matrix manipulation has proved very useful for a wide variety of operations. The 100 channels of the memory are relay-selected, with relay operation thoroughly checked before each transfer of information into or out of the memory. The paper also will discuss design features and operating experience.

The Dico 20 Digital Differential Analyzer—Floyd Steele. Dico 20 is a twenty integrator magnetic drum differential analyzer. The four information channels are interplexed to form two recirculating channels. Trapezoidal integration is used. Integrator numbers are 20 significant digits long. Complete integrator communication exists and either the dx or dy integrator inputs may be multiple. Decoding is accomplished for multiple inputs by non-numerical addition. Variables are represented and processed as difference numbers. Decision is achieved by an ordinary integrator.

Dico 20 has 6 logical flip-flops and 4 memory flip-flops. There are 180 logical diodes. Conventional computer techniques are used in electronic design. The drum speed is 30 rps. and the clock rate is 50 kc.

The Bendix General Purpose Computer—D. C. Evans and H. D. Huskey. The Bendix Model G-15 is a general purpose, stored program computer having exceptional computing efficiency. The logical design incorporates innovations which permit a substantial reduction in the physical size and complexity.

The novel command structure permits effective use of sub-routines and minimum access coding. It has commands for addition, subtraction, multiplication, and division of both single and double length words. Commands for branching and other logical operations are incorporated. The memory has a capacity of more than 2000 words including a section with 0.6 millisecond average access time.

Input-output facilities may operate concurrently with computation and include electric typewriter, punched paper tape with photo-electric reader, magnetic tape, and adapter for punch cards.



Abstracts and References

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NOTE: The Institute of Radio Engineers does not have available copies of the publications mentioned in these pages, nor does it have reprints of the articles abstracted. Correspondence regarding these articles and requests for their procurement should be addressed to the individual publications, not to the I.R.E.

Acoustics and Audio Frequencies.....	516
Antennas and Transmission Lines.....	517
Automatic Computers.....	518
Circuits and Circuit Elements.....	518
General Physics.....	519
Geophysical and Extraterrestrial Phenomena.....	521
Location and Aids to Navigation.....	522
Materials and Subsidiary Techniques.....	522
Mathematics.....	526
Measurements and Test Gear.....	526
Other Applications of Radio and Electronics.....	527
Propagation of Waves.....	528
Reception.....	528
Stations and Communication Systems.....	528
Subsidiary Apparatus.....	529
Television and Phototelegraphy.....	529
Transmission.....	529
Tubes and Thermionics.....	529
Miscellaneous.....	530

The number in heavy type at the upper left of each Abstract is its Universal Decimal Classification number and is not to be confused with the Decimal Classification used by the United States National Bureau of Standards. The number in heavy type at the top right is the serial number of the Abstract. DC numbers marked with a dagger (†) must be regarded as provisional.

U.D.C. CHANGES

In anticipation of a new edition of the Universal Decimal Classification Abridged English Edition (BS 1000 A), certain changes in U.D.C. numbers will be made in this and subsequent issues. The new numbers used will be:

Radio astronomy: 523.16

Ultrasonics: 534 subdivisions with the special analytical subdivision -8 attached

Sound recording and reproducing: 534.85

Electroacoustic problems, transduction, etc.: 534.86

ACOUSTICS AND AUDIO FREQUENCIES

534.121.2:621.395.61 602
Theory of the Effect of a Thin Air Film on the Vibrations of a Stretched Circular Membrane—D. H. Robey. (*Jour. Acous. Soc. Amer.*, vol. 26, pp. 740-745; September, 1954.) Analysis generally applicable to conditions in capacitor microphones is presented. Application of the results to the case of a particular miniature microphone with titanium diaphragm shows that the effective stiffness at 20 kc is 15 times that at 5 cps.

534.132 603
Radial Vibrations of Thick-Walled Hollow Cylinders—J. A. McFadden. (*Jour. Acous. Soc. Amer.*, vol. 26, pp. 714-715; September, 1954.) Approximate formulas are given for the natural wavelengths.

534.213.4 604
The Propagation of Elastic Waves in Thin-Walled Cylindrical Shells—M. C. Junger and F. J. Rosato. (*Jour. Acoust. Soc. Amer.*, vol. 26, pp. 709-713; September, 1954.) Analysis indicates that there are two possible modes of propagation in the axial direction; at high frequencies the one corresponds to flexural waves and the other to longitudinal waves. The dis-

persion curves are in good agreement with published experimental results. The results are relevant to investigations of cylindrical BaTiO₃ transducers and to the problem of sound insulation by lightweight partitions.

534.213.4:534.6 605
Propagation of Sound in Narrow Tubes—H. W. Helberg. (*Akus. Beihefte*, no. 2, pp. 578-586; 1954.) Theory is developed yielding the formulas of Rayleigh and Kirchhoff as limiting cases. The accuracy of Kirchhoff's expression for the attenuation is improved by addition of a constant. Using a tube of cross section 1.03 × 19.4 mm with suspended particles as indicator, measurements were made of attenuation, sound velocity, and transverse distribution of particle velocity; the results give considerable support to the theory.

534.213.4:534.833.4 606
Propagation of Sound over Single Absorptive Strips in Ducts—J. E. Young. (*Jour. Acous. Soc. Amer.*, vol. 26, pp. 804-818; September, 1954.) A method is developed for predicting the absorptive effect of the strip, with accuracy sufficient for practical purposes, by a combination of approximations involving assumptions about the potential distribution at the surface of the strip. In the case of resonant absorbers with low dissipation, the attenuation can be calculated simply and accurately for strips of sufficient length, found experimentally to be about four times the duct diameter. Predictions can be made for porous strips in the same length range if the appropriate phase parameter is determined.

534.23 607
An Exact Method for determining the Directivity Index of a General Three-Dimensional Array—E. Rhian. (*Jour. Acous. Soc. Amer.*, vol. 26, pp. 704-706; September, 1954.)

534.26 608
Diffraction of an Acoustical Wave Obliquely Incident upon a Circular Disk—H. S. Heaps. (*Jour. Acous. Soc. Amer.*, vol. 26, pp. 707-708; September, 1954.) Calculated values of the sound pressure behind the disk with an obliquely incident wave are compared with those with a normally incident wave.

534.61:621.395.61 609
Ceramic Probe Microphones—E. Ackerman and W. Holak. (*Rev. Sci. Instr.*, vol. 25, pp. 857-861; September, 1954.) Report of a study of two types of microphone used for measuring intense sound fields at frequencies in the range 6-100 kc. The sensitive element is a BaTiO₃ cylinder 1/16 inch in length and in over-all diameter. Calibration procedures are described.

534.612.4 610
Pressure Calibration of Condenser Microphones above 10,000 c/s—B. D. Simmons and F. Biagi. (*Jour. Acous. Soc. Amer.*, vol. 26, pp.

693-695; September, 1954.) "A 'plane wave' acoustic coupler and an electrical admittance method are described for the pressure calibration of condenser microphones in the ultrasonic frequency range."

534.614-8 611
Rapid-Indication Ultrasonic Interferometer—L. Bergmann. (*Akus. Beihefte*, no. 2, pp. 591-593; 1954.) Measurements of the velocity of sound in gases and liquids are made quickly using a decade counter tube to count the number of maxima traversed as the interferometer reflector is shifted through a given distance.

534.62+621.317.3.029.63 612
Construction of a Reflection-Free Room for Sound Waves and Decimetre Electrical Waves—G. W. Epprecht, G. Kurtze and A. Lauber. (*Akus. Beihefte*, no. 2, pp. 567-577; 1954.) Description of a room constructed at Berne, having inner dimensions 5×4.4×2.6 m, lining depth 60 cm and acoustic cut-off frequency 120 cps. To make the lining absorbent for em waves, steel wool is used rather than the graphite used in the anechoic chamber at Göttingen [942 of 1954 (Meyer et al.)]. The floor is netting made of perlon cables of diameter 4 mm.

534.785 613
Some Factors affecting Multichannel Listening—J. P. Egan, E. C. Carterette and E. J. Thwing. (*Jour. Acous. Soc. Amer.*, vol. 26, pp. 774-782; September, 1954.) An experimental investigation is reported of the intelligibility of a wanted speech message in the presence of an unwanted speech message. Use of a high-pass filter in either of the two channels improved intelligibility. The advantages of dichotic presentation were demonstrated. Masking of speech by noise was also investigated.

534.832:534.121.1 614
Sound Radiation from a Wall excited to Flexural Vibrations—W. Westphal. (*Akus. Beihefte*, no. 2, pp. 603-610; 1954.) By using the "radiation coefficient," whose value can be found approximately from theory developed by Gösele (949 of 1954), an estimate can be made of the sound energy radiated from a plate from measurements of the amplitude of flexural vibrations. A determination of the lateral transmission in building acoustics can hence be made.

534.833.4 615
The Multiple-Panel Sound Absorber—E. C. H. Becker. (*Jour. Acous. Soc. Amer.*, vol. 26, pp. 798-803; September, 1954.) Equivalent-circuit analysis is presented for multiple-panel absorbers; increasing the number of elements increases the absorption bandwidth, particularly at low frequencies. An example is described of a particular construction giving satisfactory results.

534.84 616
Review of Architectural Acoustics during the Past Twenty-Five Years—V. O. Knudsen. (*Jour. Acous. Soc. Amer.*, vol. 26, pp. 646-650; September, 1954.)

534.84 617
Definition and Diffusion in Rooms—E. Meyer. (*Jour. Acous. Soc. Amer.*, vol. 26, pp. 630-636; September, 1954.) The importance of diffusivity as a criterion of acoustic quality is indicated in an account of architectural acoustics research at Göttingen. The product of definition and diffusivity may prove to be a useful figure of merit.

534.84 618
The Statistical Parameters of the Frequency Response Curves of Large Rooms—M. Schröder. (*Akus. Beihefte*, no. 2, pp. 594-600; 1954.) Frequency response curves are obtained using a variable-pure-tone generator and a microphone at different pairs of separated points in the room. Calculations are made of the rms response fluctuation, the mean height of the peaks, the mean spacing of the zero points (intersections of response curve with mean level), the mean spacing of the peaks, the mean phase rotation per cps, and the "frequency irregularity."

534.84 619
The Frequency Dependence of the Sound Pressure in Rooms—H. Kuttruff and R. Thiele. (*Akus. Beihefte*, no. 2, pp. 614-617; 1954.) Measurements made on 19 rooms under different conditions are reported, using stationary excitation. Analysis of the frequency response curves over the range 70-4000 cps indicates that the mean difference of level between successive maxima and minima amounts to 9-10 db, irrespective of room volume or reverberation time. The total number of maxima is proportional to mean reverberation time, with a mean fluctuation of 10 per cent. The "frequency irregularity" is thus also proportional to reverberation time.

534.84 620
Experiments for the Determination of Optimum Reverberation Time for Large Music Studios—W. Kuhl. (*Akus. Beihefte*, no. 2, pp. 618-634; 1954.) Estimates of optimum reverberation time were made on the basis of over 13,000 individual judgments on different records of three pieces of orchestral music recorded in a number of different rooms with volumes ranging from 2,000 to 14,000 m³. Reverberation-time/frequency curves derived from the records are presented, together with the frequency analysis of a typical loud chord for each record. Values of reverberation time judged to be optimum for the different types of music are quoted; these values do not depend on studio volume. For an occupied studio the best compromise is 1.7 second; for a small unoccupied studio the best value is considerably higher.

534.84 621
Electroacoustic Characteristics of the Palais des Festivals at Cannes—C. Soulé. (*Rev. Son.*, no. 18, pp. 251-252; September/October, 1954.) The measured reverberation-time/frequency characteristic is in general agreement with the theoretical curve for an auditorium of volume 2,000 m³; the theoretical curve corresponding to the actual volume of 10,000 m³ lies rather higher. This slight deadening was achieved deliberately by acoustic treatment. Curves of recorded level/frequency in stalls and circle are also given.

534.84:621.395.623.8 622
Sound Systems for Large Auditoriums—L. L. Beranek. (*Jour. Acoust. Soc. Amer.*, vol. 26, pp. 661-675; September, 1954.) This comprehensive review includes consideration of auditorium acoustics, loudspeaker types and arrangements, and psycho-acoustic factors.

Reference is made to the sound systems in the University City Hall, Caracas, Venezuela, in the United Nations Headquarters Hall, New York, in a municipal theater, and in the Holy Cross Cathedral.

534.845 623
Advances since 1929 in Methods of Testing Acoustical Performance of Acoustical Materials—F. G. Tytzer and H. A. Leedy. (*Jour. Acous. Soc. Amer.*, vol. 26, pp. 651-656; September, 1954.)

534.845 624
Possibilities of Error in Measurements of Sound Insulation at Low Frequencies: Part 1—W. Kuhl. (*Akus. Beihefte*, no. 2, pp. 611-614; 1954.) Measurements subsequent to those reported by Becker et al. (311 of 1953) indicate that the methods used may involve appreciable errors at low frequencies, due to (a) standing waves, (b) insufficient size of specimen, (c) natural resonances of the test rooms, and (d) incorrect determination of the absorption surface in the receiving space.

534.845 625
On Sound Absorption by Cylindrical Diffusers—G. Parolini. (*Jour. Acous. Soc. Amer.*, vol. 26, pp. 795-797; *Ricerca sci.*, vol. 24, pp. 1465-1470; July, 1954.) Measurements were made on the acoustic absorption of plywood cylindrical diffusers coated with porous materials, such as glass wool. High absorption was obtained even in the low frequency range, owing to the resonance of the cylindrical plywood frame, and to sound scattering. The experimental values have been found in good agreement with those computed according to Cook and Chirzanowsky's theory.

534.845 626
A Nomogram for Simplification of the Determination of Sound Absorption by the Reverberation-Chamber Method—W. Händler and G. Venzke. (*Akus. Beihefte*, no. 2, pp. 587-590; 1954.) A nomogram based on the Sabine formula is presented and the method of use explained.

534.845.1 627
Comparative Measurements of the Absorption of Sound by Absorptive Materials, using the Tube and Reverberation-Chamber Methods—G. Kurtze and A. Lauber. (*Tech. Mitt. Schweiz. Telegr.-Teleph. Verw.*, vol. 32, pp. 249-253; July 1, 1954. In German.) Report of measurements made on some commonly used porous materials, with the object of determining the range of validity of the cosine law of variation of absorption coefficient with angle of incidence, and hence determining the degree of reliance to be placed on results obtained by the simple tube method. Large discrepancies observed between the results by the two methods are probably due to velocity components parallel to the boundary surface and not taken into account in the calculations.

534.845.1 628
Resonance Reverberation Method for Sound Absorption Measurements—J. Karpovich. (*Jour. Acous. Soc. Amer.*, vol. 26, pp. 819-823; September, 1954.) The method described is suitable for measurements on liquids at frequencies from about 20 kc to over 600 kc. Sound absorption coefficients of 44 liquids are tabulated. Application of the method to measurements on solids is mentioned.

534.85 629
A Review of Twenty-Five Years of Sound Reproduction—H. F. Olson. (*Jour. Acous. Soc. Amer.*, vol. 26, pp. 637-643; September, 1954.)

534.85 630
Noise Level and Mechanical Stresses in Plastic Sound Records—E. A. Keller. (*Jour. Acous. Soc. Amer.*, vol. 26, pp. 685-687; September, 1954.) "Residual mechanical stresses in press-polished plastic film material used for

the embossing type sound recording are in many cases responsible for relatively high noise levels. Results of noise measurements of different plastic materials on a 400 grooves per inch recorder are presented. The relation between optically observed stresses and recorded noise level is given. Practical consequences are discussed."

534.85 631
Contribution to Analysis of Recording Process in H.F. Magnetic Recorders—O. Schmidbauer. (*Funk u. Ton*, vol. 8, pp. 341-360; July, 1954.) The hf bias recording method is analyzed and the influence of tape type, tape speed, width of recording-head gap and other parameters is discussed. A brief comparison with experimental results is made and reasons for discrepancies are noted. The possibility of indicating the quality of commercial tapes by a code is examined.

534.85:621.395.625 632
Fluctuations of Pitch of Recorded Sounds and Determination of Permissible Limits for Broadcasting—P. H. Werner. (*Tech. Mitt. Schweiz. Telegr.-Teleph. Verw.*, vol. 32, pp. 360-362; September 1, 1954. In French.) Methods of measurement of wow and flutter are outlined. Subjective tests have been made of the threshold of audibility for these two effects, using sounds produced by violin, piano and organ, recorded on magnetic tape. Frequency variations of 4 per cent were perceived by highly sensitive listeners.

621.395.61/.62 633
Loudspeakers and Microphones—L. L. Beranek. (*Jour. Acous. Soc. Amer.*, vol. 26, pp. 618-629; September, 1954.) Illustrated survey of developments from 1915 to date.

621.395.61 634
Unidirectional Microphone utilizing a Variable Distance between the Front and Back of the Diaphragm—A. M. Wiggins. (*Jour. Acous. Soc. Amer.*, vol. 26, pp. 687-692; September, 1954.) A gradient microphone is described in which the force acting on the diaphragm is made frequency independent by providing two frequency-selective sound inlets at the back of the diaphragm so that the distance between the front and back inlets is greater for the lower frequencies. Under these conditions the sound-responsive element can be constructed as for a pressure microphone.

ANTENNAS AND TRANSMISSION LINES

621.372.2 635
The Elliptic Surface Wave—A. E. Karbowiak. (*Brit. Jour. Appl. Phys.*, vol. 5, pp. 328-335; September, 1954.) Formulas are derived for E-mode propagation on the surface of an elliptic cylinder; in this mode the field decays monotonically as the radial distance in the transverse plane increases. The relation between this case and that of the surface-wave line with circular cross section is examined; the performance of the latter is shown to be substantially unaffected by slight deformation. Particular types of elliptical guide discussed include (a) the dielectric-coated conducting rod, (b) the rod with rectangular corrugations, (c) the homogeneous metal rod, and (d) the dielectric rod.

621.372.2.029.6 636
High-Frequency Phenomena—W. A. Tripp. (*Wireless Eng.*, vol. 32, pp. 19-25; January, 1955.) It is shown that the introduction of em field analysis is not essential for dealing with hf phenomena, which can be discussed satisfactorily in terms of current and voltage in the same way as lf phenomena; these concepts are applied to examination of the operation of transmission lines, including waveguides.

621.372.8 637
Propagation of Electromagnetic Waves in

Cylindrical Waveguides with Imperfectly Conducting Walls—V. M. Papadopoulos. (*Quart. Jour. Mech. Appl. Math.*, vol. 7, pp. 326-334; September, 1954.) Calculations are made by a perturbation method, using approximate boundary conditions. The imperfectly conducting walls have the effect of removing a particular type of degeneracy occurring in the ideal guide, in which definite linear combinations of E - and H -modes are propagated with definite propagation constants; this type of degeneracy occurs in rectangular guides. It follows that the power-loss method of calculating the attenuation constant, which assumes pure E - or H -mode propagation, is not correct in these cases. The perturbation method gives not only the attenuation constant for each combination of modes, but also the value of the phase velocity, and the corresponding field components.

621.372.8 638

An Approximate Method for the Calculation of Propagation Constants for Inhomogeneously Filled Waveguides—L. G. Chambers. (*Quart. Jour. Mech. Appl. Math.*, vol. 7, pp. 299-316; September, 1954.) A variational method is developed which is applicable to waves whose field components are given in terms of one scalar potential, for cases where the electrical constants vary across the waveguide but not along it. The conditions for the existence of pure TE or pure TM waves in such a guide are considered.

621.372.8 639

Two-Section Transmission-Line Transformer—M. S. Wheeler. (*Wireless Eng.*, vol. 32, pp. 15-18; January, 1955.) Analysis is presented for a waveguide-type impedance transformer convenient for coupling a magnetron to a standard waveguide and comprising two sections of different transverse dimensions. The cases of equal-length and nearly-equal-length sections are treated.

621.372.8:621.317.333.6 640

Electrical Breakdown in Waveguides at 3000 Mc/s—J. W. Sutherland. (*Electronic Eng.*, vol. 26, pp. 538-540; December, 1954.) Development of methods for testing high-power waveguides is described. The power-handling capacity can be improved by using other gases, notably "arcton 6" in place of air.

621.372.8:621.39 641

Waveguide as a Communication Medium—S. E. Miller. (*Bell Sys. Tech. Jour.*, vol. 33, pp. 1209-1265; November, 1954.) The circular-electric mode is particularly advantageous for long-distance communication, since the attenuation coefficient for this mode decreases as the carrier frequency increases; the theoretical value is 2 db per mile for round guide of diameter 2 inches, for frequencies around 50 kmc. Scale-model experiments using guide of internal diameter 4.73 inches and a frequency of 9 kmc are reported. Under favorable conditions the observed losses are within 25 per cent of the theoretical values; the discrepancy is attributed partly to guide roughness and partly to mode conversion at irregularities along the guide. Distortion and crosstalk due to undesired modes can be reduced by means of mode filters; constructions providing continuous filtering are described. A suitable structure comprises a helix or series of metal rings supported in a lossy housing. Problems associated with bends in the guide are examined. Base bandwidths of the order of 500 mc should be possible, using pcm. Regeneration would probably be required at the repeaters; the spacing between repeaters should be about 25 miles.

621.396.67 642

On Isotropic Antennas—H. F. Mathis. (*Proc. I.R.E.*, vol. 42, p. 1810; December, 1954.) Addendum to 37 of 1952.

621.396.67.012.12 643
Radiation from a Point Dipole located at the

Tip of a Prolate Spheroid—E. C. Hatcher, Jr., and A. Leitner. (*Jour. Appl. Phys.*, vol. 25, pp. 1250-1253; October, 1954.) Calculations are made of the radiation patterns for spheroidal conductors of various thicknesses, with major axes equal to λ/π , $2\lambda/\pi$ and $3\lambda/\pi$.

621.396.67.029.62 644

Internally Accessible Tubular Masts as Supports for U.S.W. and Television Aerials—W. Berndt. (*Funk u. Ton*, vol. 8, pp. 288-294 and 481-489; June and September, 1954.) Mechanical construction features of support masts for use with antennas for bands I-IV are described. Slot-antenna masts are also discussed and the radiation characteristics of various systems are presented graphically. The account deals mainly with modern German practice.

621.396.674.3 645

Aerials for U.S.W. Broadcasting and Television—W. Stöhr. (*Frequenz*, vol. 8, pp. 240-248; August, 1954.) The application of dipole units for directional and omnidirectional transmissions is illustrated in descriptions of three particular systems: (a) a vhf Yagi with relative bandwidth 1.15:1; (b) a band-III television broadcasting array consisting of stacked groups of four coupled full-wave dipoles, relative bandwidth 1.5:1; (c) a vhf broadcast array built up of units comprising two full-wave dipoles arranged with their two halves at right angles so as to form a square, fitting readily around or within a mast and secured to it at the voltage nodes. Antenna systems at Langenberg and Bogota and those of the Milan-Rome television link are described. Results of an investigation of the effects of ice formation on exposed antennas are noted.

621.396.676.029.62 646

Homing Aerials for Aircraft—S. Zisler and G. Dubost. (*Ann. Télécommun.*, vol. 9, pp. 226-236; September, 1954.) Vhf systems comprising identical parallel cylindrical antennas are discussed. General relations are derived for a quadriple arrangement. Calculations are made of optimum antenna length for homing, and of the ratio between the antenna currents; these calculations are based on defining limits for the form of the radiation pattern. Factors taken into account include antenna sensitivity, represented by the variation of the radiation with direction; precision of indication of the axis, which depends on the symmetry of the system; and the need to avoid spurious indications corresponding to equality of the diagrams for directions other than the true axis.

AUTOMATIC COMPUTERS

681.142 647

DYSEAC—the New N.B.S. Electronic Computer—(*Tech. News. Bull. Nat. Bur. Stand.*, vol. 38, pp. 134-141; September, 1954.) See 39 of 1954 (Elbourn and Witt) and 2320 of 1954 (Leiner and Alexander).

681.142 648

The Effect of Interpretive Techniques on Functional Design of Computers—T. Pearcey, G. W. Hill and R. D. Ryan. (*Aust. Jour. Phys.*, vol. 7, pp. 505-519; September, 1954.) Analysis of the programs for a number of computations performed by the C.S.I.R.O. Mark I computer indicates the feasibility of designing an adaptable and reliable computer having only a relatively small amount of rapid-access erasable store and a larger amount of rapid-access non-erasable store in which would be held all interpretation routines, function blocks, etc. The operator would require no knowledge of the actual machine code, but would place his hyper-programs and data into a slow-speed backing store.

681.142 649
Program Design for the C.S.I.R.O. Mark I Computer: Part 3—Adaptation of Routines for

Elaborate Arithmetical Operations—T. Pearcey and G. W. Hill. (*Aust. Jour. Phys.*, vol. 7, pp. 485-504; September, 1954.) Discussion of the extension of the library routine system to deal with floating point, multiple precision and complex arithmetic and certain combinations of these. The "interpretive" method of program organization [1655 of 1952 (Wilkes et al.)] is used. Part 2: 641 of 1954.

CIRCUITS AND CIRCUIT ELEMENTS

621.314.7: [621.37 + 621.396.621] 650

Some Transistor Circuits—A. J. W. M. van Overbeek. (*Tijdschr. ned. Radiogenoot.*, vol. 19, pp. 231-260; September, 1954.) The characteristics of junction transistors are discussed; at frequencies of 1-10 mc the equivalent circuit is already as complicated as that of a thermionic tube at frequencies a hundred times higher. Particular circuits examined include one for a medium-wave broadcast receiver in which the selectivity is automatically increased as the signal strength decreases. It is pointed out that for a peak output power of e.g. 250 mw the consumption is 150-200 mw. Nonlinear distortion in transistors is compared with the corresponding effect in thermionic tubes. The upper frequency limit set by cut-off of current amplification is considered in relation to trigger circuits. A circuit including a p - n - p and an n - p - n transistor is shown which has properties resembling those of a gas-filled tube with adjustable ignition voltage, short ignition time, very low discharge voltage drop and low discharge noise.

621.316.722.4:537.226 651

Dielectric Potentiometers—G. E. Pihl. (*Proc. I.R.E.*, vol. 42, pp. 1758-1761; December, 1954.) A voltage divider is described comprising a movable electrode and a system of fixed electrodes all immersed in a lossy liquid dielectric, so that the paths between the electrodes are both resistive and capacitive. The arrangement is suitable for wide-band operation (e.g. 20 cps-1 mc), since the product of the equivalent parallel resistance and capacitance is a constant depending only on the nature of the dielectric. Various practical embodiments are described. Desirable characteristics for the dielectric are indicated.

621.316.86:621.396.822 652

Current Noise in Carbon-Film Resistors—K. E. Doering. (*Funk u. Ton*, vol. 8, pp. 378-385 and 422-429; July and August, 1954.) An experimental investigation at audio frequencies is reported. The results are consistent with an empirical formula according to which the current-noise voltage varies approximately directly with current. In a few specimens the noise was found to depend on the direction of current. 45 references.

621.316.89 653

The Resistivity of "Composition" Resistors at Radio Frequencies—U. Tiberio. (*Proc. I.R.E.*, vol. 42, pp. 1812-1813; December, 1954.) Continuation of the previous discussion (654 of 1954) of the mechanism causing the drop of resistance of composition resistors at hf. Experimentally obtained resistance/frequency curves are presented for resistors comprising (a) a cylindrical column of an aqueous solution of sodium chloride, (b) a cylindrical column of an aqueous solution of copper sulphate, and (c) a cylindrical column of carbon/resin composition. The resistance of type (a) is practically constant at frequencies up to 100 mc, while that of types (b) and (c) drops considerably. The drop in the resistance of the composition resistor is ascribed partly to the "resistivity factor," and partly to the "external capacitance factor." It is suggested that a frequency-independent resistor could be obtained by mixing carbon with a material of high dielectric constant.

621.318.4 654
Component Design Trends—High-Fre-

quency Coils use New Core Materials—F. Rockett. (*Electronics*, vol. 27, pp. 140-143; December, 1954.) Points discussed include use of ferrite cores for coils operating at frequencies up to about 100 mc, use of glass and other low-loss materials for formers and use of toroidal constructions.

621.319.4 **655**
The Capacity and Field of a Split Cylindrical Condenser, using the Method of Inversion—H. J. Peake and N. Davy. (*Brit. Jour. Appl. Phys.*, vol. 5, pp. 316-321; September, 1954.) "The complex potential of a split cylindrical condenser is obtained by inversion of a known, simpler case. Expressions are obtained for the value of the electrostatic field at points on the axes of symmetry, the surface density of charge on a conductor and the capacity of the condenser. The expressions obtained by Adams, using another method, are deduced as one of three special cases for which tables and graphs are provided. The results should prove of value in the design of electrode systems for various purposes."

621.319.42 **656**
Miniature Lacquer-Film Capacitors—D. A. McLean and H. G. Wehe. (*Proc. I.R.E.*, vol. 42, pp. 1799-1805; December, 1954.) A manufacturing process is described in which a thin film is cast on a supporting base and is metallized and slit while still supported, after which it is stripped and wound into capacitor units. Metallized films 0.1 mil thick have been produced; the resulting capacitors are about a seventh the size of the smallest metallized-paper types. The support may be left in if the film is extremely fragile or if the capacitor is to operate at voltages below about 15 v. A formula is derived for the effective series resistance.

621.372.412 **657**
Thickness-Shear and Flexural Vibrations of a Circular Disk—R. D. Mindlin and H. Deresiewicz. (*Jour. Appl. Phys.*, vol. 25, pp. 1329-1332; October, 1954.) Antisymmetrical modes of vibration in an AT-cut quartz disk are investigated by considering the simpler corresponding case of an isotropic disk. Differences between the frequency spectrum in this case and in that of the rectangular plate [1861 of 1951 (Mindlin)] are due to the presence of the thickness-twist modes in addition to the thickness-shear and flexural modes.

621.372.413 **658**
Theory of Coupled Endovibrators [cavity resonators]—A. I. Akhiezer and G. Ya. Lyubarski. (*Zh. Tekh. Fiz.*, vol. 24, pp. 1697-1706; September, 1954.) A system of two cavity resonators coupled by means of a narrow slot in their common wall is considered theoretically. Two classes of oscillations are considered: (a) those whose frequencies are determined primarily by the length of the slot and are nearly independent of the shape of the resonators; (b) those whose frequencies are near the frequencies of the oscillations in the resonators when not coupled.

621.372.5 **659**
The Wien Bridge as a Phase Shifter—J. M. Diamond. (*Proc. I.R.E.*, vol. 42, pp. 1807-1808; December, 1954.) Several circuits are presented illustrating the use of the Wien bridge to provide phase shift with small amplitude change.

621.372.5 **660**
Networks Attenuation and Input Impedance—R. Talks. (*Wireless Eng.*, vol. 32, pp. 29-30; January, 1955.) Useful formulas are presented.

621.372.54 **661**
Four-Terminal Networks with Transfer Function having Zeros with Small Real Part—W. Krägeloh. (*Frequenz*, vol. 8, pp. 249-256; August, 1954.) Analysis of low-pass filter net-

works based on insertion-loss principles [2940 of 1940 and 3226 of 1942 (Bader)] to investigate the effect of "critical" zeros in the transfer-function plot on the reactance to be developed, the accuracy required in the calculation, and the component values.

621.372.622 **662**
Some Aspects of Mixer-Crystal Performance—P. D. Strum. (*Proc. I.R.E.*, vol. 42, pp. 1806-1807; December, 1954.) Correction to paper abstracted in 2941 of 1953. Please note change of U.D.C. number.

621.373.421.11.016.35 **663**
Criteria for the Amplitude Stability of a Power Oscillator—W. R. MacLean. (*Proc. I.R.E.*, vol. 42, pp. 1784-1791; December, 1954.) Stability criteria for a tuned-anode oscillator are established in the form of two inequalities derived from differential equations expressing the voltage variations of the grid and anode, and involving the ratio of feedback power to anode power, the ratio of the time constant of the grid-leak and capacitor combination to the time constant of the tank circuit, and functions of the angles of current flow for grid and anode. Experimental verification of the results using a Type-3C24 triode is reported.

621.373.43:517.93 **664**
Investigation of the Dependence of Natural Frequency of Oscillation on Spectral Composition—I. I. Minakova. (*Zh. Tekh. Fiz.*, vol. 24, pp. 1677-1686; September, 1954.) A particular case of the theory of nonlinear electric oscillations is considered. Writing the solution of the equation $\ddot{x} + \psi(x)\dot{x} + \omega_0^2 x = 0$ as the sum of $A_k \sin(k\omega t + \phi_k)$, where A_k is the amplitude of the k th harmonic given by the Fourier series, the ratio $\omega^2/\omega_0^2 = \sum_1^n A_k^2 / \sum_1^n k^2 A_k^2$. Using this relation, the dependence of ω on the amplitudes of the spectrum can be calculated. The measured and calculated frequency characteristics of relaxation oscillators agreed well. The results are presented graphically. See also van der Pol, *Proc. I.R.E.*, vol. 22, pp. 1051-1086; September, 1934, for a general introduction.

621.374.4:621.314.63 **665**
Crystal Frequency Multipliers for Centimetre and Millimetre Waves—L. Grifone. (*Ricerca sci.*, vol. 24, pp. 1870-1879; September, 1954.) Circuits are described for generating harmonics at frequencies >30 kmc, using Type-1N23B crystals. The generator is coupled by a cross-bar transition to the crystal, which can be shifted axially for purposes of impedance matching. Higher frequencies can be achieved by use of Type-1N26 crystals in a similar arrangement.

621.375.2.018.75 **666**
On the Faithful Reproduction of the Flat Top of a Pulse in a High Fidelity Pulse Amplifier: Part 2—B. K. Bhattacharyya. (*Indian Jour. Phys.*, vol. 27, pp. 565-577; November, 1953.) An experimental verification is reported of analysis presented previously (3566 of 1953). It was demonstrated that the anode current may sag appreciably if the time constants of the cathode and screen-grid RC circuits are not properly chosen.

621.375.223+621.373.421 **667**
Resonance Circuits comprising RC or RL Elements, and some Applications—H. Müller. (*Funk u. Ton*, vol. 8, pp. 471-479; September, 1954.) The frequency characteristics of the Wien bridge and of an analogous inductance bridge are investigated theoretically. Pseudo-resonance phenomena occurring in the neighborhood of bridge balance are discussed. True resonance is obtained when the bridge is associated with a feedback circuit to act as an oscillator. The tuned RC amplifier and the RC and RL oscillators are described.

621.375.23 **668**
Multistage Amplifier Output Impedance—

J. B. Earnshaw. (*Electronic Eng.*, vol. 26, p. 553; December, 1954.) Based on the result that the ratio of the net amplification to the parallel output impedance of a simple amplifier is the same with and without voltage feedback, a simple expression is obtained relating the parallel output impedance to the amplifier constants for the multistage amplifier with over-all feedback. The design of an amplifier with a gain of 1,000 and a parallel output impedance of 1Ω is discussed briefly.

621.375.3 **669**
Magnetic Amplifiers—G. M. Ettinger. [*Elec. Rev. (London)*, vol. 155, pp. 348-352; September 3, 1954.] A general survey of types and applications.

621.375.3 **670**
Three-Phase High-Speed Magnetic Amplifiers—A. E. Maime. (*Electronic Eng.*, vol. 26, pp. 514-521; December, 1954.) The principle of the "half-wave" magnetic amplifier is extended to three-phase circuits. Various arrangements are described. Applications in the field of high-power control systems are discussed.

67. 375.4.026 **671**
The Transistor as a D.C. Amplifier for use in Microwave Measurements—C. F. Davidson. (*Electronic Eng.*, vol. 26, pp. 548-549; December, 1954.) A junction-type transistor with emitter earthed may have a current amplification as high as 50, with low input impedance and high output impedance. Such an arrangement is useful for amplifying the current from a Si rectifier before application to a meter, and enables the usual galvanometer to be replaced by a robust microammeter.

621.375.5 **672**
Analyses of Basic Dielectric Amplifier Circuits—Shou-lisen Chow. (*Jour. Appl. Phys.*, vol. 25, pp. 1297-1301; October, 1954.) Analysis is based on a simplified charge/voltage characteristic neglecting hysteresis; both parallel- and series-connected arrangements are studied. The steady-state response can be found with a high degree of accuracy by a method involving successive approximations; transient response is also discussed.

GENERAL PHYSICS

535.215 **673**
Photoelectric Emission in the Extreme Ultraviolet—H. E. Hiltner. (*Phys. Rev.*, vol. 96, pp. 538-539; October 15, 1954.) Results of experimental studies on photoelectric emission from various metals for quantum energies up to 21.2 ev cannot even qualitatively be accounted for by the common "free-electron" "surface effect" representation. A new theoretical model capable of explaining the observations at high photon energies is presented.

537.226:537.52 **674**
The Influence of the Cathode Material on Measured Breakdown Strengths of Solid and Liquid Dielectrics—J. J. O'Dwyer. (*Aust. Jour. Phys.*, vol. 7, pp. 400-409; September, 1954.)

537.311.31 **675**
Kinetic Equation for Electrons in Metals in Strong Fields—V. P. Shabanski. (*Zh. Eksp. Teor. Fiz.*, vol. 27, pp. 142-146; August, 1954.) See also 676 below.

537.311.31 **676**
On Deviations from Ohm's Law in Metals—V. P. Shabanski. (*Zh. Eksp. Teor. Fiz.*, vol. 27, pp. 147-155; August, 1954.) It is shown, on the basis of the kinetic equations of electrons in metals in strong fields (675 above), that the deviations observed are due primarily to the delayed transmission of energy at the collision of electrons with the lattice. At sufficiently low temperatures, the resistance should pass through a minimum at a given current.

- 537.52 677
On the Dependence of the Decay Times of Space Charges by the Static Characteristic in Intermittent Discharges—D. Brini, O. Rimondi and P. Veronesi. (*Nuovo Cim.*, vol. 12, pp. 413-424; September 1, 1954. In English.) The validity of the hypothesis previously formulated [2068 of 1954 (Brini and Veronesi)] has been investigated experimentally. Measured decay times depend on the external circuit associated with the discharge tube, but the existence of an inherent decay time dependent only on the static characteristic of the tube is indicated. An empirical method is developed for calculating decay times.
- 537.52 678
Space Charge Formation and the Townsend Mechanism of Spark Breakdown in Gases—R. W. Crowe, J. K. Bragg and V. G. Thomas. (*Phys. Rev.*, vol. 96, pp. 10-14; October 1, 1954.)
- 537.523.4 679
Measurement of the Current during the Formative Time Lag of Sparks in Uniform Fields in Air—H. W. Bandel. (*Phys. Rev.*, vol. 95, pp. 1117-1125; September 1, 1954.) Measurements were made of the current in a parallel plane gap during the period between application of a voltage and occurrence of breakdown. The current increased from 10^{-6} a few microseconds after application of the voltage to 10^{-2} a just before breakdown, for time lags between 10 and 100 μ s. The results are in agreement with previously developed theory, except for an observed delay in the initial current rise; a possible explanation of this is discussed.
- 537.525.72:537.562 680
Determination of the Electronic Temperature in U.H.F. Gas Discharges—M. Bayet and F. Guérineau. [*Compt. Rend. Acad. Sci. (Paris)*, vol. 239, pp. 1029-1031; October 27, 1954.] Measurements made in an electrodeless discharge over a pressure range 0.13-11 mm Hg and an electron-concentration range of ratio 1 to 80 indicate that the electronic temperature remains practically constant, its value being 30,000 degrees K to within 10 per cent in dry air.
- 537.533:537.534.8 681
Auger Ejection of Electrons from Tungsten by Noble Gas Ions—H. D. Hagstrum. (*Phys. Rev.*, vol. 96, pp. 325-335; October 15, 1954.) Report of an experimental investigation using atomically clean tungsten and ions with various charges. The results indicate a value of about 6.3 ev for the energy of the Fermi level above the ground state in the conduction band in tungsten.
- 537.533:537.534.8 682
Theory of Auger Ejection of Electrons from Metals by Ions—H. D. Hagstrum. (*Phys. Rev.*, vol. 96, pp. 336-365; October 15, 1954.)
- 537.533.73/.74 683
Diffraction and Inelastic Scattering of Electrons [in Metals]—A. Ya. Vyatskin. (*Zh. Eksp. Teor. Fiz.*, vol. 27, pp. 162-170; August, 1954.)
- 537.533.8:546.45 684
Secondary Electron Emission from Thin Layers of Be: Part 1—I. M. Bronshtein and T. A. Smorodina. (*Zh. Eksp. Teor. Fiz.*, vol. 27, pp. 215-223; August 1954.) An experimental investigation is reported of the change of the secondary emission coefficient (σ) and the electron energy distribution with the adsorption of Be atoms on Ni. σ decreases monotonically with adsorption of pure Be, but increases at first with impure Be, decreasing finally to σ_{Be} . The emission depth of secondary electrons depends linearly on the energy of the primary electrons in the range 100-600 ev.
- 537.533.8:546.561 685
Investigation of Energy-Distribution Function of Secondary Electrons from Cu Single Crystal covered with a Single-Crystal Cu_2O Layer using the Method of Electrical Differentiation—N. B. Gornyi. (*Zh. Eksp. Teor. Fiz.*, vol. 27, pp. 171-179; August, 1954.) The subsidiary maxima observed previously (2918 of 1954) are confirmed and discussed. See also 3527 of 1954 (Gornyi and Rakhovich).
- 537.562:538.561 686
Dielectric Constant of Plasma in Stationary Magnetic Field—M. E. Gertsenshtein. (*Zh. Eksp. Teor. Fiz.*, vol. 27, pp. 180-188; August, 1954.) The tensor of the complex dielectric constant is calculated taking into account the thermal motion of the electrons. It is shown that in a system of coordinates connected with the moving electron, the field of a monochromatic plane electric wave is frequency modulated, hence resonance effects occur at higher harmonics. Gaps in the plasma oscillation spectrum at multiples of the gyro-magnetic frequency [2151 of 1951 (Gross)] can only occur if two conditions are satisfied; the first is the condition for resonance effects to occur with all electrons the second gives the condition for high intensity of higher harmonics. With radio waves the effect of higher harmonics is negligible and hence there are no gaps in the radio wave spectrum. The conditions are satisfied in the case of sound waves in ionized gas and the effects due to resonance at multiples of the gyro-magnetic frequency can be considerable.
- 537.581:546.56 687
Thermionic Emission from Copper at the Melting Point—V. G. Bol'shov and L. I. Dobretsov. [*Compt. Rend. Acad. Sci. (URSS)*, vol. 98, pp. 193-196; September 11, 1954. In Russian.] An experimental determination is reported of the constants A and ψ in the Richardson-Dushman equation $j = A T^2 \exp(-\psi/kT)$ where ϵ is the electron charge and ψ is the effective or isothermal work function. For solid Cu at the melting point the values of A and ψ were 16.7 a. cm $^{-2}$. deg $^{-2}$ and 4.4 v, respectively, for liquid Cu 3.2×10^6 a. cm $^{-2}$. deg $^{-2}$ and 5.5 v respectively. The discontinuity at the melting point is discussed.
- 538.1/.2 688
Recent Developments in Magnetism—E. P. Wohlfarth. [*Research (London)*, vol. 7, pp. 360-367; September, 1954.] A survey with reference to 16 publications dealing with various aspects of the subject.
- 538.122 689
Magnetic Flux produced by a Dipole Located inside a Ferromagnetic Circular Wire—S. M. Rytov. (*Zh. Eksp. Teor. Fiz.*, vol. 27, pp. 307-312; September, 1954.) An approximate formula is derived for the magnetic flux through a coaxial plane circular area normal to the wire, due to a magnetic dipole element inside the wire. The radii of the circular area and the wire are assumed to be small compared with the axial distance between the circular area and the dipole element. For experimental confirmation of the formula see Grachev et al. (690 below).
- 538.122 690
Experimental Investigation of Change of Magnetic Flux in a Wire when One Domain is Remagnetized—A. A. Grachev, K. A. Goronina, N. N. Kolachevskii and I. A. Andrianova. (*Zh. Eksp. Teor. Fiz.*, vol. 27, pp. 313-317; September, 1954.) The problem, which is of importance in magnetic recording, was investigated theoretically by Rytov (689 above); the calculated and experimental results are in good agreement.
- 538.221:538.566 691
Theory of Strong Electromagnetic Waves in Massive Iron—W. MacLean. (*Jour. Appl. Phys.*, vol. 25, pp. 1267-1270; October, 1954.) Maxwell's equations are solved for propagation in material having a rectangular hysteresis curve. Formulas are derived for depth of penetration, wave impedance and power input. The results are compared with those of Rosenberg (*Electrician*, vol. 91, p. 188; August, 1923). See also 2006 of 1954 (Papoulis).
- 538.566 692
Reflection from a Wire Grid Parallel to a Conducting Plane—J. R. Wait. (*Canad. Jour. Phys.*, vol. 32, pp. 571-579; September, 1954.) The parallel-wire grid backed by a conducting plane can be represented by an impedance shunted across a transmission line as in the case of a single wire [2589 of 1948 (Macfarlane)]. The value of this impedance depends on the angle of incidence, the spacing of the grid wires, and the distance between grid and backing plane. Conditions are derived for the reflection coefficient to become zero.
- 538.566:535.42 693
Diffraction of Electromagnetic Waves by an Aperture in a Plane Screen—R. D. Kodis. (*Jour. Appl. Phys.*, vol. 25, pp. 1342-1343; October, 1954.) Discussion of 709 of 1954 (Bekefi) and 2078 of 1954 (Crysdale).
- 538.566:535.42 694
The Diffraction of Waves by an Irregular Refracting Medium—E. N. Bramley. (*Proc. Roy. Soc. A*, vol. 225, pp. 515-518; September 22, 1954.) "A method is described of calculating the diffraction effects produced by a thick stratum of an irregular refracting medium. It consists of evaluating the statistics of the phase irregularities in the wave-front after traversing the medium, and treating these irregularities as having been produced by a thin phase-changing screen. For a particular statistical model of the irregularities in the medium, the result is shown to be identical with that obtained by Fejer [1730 of 1954] using a different method."
- 538.566:535.42:517.942.9 695
A Further Note on Dual Integral Equations and an Application to the Diffraction of Electromagnetic Waves—C. J. Tranter. (*Quart. Jour. Mech. Appl. Math.*, vol. 7, pp. 314-325; September, 1954.) A solution obtained previously (1614 of 1951) is extended to cover cases in which the order of the Bessel-function kernel is not zero. As an example, the solution is applied to a problem in the diffraction of em waves by a plane slit; Groschwitz and Hönl's discussion of the problem (2183 of 1952) is criticized. Results obtained are in agreement with those of Müller and Westpfahl (1971 of 1953).
- 538.632:537.525 696
Hall Effect in Positive Column—K. Takayama, T. Suzuki and T. Yabumoto. (*Phys. Rev.*, vol. 96, pp. 531-532; October 15, 1954.) Report of measurements of Hall voltage in the positive column of dc gas discharge tubes as a function of magnetic field, tube current, distance between probes and gas pressure.
- 538.632:538.221 697
Theory of Hall Effect in Ferromagnetics—N. S. Akulov and A. V. Cherenushkina. [*Compt. Rend. Acad. Sci. (URSS)*, vol. 98, pp. 35-38; September 1, 1954. In Russian.] Assuming s - d -phonon interaction, the Hall voltage is given by the equation $e = [a_0\gamma_0 + a_2(c_F - \rho_0)] I_s i$, where a_0 and a_2 are constants, ρ_0 and c_F are the resistances of the specimen at absolute zero and at temperature T , respectively, I_s is the intensity of magnetization at saturation and i is the current density. Comparison with experimental results obtained from measurements on a bar-shaped 45 per cent-Ni/55 per cent-Fe specimen (containing stated impurities) shows close agreement.
- 538.632:538.221 698
Hall Effect in Ferromagnetics—R. Karplus and J. M. Luttinger. (*Phys. Rev.*, vol. 95, pp. 1154-1160; September 1, 1954.) Anomalous effects are explained in terms of the spin-orbit interaction of polarized conduction electrons.

- 548.0:53 699**
Wave Functions for Impurity Levels—G. F. Koster and J. C. Slater. (*Phys. Rev.*, vol. 95, pp. 1167–1176; September 1, 1954.) A general method of solving difference equations arising in impurity calculations is developed.

GEOPHYSICAL AND EXTRATERRESTRIAL PHENOMENA

- 523.16 + 621.396.11 + 621.396.9 700**
Propagation of Electromagnetic Waves, Radio Location and Radio Astronomy—E. Roessler. (*Elektrotech. Z., Edn A*, vol. 75, pp. 632–635; September 11, 1954.) A brief survey of recent progress, particularly since 1951. 73 references.

- 523.16 701**
Detection of Discrete Radio Sources at 21 cm Wavelength—J. P. Hagen, E. F. McClain and N. Hepburn. (*Proc. I.R.E.*, vol. 42, p. 1811; December, 1954.) Details are tabulated of 20 sources observed at the U.S. Naval Research Laboratory. The equipment used is briefly described.

- 523.16 702**
Nature of Discrete Sources of Cosmic R.F. Radiation—I. S. Shklovski. [*Compt. Rend. Acad. Sci. (URSS)*, vol. 98, pp. 353–356; September 21, 1954. In Russian.] Energy considerations of the colliding gas "coronas" in Cygnus A show that a considerable proportion of the kinetic energy of the colliding masses is transferred to a relatively small number of relativistic particles and hence into rf radiation. The cut-off frequency is estimated to be about 2×10^{10} cps assuming the particles to be electrons.

- 523.16 703**
Observations of Cosmic Noise at 9.15 mc—C. S. Higgins and C. A. Shain. (*Aust. Jour. Phys.*, pp. 460–470; September, 1954.) "From observations made at a frequency of 9.15 mc, with an aerial of beam width 29 degrees between half-power points and directed to Dec. –32 degrees, a curve of equivalent aerial temperature, as a function of sidereal time, is derived. The temperatures observed were of the order of 10^6 degrees K. The curve is compared with curves derived for similar conditions by calculation from the results of observations at 18.3 mc and at 100 mc. It is found that the equivalent temperatures increase rapidly with decreasing frequency, but the ratio of maximum to minimum temperature decreases with frequency. It is shown that 'atmospheric' noise levels observed by the standard techniques sometimes contain a large contribution from cosmic noise at this frequency."

- 523.72:621.396.822 704**
Harmonics in the Spectra of Solar Radio Disturbances—J. P. Wild, J. D. Murray and W. C. Rowe. (*Aust. Jour. Phys.*, vol. 7, pp. 439–459; September, 1954.) Detailed account of observations reported previously (391 of 1954). Investigations over the frequency range 40–240 mc indicate that spectral features of solar noise bursts are commonly duplicated at or below the frequency of the second harmonic. The results are consistent with the hypothesis that the fundamental frequency corresponds to the natural plasma frequency of the corona in the vicinity of the source. By applying this result to a standard model of the corona, information is deduced regarding the position, velocity and size of the sources. Velocities of 500 and 4,000 km were found for two long-duration outbursts, and velocities as great as 10^6 km for short-lived type-III bursts. The generation of bursts may be associated with longitudinal plasma oscillations excited by fast streams of charged particles.

- 523.746:621.396.822 705**
The Emission Polar Diagram of the Radio-

Frequency Radiation from Sunspots—K. E. Machin and P. A. O'Brien. (*Phil. Mag.*, vol. 45, pp. 973–979; September, 1954.) The variation in received sunspot radiation with solar rotation was determined from a statistical analysis of observations made over a number of years. Half-power widths of the radiation pattern of an average sunspot, derived from this analysis, are 15 degrees, 20 degrees and 36 degrees for frequencies of 81.5, 175 and 500 mc respectively. Results indicate that the lifetime of the radiation sources is shorter for the lower frequencies; the lifetime at 175 mc is comparable with that of a visible spot.

- 551.510.535 706**
Abnormal Amplitude of Seasonal Effects in the Ionosphere at the Equator, and Structure of the Upper Atmosphere—F. Delobbeau and R. Gallet. [*Compt. Rend. Acad. Sci. (Paris)*, vol. 239, pp. 1067–1069; October 27, 1954.] D-layer absorption, maximum ionization and structure of the F_2 layer have been studied. Seasonal variations observed in equatorial regions are much greater than expected from the geometrical variation of the sun's position. Known theories assume simple variation of ionosphere parameters with $\cos \chi$. The deviations observed are interpreted as indicating seasonal variations of the structure of the upper atmosphere, particularly as regards temperature and its gradient, molecular dissociation and movements of air masses. At an equatorial station at the solstices, not only is the solar radiation incident at an angle of 23 degrees, but the structure of the upper atmosphere is that appropriate to a latitude of about 23 degrees.

- 551.510.535 707**
Motion of a Single Cloud in the Ionosphere—S. N. Mitra. (*Indian Jour. Phys.*, vol. 27, pp. 562–564; November, 1953.) The system comprising cloud and ionosphere is treated as analogous to the system formed by a horizontal wire antenna and the ground. The radiation pattern consists of minor lobes arranged symmetrically with respect to a central major lobe; as this pattern moves with the cloud, periodic fading of transmitted pulses is observed at the receiver. Periodic fading preceded and followed by a steady signal can thus, with appropriate reservations, be interpreted as due to a single cloud. An example is discussed in which 14 maxima were observed; the wavelength being 75 m, the horizontal length of the cloud is estimated to be about 1 km.

- 551.510.535 708**
A Monochromatically Ionized Layer in a Non-Uniformly Recombinant Atmosphere; with Applications to the D and E Ionospheric Regions—S. Chapman. (*Proc. Phys. Soc.*, vol. 67, pp. 717–727; September 1, 1954.) The recombination α of the atmosphere is assumed to be given by a term α_0 independent of height together with a term $\alpha_1 e^{-h/b}$ which varies exponentially with height. A recombination datum level is defined as that at which the two terms are equal, and heights are measured from this level, in scale-height units. "The level of the absorption peak being z_χ (when the sun's zenith distance is χ , or z_0 when $\chi = 0$), the level z_m of the electron peak and the height distribution of the electron density n_e are considered, for different values of z_0 and of $c = b/H$, particularly $c = 1, 2, 3$. When $c > 0$ the electron peak is always above the absorption peak, and for $c \geq 2$ it is always above the recombination datum level: the electron peak for $c = 1$, when the absorption peak is below the recombination datum level, is about half way between the two. The decrease of n_e (from its maximum value n_{em}) on the underside (or incline) of the electron layer can be much less steep than for a Chapman layer, if the absorption peak is below the recombination datum level. The results for the model atmospheres considered are tentatively discussed with reference to the

E and D ionospheric regions, but their potential value may be realized only when better data for the D region become available."

- 551.510.535:538.566 709**
Focusing Phenomena due to Undulations of the Ionosphere, and Determination of Collision Number—Rawer and Argence. (See 842.)

- 551.510.535:551.594.5 710**
Electron Density in the E-Layer during Auroral Displays deduced from Measurements of Absolute Brightness of the Auroral Luminosity—A. Omholt. (*Jour. Atmos. Terr. Phys.*, vol. 5, pp. 243–244; September, 1954.) Values of electron density calculated from the photon emission for different auroral forms range from 1.6×10^9 to 12×10^6 electrons/cm³. See also 716 below (Seaton).

- 551.510.535:551.594.5 711**
The Association of Pulsating and Flaming Auroras with Complete Ionospheric Absorption at Macquarie Island—G. Major. (*Aust. Jour. Phys.*, vol. 7, pp. 471–476; September, 1954.) Simultaneous records show that pulsating or flaming auroras are frequently accompanied by complete absorption of waves incident vertically on the ionosphere, but the nocturnal variations of frequency of occurrence of the two phenomena are markedly different in form.

- 551.510.535:621.3.087.4 712**
Equipment for Accurate Measurement of Height of Ionosphere Layers—S. J. Bauer. (*Öst. Z. Telegr. Teleph. Funk Fernsehstech.*, vol. 8, pp. 122–125; September/October, 1954.) Use of a 200- μ s timebase, corresponding to a height range of 30 km, enables measurements to be made accurate to within 1 km. The timebase triggering is controlled by means of a phase shifter calibrated in height, so that any desired height range can be selected for close examination. Because of the pulse widening involved, a differentiator stage with wide-band amplifier is interposed between receiver output and indicator. Measurement procedure is described.

- 551.510.535:621.3.087.4 713**
Ionospheric Height Measurement by the Method of Delayed Coincidence—H. Rakshit and S. D. Chatterjee. (*Naturwiss.*, vol. 41, pp. 401–402; September, 1954. In English.) An outline description, with block diagram, is given of equipment in use at an Indian station for regular observation of lunar tides in the upper atmosphere. The apparatus can be readily adapted for automatic recording of $h'f$ curves. The method is basically as described previously (*Science and Culture*, vol. 17, p. 520; 1952), but the technique has been improved, giving an accuracy within ± 0.1 km. Good results are obtained even in the presence of heavy atmospherics.

- 551.510.535:621.396.812.3.029.53 714**
Periodic Fading of Medium-Wave Radio Signals—B. R. Rao and N. V. G. Sarma. (*Current Sci.*, vol. 23, pp. 287–288; September, 1954.) Slow fading was found to correspond to interference between $1 \times E$ and $2 \times E$ paths and between $2 \times E$ and $3 \times E$ paths, rapid fading to interference between $1 \times E$ and $3 \times E$ paths. The values of the vertical drift velocity of the E layer calculated on the basis of observations of the three types of interference at an Indian station, for the period between 0700 and 0800 hours, were 2.15 m, 2.34 m, and 2.29 m respectively. Ground-wave fading indicates a drift velocity of 2.37 m. Analysis of records shows that the drift velocity decreases during the morning; depth of fading also decreases, due to increasing D-layer absorption.

- 551.594.5 715**
Variations of Intensity of the Aurora at Macquarie Island—F. Jacka. (*Aust. Jour. Phys.*, vol. 7, pp. 477–484; September, 1954.)

- 551.594.5 716**
Excitation Processes in the Aurora and Air-

glow: Part 1—Absolute Intensities, Relative Ultraviolet Intensities and Electron Densities in High-Latitude Aurorae. Part 2—Excitation of Forbidden Atomic Lines in High-Latitude Aurorae—M. J. Seaton. (*Jour. Atmos. Terr. Phys.*, vol. 4, pp. 285–313; January, 1954.) The excitation processes are evaluated on the basis of optical and rf observations. Calculations show that electron densities of 10^7 – 10^8 cm^{-3} occur in bright high-latitude auroras. A summary is given of the various excitation and deactivation processes which may occur and an attempt is made to decide which of those will be of major importance.

551.594.6 717

Atmospherics with Long Trains of Pulses—F. Hepburn and E. T. Pierce. (*Phil. Mag.*, vol. 45, pp. 917–932; September, 1954.) "The waveforms of atmospherics having long-continued trains of pulses and the systematic modifications associated with time of recording, storm distance and the presence of a low-frequency component, are described. Their interpretation is discussed and the results of analysis—assuming the simple ionospheric reflection mechanism—are presented. Estimates of reflection height and storm distance show the applicability of the theory to the temporal parameters of the waveforms, and the origins of two groups of atmospherics having calculated ranges of 4500 and 7000 km are considered. The variation of pulse amplitude with reflection order is shown to lead to the postulation of horizontal radiating elements in the channel during the later stages of the return stroke, although difficulties arise in reconciling this concept with considerations of the magnitude and orientation of the horizontal elements." See also 2771 of 1953.

LOCATION AND AIDS TO NAVIGATION

621.396.9+621.396.11+523.16 718

Propagation of Electromagnetic Waves, Radio Location and Radio Astronomy—E. Roessler. (*Elektron. Z., Edn. A*, vol. 75, pp. 632–635; September 11, 1954.) A brief survey of recent progress, particularly since 1951. 73 references.

621.396.96.012.3 719

Radar Doppler Nomograph—A. H. Schooley. (*Electronics*, vol. 27, p. 180; December, 1954.) A nomogram is presented relating the Doppler frequency shift to transmitter frequency and target velocity.

621.396.963.325 720

P.P.I. Light-Spot Brightness Probability Distributions—G. C. Sponsler and F. L. Shadler. (*Jour. Appl. Phys.*, vol. 25, pp. 1271–1277; October, 1954.) A study is made using the statistics of noise theory. For a Type-10KP7 cathode-ray tube, the intensifier electrode has an approximately 2.5-power-law characteristic; this combines with the square-law second detector to give an over-all 5th-power detection characteristic. The spot brightness is investigated for integration of seven individual returns. "Various mathematical methods of handling the problem are considered. The Edgeworth series approximation is found to give poor results compared with the Laguerre polynomial approximation. By the latter method the light brightness probabilities are found to be obtainable by interpolation from a table of the incomplete gamma function. Ancillary tables of statistical moments and selected values of the confluent hypergeometric function, ${}_1F_1(-5n/2; 1; -x)$, are included in the text."

621.396.969:551.577/.578 721

Radar Echoes from Monsoon Rain—L. S. Mathur, A. C. De, B. N. Dutta and H. Mitra. (*Indian Jour. Met. Geophys.*, vol. 5, pp. 173–186; April, 1954.) Account of observations made at New Delhi, using modified Type-AN/APQ-13 3-cm radar equipment. PPI displays

corresponding to different weather conditions are reproduced.

621.396.969.33/.34:621.396.822 722

The Reduced Range in a Radar Subjected to an External Noise Generator—U. Tiberio. (*Proc. I.R.E.*, vol. 42, pp. 1791–1798; December, 1954.) An analytical method is described for calculating the reduction of range due to a noise generator (a) carried by the target, or (b) at some fixed location. A "reduced range index" is determined from consideration of free-space operation against aircraft, and the effect of reflection from the sea is investigated in relation to operation at low height against ships. The effect of noise on the visibility factor is briefly discussed.

MATERIALS AND SUBSIDIARY TECHNIQUES

535.5 723

Ion Pump—P. F. Váradí. (*Acta Tech. Acad. Sci. Hungaricae*, vol. 9, pp. 343–353; 1954. In English.) The ultimate pressure and pumping speed calculated for the case of a simple model are in agreement with experimental results of Foster et al. (*Rev. Sci. Instr.*, vol. 24, pp. 387–390; May, 1953.), who described an ion pump with pumping speeds between 3,000 and 7,000 l/sec at a base pressure of about 10^{-6} mm Hg. The present experimental result indicates that the reduction of gas pressure is attributable not solely to adsorption but also to a true pumping effect.

535.215:[537.311.33+537.226 724

Photoeffect from Surface Levels—G. E. Pikus. (*Zh. Eksp. Teor. Fiz.*, vol. 27, pp. 369–381; September, 1954.) The external photoeffect in semi-conductors and dielectrics, corresponding to removal of electrons from the surface zone, is considered theoretically. Expressions are derived for the energy distribution of the emitted photoelectrons and the quantum output and its dependence on the frequency of the incident light.

535.215:546.482.21 725

Photovoltaic Effect in Cadmium Sulfide—D. C. Reynolds, G. Leies, L. L. Antes and R. E. Marburger. (*Phys. Rev.*, vol. 96, pp. 533–534; October 15, 1954.) Brief report of observations.

535.215:546.482.21 726

Absorption and Conductivity Measurements on CdS in the Soft-X-Ray Region—E. Schnürer. (*Ann. Phys. (Lpz.)*, vol. 15, pp. 15–20; September 15, 1954.)

535.215:546.817.241:539.23 727

Effect of Oxygen on the Electrical Properties of Lead Telluride Films—D. E. Bode and H. Levinstein. (*Phys. Rev.*, vol. 96, pp. 259–265; October 15, 1954.) Account of an experimental investigation. Exposure to oxygen produces first an increase and then a decrease in the film resistance. The nature of the material changes from *n*-type to *p*-type in the neighborhood of the resistance maximum. The magnitudes of the photoconductive and photovoltaic effects depend on the amount of oxygen adsorbed. The observed results are explained on the basis of a model in which the oxygen removes electrons first from the conduction band, then from trapping states, and finally from the valence band.

535.37 728

On the Infrared-Sensitive Behaviors of Some Doubly Activated ZnS Phosphors—S. Asano. (*Jour. Phys. Soc. (Japan)*, vol. 9, pp. 580–594; July/August, 1954.)

535.37 729

Investigations of the Stimulation of Phosphorescence in Calcium Oxide—A. Crozet and J. Janin. [*Compt. Rend. Acad. Sci. (Paris)*, vol. 239, pp. 1031–1034; October 27, 1954.] The

effect of different activators is discussed. If rare earths are used there is a large recapture of liberated electrons.

537.226.2 730

The Dielectric Behaviour of Acetaldehyde Vapour at 9000 Mc/s—Krishnaji and P. Swarup. (*Z. Phys.*, vol. 138, pp. 550–556; September 18, 1954. In English.)

537.226.31 731

Investigation of Dielectric Losses due to Low-Frequency Relaxation in Polyethylene—G. P. Mikhailov, A. M. Lobanov and B. I. Sazhin. (*Zh. Tekh. Fiz.*, vol. 24, pp. 1553–1560; September, 1954.) Experimental results show that the losses are connected with the presence of $C=O$ polar groups and their orientation. The relaxation time associated with lf ($\sim 10^2$ cps) losses decreases with extension of the specimen, that of hf ($\sim 10^9$ cps) losses increases. The dependence of the loss angle on temperature, frequency, and percentage of crystalline phase is shown graphically.

537.226.33 732

Transient State in Dielectrics—J. Granier and P. Caillon. [*Compt. Rend. Acad. Sci. (Paris)*, vol. 239, pp. 1025–1027; October 27, 1954.] If only one hysteresis effect came into play, it would be possible to predict ac dielectric properties from dc properties. Examination of experimental results on high polymers confirms the existence of two independent hysteresis phenomena, the one related to the dipole orientation and the other to ionic polarization. See also 2963 of 1954.

537.227:546.431.824-31 733

Electromechanical Activity of BaTiO₃ Ceramic subjected to Opposing Polarization—T. F. Hueter and D. P. Neuhaus. (*Naturwiss.*, vol. 41, p. 424; September, 1954.) Over a range of field strength within the coercive field strength, second-harmonic oscillations become pronounced while the fundamental and the third harmonic disappear. Curves showing the variation of fundamental and second-harmonic amplitude with field strength are presented for a disk of thickness 0.1 cm, and an interpretation is provided in terms of domain processes.

537.227:546.431.824-31 734

An Experimental Study of Polarization Effects in Barium Titanate Ceramics—T. F. Hueter, D. P. Neuhaus and J. Kolb. (*Jour. Acous. Soc. Amer.*, vol. 26, pp. 696–703; September, 1954.) The following points were investigated: (a) transducer performance as a function of polarizing bias; (b) relative role of mechanical and dielectric losses; (c) coercivity of pre-polarized BaTiO₃; (d) effect of bias on subsidiary transducer responses; (e) constriction of dielectric hysteresis loops. The value of the experiments for elucidating the relation between the properties of the ceramic and single-crystal forms of the material is discussed.

537.227:546.431.824-31 735

Dielectric-Constant Behavior of Single-Domain, Single Crystals of Barium Titanate in the Vicinity of the Curie Point—M. E. Drougard and D. R. Young. (*Phys. Rev.*, vol. 95, pp. 1152–1153; September 1, 1954.) Brief report of measurements which confirm earlier observations by Cross (751 of 1954) of a discontinuity in the value of the dielectric constant at the Curie point.

537.227:546.431.824.831.4-31 736

Ferroelectric Properties of Solid Solutions of Barium Zirconate in Barium Titanate—G. A. Smolenski, N. P. Tarutin and N. P. Grudtsin. (*Zh. Tekh. Fiz.*, vol. 24, pp. 1584–1593; September, 1954.) An experimental investigation of solutions containing up to 40 per cent (molar) of BaZrO₃. Results, which are presented graphically, show (a) the highest value of dielectric constant ($>12,000$) at a frequency of 1 kc occurs for 18–20 per cent BaZrO₃ content,

(b) the Curie temperature is displaced downwards more slowly than in the BaSnO_3 -in- BaTiO_3 solutions due to the different character of the bonds of Zr and Sn ions with oxygen ions, (c) the dielectric constant of solutions with low electrostriction falls considerably following polarization at high field strengths, (d) the dependence of resonance frequencies on field strength decreases with increase of the zirconate content, and (e) the piezoelectric-modulus maximum occurs at a temperature slightly lower than that corresponding to the dielectric-constant maximum. Some properties of pure BaTiO_3 were also investigated.

537.311.31:[538.632+537.312.8] 737
Hall Effect and Change of Resistance of Pb, Cu, and Mg in a Magnetic Field—E. S. Borovik. (*Zh. Eksp. Teor. Fiz.*, vol. 27, pp. 355-368; September, 1954.) An experimental investigation is reported on pure polycrystalline specimens in fields of strengths up to about 25,000 oersted at temperatures between 2 degrees and 300 degrees K. From a comparison of the experimental results with results calculated on the basis of an isotropic model of a metal with overlapping energy bands values are obtained for the mobilities and concentrations of charge carriers. The magnitude of the mean free path is compared with the values obtained by other methods. Results are tabulated and presented graphically.

537.311.31:621.3.029.64 738
Surface Loss of Silver-Plated Metal Plates at 9000 Mc/s and its Correlation with Surface Roughness—S. Saito. (*Proc. I.R.E.*, vol. 42, p. 1810; December, 1954.) A cavity-resonator method is outlined for comparing the surface loss of metal plates. The surface roughness was simultaneously observed by a mechanical-stylus method and by means of electron micrographs. The results confirm that surface loss increases rapidly with increase in surface roughness. Plating defects in the silver-plated samples appear to be responsible for abnormally high losses.

537.311.33 739
Magnetoresistance Effect in Cubic Semiconductors with Spheroidal Energy Surfaces—M. Shibuya. (*Phys. Rev.*, vol. 95, pp. 1385-1393; September 15, 1954.) "The collision frequency of electrons having a spheroidal energy surface with acoustical modes of vibration is calculated without neglecting phonon energy. Using an asymptotic form in which the collision frequency is proportional to the square root of their energy, the electronic current in a semiconductor in combined magnetic and weak electric fields can be calculated in a closed form by the formal theory of conductivity. The results are compared with those obtained experimentally by Pearson and Suhl (166 of 1952)."

537.311.33 740
Mathematical Methods for Zone-Melting Processes—H. Reiss. (*Jour. Metals (New York)*, vol. 6, pp. 1053-1059; September, 1954.) The mechanism of zone melting [2125 of 1954 (Pfann)] is discussed in terms of a transport process including diffusive and convective flows. This approach provides a basis on which equations are developed for the solute concentration in the ingot as a function of the number of zone passes.

537.311.33 741
Quantum Theory of Cyclotron Resonance in Semiconductors—W. Kohn and J. M. Luttinger. (*Phys. Rev.*, vol. 96, pp. 529-530; October 15, 1954.) For the electrons in semiconductors, the quantum theory is identical with the classical theory [2479 of 1950 (Shockley)]. For holes, the situation is complicated due to degeneracy at the top of the valence band; the quantum theory leads to different energy levels and selection rules for low quan-

tum numbers, but for high quantum numbers the classical theory is again valid.

537.311.33:537.323 742
Temperature Dependence of Thermoelectric Power of Impurity Semiconductors—T. A. Kontorova. (*Zh. Tekh. Fiz.*, vol. 24, pp. 1687-1696; September, 1954.) Theoretical considerations show that the large values of the thermoelectric power observed by Frederikse (1093 of 1954) and others at very low temperatures can be accounted for by accepted theory, assuming the electron gas to be highly degenerate in that region. The maximum occurs at the transition from the degenerate state to the "classical" state.

537.311.33:537.323:546.289 743
Theory of Thermoelectric Power in Semiconductors—J. Tauc. (*Phys. Rev.*, vol. 95, p. 1394; September 15, 1954.) A method is suggested for calculating the thermoelectric power which gives a more correct expression than that presented by Johnson and Lark-Horowitz (1092 of 1954).

537.311.33:546.23 744
Some Investigations on the Electrical Properties of Hexagonal Selenium—L. M. Nijland. (*Philips Res. Rep.*, vol. 9, pp. 259-294; August, 1954.) Known properties of Se are reviewed with references to published data. A method of purifying 99.9 per cent-pure Se by evaporation at a temperature near its melting point is described. Results of measurements of hf conductivity confirm the assumption that polycrystalline Se consists of crystals of fairly good conductivity embedded in poorly conducting layers of amorphous Se. The inclusion of thallium increases the resistance of the layers without greatly affecting that of the crystals. Measurements of Hall effect and of equivalent shunt resistance as a function of frequency are also reported for pure and bromine-containing samples; results indicate the same layer structure, but inclusion of Br lowers the resistance of the layers. Conduction mechanism is discussed with reference to detailed experimental results.

537.311.33:[546.28+546.289] 745
Etch Pits and Dislocations in Germanium and Silicon—J. J. Oberly. (*Jour. Metals (New York)*, vol. 6, pp. 1025-1026; September, 1954.) Brief illustrated discussion of conical etch pits observed while examining lineage boundaries described by Vogel et al. (2693 of 1953).

537.311.33:[546.28+546.289] 746
Theory of Electron Multiplication in Silicon and Germanium—P. A. Wolff. (*Phys. Rev.*, vol. 95, pp. 1415-1420; September 15, 1954.) Multiplication of electrons and holes at junctions in Si and Ge is explained in terms similar to those of gas-discharge theory. The calculated ionization-rate/field characteristic for Si is in agreement with that obtained experimentally, assuming a mean free path of 200 Å for interactions between electrons and optical phonons.

537.311.33:[546.28+546.289] 747
Mobility of Impurity Ions in Germanium and Silicon—C. S. Fuller and J. C. Severiens. (*Phys. Rev.*, vol. 96, pp. 21-24; October 1, 1954.) The diffusivity D of Li in Ge and Si was investigated by measuring the mobilities of the Li^+ ions on applying an electric field. Using the Einstein formula relating the two properties, the value of D was found to be $25 \times 10^{-4} \exp(-11,800/RT)$ for Ge and $23 \times 10^{-4} \exp(-15,200/RT)$ for Si, in satisfactory agreement with previously published results. The region into which the Li diffuses in the p -type Ge changes to n -type, but a small region round the injection area reverts to p -type. Similar experiments with Cu in n -type Ge are mentioned.

537.311.33:[546.28+546.289]:536.2 748
The Thermal Conductivity of Germanium and Silicon at Low Temperatures—H. M. Rosenberg. (*Proc. Phys. Soc.*, vol. 67, pp. 837-840; September 1, 1954.) "The thermal conductivity of a single crystal of Ge and a polycrystalline specimen of Si have been measured in the range 2 to 100 degrees K. Both specimens were very pure. The results indicate that, at low temperatures at least, the lattice waves are not scattered by the conduction electrons, but that their mean free path is limited either by the size of the specimen (for the germanium) or by the crystallite size (for the silicon)."

537.311.33:546.28 749
Electron Spin Resonance of an Impurity Level in Silicon—A. Honig and A. F. Kip. (*Phys. Rev.*, vol. 95, pp. 1686-1687; September 15, 1954.) Electron spin resonance has been observed in a Si sample containing Li at a concentration of 7×10^{16} atoms/cm³. The ionization energy of an electron in the impurity level is 0.033 ev. A single resonance line was observed over the temperature range 4 degrees-20 degrees K using an applied frequency of about 8.8 kmc and a magnetic field of about 3,200 oersted; the same line was observed at 300 mc. The evidence indicates that the electron is bound to the impurity atom rather than associated with an impurity band.

537.311.33:546.28 750
Electrical Properties of Silicon containing Arsenic and Boron—F. J. Morin and J. P. Maita. (*Phys. Rev.*, vol. 96, pp. 28-35; October 1, 1954.) Measurements were made of the conductivity and Hall constant of single-crystal specimens over the temperature range 10 degrees-1,100 degrees K. Analysis of the extrinsic carrier concentration values, as computed from the Hall constant, indicates the ionization energy of As donor levels to be 0.049 ev, and of B acceptor levels to be 0.045 ev for low impurity concentrations. Fermi degeneracy is found to occur in the impurity concentration range 10^{18} - 10^{19} per cm³. A formula is derived for the variation of carrier concentration with temperature up to 700 degrees. Mobility values are computed.

537.311.33:546.28 751
Effective Masses of Holes in Silicon—R. N. Dexter and B. Lax. (*Phys. Rev.*, vol. 96, pp. 223-224; October 1, 1954.) Cyclotron-resonance experiments are reported, the carriers being excited by infrared radiation chopped at 900 cps. The effective mass of holes is plotted as a function of direction of magnetic field.

537.311.33:546.28 752
Effective Masses of Electrons in Silicon—R. N. Dexter, B. Lax, A. F. Kip and G. Dresselhaus. (*Phys. Rev.*, vol. 96, pp. 222-223; October 1, 1954.) Results of cyclotron resonance experiments using optical excitation of carriers are reported. Curves are shown of the effective electron mass as a function of direction of magnetic field.

537.311.33:546.28 753
Polarization of Arsenic Nuclei in a Silicon Semiconductor—A. Honig. (*Phys. Rev.*, vol. 96, pp. 234-235; October 1, 1954.) A mechanism capable of producing nearly 100 per cent polarization of nuclear spins at moderate values of field strength and temperature has been found in the course of electron spin resonance studies of the type previously reported [3254 of 1954 (Fletcher et al.)].

537.311.33:546.289 754
Electron Multiplication in Germanium at Low Temperature—E. J. Ryder, I. M. Ross and D. A. Kleinman. (*Phys. Rev.*, vol. 95, pp. 1342-1343; September 1, 1954.) In response to Conwell's suggestion of an experimental check (2976 of 1954), measurements were made of the current density in a small bar of n -type Ge as

a function of electric field strength at several temperatures in the range 12.1 degrees–300 degrees K. The curves for the lower temperatures exhibit a steep rise over part of the field-strength range; this is interpreted as evidence of electron multiplication.

537.311.33:546.289 755

Distribution of the Mass Transported from a Collector into a Germanium Crystal by the Forming Process—W. M. Aarons, M. Pobereskin, J. E. Gates and E. B. Dale. (*Phys. Rev.*, vol. 95, p. 1345; September 1, 1954.) Measurements were made using a radioactive isotope of Au as tracer. The Au was plated on a W needle which was used to form the crystals. Results obtained with successive lappings of the surface indicate that the concentration of the transferred Au atoms is high in a region near the surface, then falls, rises a little, and finally drops abruptly.

537.311.33:546.289 756

The Interaction of Impurity Atoms with Dislocations in Germanium—A. D. Kurtz and S. A. Kulin. (*Acta metallurgica*, vol. 2, pp. 352–354; March, 1954.) It is suggested that the existence of dislocations in Ge gives rise to certain specific distributions of solute atoms. Results of approximate calculations give some support to this view; the theory enables some of the electrical properties of Ge to be predicted.

537.311.33:546.289 757

Effect of Dislocations on Minority-Carrier Lifetime in Germanium—S. S. Kulin and A. D. Kurtz. (*Acta metallurgica*, vol. 2, pp. 354–356; March, 1954.) The density of randomly distributed dislocations in Ge, as determined by two independent methods, varies between 10^6 and 10^8 per cm^2 . The lifetime of minority carriers decreases hyperbolically as the dislocation density increases. The recombination efficiency per dislocation is about 2×10^{-3} per cm per second. The change in energy gap width is calculated as a function of position in relation to a dislocation.

537.311.33:546.289 758

New Minority-Carrier Phenomenon in Germanium—S. J. Angello and T. E. Ebert. (*Phys. Rev.*, vol. 96, pp. 221–222; October 1, 1954.) An experiment is described in which minority carriers were withdrawn from a bar of n -type Ge at an In-alloyed junction biased in the high-resistance direction, and the deficit was propagated along the bar by means of an electric field. The effect is the inverse of that described by Haynes and Shockley (2109 of 1949).

537.311.33:546.289 759

Injection Breakdown in Iron-Doped Germanium Diodes—W. W. Tyler. (*Phys. Rev.*, vol. 96, pp. 226–227; October 1, 1954.) Brief description of an experiment providing evidence of hole traps in high-resistivity n -type Fe-doped Ge.

537.311.33:546.289 760

Properties of Zinc-, Copper-, and Platinum-Doped Germanium—W. C. Dunlap, Jr. (*Phys. Rev.*, vol. 96, pp. 40–45; October 1, 1954.) Measurements of Hall constant and resistivity over the temperature range 15 degrees–400 degrees K indicate that Zn, Cu and Pt are all acceptors, with ionization energies of 0.029, 0.036 and 0.040 eV respectively. The temperature variation observed could be due to (a) surface conductivity with low activation energy, (b) traces of low-ionization-energy acceptors, or (c) internal leakage due to imperfections or dislocations. Evidence was found of a Pt acceptor level 0.2 eV below the conduction band and of a Cu acceptor level just below the middle of the forbidden band.

537.311.33:546.289 761

Thermal Effects on Lifetime of Minority Carriers in Germanium—R. A. Logan and M.

Schwartz. (*Phys. Rev.*, vol. 96, p. 46; October 1, 1954.) Practical precautions are described which enable Ge to be heated to temperatures as high as 875 degrees C. without causing a decrease in the lifetime of the minority carriers.

537.311.33:546.289 762

Precision Wavelength and Isotopic Shift Measurements of Germanium Arc Lines—G. V. Deverall, K. W. Meissner and G. J. Zissis. (*Phys. Rev.*, vol. 95, pp. 1463–1468; September 15, 1954.)

537.311.33:546.623.86 763

Some Electrical Properties of AlSb—W. Sasaki, N. Sakamoto and M. Kuno. (*Jour. Phys. Soc. Japan*, vol. 9, p. 650; July/August, 1954.) Measurements of resistivity, Hall constant and thermoelectric power as a function of temperature are reported.

537.311.33:[546.682.86+546.682.19] 764

Anomalous Optical Behavior of InSb and InAs—H. J. Hrostowski, G. H. Wheatley and W. F. Flood, Jr. (*Phys. Rev.*, vol. 95, pp. 1683–1684; September 15, 1954.) Observations have been made of the room-temperature transmission spectra of degenerate n -type InSb samples doped so as to have different values of electron concentration. The absorption edge is displaced to shorter wavelengths as the electron concentration is increased. The variation of the energy gap E_0 with electron concentration is compared with the curve obtained by calculation from the data of Tannenbaum and Maita (758 of 1954). Similar but smaller effects have been observed with InAs. The results indicate that the anomalous variation of E_0 is unlikely to be due to a specific impurity.

537.311.33:546.682.86 765

Neutron Irradiation of Indium Antimonide—J. W. Cleland and J. H. Crawford, Jr. (*Phys. Rev.*, vol. 95, pp. 1177–1182; September 1, 1954.) Measurements were made of Hall coefficient and resistivity of n -type and p -type single crystals of InSb after exposure to neutron irradiation. The results indicate that bombardment by fast neutrons converts p -type material to n -type and produces shallow electron traps in n -type material. Reduction of carrier mobility and changes of carrier concentration resulting from the bombardment can be removed by heat treatment.

537.311.33:546.817.241 766

Preparation and Properties of Lead Telluride—E. L. Brady. (*Jour. Electrochem. Soc.*, vol. 101, pp. 466–473; September, 1954.) "Single crystals of lead telluride, PbTe, have been prepared and their resistivity and Hall coefficients determined. Both n - and p -type lead telluride have been produced, but they were not of high resistivity. Charge carrier concentration in every case has been $1\text{--}5 \times 10^{18}/\text{cm}^3$. Hall mobility of n - and p -type carriers was found to be about 2,240 and 860 $\text{cm}^2/\text{volt-sec}$, respectively. Material of p -type was converted to n -type by allowing lead to diffuse into the crystal at 500 degrees C. The value of the diffusion coefficient of Pb in PbTe at this temperature is estimated to lie between 5.6×10^{-4} and $9.2 \times 10^{-8} \text{ cm}^2/\text{sec}$."

537.311.33:546.817.241:539.234:535.3 767

Optical Properties of Lead Telluride—M. E. Lasser and H. Levinstein. (*Phys. Rev.*, vol. 96, pp. 47–52; October 1, 1954.) Evaporated films were prepared having a density about 10 per cent less than that of the bulk material. The optical constants were calculated from curves of reflection and transmission plotted against λ . Addition of oxygen caused an increase in absorption and a slight increase in refractive index; the optical properties were then strongly dependent on the film temperature. A possible explanation of the results is presented.

537.311.33:621.314.63 768

Influence of Recombination at Contact on

the Volt/Ampere Characteristics of a Rectifier—A. V. Rzhanov. [*Compt. Rend. Acad. Sci. (URSS)*, vol. 98, pp. 389–390; September 21, 1954. In Russian.] An expression is derived for the current flowing in an n -type semiconductor bounded on one side by p -type material and on the other by a nonactive contact. The electron flow across the p - n junction is neglected. Surface recombination takes place at the other boundary. If the thickness of the semiconductor is small compared with the diffusion path of holes, and the velocity of surface recombination is large, then the effect of recombination is to increase the saturation current. This result is derived on the assumption of a concentration of holes which is small in comparison with the equilibrium concentration of electrons, i.e. applies to reverse and small direct currents through the rectifier. An expression for the current in the case of large hole concentrations is also given.

537.311.33:621.314.63 769

Theory of Rectification at a Metal/Semiconductor Contact—V. Schultz. (*Z. Phys.*, vol. 138, pp. 598–612; September 18, 1954.) A study is made particularly of the influence of the inversion layer. The analysis is presented for an excess semiconductor, but corresponding arguments hold for a defect semiconductor. On making simplifying assumptions which are generally valid for semiconductors with high mobility and long diffusion path, the rectification process can be described using diode theory for the electron current and Shockley's theory of p - n junctions for the hole current. An expression is derived for the blocking-layer capacitance as a function of bias voltage and of a parameter V which varies slightly with the bias. The temperature dependence of V is discussed.

537.311.33:621.314.632 770

Theory of Contact Phenomena—G. M. Abak'yants. (*Zh. Eksp. Teor. Fiz.*, vol. 27, pp. 333–346; September, 1954.) This theory of metal-semiconductor contacts takes into account the change of the mean kinetic energy ("heating") of electrons in an electric field. The contact resistance, at constant current, is calculated for both Schottky- and Davydov-type layers. In a discussion of a note by Burgess (2697 of 1953) it is pointed out that the Einstein equation does not apply and the diffusion coefficient D is approximately proportional to the product of the mobility and the mean energy of the electrons in the electric field. The expression for drift velocity should take into account the difference between the temperature of the electrons and the lattice, which gives rise to thermal currents, and also the effect of a non-uniform electric field in the contact region. Krömer's paper (2821 of 1953) is also briefly commented on.

537.311.33:621.314.632 771

Flow of Electrons and Holes through the Surface-Barrier Region in Point-Contact Rectification—M. Cutler. (*Phys. Rev.*, vol. 96, pp. 255–259; October 15, 1954.) Equations are derived for emission-controlled flow, taking account of nonequilibrium concentration of carriers on the semiconductor side of the barrier. A solution based on the assumption that part of the voltage drop occurs between the metal and semiconductor surfaces, rather than entirely in the barrier, leads to improved agreement between theoretical and observed current/voltage characteristics. The part played by diffusion is also discussed.

537.311.33:621.396.822 772

Some Notes on Gisolf's Theory of Electron Fluctuation Phenomena in Semiconductors—K. W. Böer. [*Ann. Phys. (Lpz.)*, vol. 15, pp. 55–56; September 15, 1954.] Correction to paper abstracted in 2139 of 1954.

537.529:621.315.61 773

The Statistical Time Lag of the Dielectric

Breakdown of Mica, Glass and KCl—H. Kawamura, H. Ohkura and T. Kikuchi. [*Jour. Phys. Soc. (Japan)*, vol. 9, pp. 541-545; July/August, 1954.] Results of pulse measurements indicate that the statistical time lag at 10 per cent overvoltage is up to 10^{-1} second for mica but $>10^{-7}$ second for glass and KCl. Theories of the breakdown mechanism are discussed in the light of these figures.

538.221 774

Ferromagnetism of Certain Manganese-Rich Alloys—E. R. Morgan. [*Jour. Metals (New York)*, vol. 6, pp. 983-988; September, 1954.] Report of an investigation of a series of alloys based on the composition $(\text{MnX})_2\text{C}$, where X is a metallic element which has both a positive size factor with respect to Mn and a high positive valence, e.g. Al, In and Sn. Measurements indicate that the effective magnetic moment of Mn in the alloys is at least 1.0 Bohr magneton per atom.

538.221 775

Study of Strip Ferronickels around the Curie Point, using Weak Alternating Fields—A. Marais. [*Compt. Rend. Acad. Sci. (Paris)*, vol. 239, pp. 873-875; October 11, 1954.] Curves are given showing the variation of initial permeability with temperature for some Ni-Fe-Cu alloys containing either Mo or Cr in addition. The influence of specimen thickness and duration of heat treatment is indicated.

538.221:537.533.8 776

Nickel Alloys with High Secondary Emissivity—A. Bobenrieth, J. Millet and S. Teszner. [*Compt. Rend. Acad. Sci. (Paris)*, vol. 239, pp. 794-796; October 4, 1954.] Ni-Be alloys with Be content up to 5 per cent by weight were prepared in a lif oven at low pressure. Measurements of the secondary emissivity are reported on specimens with 4 per cent and with 3 per cent Be, using primary electron voltages between 300 and 800 v and collector voltages between 100 and 500 v. Values up to 96 were recorded for the 4 per cent alloy and up to 45 for the 3 per cent alloy. With secondary electron currents >1 ma, no signs of fatigue were observed in the 4 per cent alloy over test periods of 8 hours. Ni-Mg alloys were prepared containing up to 1.2 per cent Mg. The highest value of secondary emissivity obtained was 2.

538.221:538.632 777

Hall Effect in Ferromagnetics—C. Kooi. [*Phys. Rev.*, vol. 95, pp. 843-844; August 1, 1954.] Measurements on Si-Fe alloys are reported briefly; the results are in good agreement with values predicted theoretically by Karplus and Luttinger (698 above).

538.221:621.318.134 778

Some Properties of Nickel-Zinc Ferrites—L. I. Rabin and B. Sh. Epshtein. (*Zh. Tekh. Fiz.*, vol. 24, pp. 1568-1578; September, 1954.) Experimental investigation is reported of the dependence of the magnetic properties of ferrites with mean permeabilities ranging from 40 to 2,500 on the frequency of the magnetic field (up to >10 mc), temperature between about -80 degrees C. and the Curie points (lying between 250 degrees C. and 80 degrees C.), and field strength up to ~ 1 oersted. The permeability and losses in weak pulsed fields and the dielectric properties were also investigated. Results are presented graphically; the code numbers denote the mean permeability in gauss/oersted.

538.221:621.318.134 779

Temperature Dependence of the Magnetic Properties of Nickel-Zinc and Copper-Zinc Ferrites—A. I. Suchkov. (*Zh. Tekh. Fiz.*, vol. 24, pp. 1579-1583; September, 1954.) Magnetic properties investigated experimentally include the saturation magnetization, saturation magnetostriction, coercive force, and initial and maximum permeabilities. Curie points of $\text{CuO-ZnO-Fe}_2\text{O}_3$ ferrites, of the molar compo-

sitions stated, lie between 30 degrees and 180 degrees C. Results are presented graphically and indicate that the general theory and the quantum theory of ferromagnetism also cover the temperature dependence of the magnetic properties of ferrites.

538.221:621.318.134 780

Neutron Diffraction Studies of a Nickel Zinc Ferrite—V. C. Wilson and J. S. Kasper. [*Phys. Rev.*, vol. 95, pp. 1408-1411; September 15, 1954.]

538.221:621.318.134.029.64/.65 781

On the Internal Field of the Microwave Resonance in Ferrites—N. Tsuya. [*Jour. Phys. Soc. (Japan)*, vol. 9, pp. 644-645; July/August, 1954.] A possible mechanism is proposed which may cause the additional internal field.

538.221:669.14.018.58 782

Nature of Change of Coercive Force due to Tempering of Hardened Low-Carbon Steel—I. A. Bil'dzyukevich, Ya. M. Golovchiner and G. V. Kurdymov. [*Compt. Rend. Acad. Sci. (URSS)*, vol. 98, pp. 385-387; September 21, 1954. In Russian.] The effect of tempering of 0.1-0.12 per cent C steel, hardened by quenching in water from 1,100 degrees C. was investigated experimentally. Results indicate that the decrease of coercivity with increase of temperature (up to 600 degrees C.) is primarily due to the removal of strains in the steel.

538.652 783

Derivation of Magnetostriction and Anisotropic Energies for Hexagonal, Tetragonal, and Orthorhombic Crystals—W. P. Mason. [*Phys. Rev.*, vol. 96, pp. 302-310; October 15, 1954.] "In order to determine the measurements necessary to characterize the anisotropic energy and the saturation magnetostriction in hexagonal cobalt, a phenomenological derivation has been given for the equations which characterize the effects. Out to fourth rank tensors, the results are the same as those for circular symmetry and it requires two constants to specify the anisotropic energy and four to specify the magnetostriction. When sixth rank tensors are evaluated, a characteristic hexagonal symmetry appears. It requires four constants to characterize the anisotropic energy and nine to characterize the magnetostriction. These constants can be measured by using two oriented slabs. Four of the constants can be determined by measurements parallel to the saturation magnetization, four when the magnetostriction is perpendicular to the magnetization and one when they are 45 degrees apart. In the appendix the first approximations for the magnetostrictive and anisotropy energies are derived for tetragonal and orthorhombic crystals."

538.652 784

Magnetostriction and Crystal Anisotropy of Single Crystals of Hexagonal Cobalt—R. M. Bozorth. [*Phys. Rev.*, vol. 96, pp. 311-316; October 15, 1954.] Measurements at field strengths up to 25,000 oersted are reported. Results are discussed in terms of theory developed by Mason (783 above). A volume contraction associated with domain orientation was observed, its value being as great as 26×10^{-6} for the most effective direction of magnetization. Superposed on this contraction is an isotropic increase of volume of 0.6×10^{-9} per oersted.

538.652:538.221 785

The Magnetostriction Constants of Silicon Steel: Part I.—H. Takaki and Y. Nakamura. [*Jour. Phys. Soc. Japan*, vol. 9, pp. 507-511; July/August, 1954.] Magnetization measurements using a ballistic method and magnetostriction measurements using a mechano-optical method were made on long single-crystal specimens of 0.7 per cent and 1.8 per cent Si steel; the magnetostriction constants λ_{100} and λ_{111} and the crystal-anisotropy constant

K_1 were hence determined. The results confirm that these three constants decrease as the Si content increases.

538.652:538.221 786

Temperature Dependence of Magnetostriction of Ferromagnetic Alloys—D. I. Volkov and V. I. Chechernikov. (*Zh. Eksp. Teor. Fiz.*, vol. 27, pp. 208-214; August, 1954.) The saturation-magnetostriction/temperature characteristics of Ni-Cu, Ni-Mn, and Ni-Fe alloys were determined by means of a tensometer method at temperatures up to the Curie point. Results, which are presented graphically, are in good agreement with theory.

539.23:537.311.3 787

Resistance of Thin Metal Films at High Frequency and Low Temperature—S. Offert and B. Vodar. [*Compt. Rend. Acad. Sci. (Paris)*, vol. 239, pp. 1027-1029; October 27, 1954.] Results of measurements indicate that short-circuiting by intergranular capacitances is probably responsible for the decrease of resistance at high frequency and for the increase in the algebraic value of the temperature coefficient of resistance. At sufficiently high frequencies it should be possible to measure the resistance of the metal grains themselves.

621.315.613.1 788

Synthetic Mica Investigations: Part 5—A Low-Shrinkage Machinable Ceramic of Phosphate-Bonded Synthetic Mica—J. E. Comeforo. [*Jour. Amer. Ceram. Soc.*, vol. 37, pp. 427-432; September 1, 1954.] Total shrinkage is easily maintained at <3 per cent with a material formed by pressing powdered synthetic mica with phosphoric acid as binding agent. The dielectric properties approximate to those reported [784 of 1954 (Comeforo et al.)] for hot-pressed phosphate-free material of similar porosity. Loss factor is 1-2 per cent at room temperature and 4-8 per cent at 300 degrees C. at a frequency of 1 mc. The material is suggested as a substitute for natural block talc.

621.315.616:537.533.9 789

Irradiated Polyethylene—J. B. Campbell. (*Mater. and Meth.*, vol. 40, pp. 91-95; September, 1954.) The effect on polyethylene of bombardment by high-voltage electrons is described; details are given of commercially available material in tape form. Unlike ordinary polyethylene, the irradiated material is not thermoplastic; its use may permit reduction in size of electrical equipment operating at high temperatures. Dielectric strength at high temperature may be improved by the treatment.

621.372.412:549.514.51 790

Effect of Acoustic Impedance and Viscosity of Gases on the Electrical Constants of Quartz—S. Parthasarathy and V. Narasimhan. [*Ann. Phys. (Lpz.)*, vol. 15, pp. 6-14; September 15, 1954. In English.] Experimental results show that the equivalent electrical resistance of a quartz crystal is proportional to the acoustic impedance of the gas in which it is vibrating; the Q factor decreases exponentially with increasing viscosity, as in liquids, but the relation is approximately linear over the small range involved. The natural frequency of the crystal used was 414,216 kc.

621.39:621.791.3 791

Investigations on Soldered Joints—H. Künzler and H. Bohren. (*Tech. Mitt. Schweiz. Telegr.-Teleph. Verw.*, vol. 32, pp. 329-351; September 1, 1954. In German.) Chemical and physico-chemical problems are discussed. Tests were made to find a solder with a noncorrosive flux. The influence of temperature on the structure of soldered materials and the quality of the joints was studied. The wear of Cu soldering irons is much reduced if the solder contains some Cu.

666:539.61:669

792

Frictional Adhesion of Metal to Glass, Quartz, and Ceramic Surfaces—R. B. Belser. (*Rev. Sci. Instr.*, vol. 25, pp. 862-864; September, 1954.) Experiments were made using small disks of various metals rotated at high speed in contact with the various insulator surfaces; by suitably adjusting speeds and pressures a layer of metal was made to adhere securely. The process may be useful for making glass-to-metal joints and electrically conducting lines, and for glass cutting.

MATHEMATICS

512.393

793

Solution of Cubics and Quartics—G. Millington; A. C. Sim. (*Wireless Eng.*, vol. 32, p. 30; January, 1955.) Comment on 187 of January and author's reply. An error in the original paper is corrected.

519.281

794

Method of Averages and its Comparison with the Method of Least Squares—M. Morduchow. (*Jour. Appl. Phys.*, vol. 25, pp. 1260-1263; October, 1954.) Comparison of the two methods is illustrated by treating the problem of fitting a straight line to a number of points. It is shown that the standard deviation of the residuals by the method of averages is at most $2/\sqrt{3}$ times as great as that by the method of least squares.

517.9

795

Relaxation Methods [Book Review]—D. N. de G. Allen. Publishers: McGraw-Hill Book Co., New York and London, 1954, 257 pp., \$7.50. (*Science*, vol. 120, pp. 423-424; September 10, 1954.) A strongly recommended textbook, giving descriptions of the basic operations and their application.

MEASUREMENTS AND TEST GEAR

53.088

796

Sensitivity—a Criterion for the Comparison of Methods of Test—J. Mandel and R. D. Stiehler. (*Jour. Res. Nat. Bur. Stand.*, vol. 53, pp. 155-159; September, 1954.) "If M is a measure of some property Q , and σ_M its standard deviation, the sensitivity of M , denoted ψ_M is defined by the relation $\psi_M = (dM/dQ)/\sigma_M$. It follows from this definition that the sensitivity of a test method may or may not be constant for all values of the property Q . A statistical test of significance is derived for the ratio of sensitivities of alternative methods of test. Unlike the standard deviation and the coefficient of variation, sensitivity is a measure of merit that is invariant with respect to any functional transformation of the measurement, and is therefore independent of the scale in which the measurement is expressed."

621.316.842(083.74)

797

Recent Development of Standard Resistors—A. Schulze. (*Elekrotech. Z., Edn A*, vol. 75, pp. 547-550; September 1, 1954.) The construction and manufacture of sealed-in standard resistors is described. The temperature coefficient of resistance has been reduced to 1×10^{-6} in 97.95 Au/2.05 Cr alloy resistors by heat pre-treatment, and to 10×10^{-6} in manganin-wire resistors. The change of resistance of the Au/Cr standards is of the order of 1×10^{-6} over a period of 3 to 5 years.

621.317.3.029.63:534.62

798

Construction of a Reflection-Free Room for Sound Waves and Decimetre Electrical Waves—Epprecht, Kurtze and Lauber. (See 612.)

621.317.33

799

A Two-E.M.F. Method for the Comparison of Resistances—H. J. Hoge. (*Rev. Sci. Instr.*, vol. 25, pp. 902-907; September, 1954.) Two sources of emf, each with an associated dropping resistor, are respectively connected in series with two resistors to be compared in such a way that the latter are not adjacent; the

circuit is adjusted so that the potentials at the corresponding ends of the two resistors are equal when they carry equal currents. The method permits continuous observation or recording without error due to uncontrolled change of resistance in the connections.

621.317.333.6:621.372.8

800

Electrical Breakdown in Waveguides at 3000 Mc/s—Sutherland. (See 640.)

621.317.335

801

Measurement of Permittivity and Tangent of Loss Angle of Dielectrics with High Absorption at Centimetre Wavelengths—E. Briganti. (*Poste e Telecomun.*, vol. 22, pp. 327-332; July, 1954.) A cavity-resonator method suitable for liquid specimens is described. Rigorous but simple formulas are derived for determining the permittivity and loss angle from measurements of wavelength and power.

621.317.337:621.372.413

802

Nomogram for Q of a Cavity—J. D. Harmer. (*Wireless Eng.*, vol. 32, pp. 25-27; January, 1955.) A nomogram is presented with the aid of which the Q factor can be determined from a small number of measurements of the standing waves in a feeder coupled to the cavity.

621.317.41/.42:621.395.625.3

803

Experimental Determination of [magnetic] Parameters of Magnetic-Recording Media—G. S. Veksler and P. S. Tomashevski. (*Zh. Tekh. Fiz.*, vol. 24, pp. 1594-1598; September, 1954.) Description, with circuit diagram, of cro apparatus for tracing $4\pi I/H$ curves at frequencies up to $>10^4$ cps. Photographs of typical hysteresis loops are presented including one showing also the initial part of the curve.

621.317.7:621.396.822:621.385

804

A Multichannel Noise Spectrum Analyzer for 10-1000 c/s—E. G. Nielsen and A. van der Ziel. (*Rev. Sci. Instr.*, vol. 25, pp. 899-902; September, 1954.) An instrument for rapid analysis of flicker noise in tubes is described. Values of either the equivalent saturated diode current or the equivalent noise resistance at 10, 30, 100, 300, 1,000, 3,000 and 10,000 cps are obtained simultaneously by feeding the signal into seven parallel channels each incorporating a wide-band RC filter and a thermistor square-law detector.

621.317.715:621.375.13.029.424

805

A Tuned Galvanometer Amplifier—J. R. Beattie and G. K. T. Conn. (*Rev. Sci. Instr.*, vol. 25, pp. 888-891; September, 1954.) Description of a system consisting of a galvanometer tunable by feedback over the range 1-3 cps, followed by an electronic amplifier covering the same frequency range (2950 of 1953), for amplifying signals from a radiation thermocouple used with a low-frequency interrupter. The output is presented as a rectified smoothed meter deflection. The amplification is sufficient to reveal thermal noise in the thermocouple circuit.

621.317.73

806

A Goniometer Quotient-Measurement Method for the Direct Determination of Admittance—H. Fricke. (*Funk u. Ton*, vol. 8, pp. 225-238 and 369-377; May and July, 1954.) Voltages proportional respectively to the current through the unknown impedance and the voltage across it are applied to the crossed coils of a goniometer, phase equalization being effected by a tuned circuit or a short-circuited transmission line across the unknown impedance. Calibration curves for 1-mc operation are given; error can be kept within a given limit, say 1 per cent, over different ranges by adjustment of series resistance and goniometer coupling; sensitivity is improved by a two-goniometer system. Circuits and goniometer arrangement for operation at uhf are described. A possible application of the

equipment with a cro for displaying locus curves is outlined.

621.317.733

807

Modern Bridge Techniques—P. M. Ratcliffe. (*Marconi Instr.*, vol. 4, pp. 167-175; September, 1954.) A concise review.

621.317.733:621.316.86:537.312.6

808

A Self-Balancing Thermistor Bridge—A. F. Standing. (*Jour. Sci. Instr.*, vol. 31, pp. 343-344; September, 1954.) A direct reading of power in the range 0.1-10 mw is obtained by including the thermistor in an oscillatory feedback circuit such that the thermistor resistance is held constant. The oscillator output is balanced against a direct voltage to give a zero meter reading in the absence of rf power.

621.317.74:621.315.212

809

R.F. Cable Characteristics measured with a Q -Meter—J. Shekel. (*Electronic Eng.*, vol. 26, pp. 540-542; December, 1954.) Theory and practical details are given for a method in which a Q -meter determination of the frequency at which a section of the cable becomes a $\lambda/2$ resonator enables the propagation velocity, the characteristic impedance and the attenuation of the cable to be computed. A numerical example is included.

621.317.74 + 621.317.772:621.397.5:535.623

810

Differential Phase and Gain Measurements in Color-Television Systems—Kelly. (See 871.)

621.317.742:621.3.018.756

811

Direct V.S.W.R. Readings in Pulsed R.F. Systems—L. A. Rosenthal and G. M. Badoyannis. (*Electronics*, vol. 27, pp. 162-165; December, 1954.) Development of the ratio meter previously described [466 of 1953 (Rosenthal et al.)] to deal with pulse systems such as radar. Thermionic diodes are used as detectors. Pulse stretching is found necessary; a suitable circuit is shown. Operating procedure is outlined and measurements with standard and other loads are reported.

621.317.742:621.315.212

812

A Coaxial Standing-Wave Indicator for Frequencies near 10,000 Mc/s—F. A. Benson and G. V. G. Lusher. (*Electronic Eng.*, vol. 26, pp. 534-537; December, 1954.) Details are given of the construction of a precision instrument.

621.317.75:631.396.3

813

Response of Radio Spectrometers to Non-periodic Morse Signals—J. Marique. (*Ann. Télécommun.*, vol. 9, pp. 215-223 and 247-255; July-September, 1954.) The analysis is presented for the same arrangement of cascaded tuned circuits as discussed in previous studies (e.g. 2034 of 1954); expressions are derived for the currents in these circuits. The envelope-response of the circuit is defined and the conditions are investigated for this response to be sufficiently independent of the circuit design for practical purposes.

621.317.755

814

Measurement of an Impedance by means of a Double-Trace Cathode-Ray Oscillograph: Coincidence Method—A. Grumbach. (*Compt. Rend. Acad. Sci. (Paris)*, vol. 239, pp. 869-871; October 11, 1954.) The unknown impedance is connected in series with a known impedance across the output terminals of a sine-wave generator, one of these terminals being earthed while the other is connected to one of the cro deflection systems. The second deflection system is connected to the junction between the impedances. The known impedance is varied to produce phase coincidence between the two curves; amplitude coincidence is produced by means of amplifiers. The sensitivity of the method is discussed in relation to meas-

urement of the permittivity and conductivity of a capacitor dielectric. Use of the arrangement for measuring frequency is also indicated. The frequency range is up to 100 kc.

621.317.755 815
Wide-Band Amplitude Distribution Analysis of Voltage Sources—L. W. Orr. (*Rev. Sci. Instr.*, vol. 25, pp. 894-898; September, 1954.) Signals to be analyzed are applied to a cro fitted with a slotted mask through which the luminescence output is fed to a photocell, the required amplitude distribution function being displayed on a second cro. The apparatus can be set up and operated quickly, and the accuracy is within 5 per cent. Analyses of two noise sources are shown.

621.317.755:621.385.3/5 816
Development of a Characteristic-Curve Tracer for Transmitting Valves—J. Kammerloher and H. Krebs. (*Funk u. Ton*, vol. 8, pp. 453-470; September, 1954.) Description of a dc-pulse cro suitable for dealing with positive grid voltages. Circuit diagrams and some details of the switching arrangements are included. Typical curves obtained are reproduced; these include I_a/V_a curves for I_a up to 2 A, V_a up to 1,000 v and grid voltages of 0, 60, ... 300 v. An ac-pulse instrument is briefly mentioned.

621.317.761:621.374.3 817
A Pulse-Interval Meter for measuring Pulse Repetition Frequency—A. M. Andrew and T. D. M. Roberts. (*Electronic Eng.*, vol. 26, pp. 469-474 and 543-547; November and December, 1954.) The instrument described is suitable for use in cases such as neurophysiological measurements where the pulse repetition frequency varies rapidly. The output voltage at any instant is determined either by the duration of the preceding pulse interval, or by the duration already attained by the current interval, whichever is the longer. Frequency is indicated on an approximately linear scale, either by oscilloscope or by moving-coil meter.

621.396.001.4 818
Prediction of Electronic Failures—(*Elect. Jour.*, vol. 153, pp. 717-718; September 3, 1954.) The N.B.S. experimental failure-prediction unit is based primarily on the detection of a decrease in tube transconductance in successive stages of the equipment under test, which requires slight modification to enable stages to be tested separately. Provision for checking capacitors for leakage, and for voltage and current measurements is also made.

OTHER APPLICATIONS OF RADIO AND ELECTRONICS

534.1-8:669 819
Metallurgical Effects of Ultrasonic Waves—E. A. Hiedemann. (*Jour. Acous. Soc. Amer.*, vol. 26, pp. 831-842; September, 1954.) A survey with 121 references.

534.88 820
Echo-Location for the Blind—C. M. Witcher and L. Washington, Jr. (*Electronics*, vol. 27, pp. 136-137; December, 1954.) Developments are described in devices of the type emitting high-frequency clicks which are reflected by obstacles. The sound projector is rotated through an angle of 60 degrees in a period of 0.7-1 second by means of a miniature motor. A model for attachment to clothing weighs $1\frac{3}{4}$ pound.

621.314.214.5:621.317.39 821
Differential Transformers for Mechanical Measurements—L. W. Blick. (*Jour. Brit. IRE*, vol. 14, pp. 603-610; December, 1954. Discussion, p. 611.) Circuits for use with this type of transformer are described and applications as transducers are indicated.

621.316.7 822
A Design Philosophy for Man-Machine

Control Systems—H. P. Birmingham and F. V. Taylor. (*Proc. I.R.E.*, vol. 42, pp. 1748-1758; December, 1954.) Methods are described for designing control systems so that the human operator is required to act only as a simple amplifier. Aided tracking is discussed in relation to efforts to improve the stability of man-machine systems by the use of special equalization networks.

621.316.718:534.85:621.94 823
Magnetic Tape controls Machine Tools—J. W. Hlogan. (*Electronics*, vol. 27, pp. 144-147; December, 1954.)

621.317.39 824
Three Electronic Thickness Gages for Metallic Coatings—(*Tech. News Bull. Nat. Bur. Stand.*, vol. 38, pp. 127-132; September, 1954.) Three instruments are described, in all of which operation depends on the difference in electrical conductivity between the plating and the support metal. The "dermitron" and the phase-angle thickness meter are electromagnetically coupled to the specimen and make use of the reflected field from eddy currents induced in the specimen. The "wageguide plating quantity indicator" makes a direct measurement of conductivity, using point electrodes.

621.317.39 825
Wire Strain-Gauge Transducers for the Measurement of Pressure, Force, Displacement, and Acceleration—J. L. Thompson. (*Jour. Brit. IRE*, vol. 14, pp. 583-600; December, 1954. Discussion, p. 600-601.) Construction, theory of operation, and applications are discussed.

621.317.39 826
Load-Cell Force Transducers—D. L. Johnston. (*Jour. Brit. IRE*, vol. 14, pp. 613-620; December, 1954. Discussion, p. 620.) A chart is presented indicating ranges of power level and conversion efficiency for various types of transducer; the strain-gauge load-cell type is particularly useful for dealing with large forces. Associated measuring circuits are discussed.

621.317.79:621.385:531.717.3 827
Electronic-Mechanical Transducers—L. A. Goncharski. (*Zh. Tekh. Fiz.*, vol. 24, pp. 1711-1723; September, 1954.) Theory and practical considerations are discussed of a sensitive electron-tube-type transducer consisting basically of a thin, electrically heated filament between a pair of plate electrodes, one of which is operated as anode. The two plates are rigidly attached by through leads to a glass pinch, flexibly sealed in the container wall, which picks up the displacements to be measured. Using a constant anode current of 1.5 ma, with >90 v on the anode and -10 v on the other plate, the voltage sensitivity is $\sim 6,000$ v per cm displacement of the plates. The current sensitivity is of the order of 0.1 a/cm-displacement. References to previous papers on the applications of this transducer are given.

621.365.55:674 828
Electronic Heating and the Woodworking Industry—M. T. Elvy. (*Jour. Brit. IRE*, vol. 14, pp. 547-566; November, 1954. Discussion, p. 567.) A review of hf dielectric heating methods with particular reference to the use of synthetic resin glues. The design of oscillation generators and coupling systems, jigs and electrodes is discussed.

621.384.612 829
Suppression of Coherent Radiation by Electrons in a Synchrotron—J. S. Nodvick and D. S. Saxon. (*Phys. Rev.*, vol. 96, pp. 180-184; October 1, 1954.)

621.384.612 830
Phase Oscillations in the Strong-Focusing Synchrotron—E. Bodendstedt. [*Ann. Phys. (Lpz.)*, vol. 15, pp. 35-54; September 15, 1954.]

Phenomena connected with the acceleration of particles beyond the critical-energy region in the strong-focusing synchrotron [1454 of 1953 (Courant et al.)] were investigated using a mechanical analog machine. Acceleration of particles beyond the critical energy can be accomplished by means of an odd number of jumps of the oscillator phase or by modulation of the oscillator voltage.

621.384.612 831
Project for an Electron Synchrotron in Italy—G. Salvini. (*Nuovo Cim.*, Supplement to vol. 12, pp. 77-100; 1954.) Plans are discussed in some detail for a 600-mev machine to be available for all nuclear physics work in Italy; the desirability of raising the energy to 1,000 mev is indicated.

621.384.612 832
Amplitudes of Oscillations in the Strong-Focusing Synchrotron—J. Seiden. [*Compt. Rend. Acad. Sci. (Paris)*, vol. 239, pp. 798-800; October 4, 1954.] Simple formulas are presented.

621.384.612 833
Effects on Orbits of Correlation between Lens Alignment Faults in the Synchrotron—J. Seiden. [*Compt. Rend. Acad. Sci. (Paris)*, vol. 239, pp. 966-968; October 18, 1954.]

621.384.622.1:537.533.9 834
Applications of High-Energy Electrons to the Sterilization of Pharmaceuticals and the Irradiation of Plastics—C. W. Miller. (*Jour. Brit. IRE*, vol. 14, pp. 637-652; December, 1954. Discussion, p. 652.) For irradiating the materials discussed, the available energy can be used more efficiently by electron bombardment than by X- or γ -rays. The required high-energy electrons should be obtained from a linear accelerator rather than a radioactive source. Practical and economic aspects of the use of linear accelerators are discussed. Over 50 references.

621.384.622.2 835
A Theory of Electron-Beam Loading in Linear Accelerators—G. Saxon. (*Proc. Phys. Soc.*, vol. 67, pp. 705-716; September 1, 1954.) Analysis is presented for the waveguide circuit of a linear accelerator using rf power feedback, to determine how the power flowing into such an accelerator varies with the beam loading. The relations derived are used in conjunction with formulas for the energy gain and beam power output of a length of accelerator waveguide to calculate the performance to be expected under feedback conditions as the beam current is varied. The calculated results are in reasonable agreement with measurements.

621.385.833 836
The Application and Limitations of the Edge-Diffraction Test for Astigmatism in the Electron Microscope—M. E. Haine and T. Mulvey. (*Jour. Sci. Instr.*, vol. 31, pp. 326-332; September, 1954.)

621.385.833 837
Rigorous Calculation for a Typical Electrostatic Unipotential Lens—W. Glaser and P. Schiske. [*Optik (Stuttgart)*, vol. 11, nos. 9 and 10, pp. 422-443 and 445-467; 1954.]

621.385.833 838
Centering of Magnetic Electron Lenses—S. Leisegang. [*Optik (Stuttgart)*, vol. 11, no. 9, pp. 397-406; 1954.]

621.387.4 839
The Reliability of Nucleonic Instruments—D. Taylor. (*Jour. Brit. IRE*, vol. 14, no. 11, pp. 570-580; November, 1954. Discussion, p. 580.) The servicing of nucleonic instruments under factory conditions is discussed; problems encountered in the British Atomic Energy Project are considered. Annual failure rates of some of the standard instruments are given; these are highest for the more orthodox com-

ponents, such as tubes and resistors. U. S. and Canadian figures are also quoted. Design principles are derived from the analyses presented.

- 621.387.424 840
Secondary Electron Emission by Photoelectric Action and Ion Bombardment at the Cathode in Corona Breakdown of Argon—L. Colli and U. Facchini. (*Phys. Rev.*, vol. 96, pp. 1-4; October 1, 1954.) Discussion of secondary-emission mechanisms in cylindrical argon-filled counter tubes with positive axial wire.

- 621.398.029.62:525.6 841
Radio-Linked Unattended Tide Gauge for Persian Gulf—[*Engineer (London)*, vol. 198, p. 305; August 27, 1954.] A 27-w vhf transmitter is keyed by a series of half-second pulses generated by a balance wheel system; the number of pulses is controlled by a plunger system operated by a tide drum and gives a coded indication of tide height over a range -1 to +13 feet. The transmitter is supplied by NiFe cells charged automatically. The output of a receiver 12 miles away is fed to an integrator which operates an indicator scaled in divisions of 0.1 foot.

PROPAGATION OF WAVES

- 538.566:551.510.535 842
Focusing Phenomena due to Undulations of the Ionosphere, and Determination of Collision Number—K. Rawer and É. Argence. [*Compt. Rend. Acad. Sci. (Paris)*, vol. 239, pp. 1066-1067; October 27, 1954.] An approximate formula is given for the amplitude of echoes for a reflecting surface with sinusoidal undulations. The effect of the curvature becomes preponderant near the focus state $r = \rho h_0$, where r is the radius of curvature, ρ the order of the echo, and h_0 the height of the ionized layer. The parameters involved can be calculated from the amplitudes of three successive echoes, most conveniently observed at night. The curvature effect is useful for re-interpreting old observations giving reflection coefficients greater than unity. Adopting the hypothesis that collisions are mostly between electrons and neutral molecules, a collision number of about 400 is found for the middle of the F_2 layer during winter nights.

- 621.396.11+621.396.9+523.16 843
Propagation of Electromagnetic Waves, Radio Location and Radio Astronomy—E. Roessler. (*Elektrotech. Z., Edn. A*, vol. 75, pp. 632-635; September 11, 1954.) A brief survey of recent progress, particularly since 1951. 73 references.

- 621.396.11:551.510.5 844
Central Radio Propagation Laboratory of the N.B.S.—[*Engineer (London)*, vol. 198, pp. 401-404; September 17, 1954.] See 3018 of 1954.

- 621.396.11.029.62:551.51 845
A Study of Some of the Meteorological Effects on Radio Propagation at 96.3 Mc/s between Richmond, Va. and Washington, D. C.—D. L. Randall. (*Bull. Amer. Met. Soc.*, vol. 35, pp. 56-59; February, 1954.) "For meteorological observations during which the wind speeds were equal to or greater than 10 m.p.h., and when fronts, low overcast clouds (less than 5,000 feet), rain, thunderstorms and fogs were excluded, a 0.70 correlation coefficient was found between hourly surface refractive index and hourly median field strength."

- 621.396.81 846
Theoretical Field Strengths and Angles of Incidence of WWV Transmissions at the Châtonnaye Receiving Station—C. Glinz. (*Tech. Mitt. Schweiz. Telegr.-Teleph. Verw.*, vol. 32, pp. 253-267; July 1, 1954. In German.) German version of paper abstracted in 3668 of 1954.

- 621.396.81 847
Some Observations on Rayleigh Fading—B. van der Pol. (*Tijdschr. ned. Radiogenoot.*, vol. 19, pp. 223-229; September, 1954. In English.) Fading due to multipath propagation is discussed on the basis of the random-walk problem. A formula is presented from which (a) the most probable, (b) the median, and (c) the mean signal amplitude is evaluated. The method of obtaining the Rayleigh distribution curve from a typical fading record is shown. The distinction between Rayleigh fading and fading due to variation of ionospheric absorption is emphasized.

- 621.396.812.3.029.53:551.510.535 848
Periodic Fading of Medium-Wave Radio Signals—Rao and Sarma. (See 714).

RECEPTION

- 621.376.232.2.015.7 849
Pulse Response of Signal Rectifiers—M. V. Callendar. (*Wireless Eng.*, vol. 32, pp. 3-14; January, 1955.) An examination is made of the loading imposed by a diode rectifier on a tuned circuit from which it is fed. The response time for the circuit together with the diode is investigated, both directly and via the modulation-frequency response characteristic. Measurements are reported confirming the theory for the simplest case of a single circuit centrally tuned and yielding information on the response with more complex wideband amplifiers. Comparison is made with the performance of a triode detector. The most important practical conclusion is that no advantage is gained by using a shunt resistor across the circuit to provide damping.

- 621.396.62(43) 850
New German Broadcast Receivers—(I. Limann. (*Elektrotech. Z., Edn. B*, vol. 6, pp. 347-351; September 21, 1954.) Improved sensitivity, selectivity and fidelity are features of the 1954 models. The triode Type-ECC85 is used almost universally as low-noise USW first-stage tube; the medium-slope Type-EF89 pentode has advantages for the IF stage. Distortion in the af stages has been reduced by two-channel amplification and other methods; side loudspeakers are used in some models.

- 621.396.621+621.37:621.314.7 851
Some Transistor Circuits—van Overbeek. (See 650.)

- 621.396.621.54 852
A.M./F.M. Communications Receiver—(*Wireless World*, vol. 61, pp. 41-43; January, 1955.) Description and performance report of a commercial model with continuous tuning over the range 19-165 mc in six bands. The IF is 5.2 mc. Rf, oscillator and mixer stages form a single unit; the six-position rotary-coil turret, three-gang split-stator capacitors and other components are arranged so as to minimize the amount of wiring.

- 621.396.821:519.2 853
Study of Statistical Models suggested by Consideration of the Effects of Atmospherics on Amplifiers—A. Blanc-Lapierre, M. Savelli and A. Tortrat. (*Ann. Télécommun.*, vol. 9, pp. 237-245; September, 1954.) An analysis is made of the output of a linear amplifier to whose input is applied (a) Gaussian noise with uniform spectrum, and (b) random pulses with a Poisson distribution in time. Consideration is given to the limiting condition when the density of the Poisson distribution is very high. A method of calculating the instantaneous amplitude distribution is developed. The case of a train of very short pulses is dealt with, and properties of the detected voltage are discussed.

- 621.396.621 854
Radio Receiver Design—Part I [Book Review]—K. R. Sturley. Publishers: Chapman &

Hall, London, Eng., 2nd ed., 667 pp., 58s. (*Jour. Sci. Instr.*, vol. 31, p. 347; September, 1954.) Revised and augmented to a length about 50 per cent greater than that of the first edition. "The book should be a useful and reliable guide to those concerned with circuits for radio frequencies or with the characteristics of radio receivers."

STATIONS AND COMMUNICATION SYSTEMS

- 621.376.56 855
Experimental Model for Pulse-Number Modulation—J. Holzer, G. Missriegler, E. Niedermayr, H. Putsch and H. Zemanek. (*Öst. Z. Telegr. Teleph. Funk Fernschtech.*, vol. 8, pp. 125-132; September/October, 1954.) Description of equipment installed at the Technische Hochschule at Vienna. The principle of operation is to convert the lf signal into width-modulated pulses which are in turn converted into groups containing corresponding numbers of narrow equal-width pulses; these are counted in a binary system and the results give the code group. 32-step quantization is used. Decoding is performed by causing the received pulses to charge a capacitor which is discharged through a resistor, the time constant of the system being made equal to $t_0/\log 2$, where t_0 is the time spacing of the pulses in the code group.

- 621.376.56 856
Reception of Code-Modulation Signals by Integration—H. Harnuth. (*Fernmeldetechn. Z.*, vol. 7, pp. 461-464; September, 1954.) See 2505 of 1954.

- 621.39:621.372.8 857
Waveguide as a Communication Medium—Miller. (See 641.)

- 621.39.001.11 858
1954 Symposium on Information Theory—(*Trans. I.R.E.*, no. PGIT-4, pp. 1-227; September, 1954.) The text is given of the following papers presented at the symposium held at the Massachusetts Institute of Technology in September, 1954:

- "A New Basic Theorem of Information Theory,"—A. Feinstein (pp. 2-22).
 "Binary Coding,"—M. J. E. Golay (pp. 23-28).
 "Error-Free Coding,"—P. Elias (pp. 29-37).
 "A Class of Multiple-Error-Correcting Codes and the Decoding Scheme,"—I. S. Reed (pp. 38-49).
 "Coding for Constant-Data-Rate Systems,"—R. A. Silverman and M. Balser (pp. 50-63).
 "Information, Organization and Systems,"—J. Rothstein (pp. 64-66).
 "An Information-Theoretical Model of Organizations,"—M. Kochen (pp. 67-75).
 "Simulation of Self-Organizing Systems by Digital Computer,"—B. G. Farley and W. A. Clark (pp. 76-84).
 "A Study of Ergodicity and Redundancy based on Intersymbol Correlation of Finite Range,"—S. Watanabe (pp. 85-92).
 "Multivariate Information Transmission,"—W. J. McGill (pp. 93-111).
 "Choice and Coding in Information Retrieval Systems,"—C. N. Moore (pp. 112-118).
 "Modern Statistical Approaches to Reception in Communication Theory,"—D. Van Meter and D. Middleton (pp. 119-145).
 "A Nonlinear Prediction Theory,"—R. F. Drenick (pp. 146-162).
 "The Detection of Signals perturbed by Scatter and Noise,"—R. Price (pp. 163-170).
 "The Theory of Signal Detectability,"—W. W. Peterson, T. G. Birdsall and W. C. Fox (pp. 171-212).
 "The Human Use of Information: Part 1—Signal Detection for the Case of the Signal Known Exactly,"—W. P. Tanner, Jr., and J. A. Swets (pp. 213-221).
 "The Human Use of Information: Part 2—

Signal Detection for the Case of an Unknown Signal Parameter,"—W. P. Tanner, Jr., and R. Z. Norman (pp. 222-227).

621.39.001.11 859

Note on a Theorem of Shannon—S. De Francesco. (*Ann. Geofis.*, vol. 7, pp. 195-207; April, 1954.) The formula given in Shannon's "sampling theorem" (1649 of 1949) is shown to correspond to complete convergence of a generalized Fourier-series expansion. An expression is derived for the inherent error

621.395.44:621.375.2 860

Line Amplifiers for Symmetrical Carrier Frequency Cables—F. Feil. (*Fernmeldetechn. Z.*, vol. 7, pp. 454-460; September, 1954.) Amplifier types V12, V60 and V120 for the German Post Office system are described. As a result of advances in design, the V120 is smaller and consumes less power than the V12. A comparison table of the most important data is presented.

621.396.3 861

Predicted-Wave Radio Teleprinter—M. L. Doelz. (*Electronics*, vol. 27, pp. 166-169; December, 1954.) The "predicted-wave" system is a particular form of frequency-shift. The detection circuits accumulate over each pulse period the signal and noise from each of the two frequency channels in high-Q magnetostrictive resonators; a mark or a space is registered according as the accumulated amplitude is greater in the mark channel or in the space channel. A synchronizing signal is transmitted on a frequency of 23.04 kc, midway between mark and space frequencies. Performance figures are given; an error rate of 0.1 per cent was obtained with signals 6 db below noise level in the IF band; this constitutes a considerable improvement over the performance of a conventional frequency-shift system.

621.396.3:621.317.75 862

Response of Radio Spectrometers to Non-periodic Morse Signals—Marique. (See 813.)

621.396.712+621.397.743 863

Plans for F.M.-Radio and Television Networks in Denmark—G. Pedersen. [*Teleteknik* (Copenhagen), vol. 5, pp. 220-230; July, 1954.]

621.396.932 864

Radio Communication in the Merchant Marine—W. E. Steidle (*Elektrotech. Z.*, *Edn A*, vol. 75, pp. 584-587; September 11, 1954.) A brief account is given of the use made of the several frequency bands allocated to shipping, and of modern radio equipment used at sea.

621.39.001.11 865

Information Theory [Book Review]—S. Goldman. Publishers: Syracuse University, Prentice-Hall Inc., New York, N.Y., 1953, 385 pp., \$9.00. (*Electronics*, vol. 27, pp. 360-362; December, 1954.) Intended for graduate students in electrical engineering. The treatment is based on Shannon's work; liberal use is made of examples.

SUBSIDIARY APPARATUS

621.314.63:546.28 866

Silicon Power Rectifier handles 1200 Watts—E. F. Losco. (*Electronics*, vol. 27, pp. 157-159; December, 1954.) A *p-n* fused-junction Si-Al rectifier with a junction area of 0.05 cm² is mounted in a copper radiator with a slotted periphery. High power-handling capacity is obtained by use of forced-air cooling. Characteristic curves are shown.

621.316.722 867

Direct-Voltage Stabilizers in the Range 10-100 kV with Particular Reference to Degenerative Systems—M. W. Jervis. (*Jour. Brit. IRE*, vol. 14, pp. 629-636; December, 1954.) The maximum usable loop gain depends on the

frequency response of the system; in practice, loop gains of the order of 1,000 are possible, i.e. the effects of mains-voltage and load-current fluctuations are reduced by this factor. Reference elements are reviewed; wire-wound resistance potential dividers are the most stable, but electron-energy analysers give comparable stability with smaller size and current drain.

621.316.722.1 868

A 2 kVA A.C. Voltage Stabilizer—R. G. Ackland. (*Aust. Jour. Instr. Tech.*, vol. 10, pp. 98-101; August, 1954.) A motor-operated variac type of stabilizer is described, in which the voltage-sensing unit contains only one tube, of cold-cathode type, and incorporates biasing to hold the output voltage near the center of the control range. The output is maintained at 230 ± 1 per cent (or ± 1 v if required). The correction rate is 10 v/second.

TELEVISION AND PHOTOTELEGRAPHY

621.397.5 869

Television Engineering [in Western Germany]—F. Kirschstein. (*Elektrotech. Z.*, *Edn A*, vol. 75, pp. 638-640; September 11, 1954.) Brief survey of studio and industrial techniques, transmitters, receivers and microwave and cable links. 55 references.

621.397.5:061.3 870

Technical Conference of West German Broadcasting Authorities held at Munich, 24th-28th May 1954—(*Tech. Hausmitt. NordwDtsch. Rdfunks*, vol. 6, pp. 149-176; 1954.) Summaries are given of 29 papers dealing with various aspects of television. Subjects include transmission, studio equipment and techniques, television links, and measuring equipment.

621.397.5:535.623:[621.317.74+621.317.772 871

Differential Phase and Gain Measurements in Color-Television Systems—H. P. Kelly. [*Elec. Eng. (New York)*, vol. 73, pp. 799-802; September, 1954.] Portable apparatus suitable for color-carrier-frequency measurements in N.T.S.C. systems is described with the aid of block diagrams. Simplified circuit diagrams are also given of sections of the test transmitter and receiver.

621.397.5:535.767 872

Stereo-Television—H. Dewhurst. (*Jour. Telev. Soc.*, vol. 7, pp. 279-285; July/September, 1954.) A review of possible methods. Only those involving viewer aids appears to be practicable. 33 references.

621.397.61 873

An Experimental Camera Circuit for 405 Lines—C. H. Banthorpe. (*Jour. Telev. Soc.*, vol. 7, pp. 300-303; July/September, 1954.) A simple camera for scanning still pictures and captions makes use of a Type-5527 iconoscope, with es focusing and deflection. Brief descriptions are given of the vision pre-amplifier, black-level clamp, spurious-signal remover, picture and synchronizing-signal mixer, blanking-pulse amplifier, scanning circuits and power supply.

621.397.62 874

Frame Flyback Suppression—W. T. Cocking. (*Wireless World*, vol. 61, pp. 33-35; January, 1955.) An explanation is given of the fact that the flyback trace is commonly visible at the receiver notwithstanding that the form of the television signal is designed to render this part of the trace invisible. Circuits for deriving a suppression pulse from the frame timebase are described, this pulse can be applied, with appropriate polarity, to either grid or cathode of the picture tube.

621.397.62:621.397.335 875

A Critical Review of Synchronizing Sepa-

rators with Particular Reference to Correct Interlacing—G. N. Patchett. (*Jour. Brit. IRE*, vol. 14, p. 621; December, 1954.) Discussion on 2527 of 1954.

621.397.621.2:621.385.832 876

Transfer Characteristics and Mu Factor of Picture Tubes—H. Moss: K. Schlesinger. (*Proc. I.R.E.*, vol. 42, p. 1809; December, 1954.) Comment on 2171 of 1953 and author's reply.

621.397.7 877

The Television Centre at Hamburg-Lokstedt—E. Schwartz. (*Fernmeldetechn. Z.*, vol. 7, pp. 468-472; September, 1954.) A brief description. For a detailed account see *Tech. Hausmitt. NordwDtsch. Rdfunks*, vol. 5, nos. 7/8; 1953.

621.396.712+621.397.74.3 878

Plans for F.M.-Radio and Television Networks in Denmark—G. Pedersen. [*Teleteknik* (Copenhagen), vol. 5, pp. 220-230; July, 1954.]

TRANSMISSION

621.396.61:621.396.3 879

Keying V.L.F. Transmitters at High Speed—M. I. Jacob and H. N. Brauch. (*Electronics*, vol. 27, pp. 148-151; December, 1954.) Naval communication technique is discussed; transmitters with power ranging from 250 kw to 1 mw and operating at frequencies from 15 to 35 kc are used. High-speed frequency-shift keying is made possible by varying the resonance frequency of the antenna in synchronization with the signal-frequency variation. This is done by connecting a saturable reactor across the antenna. Successful teleprinter transmissions have been made over distances > 5,000 miles, using an output of 450 kw.

TUBES AND THERMIONICS

621.314.632:546.289 880

Transient Phenomena in the Backward Direction of Germanium Crystal Rectifiers—M. Kikuchi and Y. Tarni. (*Jour. Phys. Soc. Japan*, vol. 9, pp. 642-644; July/August, 1954.) Theory based on the heating process involved is developed to account for the step variation of the current observed in point-contact rectifiers on sudden application of reverse voltage. The step occurs at the instant when the contact temperature reaches a critical value, estimated to be about 80 degrees C. This result is supported by observations.

621.314.632:546.289:537.312.6 881

An Additional Observation on Thermal Effects in Point Contact Rectifiers—H. L. Armstrong (*Jour. Appl. Phys.*, vol. 25, p. 1345; October, 1954.) Addendum to 1242 of 1954.

621.314.7 882

The Effect of a Transverse Electric Field on Carrier Diffusion in the Base Region of a Transistor—J. S. S. Kerr, J. S. Schaffner and J. J. Suran. (*Jour. Appl. Phys.*, vol. 25, pp. 1293-1297; October, 1954.) Analysis is presented for a junction transistor with parallel emitter and collector junctions and a field applied across the base in a direction parallel to the junctions. An expression is obtained for the current gain as the sum of terms each having the same form as that found by Steele (881 of 1953) for the one-dimensional case.

621.314.7 883

Expression for the "α Cut-Off" Frequency in Junction Transistors—D. Haneman. (*Proc. I.R.E.*, vol. 42, pp. 1808-1809; December, 1954.) The expression $\omega_c = \kappa DM/W^2$, where DM is the diffusion constant for minority carriers in the base region of width W , has been used by a number of workers, and various values have been proposed for κ . A method of solution is indicated for nonzero values of W/L_m , where L_m is the diffusion length; in this case κ is a

slowly varying function of W/I_M , with a value around 2.5. Results obtained are in qualitative agreement with observations, in particular the prediction that ω_c is more sensitive to variations of collector voltage if ω_c is initially high.

621.314.7:621.318.57 884

Large-Signal Behavior of Junction Transistors—J. J. Ebers and J. L. Moll. (Proc. I.R.E., vol. 42, pp. 1761-1772; December, 1954.) Analysis relevant to the use of transistors for switching is based on recognition of three distinct operating conditions defined by Anderson (652 of 1953) and corresponding to the on, off, and transition states. Expressions for the impedance in the open and in the closed state are derived in terms of easily measurable transistor parameters. The influence on switching time of the alpha cut-off collector capacitance and minority carrier storage is considered.

621.314.7:621.318.57 885

Large-Signal Transient Response of Junction Transistors—J. L. Moll. (Proc. I.R.E., vol. 42, pp. 1773-1784; December, 1954.) Analysis is based on the three distinct operation states defined previously (884 above). A calculation is made of carrier storage time, or time required for the operating point to move from the collector-current-saturation state to the intermediate state. The alpha cut-off frequency ω_α is the most important parameter affecting switching speed. It is possible with moderate driving current to switch the operating point from collector-current-cut-off to collector-current-saturation in a period of the order of $3/\omega_\alpha$ sec; to permit switching at this speed in the opposite direction, carrier storage effects must be avoided.

621.314.7.002.2 886

Manufacturing Grown Junction Transistors—F. H. Bower. (Electronics, vol. 27, pp. 130-134; December, 1954.) A step-by-step account of the procedure.

621.385.029.6 887

Some Recent Advances in Microwave Tubes—J. R. Pierce. (Proc. I.R.E., vol. 42, pp. 1735-1747; December, 1954.) A review covering high-power klystrons, double-tuned circuits for reflex klystrons, reduction of noise in traveling-wave tubes by velocity-jump or space-charge-wave de-amplification, periodic magnetic focusing of the beam, and the backward-wave oscillator. 23 references.

621.385.029.6 888

The Wave Picture of Microwave Tubes—J. R. Pierce. (Bell. Sys. Tech. Jour., vol. 33, pp. 1343-1372; November, 1954.) The low-level operation of long-beam tubes such as klystrons, resistive-wall amplifiers, cavitrons, space-charge-wave amplifiers, traveling-wave tubes and double-stream amplifiers is discussed in terms of the waves propagated.

621.385.029.6 889

Equations for the Oscillations in Uniform Electron Beams—Yu. A. Katsman. (Zh. Tekh. Fiz., vol. 24, pp. 1359-1369; July, 1954.) Addendum to 3144 of 1953.

621.385.029.63/64 890

Focusing of a Long Cylindrical Electron Stream by means of Periodic Electrostatic Fields—Ping King Tien. (Jour. Appl. Phys., vol. 25, pp. 1281-1288; October, 1954.) Space-periodic fields produced by a bifilar helix or a series of annular rings are considered. The potential distribution is given in the form of a power series and the equation of electron motion is derived. The solution indicates that the flow is essentially parallel provided the electrons have low transverse velocity on entering the focusing structure and are distributed transversely so that the transverse distribution of space-charge field is similar to that of the

focusing field. Numerical examples relevant to the design of traveling-wave tubes are presented. The use of a bifilar helix to provide both retardation and focusing is discussed.

621.385.029.63 891

Scalloped Beam Amplification—T. G. Mühran. (Jour. Appl. Phys., vol. 25, p. 1341; October, 1954.) Brief account of experiments on a tube in which electron bunching is produced by the usual control grid close to the cathode, to which a 1-kinc signal is applied, and alternate debunching and bunching occur along the beam under the influence of a longitudinal magnetic field only. Rf power gain and beam diameter are plotted against beam drift distance. An over-all gain of 24 db was measured, of which 10.6 db resulted from the "scallop-beam" amplification.

621.385.032.216:621.396.822 892

On the Flicker Noise Generated in an Interface Layer—H. J. Hannam and A. van der Ziel. (Jour. Appl. Phys., vol. 25, pp. 1336-1340; October, 1954.) The equivalent noise resistance R_n of a tube due to the oxide-cathode interface layer is proportional to the square of the interface resistance R_i and inversely proportional to f^α , where f is the frequency and α ranges from 1.1 for low values of R_i to 1.5 for high values of R_i , at af. Experiments are reported proving that the noise originates in the interface layer. Measurements at hf indicate an increase in the value of α .

621.385.2:621.376.232.2.029.63/64 893

Mechanism of Rectification in Vacuum-Tube Diodes at Microwave Frequencies—G. Papp. (Elec. Commun., vol. 31, pp. 215-219; September, 1954.) The equations of motion of the electrons are derived. Calculation of the diode current indicates that appreciable rectification is obtained; this is confirmed by measurements. Similar work has been reported by Bronwell et al. (3096 of 1954).

621.385.2.032.216 894

Effects of Cathode and Anode Resistance on the Retarding-Potential Characteristics of Diodes—G. C. Dalman. (Jour. Appl. Phys., vol. 25, pp. 1263-1267; October, 1954.) Anomalous retarding-potential characteristics of diodes with oxide cathodes are explained by taking account of high-resistance layers at cathode and anode; an improvement is thereby obtained in the accuracy of estimating the cathode temperature.

621.385.3 895

Amplification Factor and Perveance of an Elliptic Triode—S. Deb and G. S. Sanyal. (Jour. Appl. Phys., vol. 25, pp. 1196-1203; September, 1954.) Expressions for the amplification factor μ , interelectrode capacitance, and perveance P of the triode are derived using a conformal transformation to reduce the elliptic geometry to plane geometry. Results indicate that μ depends on the parametric angle θ of the ellipse and that the average value of μ increases as the grid eccentricity increases and as the anode eccentricity decreases. Curves are presented for finding μ when the value of θ is known. The value of P depends mainly on the grid eccentricity and the focal distance. The theory is illustrated by considering the design of a Type 6CS-GT/G tube.

621.385.3/.5:621.317.755 896

Development of a Characteristic-Curve Tracer for Transmitting Valves—Kammerloher and Krebs. (See 816.)

621.385.3.029.63 897

The Design of Triodes for U.H.F. Medium-Level Power Amplifiers—W. E. Rowlands. (Electronic Eng., vol. 26, pp. 522-527; December, 1954.) Tubes with parallel-plane electrode systems are considered, for operation at about

2 kmc with output of about 10 w. An expression is derived for radiant-heat dissipation in the grid, and an estimate is made of the influence of the heat-reflection coefficients of the electrodes. Particular attention is devoted to problems arising from cathode evaporation in conjunction with the close spacing of electrodes. A relation between evaporation rate and increase of grid-wire diameter is derived, and the effect of grid growth on tube operating parameters and on tube life is investigated. A life of about 10,000 h appears possible.

621.385.5:621.396.822 898

Partition Components of Flicker Noise—T. B. Tomlinson. (Jour. Brit. IRE, vol. 14, pp. 515-526; November, 1954.) In a pentode, the reduction of flicker noise obtained by operating under space-charge-limited conditions (281 of 1953) is to some extent canceled as a result of the partition of the current between anode and screen grid. Noise measurements made with the tube connected (a) normally and (b) as a triode confirm the existence of this partition component and provide information on the origin of the flicker noise. Defects giving rise to excessive noise at low frequencies were encountered in standard type tubes examined.

621.385.832 899

Dark-Trace Display Tube has High Writing Speed—S. Nozick, N. H. Burton and S. Newman. (Electronics, vol. 27, pp. 154-156; December, 1954.) A cathode-ray tube is described in which a high value of beam current for a given spot size, and hence high writing speed, is obtained by using an auxiliary focusing system to reduce the beam diameter in the region of the deflection coils. Writing speeds up to 15 km have been attained.

621.387 900

A Magnetic Gas-Discharge-Tube Oscillator—J. M. Somerville. (Jour. Sci. Instr., vol. 31, pp. 279-284; August, 1954.) If a glow discharge is established between an axial cylindrical anode and an outer cylindrical cathode split transversely into two differently biased sections, then, in the presence of a magnetic field, the conductance between the two cathode sections will, under particular operating conditions, be negative. The effects of various factors on the conductance and the efficiency of the tube as an oscillator are discussed. The frequency range is up to about 50 kc, and efficiencies up to 70 per cent are attainable. A mercury-vapor split-cathode tube suitable for commercial production is also described.

621.314.7 901

Transistoren [Book Review]—M. J. O. Strutt. Publishers: S. Hirzel Verlag, Zurich and Stuttgart, 1954, 166 pp., DM21. (Arch. elekt. Übertragung, vol. 8, pp. 371-372; August, 1954.) "... the main emphasis of the book is on the technical application of transistors."

MISCELLANEOUS

061.4:621.3 902

Ninth Annual Electronics Exhibition, Manchester—(Instrum. Practice, vol. 8, pp. 703-711; August, 1954.) Illustrated account of the exhibition held in July, 1954.

621.38+621.39 903

Fortschritte der Hochfrequenztechnik, Band 3. [Book Review]—F. Vilbig and J. Zenneck (Eds.). Publishers: Akademische Verlagsgesellschaft Geest und Portig K.-G., Leipzig, 1954, 718 pp., DM49. (Frequenz, vol. 8, p. 198; June, 1954.) Progress reports include surveys on radio-propagation conditions in various wavebands, the sun and ionosphere, traveling-wave tubes, radio-interference suppression, receiver engineering, frequency modulation, etc. A 36-page index is included.

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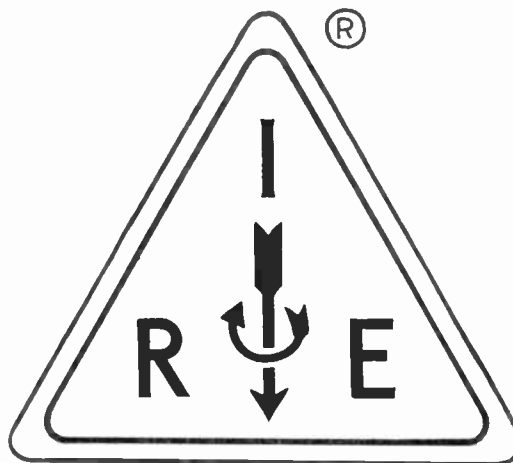
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TABLE OF CONTENTS

General Information.....	Cover II	Part 7—Broadcasting and Television Re-	
Contents of Volume II.....	Page 3	ceivers.....	Page 5
Part 1—Antennas & Propagation.....	Page 3	Part 8—Communication and Microwave....	Page 6
Part 2—Circuit Theory.....	Page 3	Part 9—Medical and Nuclear Electronics...	Page 6
Part 3—Electron Devices, Component Parts	Page 3	Part 10—Instrumentation, Industrial Elec-	
Part 4—Electronic Computers, Information		tronics.....	Page 6
Theory.....	Page 4	Part 11—Engineering Management, Quality	
Part 5—Aeronautical Electronics and Telem-		Control.....	Page 7
etry.....	Page 4	Index to Authors.....	Page 7
Part 6—Audio and Ultrasonics.....	Page 5	Index to Subjects.....	Page 8

GENERAL INFORMATION

The Institute

The Institute of Radio Engineers serves those interested in radio, allied electronics, and communications fields through the presentation of technical material, and by the monthly publication of PROCEEDINGS OF THE IRE, a technical journal. The Institute also publishes IRE Standards, and a number of Professional Group Publications, as well as a Convention Record.

Membership has grown from a few dozen in 1912 to more than 42,000 in 1955. There are several grades of membership, depending on the qualifications of the applicant, with dues ranging from \$5.00 per year for Students to \$15.00 per year for Members, Senior Members, Fellows, and Associates of more than five years' standing.

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To serve more fully the many special fields of interest, Professional Groups have been formed in each of 23 technical fields on an Institute-wide basis, with membership open only to IRE members.

Group activities include the sponsoring of symposia and conferences, and the publication of TRANSACTIONS OF THE IRE, journals containing technical material on special interest to Group members. To support Group activities, assessment fees may be levied by the Groups on their members.

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The PROCEEDINGS has been published without interruption from 1913, when the first issue appeared. Over 5,300 technical contributions have been included in its pages, portraying a currently written history of developments in both theory and practice. The contents of papers are the responsibility of the authors and are not binding on the Institute or its members. All rights of republication are reserved by the Institute.

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TRANSACTIONS have been published at intervals by Professional Groups since December, 1951, when the first issue appeared. A total of nearly 900 specialized technical papers have thus been made available to Group members during the first year of publication. All papers published in the TRANSACTIONS are procured and reviewed by the respective Professional Group. The contents of papers are the responsibility of the authors and are not binding on the Institute, the Groups, or their members. All rights of republication are reserved by the Institute. The annual contents of the TRANSACTIONS are published in a separate index. The TRANSACTIONS are sent free to paid members of the respective Groups.

Subscription rates covering all issues of TRANSACTIONS published during the twelve-month period starting July 1, 1954, for the United States, its possessions, and Canada, \$100; to colleges and public libraries, \$75.

IRE STANDARDS

IRE Standards published in the PROCEEDINGS OF THE IRE from time to time, are available in reprint form for those who wish to buy them from the Institute.

CONVENTION RECORD

In 1953, for the first time, the Institute published the first volume of the Convention Record of the IRE. In 1954, a second Convention Record was issued in eleven parts, comprising an almost complete record of the hundreds of papers presented at the annual national Convention. The subject matter of each part is devoted to a specialized field of electronics.

All members of the IRE Professional Groups receive a copy gratis of that part of the Record that contains papers within the member's specialized field of interest. Any one part of the Convention Record, or all 11 parts, can be purchased from the Institute.

CONTENTS OF CONVENTION RECORD OF THE I.R.E. VOL. 2—1954

1954 CONVENTION RECORD OF THE I.R.E.

Part 1—Antennas and Propagation

Cumulative Index Number	Page
SESSION 30: Antennas and Propagation I—General (Sponsored by the Professional Group on Antennas and Propagation.)	
214. Empirical Approximations to the Current Values for Large Dolph-Tchebycheff Arrays, <i>Louis L. Bailin, Robert S. Wehner, and Ivan P. Kaminow</i>	3
215. Gain Pattern of a Terminated-Waveguide Slot Antenna by an Equivalent Circuit Method, <i>Leopold B. Felsen</i> ...	10
216. A Four Slot Cylindrical Antenna for VOR Service, <i>Andrew Alford and R. M. Sprague</i>	12
217. Trapped Wave Antennas, <i>Herman Ehrenspeck, Werner Gerbes, and Francis J. Zucker</i>	25
218. Scattering of Electromagnetic Waves by Wires and Plates, <i>J. Weber</i>	31
SESSION 37: Antennas and Propagation II—Microwave Antennas (Sponsored by the Professional Group on Antennas and Propagation.)	
219. Reflections in Microwave Antennas and Their Harmful Effects, <i>Peter W. Hannan</i>	39
220. Surface Matching of Dielectric Lenses, <i>E. M. T. Jones and S. B. Cohn</i>	46
221. Double Parabolic Cylinder Pencil Beam Antenna, <i>Roy C. Spencer, F. Sheppard Holt, Helen M. Beauchemin, and John L. Sampson</i>	54
222. Diffuse Radiation in Pencil Beam Antennas, <i>David Carter</i>	60
223. Theoretical Gain of Flat Microwave Reflectors, <i>D. R. Crosby</i>	71
SESSION 42: Antennas and Propagation III (Sponsored by the Professional Group on Antennas and Propagation.)	
224. Isotropic Variable Index Media, <i>W. O. Puro and K. S. Kelleher</i>	76
225. The Characteristics of a Vertical Antenna with a Radial Conductor Ground System, <i>James R. Wait and W. A. Pope</i>	79
226. Some Information Theory Aspects of Propagation Through Time Varying Media, <i>Joseph Feinstein</i>	87
227. Comparative 100 Mc Measurements at Distances Far Beyond the Radio Horizon, <i>Albrecht P. Barsis</i>	98
228. The Measurement of the Polarization of Radio Waves Reflected from the Ionosphere at Non-Vertical Incidence, <i>G. T. Inouye</i>	108
SESSION 49: Antennas and Propagation IV—Symposium: UHF Television—Boom or Bust (Sponsored by the Professional Group on Antennas and Propagation.)	
229. FCC Rules and Propagation Data, <i>E. W. Allen</i>	116
230. Propagation in the UHF-TV Band (Abstract), <i>J. W. Herbstreit</i>	120
231. Overcoming the Line-of-Sight Shibboleth with the Air and High Power, <i>Thomas J. Carroll</i>	121
232. A Comparison of Antenna Problems at UHF and VHF TV, <i>Lloyd O. Krause</i>	126

1954 CONVENTION RECORD OF THE I.R.E.

Part 2—Circuit Theory

SESSION 28: Circuit Theory I—Symposium: Network Equalization (Sponsored by the Professional Group on Circuit Theory.)	
233. Limitations on Amplitude Equalizers, <i>Herbert J. Carlin</i>	3
234. Synthesis of Resistively-Terminated RLC Ladder Networks, <i>Er-Chun Ho and DeForest L. Trautman</i>	14
235. Equalization of Video Cables, <i>Philip W. Rounds</i>	16
236. Application of a Minimum Phase Matrix to Adjustable Equalizer Design, <i>W. R. Lundry</i>	25

Part 2—Circuit Theory (Cont'd.)

Cumulative Index Number	Page
237. Equalization in the Time Domain, <i>Murlan S. Corrington, T. Murakami, and Richard W. Sonnenfeldt</i>	30
SESSION 35: Circuit Theory II—Circuit Theory (Sponsored by the Professional Group on Circuit Theory.)	
238. The Group Theoretical Aspect of Linear Four-Pole Theory, <i>Wolfgang Gaertner</i>	36
239. A Mathematical Technique for the Analysis of Linear Systems, <i>John R. Ragazzini and Arthur R. Bergen</i>	41
240. Weighting Functions for Time Varying Feedback Systems, <i>J. A. Aseltine and R. R. Favreau</i>	52
241. Interconnection of Linear Transducers, <i>Herbert Kurss</i>	58
242. Dynamic Characteristics of Four-Terminal Networks, <i>W. W. Happ</i>	60
SESSION 39: Circuit Theory III—Network Synthesis (Sponsored by the Professional Group on Circuit Theory.)	
243. Some Techniques for Network Synthesis, <i>George L. Matheraei</i>	77
244. An Iterative Method for RC Ladder Network Synthesis, <i>R. E. Scott and N. DeClaris</i>	86
245. Networks Terminated in Resistance at Both Input and Output, <i>Louis Weinberg</i>	90
246. Approximating Band-Pass Attenuation and Phase Functions, <i>V. H. Grinich</i>	96
247. An Application of Modern Network Synthesis to the Design of Constant-Time-Delay Networks with Low-Q Elements, <i>Leo Storch</i>	105
SESSION 46: Circuit Theory IV—Transistor Circuits (Sponsored by the Professional Group on Circuit Theory.)	
248. A Transistor Analog (Abstract), <i>R. D. Lohman</i>	118
249. Junction-Transistor Multivibrators and Flip-Flops, <i>Eugene W. Sard</i>	119
250. A Synthesis Procedure for Linear Transistor Circuits, <i>J. R. Burnett</i>	125
251. Network Partitioning Techniques Applied to the Synthesis of Transistor Amplifiers, <i>H. Markarian</i>	130
252. A New Equivalent Circuit for Junction Transistors, <i>Ge Yao chu</i>	135

1954 CONVENTION RECORD OF THE I.R.E.

Part 3—Electron Devices and Component Parts

SESSION 6: Electronic Components I—Techniques (Sponsored by the Professional Group on Electronic Components.)	
253. The Effect of Maintenance on Reliability of Complex Military Electronic Equipment, <i>J. B. Arnold</i>	3
254. Miniaturized Computer Applications of the Hughes Diode, <i>S. G. Lutz</i>	8
255. Subminiaturization Techniques for UHF Communication Equipment (Abstract), <i>Gustave Shapiro</i>	16
256. Synthetic Quartz Crystals for the Electronic Industry, <i>Danforth R. Hale and Wilfred H. Charbonnet</i>	17
257. Application of Precise Components in Permeability Tuned Oscillators, <i>David Hodgkin</i>	22
SESSION 14: Electronic Components II—Application (Sponsored by the Professional Group on Electronic Components.)	
258. Magnetic-Core Delay Cables, <i>Dimitri R. Stein</i>	30
259. Improvements in the Field of Electrolytic Capacitors, <i>Dietrich Allenpohl</i>	35
260. An Investigation of Lowest Resonant Frequency in Commercially Available Bypass Capacitors, <i>David T. Geiser</i>	43
261. Resolution in Precision Wire-Wound Potentiometers, <i>Robert J. Sullivan</i>	48
262. Evaluation of Core Materials for Magnetic Amplifier Applications, <i>R. D. Teasdale and H. R. Brownell</i>	56

Part 3—Electron Devices and Component Parts (Cont'd.)

Cumulative Index Number	Page
SESSION 25: Electron Devices I—Electron Tubes (Sponsored by the Professional Group on Electron Devices.)	
263. The Hollow Cathode in Cylindrical Geometry, <i>B. D. Kumpfer and Herbert Brett</i>	66
264. The Machining of Tungsten and its Application in the Fabrication of Philips Dispenser Cathodes, <i>Roberto Levi Lob</i>	70
265. The GE Post Acceleration Color Tube (Abstract), <i>C. G. Lob</i>	73
266. Amperex Type E1T Decade Counter Tube, <i>Irwin Rudich</i>	74
267. A Developmental Thyatron Capable of Current Interruption by Grid Action (Abstract), <i>E. O. Johnson, J. A. Olmstead, and W. M. Webster</i>	79
SESSION 32: Electron Devices II—Transistors (Sponsored by the Professional Group on Electron Devices.)	
268. Transistors for High Power Application, <i>John S. Saby</i> ...	80
269. A New High Temperature Silicon Diode, <i>C. G. Thornton and L. D. Hanley</i>	84
270. Small-Signal Parameters of Grown-Junction Transistors at High Frequencies, <i>R. L. Pritchard and W. N. Coffey</i>	89
271. The Study and Design of Alloyed-Junction Transistors, <i>L. J. Giacoletto</i>	99
272. An Analytical Study of α , y and h Parameter Accuracies in Transistor Sweep Measurements, <i>H. G. Follingstad</i> ...	104
SESSION 40: Electron Devices III—Storage Tubes (Sponsored by the Professional Group on Electron Devices.)	
273. The Metrechon—A New Half-Tone Picture Storage Tube, <i>L. Pensak</i>	117
274. Characteristics of Viewing Storage Tubes with Halftone Display (Abstract), <i>M. Knoll, H. O. Hook, and R. P. Stone</i>	120
275. A High Writing Speed Dark Trace Tube, <i>S. Nozick, N. H. Burton, and S. Newman</i>	121
276. A Large Capacity Storage Tube for Digital Computer Applications, <i>R. B. DeLano, Jr.</i>	125
277. Noise Limitations on Storage Tube Operation, <i>Stanley Winkler and Seymour Nozick</i>	131
SESSION 47: Electron Devices IV—Microwave Tubes (Sponsored by the Professional Group on Electron Devices.)	
278. A Voltage-Tunable Magnetron for Operation in the Frequency Range 1500 to 3000 Megacycles, <i>Joseph A. Boyd</i>	139
279. Control of Electron-Beam Spread by Positive Ion Traps, <i>E. L. Ginzton and B. H. Wadia</i>	145
280. The Multipactor Effect in Klystrons, <i>Kees Bol</i>	151
281. Backward-Wave Oscillator Characteristics (Abstract), <i>H. R. Johnson</i>	156
282. The Propagation Properties of Cross-Wound Twin Helices Suitable for Traveling-Wave Tubes (Abstract), <i>M. Chodorow, E. L. Chu, and J. R. Nevins, Jr.</i>	156

1954 CONVENTION RECORD OF THE I.R.E.

Part 4—Electronic Computers and Information Theory

SESSION 2: Information Theory I—Application of Information Theory to Communication Systems (Sponsored by the Professional Group on Information Theory.)	
283. Information Theory—Past, Present and Future, <i>R. M. Fano</i>	2
284. Optical Filters: Their Equivalence to and Difference from Electrical Networks, <i>Thomas P. Cheatham, Jr. and Arthur Kohlenberg</i>	6
285. Theoretical Improvement in Signal-to-Noise Ratio of Television Signals by Equivalent Comb Filter, <i>M. J. Stateman and M. B. Riterman</i>	13
286. Information Losses in Regenerative Pulse-Code Systems, <i>Warren D. White</i>	18
287. A Gaussian Noise Generator for Frequencies Down to 0.001 Cycles per Second, <i>David F. Winter</i>	23
SESSION 12: Information Theory II—Coding and Noise (Sponsored by the Professional Group on Information Theory.)	
288. Matched Filters for Detecting Pulsed Signals in Noise, <i>John S. Rochefort</i>	30
289. An Experimental Study of the Information Rate of a Digital Computer, <i>Norman R. Scott</i>	35
290. Time-Varying Quasi-Linear Method of Speech Noise Suppression (Abstract), <i>M. J. DiToro</i>	40
291. Discriminatory Analysis Applied to Speech Sound Recognition, <i>H. L. Stubbs</i>	41

Convention Record Index—4

Part 4—Electronic Computers and Information Theory (Cont'd.)

Cumulative Index Number	Page
292. A Discussion of Auto-Correlated Error Terms in Time Series Analysis, <i>Ralph K. Weller</i>	45
SESSION 19: Information Theory III—Speed and Computation (Sponsored by the Professional Group on Information Theory.)	
293. Optimized Data Encoding for Digital Computers, <i>William H. Kautz</i>	47
294. Symbolic Methods in the Design of Delay- and Cycle-Free Logical Nets, <i>George W. Patterson</i>	58
295. Threshold Detection, <i>B. L. Basore</i>	65
296. The Nature of the Uncorrelated Component of Induced Grid Noise, <i>T. E. Talprey and A. B. Macnee</i>	69
297. Effect of Limiting on the Information Content of Noisy Signals, <i>G. O. Young and B. Gold</i>	76
SESSION 27: Electronic Computers I—Computer Design and Techniques (Sponsored by the Professional Group on Electronic Computers.)	
298. The Role of General Purpose Digital Computers in Automatic Control and Information Systems, <i>Arnold A. Cohen</i>	82
299. Design Features of Current Digital Differential Analyzers, <i>Edvard L. Braun</i>	87
300. Design Features of the JAINCOMP-C and JAINCOMP-D Electronic Digital Computers, <i>Donald H. Jacobs</i> ...	98
301. A Germanium Tape Reader (Abstract), <i>R. A. Langevin</i> ...	105
302. Electrostatic Reading of Perforated Media, <i>Samuel Lubkin</i>	106
SESSION 34: Electronic Computers II—Computer Components (Sponsored by the Professional Group on Electronic Computers.)	
303. Considerations for the Selection of Magnetic Core Materials for Digital Computer Elements, <i>O. J. Van Sant, Jr.</i>	109
304. Magnetic Core Selection Systems, <i>S. Guterman and R. D. Kodis</i>	116
305. Circuits to Perform Logical and Control Functions with Magnetic Cores, <i>S. Guterman, R. D. Kodis, and S. Ruhman</i>	124
306. Packaged Logical Circuitry for a 4-Mc Computer, <i>Norman Zimbel</i>	133
307. Transistor Shift Registers, <i>C. Huang, E. Slobodzinski, and B. White</i>	140

1954 CONVENTION RECORD OF THE I.R.E.

Part 5—Aeronautical Electronics and Telemetry

SESSION 3: Aeronautical and Navigational Electronics I (Sponsored by the Professional Group of Aeronautical and Navigational Electronics.)	
308. An Impulse Generator for Receiver Performance Measurement, <i>Joseph H. Vogelmann</i>	3
309. Aerial Methods in Microwave Survey, <i>Marc Sheldon and Lewis A. Dickerson</i>	12
310. The Development of a Production Radome Tester, <i>Robert P. Walcott</i>	31
311. A Correlation Direction Finder for Guided Missile Range Instrumentation, <i>Marvin S. Friedland and Nathan Marchand</i>	35
312. Present Status of Microwave Radiometric Receiver Development, <i>R. M. Ringoen</i>	42
SESSION 5: Radio Telemetry and Remote Control I—Systems and Elements (Sponsored by the Professional Group on Radio Telemetry and Remote Control.)	
313. Guided Missile Range Instrumentation—A New Electronic Art, <i>M. S. Friedland</i>	48
314. Interpretation of Sequential Samples from Commutated Data (Abstract), <i>Lawrence L. Rauch</i>	58
315. Comparison of Required Radio Frequency Power in Different Methods of Multiplexing and Modulation, <i>M. H. Nichols</i>	59
316. Flight Testing of an Airborne Digital Computer, <i>E. M. Grabbe, D. W. Burbeck, and S. B. Neister</i>	66
317. Evaluation for Magnetic Tape Equipment for Telemetering Instrumentation (Title Only), <i>R. E. Rawlins</i>	71
SESSION 8: Aeronautical and Navigational Electronics II (Sponsored by the Professional Group on Aeronautical and Navigational Electronics.)	
318. The Digital Airborne Digital Computer, <i>E. E. Bolles</i> ...	72
319. A New Fixed-Beam Approach System, <i>R. A. Hampshire</i> ...	77
320. The Role of Flight Directors in Present-Day Aircraft, <i>N. L. Graham</i>	84

Part 5—Aeronautical Electronics and and Telemetry (Cont'd.)

Cumulative Index Number	Page
321. Navaglobe-Navarho Long-Range Radio Navigational System, <i>C. T. Clark, R. I. Colin, M. Dishal, I. Gordy, and M. Rogoff</i>	88
322. The N-1 Compass System, <i>Robert C. Rosaler</i>	98
SESSION 10: Radio Telemetry and Remote Control II—Telemetry (Sponsored by the Professional Group on Radio Telemetry and Remote Control.)	
323. A 227 Mc Pulse Position Modulation Telemetry Unit, <i>D. G. Mazur</i>	105
324. Crystal Control Low Distortion FM Telemetry Transmitter (Title Only), <i>R. E. Rawlins</i>	112
325. A Crystal Control FM Telemetry Transmitter, <i>Foster N. Reynolds</i>	113
326. High Gain Antenna System for Multiple Operation, <i>James B. Wynn, Jr.</i>	116
SESSION 15: Aeronautical and Navigational Electronics III (Sponsored by the Professional Group on Aeronautical and Navigational Electronics.)	
327. Operational Analysis of Track-While-Scan Radars, <i>Stephen J. O'Neil</i>	123
328. A Study of the UHF Omnidirectional Aircraft Antenna Problem and Proposed Methods of Solution, <i>IV. Spanos and J. J. Nail</i>	135
329. A Modulator Technique for Producing Short Pulses in High Powered Magnetrons, <i>Thomas J. Parker</i>	142
330. The Role of Stereo in "3-D" Radar Indicating Systems, <i>Walter R. Tower</i>	152
331. An Automatic Antenna Matching Unit, <i>E. W. Schmitt</i>	163
SESSION 45: Radio Telemetry and Remote Control III—Remote Control (Sponsored by the Professional Group on Radio Telemetry and Remote Control.)	
332. A Proportional Data Transmission System, <i>W. C. Petrie</i>	169
333. A Digital Autopilot Coupler, <i>W. L. Exner and A. D. Scarbrough</i>	174
334. System Compensation with a Digital Computer, <i>John M. Salzer</i>	179
335. Binary Control System for Digital-to-Shaft Position Mechanisms, <i>Arthur H. Wulfsberg</i>	187
336. Optimization of Servosystems, <i>Richard C. Lyman and William P. Caywood, Jr.</i>	193

1954 CONVENTION RECORD OF THE I.R.E.

Part 6—Audio and Ultrasonics

SESSION 11: Audio I—High Fidelity (Sponsored by the Professional Group on Audio.)	
337. Large Area Microphones for Distant Pickup Use (Abstract), <i>T. Aamodt and F. K. Harvey</i>	3
338. The Enhancement of Music by Reverberation, <i>Daniel W. Martin</i>	4
339. Some New Developments in High Fidelity Loudspeakers (Abstract), <i>H. F. Olson and John Preston</i>	8
340. High Fidelity and the Hearing Process (Abstract), <i>W. E. Kock</i>	8
341. Some Aspects of Stereophonic Sound in Motion Picture Theaters (Abstract), <i>R. H. Ranger</i>	8
SESSION 18: Audio II—General (Sponsored by the Professional Group on Audio.)	
342. Nonlinear Communications Systems: Some Aspects of Clipped Speech, <i>R. E. Lacy and R. K. Saxx</i>	9
343. A Miniature Unidirectional Microphone, <i>B. B. Bauer and J. W. Medill</i>	12
344. A High Efficiency-High Quality Audio Power Amplifier, <i>Alexander B. Bereskin</i>	18
345. System Design Factors for Audio Amplifiers, <i>M. V. Kiebert, Jr.</i>	25
346. Driver System for Single Ended Push-Pull Amplifiers (Abstract), <i>C. T. Hall</i>	41
SESSION 23: Audio Seminar—High Fidelity in Audio Engineering (Sponsored by the Professional Group on Audio.)	
347. Microphones, <i>John K. Hilliard</i>	42
348. Loudspeakers (Title Only), <i>H. F. Olson</i>	45
349. Room Acoustics (Title Only), <i>R. L. Hanson</i>	45
350. High Fidelity in Radio Broadcasting Systems, <i>John V. L. Hogan</i>	46
SESSION 41: Ultrasonics I (Sponsored by the Professional Group on Ultrasonics.)	
351. The Ultrasonic Burglar Alarm System, <i>Samuel Bagno, Jack B. Cooper, and Eli A. Levy</i>	49

Part 6—Audio and Ultrasonics (Cont'd.)

Cumulative Index Number	Page
352. A Complex Impedance Recorder, <i>Harold M. Sharaf</i>	59
353. Ultrasonic Delay Lines, <i>David L. Arenberg</i>	63
354. Wide-Band Large-Dynamic-Range Fused Quartz Delay Lines for Increased-Capacity High Speed Computer Memories, <i>D. A. Spaeth, T. F. Rogers, and S. J. Johnson</i>	77
355. Contour Modes of Plates Excited Piezoelectrically and Determination of Elastic and Piezoelectric Coefficients, <i>R. Beckmann</i>	77
SESSION 48: Ultrasonics II (Sponsored by the Professional Group on Ultrasonics.)	
356. Applications of Ultrasonic Energy to Industrial Use (Abstract), <i>A. L. Bayles</i>	85
357. The Effects of Ultrasonic Waves on Electrolytes and Electrode Processes, <i>S. Barnartt</i>	86
358. Application of Ultrasound to the Brain, <i>P. A. Lindstrom</i>	96
359. Selective Action of Ultrasound on Nerve Tissue, <i>William J. Fry and John W. Barnard</i>	102
360. Effects of Ultrasound on Living Cell Structure, <i>Earl H. Newcomer</i>	107

1954 CONVENTION RECORD OF THE I.R.E.

Part 7—Broadcasting and Television Receivers

SESSION 13: Broadcast and Television Receivers I—General (Sponsored by the Professional Group on Broadcast and Television Receivers.)	
361. Ferrite Cored Antennae, <i>C. A. Grimmer</i>	3
362. Transistor AM Broadcast Receivers, <i>Arthur P. Stern and John A. A. Raper</i>	8
363. Wide-Band Amplification with Surface-Barrier Transistors, <i>J. B. Angell</i>	15
364. Automatic Damping for Vertical Output Circuits in TV Systems, <i>H. E. Thomas, S. A. De Mars, and M. E. Jones</i>	21
365. A Wide Range Tuning System, <i>H. T. Lyman, F. G. Mason, and H. Ross</i>	27
SESSION 20: Broadcast and Television Receivers II—Color Television (Sponsored by the Professional Group on Broadcast and Television Receivers.)	
366. A Self-Balancing Phase Detector for Color Receiver Reference Oscillators, <i>E. G. Clark</i>	31
367. Color Fidelity in TV Receiver Having Nonstandard Primaries (Abstract), <i>F. J. Bingley</i>	38
368. Color Distortion in Sequential Television Displays, <i>Donald C. Livingston</i>	39
369. Single-Gun Picture Tubes in NTSC Color Television, <i>Stephen K. Altes and Arthur P. Stern</i>	46
370. Significance of Some Receiver Errors to Color Reproduction, <i>Harold Weiss</i>	53
SESSION 26: Broadcast Transmission Systems I—Symposium: TV Broadcasting (Sponsored by the Professional Group on Broadcast Transmission Systems.)	
371. Antenna System for Station WOR-TV Channel 9, Installed on the Empire State Building in New York City, <i>G. J. Adams, Andrew Alford, H. H. Leach, Richard Rubin, and Fred Abel</i>	65
372. A Pulse Distribution System for a Network Originating Center, <i>John S. Auld and Anthony Gallonio</i>	72
373. An Improved Television Clamp Circuit Employing Feedback, <i>K. R. Wendt and W. K. Squires</i>	79
374. A High Level Plate Mixer for Use at UHF, <i>Ralph E. Western</i>	85
375. Coaxial Line Transfer Switch for Television Transmitters, <i>Carl F. Schunemann and J. B. Epperson</i>	88
SESSION 33: Broadcast Transmission Systems II—Symposium: Color TV Broadcasting (Sponsored by the Professional Group on Broadcast Transmission Systems.)	
376. Color Film Scanner—Circuits, <i>Joseph F. Fisher</i>	94
377. Color Characteristics of a Television Film Scanner, <i>Jesse H. Haines</i>	100
378. Factors in the Design of Keyed Clamping Circuits, <i>Roland N. Rhodes</i>	105
379. Photographic Simulation of Color Television Brightness Modifications, <i>J. H. Ladd and W. L. Brewer</i>	110
380. Feasibility and Technique of Storing Color Video Information on Black and White Film, <i>William L. Hughes</i>	114
381. A System for Recording and Reproducing Television Signals (Abstract), <i>H. F. Olson, W. D. Houghton, A. R. Morgan, J. Zenel, M. Artzt, J. G. Woodward, and J. T. Fischer</i>	119
382. V T R: A Video Magnetic Tape Recorder, <i>John T. Mullin</i>	120

1954 CONVENTION RECORD OF THE I.R.E.

Part 8—Communications and Microwave

- SESSION 1: Symposium: Advances in Mobile Communications (Sponsored by the Professional Group on Vehicular Communications.)
383. Transient Response of Selective Networks and Impulse Noise in Narrow-Band FM Receivers, *Stanley P. Lapin and Jerome J. Suran*..... 3
384. Advances in Petroleum Mobile Communications, *L. A. M. Barnette*..... 9
385. A New Approach to 450–470 Mc Communication Equipment, *R. W. Tuttle*..... 15
386. Operation and Planning of a Utility Radio System, *A. B. Buchanan*..... 18

- SESSION 7: Radio Communications I—Symposium: Facsimile (Sponsored by the Professional Group on Communications Systems.)
387. Facsimile Systems, *A. S. Hill*..... 24
388. Operation of International Commercial Radiophoto Circuits, *M. P. Rehm*..... 32
389. Applications of Facsimile in the USAF, *Harold R. Johnson*..... 39
390. Application of Cathode-Ray Tubes on Facsimile, *Warren H. Bliss*..... 44

- SESSION 21: Radio Communications II—General (Sponsored by the Professional Group on Communications Systems.)
391. System Aspects and Trends of Modern Communication, *I. S. Coggeshall*..... 51
392. A Predicted Wave Radio Teletype System, *Melvin L. Doelz and Earl T. Heald*..... 63
393. Design Consideration for FSK Circuits, *Walter Lyons*.... 70
394. Predicting Interference Levels in the Communication System, *Paul G. Wulfsberg*..... 74
395. UHF Diversity System for Long-Range Ship-to-Air Communication, *F. J. Altman and J. J. Nail*..... 78

- SESSION 43: Microwave Electronics I—Ferrites and Strip Lines (Sponsored by the Professional Group on Microwave Theory and Techniques.)
396. Nonreciprocal Microwave Components, *Herman N. Chait*..... 82
397. Ferrite Quarter-Wave and Half-Wave Plates at X-Band (Abstract), *N. G. Sakiotis*..... 88
398. The Radiation Conductance of a Series Slot in Strip Transmission Line (Summary), *Arthur A. Oliner*..... 89
399. New Techniques for High-Q Strip Microwave Components, *E. G. Fubini, W. E. Fromm, and H. S. Keen*.... 91
400. Microwave Applications of High-Q Strip Components, *E. G. Fubini, W. E. Fromm, and H. S. Keen*..... 98

- SESSION 50: Microwave Electronics II—Components (Sponsored by the Professional Group of Microwave Theory and Techniques.)
401. Design of Stable Tunable Microwave Oscillators, *J. G. Stephenson*..... 104
402. Microwave Measurements with a Lossy Variable Short Circuit (Summary), *H. M. Altschuler and A. A. Oliner*.... 113
403. Directional Couplers, *Peter Sferrazza*..... 115
404. Diplexing Filters, *Maurice E. Breese*..... 125
405. A High Precision Compensated Reference Cavity for C-Band, *John Hall and Frank McCarthy*..... 134

1954 CONVENTION RECORD OF THE I.R.E.

Part 9—Medical and Nuclear Electronics

- SESSION 17: Medical Electronics (Sponsored by the Professional Group on Medical Electronics.)
406. The Gamma Ray Pinhole Camera with Image Amplifier, *Robert K. Mortimer, Hal O. Anger, and Cornelius A. Tobias*..... 2
407. Expansion Chamber for Measurement of Red Cell Permeation by Water (Abstract), *A. K. Solomon and C. V. Paganelli*..... 6
408. Color X-Ray Pictures, *R. Stuart Mackay*..... 7
409. Measurement of Slow Neutron Depth Dose in Tissue, *E. Stickley*..... 9
410. Use of Charged Particles to Measure Skin Thickness and Other Surface Properties, *Franklin Hutchinson*..... 13

Convention Record Index—6

Part 9—Medical and Nuclear Electronics (Cont'd.)

- | Cumulative
Index
Number | Page |
|---|------|
| SESSION 22: Medical Electronics Symposium: Engineering Based on Biological Design (Sponsored by the Professional Group on Medical Electronics.) | |
| 411. Medical Electronics Symposium: Engineering Based on Biological Design, <i>W. R. G. Baker</i> | 17 |
| 412. Human Engineering, <i>Leonard C. Mead</i> | 19 |
| 413. Information Theory (Title Only), <i>Norbert Wiener</i> | 26 |
| 414. Biological Transducers, <i>S. S. Stevens</i> | 27 |
| 415. Biological Servomechanisms and Control Circuitry, <i>Otto H. Schmitt</i> | 34 |
| SESSION 24: Nuclear Science I—Symposium: Progress Report (Sponsored by the Professional Group on Nuclear Science.) | |
| 416. Electronics in the Nuclear Industry, <i>L. V. Berkner</i> | 42 |
| 417. Secrecy and the Electronic Engineer (Abstract), <i>J. G. Beckerly</i> | 44 |
| 418. Nonreactor Electronics at Oak Ridge (Abstract), <i>P. R. Bell</i> | 44 |
| 419. Brookhaven Electronics Instrumentation Program, <i>W. A. Higinbotham</i> | 45 |
| 420. Nonreactor Electronic Work at Argonne, <i>Thomas Brill</i> | 51 |
| 421. Nonreactor Electronics at Los Alamos, <i>R. J. Watts</i> | 61 |
| SESSION 31: Nuclear Science II—Symposium: Reactor Electronics (Sponsored by the Professional Group on Nuclear Science.) | |
| 422. Chairman's Introductory Remarks: Symposium on Reactor Electronics, <i>William M. Breazeale</i> | 67 |
| 423. Nuclear Reactor Simulators, <i>Kenneth H. Fischbeck</i> | 69 |
| 424. Safety Aspects of Control Circuitry, <i>T. E. Cole</i> | 75 |
| 425. Instruments Used with Experimental Reactors, <i>Elmer J. Wade</i> | 79 |
| 426. Synthesis of Control Systems for Nuclear Power Plants, <i>J. N. Grace</i> | 83 |

1954 CONVENTION RECORD OF THE I.R.E.

Part 10—Instrumentation and Industrial Electronics

- SESSION 29: Instrumentation I (Sponsored by the Professional Group on Instrumentation.)
427. Phase Measurements in the Video Frequency Range, *W. W. Graustein, Jr., and R. W. Houghton*..... 3
428. An X-Band Rapid-Sweep Oscillator, *Herbert H. Rickert and David Deltinger*..... 7
429. A Shielded Two-Wire Hybrid Junction and its Use as an Ultra-High-Frequency Impedance Bridge, *Edgar W. Matheus, Jr.*..... 14
430. High-Speed, High-Resolution Spectrum Analyzer, *Nesbit L. Duncan*..... 22
431. Rapid, Precision Impedance Measurements in the 400–1600 Megacycle Frequency Range, *David M. Goodman*..... 27
- SESSION 36: Instrumentation II—Symposium: High Frequency Measurement and Control (Sponsored by the Professional Group on Instrumentation.)
432. An Approach to a Company-Owned Frequency Standard, *John W. Smith*..... 34
433. Standard Frequency Controlled Wide Range Oscillator, *E. P. Felch, J. O. Israel, and O. Kummer*..... 38
434. Performance of the Bell System Frequency Standard (Abstract), *George N. Packard*..... 45
435. A Computer-Type Decade Frequency Synthesizer, *R. W. Frank*..... 46
436. A High-Speed Digital Frequency Divider of Arbitrary Scale, *Robert W. Stuart*..... 52
- SESSION 38: Industrial Electronics (Sponsored by the Professional Group on Industrial Electronics)
437. The Design of Automatic Factories, *Geoffrey Post*..... 58
438. Industrial Punch Card Automatic Control Development, *Wilfrid L. Atwood*..... 63
439. Considerations in the Automatic Assembly of Components, *Ben Warriner*..... 67
440. Electronic Flow Measurement and Control (Abstract), *Eugene Mittelman*..... 69
441. Photosensitive Germanium Devices and Some Device Applications, *Richard G. Seed*..... 70
- SESSION 44: Instrumentation III (Sponsored by the Professional Group on Instrumentation.)
442. A Novel Approach to Transistor Testing (Abstract), *N. J. Gottfried*..... 80

Part 10—Instrumentation and Industrial Electronics (Cont'd.)

Cumulative Index Number	Page
443. A Transistor Frequency Scanner, <i>O. Kummer</i>	81
444. A Simple Transistor Noise Test Set, <i>Richard W. Carlisle, Harry A. Pearson, and William H. Greenbaum</i>	88
445. Wide-Band Amplitude Distribution Analysis of Voltage Sources, <i>Lyman W. Orr</i>	92
446. A Generator of Uniformly Distributed Random Noise, <i>Robert Bernstein, Henry Bickel, and Eli Brookner</i>	97

1954 CONVENTION RECORD OF THE I.R.E.

Part 11—Engineering Management and Quality Control

SESSION 4: Quality Control and Reliability (Sponsored by the Professional Group on Quality Control.)	
447. Improving Reliability of Electronic Equipment by Effective Analysis of Field Performance, <i>Richard R. Landers</i>	2
448. A Survey of Electronic Failure Prediction Technique, <i>J. H. Muncy</i>	9
449. A New Approach to Attainment of Reliability in the Production of Airborne Electronic Systems, <i>A. Warsher</i>	13

Part 11—Engineering Management and Quality Control (Cont'd.)

Cumulative Index Number	Page
450. A Method of Testing and Evaluation of Complex Missile Systems, <i>E. J. Althaus, S. C. Morrison, and W. R. Tate</i>	23
SESSION 9: Engineering Management I (Sponsored by the Professional Group on Engineering Management.)	
451. The Engineer and Return on Investment, <i>S. C. Peck</i>	29
452. Technical Information: Communication for Research, <i>Charles DeVore</i>	35
453. A Working Philosophy for Engineering Management, <i>Thomas G. Slattery</i>	38
454. Organization for Operations Research, <i>Foster L. Weldon</i> ..	42
455. Training for Operations Research Groups, <i>Thornton Page</i>	45
SESSION 16: Engineering Management II—Symposium: Personnel Training and Selection for Engineering Management (Sponsored by the Professional Group on Engineering Management.)	
456. Personnel Selection and Training for Engineering Management from the University Viewpoint, <i>S. C. Hollister</i>	49
457. Personnel Training and Selection for Engineering Management, <i>W. R. G. Baker</i>	52
458. Selection and Training for Engineering Management in the Department of Defense, <i>James M. Mitchell</i>	55

INDEX TO AUTHORS

A

Aamodt, T., 337
 Abel, F., 371
 Adams, G. J., 371
 Alford, A., 216, 371
 Allen, E. W., 229
 Altenpohl, D., 259
 Altes, S. K., 369
 Althaus, E. J., 450
 Altman, F. J., 395
 Altschuler, A. M., 402
 Angell, J. B., 363
 Anger, H. O., 406
 Arenberg, D. L., 353
 Arnold, J. B., 253
 Artzt, M., 381
 Aseltine, J. A., 240
 Atwood, W. L., 438
 Auld, J. S., 372

B

Bagno, S., 351
 Bailin, L. L., 214
 Baker, W. R. G., 441, 457
 Barnard, J. W., 359
 Barnartt, S., 357
 Barnett, L. A. M., 384
 Barsis, A. P., 227
 Basore, B. L., 295
 Bauer, B. B., 343
 Bayles, A. L., 356
 Beachemin, H. M., 221
 Beckerly, J. G., 417
 Bechmann, R., 355
 Bell, P. R., 418
 Bereskin, A. B., 344
 Bergen, A. R., 239
 Berkner, L. V., 416
 Bernstein, R., 446
 Bickel, H., 446
 Bingley, F. J., 367
 Bliss, W. H., 390
 Bol, K., 280
 Bolles, E. E., 318
 Boyd, J. A., 278
 Braun, E. L., 299
 Breazeale, W. M., 422

Breese, M. E., 404
 Brett, H., 263
 Brewer, W. L., 379
 Brill, T., 420
 Brookner, E., 446
 Brownell, H. R., 262
 Buchanan, A. B., 386
 Burbeck, D. W., 316
 Burnett, J. R., 250
 Burton, N. H., 275

C

Carlin, H. J., 233
 Carlisle, R. W., 444
 Carroll, T. J., 231
 Carter, D., 222
 Caywood, W. P., Jr., 336
 Chait, H. N., 396
 Charbonnet, W. H., 256
 Cheatham, T. P., Jr., 284
 Chodorow, M., 282
 Chu, E. L., 282
 Chu, G. Y., 252
 Clark, C. T., 321
 Clark, E. G., 366
 Coffey, W. N., 270
 Coggeshall, I. S., 391
 Cohen, A. A., 298
 Cohn, S. B., 220
 Cole, T. E., 424
 Colin, R. I., 321
 Copper, J. B., 351
 Corrington, M. S., 237
 Crosby, D. R., 223

D

DeClaris, N., 244
 DeLano, R. B., Jr., 276
 DeMars, S. A., 364
 Dettinger, D., 428
 DeVore, C., 452
 Dickerson, L. A., 309
 Dishal, M., 321
 DeToro, M. J., 290
 Doelz, M. L., 392
 Duncan, N. L., 430

E

Ehrenspeck, H., 217
 Epperson, J. B., 375
 Exner, W. L., 333

F

Fano, R. M., 283
 Favreau, R. R., 240
 Feinstein, J., 226
 Felch, E. P., 433
 Felsen, L. B., 215
 Fischbeck, K. H., 423
 Fischer, J. T., 381
 Follingstad, H. G., 272
 Frank, R. W., 435
 Friedland, M. S., 311, 313
 Fromm, W. E., 399, 400
 Fry, W. J., 359
 Fubini, E. G., 399, 400

G

Gaertner, W., 238
 Gallonio, A., 372
 Geiser, D. T., 260
 Gerbes, W., 217
 Giacometto, L. J., 271
 Ginzton, E. L., 279
 Gold, B., 297
 Goodman, D. M., 431
 Gordy, I., 321
 Gottfried, N. J., 442
 Grabbe, E. M., 316
 Grace, J. N., 426
 Graham, N. L., 320
 Graustein, W. W., Jr., 427
 Greenbaum, W. H., 444
 Grimmett, C. A., 361
 Grinich, V. H., 246
 Guterma, S., 304, 305

H

Haines, J. H., 377
 Hale, D. R., 256
 Hall, C. T., 346
 Hall, J., 405
 Hampshire, R. A., 319
 Hanley, L. D., 269

Hannan, P. W., 219
 Hanson, R. L., 349
 Happ, W. W., 242
 Harvey, F. K., 337
 Heald, E. T., 392
 Herbstreit, J. W., 230
 Higinbotham, W. A., 419
 Hill, A. S., 387
 Hilliard, J. K., 347
 Ho, Er-Chun, 234
 Hodgins, David, 257
 Hogan, J. V. L., 350
 Hollister, S. C., 456
 Holt, F. S., 221
 Hook, H. O., 274
 Houghton, R. W., 427
 Houghton, W. D., 381
 Huang, C., 307
 Hughes, W. L., 380
 Hutchison, F., 410

I

Inouye, G. T., 228
 Israel, J. O., 433

J

Jacobs, D. H., 300
 Johnson, E. O., 267
 Johnson, H. R., 281, 389
 Johnson, S. J., 354
 Jones, E. M. T., 220
 Jones, M. E., 364

K

Kaminow, I. P., 214
 Kautz, W. H., 293
 Keen, H. S., 399, 400
 Kelleher, K. S., 224
 Kiebert, M. R., Jr., 345
 Knoll, M., 274
 Kock, W. E., 340
 Kodis, R. D., 304, 305
 Kohlenberg, A., 284
 Krause, L. O., 232
 Kummer, O., 433, 443
 Kumpfer, B. D., 263
 Kurss, H., 241

L
Lacy, R. E., 342
Ladd, J. H., 379
Lapin, S. P., 383
Landers, R. R., 447
Langevin, R. A., 301
Leach, H. H., 371
Levi, R., 264
Levy, E. A., 351
Lindstrom, P. A., 358
Livingston, D. C., 368
Lob, C. G., 265
Lohman, R. D., 248
Lubkin, S., 302
Lundry, W. R., 236
Lutz, S. G., 254
Lyman, H. T., 365
Lyman, R. C., 336
Lyons, W., 393

M
Mackay, R. S., 408
Macnee, A. B., 296
Marchand, N., 311
Markarian, H., 251
Martin, D. W., 338
Marson, F. G., 365
Matthaci, G. L., 243
Matthews, E. W., Jr., 429
Mazur, D. G., 323
McCarthy, F., 405
Mead, L. C., 412
Medill, J. W., 343
Mitchell, J. M., 458
Mittelman, E., 440
Morgan, A. R., 381
Morrison, S. C., 450
Mortimer, R. K., 406
Mullin, J. T., 382
Muncy, J. H., 448
Murakami, T., 237

N
Nail, J. J., 328, 395
Neister, S. B., 316
Nevins, J. R., 282

Newcomer, E. H., 360
Newman, S., 275
Nichols, M. H., 315
Nozick, S., 275, 277

O
Oliner, A. A., 398, 402
Olmstead, J. A., 267
Olson, H. F., 339, 348, 381
O'Neil, S. J., 327
Orr, L. W., 445

P
Packard, G. N., 434
Paganelli, C. V., 407
Page, T., 455
Parker, T. J., 329
Patterson, G. W., 294
Pearson, H. A., 444
Peck, S. C., 451
Pensak, L., 273
Petrie, W. C., 332
Pope, W. A., 225
Post, G., 437
Preston, J., 339
Pritchard, R. L., 270
Puro, W. O., 224

R
Raggazzini, J. R., 239
Ranger, R. H., 341
Raper, J. A. A., 362
Rauch, L. L., 314
Rawlins, R. E., 317, 324
Rehm, M. P., 388
Reynolds, F. N., Jr., 325
Rhodes, R. N., 378
Rickert, H. H., 428
Ringo, R. M., 312
Ritterman, M. D., 285
Rocheffort, J. S., 288
Rogers, T. F., 354
Rogoff, M., 321
Rosaler, R. C., 322
Ross, H., 365
Rounds, P. W., 235
Rubin, R., 371

Rudich, I., 266
Ruhman, S., 305

S
Saby, J. S., 268
Sakiotis, N. G., 397
Salmon, A. K., 407
Salzer, J. M., 334
Sampson, J. L., 221
Sard, E. W., 249
Saxe, R. K., 342
Scarborough, A. D., 333
Schmitt, O. H., 415
Schunemann, C. F., 375
Schwittek, E. W., 331
Scott, N. R., 289
Scott, R. E., 244
Seed, R. G., 441
Sferrazza, P., 403
Shapiro, G., 255
Sharaf, H. M., 352
Sheldon, M., 309
Slattery, T. G., 453
Slobodzinsky, E., 307
Smith, J. W., 432
Solomon, A. K., 407
Sonnenfeldt, R. W., 237
Spaeth, D. A., 354
Spanos, W., 328
Spencer, R. C., 221
Sprague, R. M., 216
Squires, W. K., 373
Stateman, M. J., 285
Stein, D. R., 258
Stephenson, J. G., 401
Stern, A. P., 362, 369
Stevens, S. S., 414
Stickley, E., 409
Stone, R. P., 274
Storch, L., 247
Stuart, R. W., 436
Stubbs, H. L., 291
Sullivan, R. J., 261
Suran, J. J., 383

T
Talpey, T. E., 296

Tate, W. R., 450
Teasdale, R. D., 262
Thomas, H. E., 364
Thornton, C. G., 269
Tobias, C. A., 406
Tower, W. R., 330
Trautman, D. L., 234
Tuttle, R. W., 385

V
Van Sant, O. J., Jr., 303
Vogelman, J. H., 308

W
Wade, E. J., 425
Wadia, B. H., 279
Wait, J. R., 225
Walcutt, R. P., 310
Warriner, B., 439
Warsher, A., 449
Watts, R. J., 421
Weber, J., 218
Webster, W. M., 267
Wehner, R. S., 214
Weinberg, L., 245
Weiss, H., 370
Weldon, F. L., 454
Weller, R. K., 292
Wendt, K. R., 373
Western, R. E., 374
White, B., 307
White, W. D., 286
Wiener, N., 413
Winkler, S., 277
Winter, D. F., 287
Woodward, J. G., 381
Wulfsberg, A. H., 335
Wulfsberg, P. G., 394
Wynn, J. B., Jr., 326

Y
Young, G. O., 297

Z
Zenel, J., 381
Zimbel, N., 306
Zucker, F. J., 217

INDEX TO SUBJECTS

A
Acoustics, Room: 349
Aerial Methods in Microwave Survey: 309
Amplifier: 262, 344, 345, 346
Audio: 344, 345
Design Factors: 345
Magnetic: 262
Core Materials: 262
Single-Ended Push-Pull: 346
Antennas: 215, 216, 217, 219, 221, 225, 232, 321, 326, 328, 331, 361, 371
Aircraft Omnidirectional: 328
Ferrite Core: 361
For Multiple Operation in Telemetry: 326
Four Slot: 216
Matching Unit, Automatic: 331
Microwave, Reflections in: 219
Pencil Beam: 221, 222
Slot: 215
Gain Pattern: 215
Trapped Wave: 217
UHF and VHF TV Problems: 232
Vertical, with Radial Conductor Ground: 225
WOR-TV Installation: 371
Argonne Nonreactor Electronics: 420

Arrays: 214
Dolph-Tchebycheff: 214
Autocorrelated Error Terms: 292
Time Series Analysis: 292
Autopilot, Digital: 333
Automatic Assembly of Components: 439
Automatic Control Systems: 298, 438
Punch Cards: 438
Role of Digital Computers: 298
Automatic-Factory Design: 437
B
Back-Wave Oscillator: 281
Biological Design Applied to Engineering: 411, 412, 413, 414, 415
Biological Servomechanisms: 415
Biological Transducers: 414
Brookhaven Instrumentation Program: 419
Burglar Alarm, Ultrasonic: 351
C
Camera, Pinhole, Gamma Ray: 406
Capacitors: 259, 260
Bypass: 260
Lowest Resonant Frequency: 260
Electrolytic: 259

Cathode-Ray Tubes: 390
Applications in Facsimile: 390
Cathodes: 263, 269
Hollow: 263
Machined Tungsten: 269
Cavity, Reference, for C Band: 405
Circuits: 248, 249, 250, 251, 252
Transistor: 248, 249, 250, 251, 252
Circuit Theory: 238, 239, 240
Analysis of Linear Systems: 239
Four-Pole Theory: 238
Time Varying Feedback Systems: 240
Clamp Circuit for Television: 373, 378
Coding: 286, 293
Optimized Data for Computers: 293
Pulse-Code Systems: 286
Information Losses: 286
Color Television: 265, 366, 367, 368, 369, 370, 376, 377, 378, 379, 380, 381, 382
Clamp Circuit: 378
Distortion in Sequential Displays: 368
Fidelity in Receivers with Nonstandard
Primaries: 367
Film Considerations: 379
Film Scanner: 376, 377
Characteristics: 377
Circuits: 376

Color Television, *Cont'd.*

Magnetic Tape Recording: 381, 382
Phase Detector for Color Reference Oscillator: 366
Picture Tube: 265, 369
Post Acceleration: 265
Single Gun: 369
Reproduction, Effects of Receiver Errors on: 370
Storing on Black-and-White Film: 380
Color X-Ray Pictures: 408
Communication of Technical Information: 452
Communication Systems: 391, 394
Predicting Interference Levels: 394
Ship-to-Air UHF Diversity System: 395
Trends: 391
Compass, N-1: 322
Complex-Impedance Recorder: 352
Computers: 254, 289, 293, 298, 299, 300, 301, 302, 303, 304, 305, 306, 307, 316, 318, 334, 354
Airborne Digitac: 318
Airborne, Flight Testing of: 316
Differential Analyzer Design: 299
Electrostatic Reading of Perforated Media: 302
Germanium Tape Reader: 301
Information Rate: 289
JAINCOMP Design: 300
Logical Circuitry: 306
Magnetic Core Circuits: 305
Magnetic Core Materials: 303
Magnetic Core Selection Systems: 304
Miniaturized Diodes: 254
Optimized Data Encoding: 293
Quartz Delay Lines: 354
Role in Automatic Control Systems: 298
System Compensation: 334
Transistor Shift Registers: 307
Control System for Digital-to-Shaft Positioning: 335

D

Data Interpretation, Telemetry: 314
Delay Cables, Magnetic-Core: 258
Delay Lines, Quartz: 354
Delay Lines, Ultrasonic: 353
Diode for Miniaturized Computer Applications: 254
Diodes: 269
Silicon: 269
Diplexing Filters: 404
Directional Couplers: 403
Direction Finder for Guided Missiles: 311
Diversity System for UHF Ship-to-Air Communication: 395

E

Electrolytic Capacitors, Improvements: 259
Electron Beams: 279
Positive Ion Traps for Control: 279
Electronics in Nuclear Industry: 416
Electron Tubes: 254, 263, 264, 265, 266, 267, 273, 274, 275, 276, 277, 278, 279, 280, 281, 282, 329, 369, 390
Backward-Wave Oscillator: 281
Cathode: 263, 264
Hollow: 263
Machined Tungsten: 264
Cathode-Ray, For Facsimile: 390
Color Picture: 265, 369
Post Acceleration: 265
Single Gun: 369
Decade Counter: 266
Diode, Miniaturized: 254
Computer Applications: 254
Klystrons: 280
Multipactor Effect: 280
Magnetron Modulator: 329
Magnetron, Voltage Tunable: 278
Positive Ion Traps for Beam Control: 279
Storage Tubes: 273, 274, 275, 276, 277
Half-Tone Picture: 273, 274
High-Speed Dark-Trace: 275
Large Capacity: 276

Electron Tubes, *Cont'd.*

Noise Limitations: 277
Thyratron: 267
Current Interruption by Grid: 267
Traveling Wave: 282
Twin Helices: 282
Electrostatic Reading of Perforated Media: 302
Engineering Based on Biological Design: 411, 412, 413, 414, 415
Engineering Management Philosophy: 453
Equalizers: 233, 234, 235, 236, 237
Amplitude, Limitations on: 233
In the Time Domain: 237
Minimum Phase Design: 236
RLC Ladder Networks: 234
Video Cables: 235
Expansion Chamber for Red Cell Measurements: 407

F

Facsimile: 387, 388, 389, 390
Applications in USAF: 389
Cathode-Ray Tube Applications: 390
International Radiophoto Operation: 388
Systems: 387
Failure Prediction Technique: 448
FCC Rules and Propagation Data: 229
Feedback: 240
Time Varying Systems: 240
Ferrite Core Antennas: 361
Ferrite Plates at X-Band: 397
Filters: 284, 285, 288, 404
Comb: 285
Television Signals: 285
Detecting Pulse Signals in Noise: 288
Diplexing: 404
Optical: 284
Fixed-Beam Approach System: 319
Flight Directors, Role of: 320
Flow Measurement and Control, Electronic: 440
Frequency Divider, Digital: 435, 436
Frequency-Shift-Key Circuit Design: 393
Frequency Standard: 432, 434
Bell System: 434
Company-Owned: 432
Frequency Synthesizer: 435

G

Generator: 287, 308, 446
Impulse: 308
Receiver Measurements: 308
Noise: 287, 446
Germanium Tape Reader: 301
Guided Missile Instrumentation: 311, 313
Direction Finder: 311

H

High Fidelity and Hearing: 340
High Fidelity in Radio Broadcasting: 350
Human Engineering: 412

I

Impedance Bridge, UHF: 429
Impedance Recorder, Complex: 352
Impedance Measurements at 400-1600 Mc: 431
Information Rate of Digital Computers: 289
Information Theory: 226
Propagation through Time Varying Media: 226
Information Theory in Biology and Engineering: 413
Information Theory, Past and Future: 283
Interference-Level Prediction: 394
Interference-Level Prediction: 394
Isotropic Variable Index Media: 224

K

Klystrons: 280
Multipactor Effect: 280

L

Lenses: 220
Surface Matching: 220

Limiting, Effects on Signal Content: 297
Logical Nets, Symbolic Methods in Design: 294
Los Alamos Nonreactor Electronics: 418
Loudspeakers: 339, 348
High-Fidelity: 339

M

Magnetic Amplifiers: 362
Core Materials: 262
Magnetic Core Circuits: 305
Magnetic-Core Delay Cables: 258
Magnetic Core Materials for Computers: 303
Magnetic Core Selection Systems: 304
Magnetic Tape Equipment for Telemetry: 317
Magnetic Tape Recording of Color TV: 381, 382
Magnetron Modulator: 329
Magnetron, Voltage Tunable: 278
Maintenance Effect on Reliability: 253
Microphones: 337, 343, 347
Distant Pickup: 337
Miniature Unidirectional: 343
Microwave Measurements with Lossy Variable Short Circuit: 402
Missile System Evaluation: 450
Mixer for UHF Television: 374
Mobile Communications: 383, 384, 385, 386
450-470 Mc: 385
Impulse Noise in FM Receivers: 383
Petroleum Industry: 384
Public Utility System: 386
Modulation: 323
Pulse-Position Unit: 323
Modulation Power Requirements: 315
Multiplexing Power Requirements: 315
Music Enhancement by Reverberation: 338

N

N-1 Compass System: 322
Navaglobe Navigation System: 321
Navigation System, Long-Range: 321
Networks: 234, 243, 244, 245, 246, 247
Band-Pass Attenuation and Phase Functions: 246
Constant Time Delay, with Low Q Elements: 247
RC Ladder Synthesis: 244
Resistance Terminated: 245
RLC Ladder Synthesis: 234
Synthesis Techniques: 243
Neutron Depth Dose in Tissue: 409
Noise: 287, 288, 290, 296, 297, 383, 445, 446
Amplitude Distribution Analysis: 445
Detecting Pulse Signals: 288
Effects of Amplitude Limiting: 297
Generator: 287, 446
Impulse: 383
In FM Receivers: 383
Induced Grid: 296
Suppression in Speech: 290
Nonreciprocal Microwave Components: 396
Nuclear Industry, Electronics in: 416
Nuclear Reactor: 422, 423, 424, 425, 426
Electronics: 422
Instruments: 425
Power Plant Control Systems: 426
Safety Aspects of Control Circuitry: 424
Simulators: 423

O

Oak Ridge Nonreactor Electronics: 418
Operations Research: 454, 455
Organization: 454
Training for: 455
Optical Filters: 284
Organization for Operations Research: 454
Oscillators: 257, 281, 401, 428, 433
Backward-Wave: 281
Precision Components for: 257
Standard-Frequency Controlled: 433
Tunable Microwave: 401
X-Band Rapid-Sweep: 428

P

Personnel Selection and Training: 456, 457, 458
 Government Viewpoint: 458
 Industry Viewpoint: 457
 University Viewpoint: 456
 Petroleum Mobile Communications: 384
 Phase Detector for Color Reference Oscillator: 366
 Phase Measurements in Video Range: 427
 Photosensitive Germanium Devices: 441
 Piezoelectric Coefficient of Plates: 355
 Pinhole Camera, Gamma Ray: 406
 Polarization: 228
 Waves Reflected from Ionosphere: 228
 Potentiometers, Precision Wire-Wound: 261
 Pulse Distribution System for TV: 372
 Punch Card for Automatic Control: 438

Q

Quartz Crystals, Synthetic: 256
 Quartz Delay Lines: 354

R

Radar: 327, 330
 Stereo "3D" Indicating Systems: 330
 Track-While-Scan: 327
 Radome Tester: 310
 Receivers: 312, 362, 367, 370, 383
 Color Television: 367, 370
 Nonstandard Primaries: 367
 Errors, Effects on Color Reproduction: 370
 FM: 383
 Impulse Noise: 383
 Microwave Radiometric: 312
 Transistor, Broadcast: 362
 Red Cell Permeation by Water, Measurements: 407
 Reference Cavity for C Band: 405
 Reflectors: 223
 Flat Microwave, Gain of: 223
 Reliability Effect of Maintenance on: 253
 Reliability Improvement by Field Performance Analysis: 447
 Reliability in Airborne Systems: 449
 Return on Investment: 451
 Reverberation for Music Enhancement: 338
 Room Acoustics: 349

S

Scattering of Waves by Wires and Plates: 218
 Secrecy and Electronic Engineering: 417
 Servomechanisms, Biological: 415
 Servosystem, Optimization: 336
 Silicon Diode: 269
 Skin Thickness Measurement: 410
 Slot Radiation Conductance: 398
 Spectrum Analyzer: 430

Speech, Clipped: 342
 Speech Noise Suppression: 290
 Speech Sound Recognition: 291
 Stereophonic Sound for Movies: 341
 Storage Tubes: 273, 274, 275, 276, 277
 Half-Tone Picture: 273, 274
 High-Speed Dark-Trace: 275
 Large Capacity: 276
 Noise Limitations: 277
 Strip Transmission Line: 398, 399, 400
 High-Q Components: 399, 400
 Slot Radiation Conductance: 398
 Subminiaturization Techniques: 255
 Symbolic Methods in Logical Net Design: 294

T

Telemetering: 314, 317, 323, 324, 325, 326, 332
 Antenna for Multiple Operation: 326
 Data Interpretation: 314
 Magnetic Tape Equipment: 317
 Proportional Data Transmission: 332
 Pulse Position Modulation Unit: 323
 Transmitter: 324, 325
 Teletype, Predicted-Wave System: 392
 Television: 229, 230, 231, 232, 265, 285, 364, 366, 367, 368, 369, 370, 371, 372, 373, 374, 375, 376, 377, 378, 379, 380, 381, 382
 Antenna Problems at UHF and VHF: 232
 Clamp Circuit: 373, 378
 Color: 265, 366, 367, 368, 369, 370, 376, 377, 378, 379, 380, 381, 382
 Clamp Circuit: 378
 Distortion in Sequential Displays: 368
 Fidelity in Receivers with Nonstandard Primaries: 367
 Film Considerations: 379
 Film Scanner: 376, 377
 Characteristics: 377
 Circuits: 376
 Magnetic Tape Recording: 381, 382
 Phase Detector for Color Reference Oscillator: 366
 Picture Tube: 265, 369
 Post Acceleration: 265
 Single Gun: 369
 Reproduction, Effects of Receiver Errors on: 370
 Storing on Black-and-White Film: 380
 Comb Filters: 285
 FCC Rules and Propagation Data: 229
 Pulse Distribution System for Network Studio: 372
 Transmitter Transfer Switch: 375
 UHF Mixer: 374
 UHF Propagation: 230, 231
 Vertical Deflection Circuits: 364
 Automatic Damping: 364

Television, *Cont'd.*

WOR-TV Antenna System: 371
 Threshold Detection: 295
 Thyatron: 267
 Current Interruption by Grid: 267
 Time Series Analysis: 292
 Autocorrelated Error Terms: 292
 Tissue, Neutron Depth Dose: 409
 Training for Operations Research Groups: 455
 Transducers: 241
 Linear, Interconnection of: 241
 Transducers, Biological: 414
 Transistors: 248, 249, 250, 251, 252, 268, 270, 271, 272, 307, 362, 363, 442, 443, 444
 Accuracies in Sweep Measurements: 272
 Alloyed-Junction: 271
 Broadcast Receiver: 362
 Circuits: 248, 249, 250, 251, 252
 Frequency Scanner: 443
 High Power: 268
 Noise Test Set: 444
 Shift Registers: 307
 Small-Signal Parameters: 270
 Surface-Barrier: 363
 Testing: 442
 Transmission Line: 398, 399, 400
 Strip: 398, 399, 400
 High-Q Components: 399, 400
 Slot Radiation Conductance: 398
 Transmitters: 324, 325
 Telemetering: 324, 325
 Traveling-Wave Tubes: 282
 Twin Helices: 282
 Tuning, Wide-Range Systems: 365
 U
 Ultrasonics: 351, 353, 356, 357, 358, 359, 360
 Action on Nerve Tissue: 359
 Application to the Brain: 358
 Burglar Alarm: 351
 Delay Lines: 353
 Effect on Electrolytes and Electrode Processes: 357
 Effects on Living Cells: 360
 Industrial Uses: 356
 W
 Wave Scattering by Wires and Plates: 218
 Wave Propagation: 226, 227, 229, 230, 231
 Far Beyond Horizon: 227, 231
 At 100 Mc: 227
 At UHF: 231
 FCC Rules and Data: 229
 Through Time Varying Media: 226
 UHF-TV Band: 230
 X
 X-Ray Pictures in Color: 408



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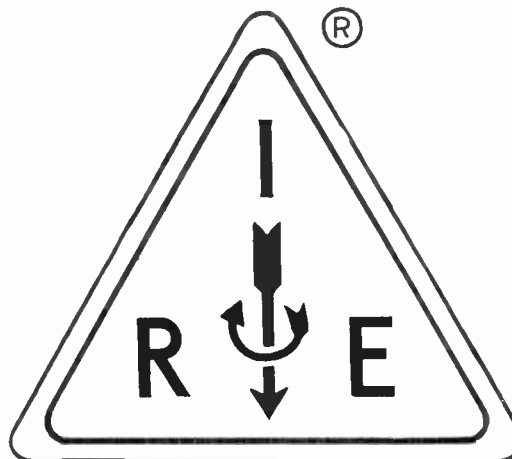
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1954 INDEX



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tronics
Antennas and Propagation
Audio
Broadcast and Television Receivers
Circuit Theory
Communications Systems
Component Parts
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TABLE OF CONTENTS

General Information.....	Cover II	Professional Group on Engineering Management.....	Page 12
Combined Index to Authors.....	Page 3	Professional Group on Information Theory.....	Page 12
Contents of IRE Transactions.....	Page 5	Professional Group on Instrumentation....	Page 13
Professional Group on Aeronautical and Navigational Electronics.....	Page 5	Professional Group on Microwave Theory and Techniques.....	Page 13
Professional Group on Antennas and Propa- gation.....	Page 5	Professional Group on Nuclear Science....	Page 14
Professional Group on Audio.....	Page 6	Professional Group on Quality Control....	Page 14
Professional Group on Broadcast and Tele- vision Receivers.....	Page 7	Professional Group on Radio Telemetry and Remote Control.....	Page 14
Professional Group on Circuit Theory.....	Page 8	Professional Group on Ultrasonic Engineer- ing.....	Page 15
Professional Group on Communications Systems.....	Page 8	Professional Group on Vehicular Communi- cations.....	Page 15
Professional Group on Component Parts....	Page 9	Index to Subjects.....	Page 16
Professional Group on Electronic Devices..	Page 10	Nontechnical Index.....	Page 20
Professional Group on Electronic Com- puters.....	Page 11		

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COMBINED INDEX TO AUTHORS

Code numbers refer to papers as given in the contents list for each Professional Group. The letters in the code number designate the particular Professional Group, as follows:

ANE: Aeronautical and Navigational Electronics
AP: Antennas and Propagation
A: Audio
BTR: Broadcast and Television Receivers
CT: Circuit Theory
CS: Communication Systems
CP: Component Parts
ED: Electronic Devices
EC: Electronic Computers

EM: Engineering Management
IT: Information Theory
I: Instrumentation
MTT: Microwave Theory and Techniques
NS: Nuclear Science
QC: Reliability and Quality Control
TRC: Radio Telemetry and Remote Control
UE: Ultrasonics Engineering
VC: Vehicular Communications

A

Adam, H. R.: CS52
Adler, R.: BTR25
Ahlert, R. H.: ED37
Albright, W. G.: AP116
Aldrich, R. W.: ED29
Alexander, S. N.: EC25
Allaire, R. P.: ED83
Allen, E. W.: CS44
Allen, J. E.: BTR26
Allison, W. M.: CP5
Alsberg, D. A.: ED95
Altman, J. L.: MTT37
Ammerman, C. R.: AP115
Amos, B.: BTR31
Anderson, E. J.: BTR34
Angelakos, D. J.: MTT39
Apelbaum, J.: ED82
Applegarth, A. R.: AE41
Armstrong, D. B.: CT29
Arndahl, G. M.: EC51
Arnett, H. D.: ED80
Artzt, M.: A81
Astrahan, M. M.: EC51
Atkinson, J. F.: VC39
Avins, J.: BTR21, BTR41
Ayres, H. F.: ED112

B

Ba Hli, F.: CT39
Bailey, S. L.: BTR38
Balamuth, L.: UE10
Balser, M.: IT55
Bangert, J. T.: ED26
Barton, L. E.: BTR17
Bauer, B. B.: A75, A82
Beck, A. C.: CS32, MTT22
Bedrosian, S. D.: CS61
Belevitch, V.: CT45, CT58
Bell, H.: EC32
Bennett, B. J.: CT43
Bereskin, A. B.: A65
Bernstein, M. S.: MTT50
Binns, J. E.: NS1
Biondi, F. J.: ED60
Birdsall, C. K.: ED94
Birdsall, T. G.: IT65
Bittman, C. A.: ED77
Bixler, O. C.: A60
Blachman, N. M.: IT48
Blake, R. F.: TRC5
Bloomsburg, R. A.: BTR45
Boff, A. F.: I-19
Bogan, L. B.: CS60
Booth, R. E.: ED54
Boughtwood, J. E.: CS66
Boyd, J. A.: ED119
Boykin, R. S.: CS61
Brewer, G. R.: ED94
Briggs, T. H.: ED42
Brooks, H. B.: I-20
Brown, R. N.: CP6

Brown, S. P.: CS70
Brueckmann, H.: CS71
Burnham, J.: CP2
Bushby, T. R. W.: AE36
Buss, R. E.: I-22
Buss, R. R.: ED85
Butterfield, F. E.: AP133

C

Cachens, J. C.: AE34
Cain, C. W.: BTR29
Callahan, J. L.: CS12
Cameron, C. F.: CP8
Capelli, M.: AE31
Caprarola, L. J.: ED66
Carlin, H. J.: CT35
Carman, J. N.: ED77
Caywood, W. P., Jr.: IT49
Cervenka, F. J.: CS51
Chaney, J. G.: AP117
Chapin, E. W.: BTR48
Chapman, R. D.: CS23, MTT23
Chittick, K. A.: BTR40
Chlavin, A.: AP128
Clark, M. A.: A57
Clark, N., Jr.: UE8
Clark, W. A.: IT58
Claypool, W. S.: VC36
Clement, L. M.: BTR37, QC12
Cohen, R. M.: ED31
Cohn, S. B.: MTT41
Coleman, A. F.: EM11
Coleman, H. P.: AP124
Collins, A. L.: ED126
Comerci, F. A.: A80
Comley, A.: EC34
Comuntzis, M. G.: CP11
Convert, G.: ED116
Corput, V. D., Jr.: CS38
Cottingham, J. G.: NS3
Crain, C. M.: AP114
Crain, C. W.: ED24
Creedon, J. E.: ED58
Cronshagen, A.: QC18
Crownover, J. W.: CP12
Cruser, V. I.: CS69
Crytzer, S.: ED35
Csepely, J. A.: CP7
Cunningham, W. J.: EC48
Cysack, F. H.: CS67

D

Dalman, C. G.: ED105
Danielson, W. E.: ED121
Danos, M.: MTT47
Danser, J. W.: CS59
Davies, G. L.: A77
Davis, L., Jr.: ED88
Davis, R. C.: IT40
Day, C.: ED52
Deschamps, G. A.: CS33, MTT33

Deutch, R.: IT44
de Wolf, F. C.: CS42
Dietsch, C. G.: CS20
Dobischek, D.: ED74
Dodrill, G. E.: VC39
Doehler, O.: ED116
Donald, D. D.: CS57
Dorr, R.: BTR30
Drenick, R. F.: IT63
Drennan, J. E.: ED57
Dropkin, H. A.: AE34
Duffendack, O. S.: ED39
Dukat, F. M.: QC19
Dwork, L.: ED27
Dyke, E.: CS29, MTT38

E

Edwards, E. V.: ED63
Elias, P.: IT53
Elliott, R. S.: AP122
Ellis, C. E.: EM4
Erath, L. W.: TRC8
Espersen, G. A.: ED128
Everitt, W. L.: A58

F

Fagen, M. D.: UE7
Fairbanks, G.: A58
Farley, B. C.: IT58
Feinstein, A.: IT51
Feinstein, J.: AP115, AP121
Felton, W. W.: AE29
Ferris, W. R.: ED110
Finison, H. J.: EM7
Fink, D. G.: BTR41
Finke, H. A.: I-21
Firle, T. E.: ED87
Fischer, J. T.: A81
Fitch, J. L.: I-22
Foss, F. A.: EC45
Fossier, M. W.: EC50
Foster, R. M.: CT60
Fowler, A. B.: ED92
Fowler, V. J.: AP130
Fox, G. W.: CS29, MTT29
Fox, W. C.: IT65
Fraser, W. C. G.: AP131
Frayne, J. G.: A70
Freas, R. R.: AE27
Freely, J.: ED74
Freeman, H.: EC26
Fried, C.: ED115
Fyler, N. F.: ED24
Fyler, N. J.: BTR29

G

Gabor, D.: CT53
Gamson, E. R.: CP14, QC20
Gates, P. E.: ED55
Gaw, N.: CP9
Gerhard, J. R.: AP114
Ghose, R. N.: AP117

Gifford, R. P.: VC37
Gillis, T. B.: ED35
Gilman, G. W.: CS40
Giodano, P. J.: MTT38
Given, I. K.: CS56
Glenn, A. B.: BTR42
Goddard, C. T.: ED84
Goetz, D.: UE9
Golay, M. J. E.: IT52
Goldberg, H.: EM2
Gordon, B. M.: EC27
Gottschalk, W. M.: ED107, ED128
Gow, J. D.: BTR30
Grammer, G.: BTR35
Gratien, J. W.: A61, A62
Green, E.: CT57
Greenhow, C. R.: ED111
Greenslit, C. L.: AE37
Grien, G. W., Jr.: AE26
Grinn, H. H.: ED109
Grinich, V.: CT50
Gruen, W. J.: BTR20
Gruenberg, H.: AP132
Guillemin, E. A.: CT25, CT48
Gyorgy, E. M.: MTT51

H

Hall, J. D.: ED55
Halliday, R. G.: CT19
Halvorsen, R. L.: CS24, MTT24, MTT29
Hammer, H.: EC49
Hanley, T. E.: ED61
Harp, M. C.: CS25, MTT25
Harris, B.: BTR21
Harris, W. A.: ED120
Hasse, A. P.: ED44, ED45
Hattery, L. H.: EM10
Haus, H. A.: ED124
Hazen, D. F.: CS59
Hedrich, A. L.: UE6
Henesian, A.: CP14, QC20
Herrick, J. F.: UE1
Hetland, G., Jr.: ED85
Hickle, J. C.: ED45
Higginson, G. M.: CS50
Hill, F. R.: CS35
Hilliard, J. K.: A84
Hoffmann, J. P.: CS63
Honey, R. C.: MTT35
Hopfer, S.: MTT54
Hopkins, M.: EM8
Horton, C. E.: ED44
Horvath, J. S.: BTR21
Houghton, W. D.: A81
Howard, R. C.: EC43
Huang, C.: ED27
Huber, G. H.: CS64
Huer, C.: BTR25

Continued on page 4

Combined Index to Authors

Huggins, W. H.: CT46, ED127
Hurley, R. B.: CP13
Huskey, H. D.: EC51

I

Ito, Y.: AE43

J

Jacobs, H.: ED74
Jaeger, R. P.: A58
James, K.: BTR22
Johnson, C. M.: MTT49
Johnston, J. H.: CS61
Jones, E. M. T.: AP129
Jones, P.: CT20
Jones, R. E.: AP125
Jones, W. R.: ED30, ED10, QC16
Joyce, M. V.: CT33

K

Kalmus, H. P.: AE34, UE6
Kaufman, W. M.: IT49
Kautz, W. H.: CT37
Kearny, T. J.: UE4
Kebby, M. H.: CS25, CT25
Kelleher, K. S.: AP124
Kellogg, E. W.: A72
Kenn, V.: ED50
Kiebert, M. V., Jr.: TRC2
King, W. C.: MTT45
Kingsley, H. F. N.: CS62
Kline, M.: CP9
Klopfenstein, R. W.: AP126
Klotter, K.: CT26, CT52
Knecht, W.: ED51
Knoblauch, A. F.: A71
Knoll, M.: ED78
Kohl, W. H.: ED31, ED68, ED70
Kovach, L. D.: EC34
Kroll, N. M.: MTT50
Krulce, R. L.: ED113
Krull, A. R.: ED99
Krusen, F. H.: UE1
Ku, Y. H.: CT51
Kumpfer, B. D.: ED62
Kyser, R. H.: ED80

L

Laemmel, A. E.: IT47
Lanciani, D. A.: MTT40
Lane, J. F.: AE33
LaPlante, R. A.: ED108, ED128
Larkin, K. I.: ED102
Lashinsky, H.: MTT47
Lawton, C. S.: CS58
Lee, L. K.: CP1
Lohan, F. W.: TRC7
Lehr, C. G.: ED126
Leiner, A. L.: EC25, EC30
Lenahan, J. J.: CS27, MTT27
Lesk, I. A.: ED29
Levine, D.: AE28
Levine, R. H.: CS68
Levy, I. E.: ED49
Lewis, N. W.: CT56
Libbey, R. L.: A83
Linville, T. M.: EM12
London, A. L.: ED86
Loveridge, L. E.: A61
Lubkin, S.: EC40
Lyman, R. C.: IT49
Lynch, W. W.: CS53

M

Mack, A.: CS68
MacQuivey, D. R.: CS45
Maggs, C.: ED59
Maginnis, W. P.: CS21, MTT21
Mallinckrodt, A. J.: IT50
Maloney, E.: ED48

Mandelbrot, B.: IT46
Mannheimer, D.: AE40
Manning, L. A.: AP123
Maron, M. E.: EC29
Martin, D. W.: A71, A73
Mason, S. J.: CT31
Mauchley, J. W.: EC51
May, J. E.: UE3
Mayo, B. R.: ED109
McBride, W. J., Jr.: ED64
McClain, E. F.: ED110
McCool, C. D.: ED46
McDonough, S. L.: AE44
McGill, W. J.: IT60
McLeod, W. W., Jr.: ED101
McHwain, K.: BTR27
McMahon, M. E.: ED87
McRuer, D. T.: CT19
Meaker, L. S. F.: CS55
Meissner, P.: ED47
Metzger, S.: CS30, MTT30
Meyer, M. A.: EC27, EC31
Michael, F. R.: ED42, ED43
Middlekamp, L. C.: BTR48
Middleton, D.: ED106, IT39, IT62

Miessner, B. F.: A76
Miles, P. D.: CS43
Miller, W. A.: CS13
Miller, W. F.: CS64
Mittra, A. P.: AP125
Mockus, E.: ED35
Moers, C. N.: IT61
Moore, E. F.: EC51
Morgan, A. R.: A81
Morgan, H. K.: A74
Morgan, K. A.: TRC5
Morgan, R. B.: ED76
Morrill, C. D.: EC35
Moulton, A. B.: CS14
Mueller, R.: ED104
Muller, D. E.: EC38
Mushiaki, Y.: AP135

N

Nail, J. J.: AP113
Needle, J. S.: ED96
Neely, G. M.: CS49
Nelson, E. C.: EC39
Nelson, J. H.: CS15
Nethercot, A. H., Jr.: MTT46
Neubauer, J. R.: VC33
Newhouse, R. C.: AE39
Nicola, R. N.: EC27
Noble, J. J.: A84
Nordahl, J. G.: CS65
Norman, R. Z.: IT67
Notz, W. A.: EC30

O

Okress, E. C.: ED65
Okun, A. M.: CP10
Olson, H. F.: A81
Olthuis, R. W.: CS34, MTT34
Orchard, H. J.: CT57
Ornstein, W.: VC35
Oswald, J.: CT59

P

Page, C. H.: CT47
Palmer, H. W.: ED27
Pan, W. Y.: BTR18, ED28
Pankove, J. I.: ED25
Pappas, N. L.: I-23
Pardue, D. R.: UE6
Parsons, E.: EC26
Partridge, G. R.: EC46
Pasek, D. M.: AE32
Paul, F. A.: CP1, CP3
Peck, S. C.: BTR19
Peeler, G. D.: AP124
Pennell, E. S.: UE5

Percival, W. S.: CT60
Peterson, W. W.: IT65
Petry, C. A.: CS54
Pettit, J. M.: CT36
Peyser, W. P.: I-24
Phinney, T. W.: A64
Pierce, J. R.: ED114
Place, H.: CS21, MTT21
Plotkin, M.: NS3
Podolsky, L.: CP4
Polimerou, L. G.: EC42
Pope, W. A.: AP134
Porter, W. A.: CS41
Post, E. A.: AE38
Powell, F. H.: QC13
Pratt, H.: CS39
Price, R.: IT64
Price, R. L.: A66
Pritchard, W. L.: ED102
Pryslak, N. E.: ED67
Pugsley, D. W.: BTR39
Purington, E. S.: A69
Purinton, H. G.: EM3
Putzrath, F. L.: A63
Pynn, R. D.: CS22, MTT22

Q

Quirk, C.: BTR31

R

Rabinowitz, S. J.: MTT48
Raka, E. C.: NS3
Read, A. H.: CS46
Records, J. K.: ED109
Reed, I. S.: IT54
Reed, R. H.: AP127
Reingold, L.: ED63
Reza, F. M.: CT24, CT29
Rhoads, A. S., Jr.: ED105
Rhodes, H. A.: CS31, MTT31
Ribe, M. L.: CS70
Richman, D.: BTR28
Riddle, F. M.: TRC6
Riddle, R. L.: AP116
Ridenour, L. N.: EC51
Rideout, V. C.: EC32
Rittner, E. S.: ED37, ED38
Rives, T. C.: EM1
Roach, J. F.: ED87
Roberts, F.: BTR47
Roberts, T. E., Jr.: AP119
Roberts, W. O.: CS16
Roberts, W. K.: BTR48
Robertson, S. D.: MTT52
Robinson, K.: ED81
Rochester, N.: EC51
Rosen, H. A.: EC50
Rothe, H.: ED125
Rothlein, B. J.: ED92
Rothstein, J.: EM5, IT56
Roveto, J. P.: BTR46
Rowe, E. G.: QC17
Rowe, W. E.: BTR29, ED24
Rubin, L. G.: ED88
Rudd, J. B.: CS19
Rudisuhle, E. J.: CS25, MTT25
Rush, J. H.: CS16
Rutledge, W. C.: ED38
Ryan, R. D.: EC37

S

Samuel, A. L.: EC51
Sandretto, P. C.: AE30
Savant, C. J.: EC43
St. John, G. E.: ED118
Schaffner, J. S.: CT28
Schiesler, A.: A59
Schlaack, N. F.: CS17
Schlesinger, K.: BTR24, BTR49
Schramm, C. W.: CS64
Schwartz, L. S.: QC15
Schwartz, R.: A80

Schwartz, R. F.: MTT42
Schwarz, R. J.: CT55, CT56
Selsted, W. T.: A78
Seybold, A. M.: ED41
Shanahan, W. J.: AE32
Shannon, C.: EC51
Shannon, W. W.: BTR23
Sharaf, H. F.: UE2
Sharpless, W. M.: MTT53
Shekel, J.: CT30
Shelton, E. J.: ED103
Shepherd, W. G.: ED33
Sherman, H.: EM6
Shiowitz, M.: TRC3
Shrader, T. M.: ED71
Sichak, W.: AP113
Siegert, A. S. F.: IT38
Silbertstein, R.: AP120
Silva, L. M.: CT23
Silver, M.: ED90, ED91
Silverman, D.: CP9
Silverman, R. A.: IT55
Skar, R. C.: BTR50
Skellett, A. M.: A61
Skinner, L. V.: CT32
Slepian, D.: IT42
Smith, F. C., Jr.: TRC8
Smith, G. A.: CT27
Smith, J. L.: EC30
Smith, O. J. M.: CT17, IT45
Smullen, L. D.: ED115
Snyder, R. H.: A78
Sollenberger, T. E.: IT50
Soltes, A. S.: EC33
Spanke, W. F.: CS48
Spencer, R. C.: CT34
Sprague, J. K.: CP4
Stansel, F. R.: I-25
Stegan, R. J.: AP127
Stello, P. E.: ED77
Stephanz, K. R.: ED69
Sterling, G. E.: BTR36
Stiles, K. P.: CS18
Stone, R. P.: ED79
Stout, T. M.: CT22
Straub, W. D.: ED88
Strum, P. D.: ED128, MTT36
Stubbs, G. S.: NS2
Swets, J. A.: IT66

T

Talkin, A. I.: CT44
Tallman, O. G.: EM13
Talpey, T. E.: MTT44
Tanaka, I.: AE43
Tanner, R. L.: AE42
Tanner, W. P., Jr.: IT66, IT67
Taub, N. B.: CT38
Tharp, N. B.: CS26, CT26
Thomas, D. E.: ED89
Thomson, W. E.: CT55
Todd, F. C.: ED36, ED37
Torsch, C. E.: BTR32, BTR33
Truxal, J. G.: CT42

U

Udelson, B. J.: ED58
Ullman, F. G.: ED75

V

Vance, R. L.: ED59
van der Ziel, A.: ED73, ED93
Van Meter, D.: IT62
Varallo, F. A.: I-26
Veronda, C. M.: ED97
Vilbig, F.: A68
Von Ohlsen, L. H.: ED117

W

Wait, J. R.: AP134
Wakefield, E. H.: ED56

Continued on page 5

Combined Index to Authors

Walters, L. G.: CT21
Wargo, P.: ED72
Watanabe, S.: IT59
Waters, W. E., Jr.: ED112
Watson, D.: EM9
Webber, S. F.: ED98
Webster, E. M.: CS47, VC34
Webster, F. D.: CS36
Webster, S. K.: A67
Weinberger, A.: EC30
Weisner, J. B.: ED100

Weiss, M. T.: MTT51
Weisz, W. S.: VC38
Welch, P.: QC17
Weller, R.: QC21
West, C. F.: TRC1, TRC4
Westcott, J. H.: CT41
Whinnery, J. R.: ED123
White, R. C.: ED54
White, W. D.: CT49
Whittier, R. J. E.: QC14
Widrow, B.: EC47

Wiesner, J. B.: EC51
Williams, C. E.: AP114
Wilpon, S.: A80
Wilson, R. D.: ED53
Wimberly, F. T.: AE33
Winkler, M. R.: A79
Winkler, S.: CT38
Woll, H. J.: A63
Woods, R. E.: ED112
Woodward, J. G.: A81
Wyler, E. N.: ED36

Y
York, D. E.: VC40
Youla, D. C.: IT43
Young, G. O.: IT41
Young, K. D.: CS60
Young, W. J.: VC41
Z
Zeirldt, C. H., Jr.: ED32
Zenel, J.: A81
Zillis, G. S.: CS59
Zweizig, J. R.: EC41

CONTENTS OF IRE TRANSACTIONS

Professional Group on Aeronautical and Navigational Electronics

1954

Vol. ANE-1, No. 1, March, 1954

Index Number	Page
Special Announcement, <i>John E. Wilkinson</i>	1
Chairman's Report, <i>K. C. Black</i>	3
AE26 The Information Content of Air-Ground Messages, <i>George W. Grier, Jr.</i>	5
AE27 Airborne Loran Receiver—the AN/APN-70, <i>R. R. Freas</i>	17
AE28 Antenna Scan Considerations, <i>Daniel Levine</i>	26
AE29 Some Channel Allocation Problems in Air-Ground Voice Communications, <i>Walter W. Felton</i>	41

Vol. ANE-1, No. 2, June, 1954

Dayton Conference, <i>K. C. Black</i>	1
AE30 The Long Quest, <i>P. C. Sandretto</i>	2
AE31 Radio Altimeter, <i>M. Capelli</i>	3
AE32 Flare-Out Unit AN/APN-1 (XA-3) an Aid to Aircraft Instrument Landing, <i>D. M. Pasek and W. J. Shanahan</i>	7
AE33 The AN/APN-22 Radio Altimeter, <i>F. T. Wimberly and J. F. Lane</i>	8
AE34 Nonquantized Frequency-Modulated Altimeter, <i>H. P. Kaluus, J. C. Cacheris and H. A. Dropkin</i>	15
AE35 Corrections.....	22

Vol. ANE-1, No. 3, September, 1954

Index Number	Page
East Coast Conference, <i>K. C. Black</i>	1
AE36 A Maintenance Plan for Airborne Radio Equipment, <i>T. R. W. Bushby</i>	2
AE37 A Thunderstorm Avoidance Radar for Civil Aircraft, <i>C. L. Greenslit</i>	7
AE38 The Operational Applications of Airborne Radar, <i>E. A. Post</i>	15
AE39 Feedback Relations in Military Weapon Systems, <i>R. C. Newhouse</i>	24

Vol. ANE-1, No. 4, December, 1954

AE40 Airborne Radar as a Navigational Aid, <i>David Mannheimer</i>	2
AE41 A New Airborne DME Interrogator Designed for Stable Operation and Ease of Maintenance, <i>A. R. Applegarth</i>	10
AE42 High-Voltage Problems in Flush and External Aircraft HF Antennas, <i>R. L. Tanner</i>	16
AE43 Development of Ring Goniometer for Radio Direction Finders, <i>Ito and Tanaka</i>	20
AE44 Electronic Simulators for Study of Aircraft Flight Paths, <i>S. L. McDonough</i>	24

Professional Group on Antennas and Propagation

1954

Vol. AP-2, No. 1, January, 1954

Index Number	Page
News and Views.....	1
AP113 UHF Omnidirectional Antenna Systems for Large Air- craft, <i>W. Sichak and J. J. Nail</i>	6
AP114 A Preliminary Survey of Tropospheric Refractive In- dex Measurements for U. S. Interior and Coastal Regions, <i>C. M. Crain, J. R. Gerhardt and, C. E. Williams</i>	15
AP115 Some Stochastic Problems in Wave Propagation—	

PGAP-2, January, 1954 (Cont'd.)

Index Number	Page
Part I, <i>Joseph Feinstein</i>	23
AP116 A Preliminary Study of Fading of 100 Megacycle FM Signals, <i>R. L. Riddle and C. R. Ammerman</i>	30
AP117 VHF Field Intensities in the Diffraction Zone, <i>R. N. Ghose and W. G. Albright</i>	35
AP118 Mutual Impedance of Stacked Rhombic Antennas, <i>J. G. Chuney</i>	39

Continued on page 6

Transactions Index—5

Vol. AP-2, No. 2, April, 1954

Index Number	Page
Why Transactions? (Editorial).....	41
AP119 News and Views.....	42
An Experimental Investigation of the Single-Wire Transmission Line, <i>T. E. Roberts, Jr.</i>	46
AP120 Sweep Frequency Backscatter—Some Observations and Deductions, <i>Richard Silberstein</i>	56
AP121 Some Stochastic Problems in Wave Propagation—Part II, <i>Joseph Feinstein</i>	63
AP122 On the Theory of Corrugated Plane Surfaces, <i>R. S. Elliott</i>	71
AP123 Meteoric Radio Echoes, <i>L. A. Manning</i>	82

Vol. AP-2, No. 3, July, 1954

News and Views.....	91
AP124 Virtual Source Luneberg Lenses, <i>G. D. M. Peeler, K. S. Kelleher, and H. P. Coleman</i>	94
AP125 A Theoretical and Experimental Study of the Recombination Coefficient in the Lower Ionosphere, <i>A. P. Mitra and R. E. Jones</i>	99
AP126 Low Frequency Waves on Transmission Lines of Composite Section, <i>R. W. Klopfenstein</i>	103
AP127 Arrays of Closely-Spaced Nonresonant Slots, <i>R. J.</i>	

PGAP-3, July, 1954 (Cont'd.)

Index Number	Page
<i>Stegen and R. H. Reed</i>	109
AP128 A. New Antenna Feed Having Equal E- and H-Plane Patterns, <i>Alvin Chlavin</i>	113
AP129 Paraboloid Reflector and Hyperboloid Lens Antennas, <i>E. M. T. Jones</i>	119

Vol. AP-2, No. 4, October, 1954

News and Views.....	129
AP130 Analysis of Helical Transmission Lines by Means of the Complete Circuit Equations, <i>V. J. Fowler</i>	132
AP131 Radiation from a Vertical Dipole over a Stratified Ground, <i>J. R. Wait and W. C. G. Fraser</i>	144
AP132 A Waveguide Array for Solar Noise Studies, <i>H. Gruenberg</i>	147
AP133 Dielectric Sheet Radiators, <i>F. E. Butterfield</i>	152
AP134 Evaluation of Errors in an Eight-Element Adcock Antenna, <i>J. R. Wait and W. A. Pope</i>	159
Communications:	
AP135 An Exact Step-Up Impedance-Ratio Chart of a Folded Antenna, <i>Y. Mushiaki</i>	163

Professional Group on Audio

1954

Vol. AU-2, No. 1, January–February, 1954

Index Number	Page
PGA Chapter News	
San Diego Chapter Activities.....	1
Philadelphia Chapter Events.....	1
Summary of Available Tape Scripts.....	2
PGA Briefs.....	3
Audio News	
Department of Defense Symposium on Magnetic Recording.....	3
RCA Records Color Television on Magnetic Tape.....	4
A57 An Acoustic Lens as a Directional Microphone, <i>M. A. Clark</i>	5
A58 Method for Time or Frequency Compression-Expansion of Speech, <i>G. Fairbanks, W. L. Everitt and R. P. Jaeger</i>	7
A59 A Device for Time Expansion Used in Sound Recording, <i>H. Schiesser</i> (translation by <i>V. Ruvalds</i>).....	12
A60 Mechanical Components for Handling Magnetic Recording Tape, <i>O. C. Bixler</i>	15
A61 Electron-Beam Head for Magnetic Tape Playback, <i>A. M. Skellett, L. E. Loveridge, and J. W. Gratian</i>	23
A62 Investigation of Core Structures for the Electron-Beam Reproducing Head, <i>J. W. Gratian</i>	27

Vol. AU-2, No. 2, March–April, 1954

A63 A Note on Noise in Audio Amplifiers, <i>H. J. Wall and F. L. Putzrath</i>	39
PGA News	
Houston Chapter PGA, <i>L. A. Geddes</i>	42
How Much Distortion Can You Hear? <i>E. M. Jones</i>	42
PGA Briefs.....	43
A64 The Vagabond Wireless Microphone System, <i>T. W. Phinney</i>	44
A65 A High-Efficiency High-Quality Audio-Frequency Power Amplifier, <i>A. B. Bereskin</i>	49
A66 The Cascode as a Low Noise Audio Amplifier, <i>R. L. Price</i>	60
A67 An All-Transistor Hearing Aid, <i>S. K. Webster</i>	65

Vol. AU-2, No. 3, May–June, 1954

PGA News	
Chairman's Report 1953–54, <i>Marvin Camras</i>	71
Treasurer's Report 1953–54, <i>B. B. Bauer</i>	72

PGAU-2, May–June, 1954 (Cont'd.)

Index Number	Page
Editorial Committee Report 1953–54, <i>D. W. Martin</i>	72
PGA Sessions at National IRE Convention.....	73
PGA Briefs.....	74
PGA Chapter News	
Cleveland Chapter Has Stereo Sound Symposium, <i>Albert Preisman</i>	75
IRE Technical Committee News	
Sound Recording and Reproducing Committee, <i>M. S. Corrington</i>	76
A68 Visible Speech—Rotary Field—Coordinate Conversion Analyzer, <i>Friedrich Villbrig</i>	76
A69 Dynamic Amplifiers for Phonograph Reproduction, <i>E. S. Purington</i>	80
A70 Components and Mechanical Considerations for Magnetic Sound on 35 mm Film, <i>J. G. Frayne</i>	86
A71 A Loudspeaker Accessory for the Production of Reverberant Sound, <i>D. W. Martin and A. F. Knoblaugh</i>	95

Vol. AU-2, No. 4, July–August, 1954

A72 Comment on Flutter Standards, <i>E. W. Kellogg</i>	99
A73 High Fidelity in Musical Tone Production? <i>D. W. Martin</i>	102
PGA News	
Philadelphia Chapter Activities, <i>M. S. Corrington</i>	104
Cincinnati Field Trip to Dayton, <i>E. M. Jones</i>	104
PGA Briefs.....	105
IRE Technical Committee News	
Recording and Reproducing Committee, <i>M. S. Corrington</i>	106
A74 Natural Sound Reproduction, <i>H. K. Morgan</i>	106
A75 Equivalent Circuit Analysis of Mechano-Acoustic Structures, <i>B. B. Bauer</i>	112
A76 Frequency Modulation Phonograph Pickups, <i>B. F. Miessner</i>	121

Vol. AU-2, No. 5, September–October, 1954

PGA News	
Houston Chapter PGA, <i>Les Geddes</i>	131
Reports on Tapescripts Committee, <i>A. B. Jacobsen</i>	131
PGA Briefs.....	132

Continued on page 7

PGAU-2, September-October, 1954 (Cont'd.)

Index Number	Page
A77 Magnetic Recorders for Data Recording Under Adverse Environments, <i>G. L. Davies</i>	133
A78 Magnetic Recording—A Report on the State of the Art, <i>W. T. Selsted and R. H. Snyder</i>	137
A79 Cathode Bias Resistor for Class A ₁ Triode, <i>M. R. Winkler</i>	145
A80 Navy Standardization of ½-inch Magnetic Tape and Recorder-Reproducers, <i>F. A. Comerci, S. Wilpon and R. Schwartz</i>	146
Vol. AU-2, No. 6, November-December, 1954	
PGA News	
PGA at WESCON	155

PGAU-2, November-December, 1954 (Cont'd.)

Index Number	Page
PGA Chapter News	155
PGA Briefs	155
1954-1955 PGA Committees	156
A81 A System for Recording and Reproducing Television Signals, <i>H. F. Olson, W. D. Houghton, A. R. Morgan, J. Zenel, M. Artzt, J. G. Woodward and J. T. Fischer</i>	159
A82 Correction: Equivalent Circuit Analysis of Mechano-Acoustic Structures, <i>B. B. Bauer</i>	164
A83 A Miniature High-Gain Audio Amplifier, <i>R. L. Libbey</i>	165
A84 The "Lipstik" Condenser Microphone System, <i>J. K. Hilliard and J. J. Noble</i>	168
Cumulative Index for 1954, <i>W. R. Ayres</i>	176

Professional Group on Broadcast and Television Receivers

1954

PGBTR-5, January, 1954

Index Number	Page
Minutes of the Meeting of the Administrative Committee of the PGBTR.....	1
BTR17 An Experimental Transistor Personal Broadcast Receiver, <i>Loy E. Barton</i>	6
BTR18 Investigation of UHF Television Amplifier Tubes, <i>Wen Yuan Pan</i>	14
BTR19 A Disc-Seal Triode as a UHF Amplifier, <i>S. Christopher Peck</i>	31
BTR20 Test Generator for Horizontal Scanning AFC System, <i>Wolf J. Gruen</i>	36
BTR21 Improving the Transient Response of Television Receivers (Abstract), <i>J. Avins, B. Harris and J. S. Horvath</i>	44
BTR22 A Practical Adaptation of the Barnes Colorimeter for Kinescope-Screen Color Determination, <i>Kenneth James</i>	45
BTR23 The Barnes Colorimeter Applied to Television Quality Control, <i>William W. Shannon</i>	49
BTR24 Pulsed Envelope Detection of Color Signals, <i>Kurt Schlesinger</i>	53
BTR25 Color Decoder Simplifications Based on a Beam Deflection Tube, <i>Robert Adler and Charles Heuer</i>	64
BTR26 Beat Between Sound Carrier and Color Signal Components in a Television Receiver, <i>John E. Allen</i>	71
BTR27 Discussion of Paper "National Television System Committee Field Tests," <i>Knox McIlwain</i>	87
BTR28 The DC Quadricorrelator: A Two-Mode Synchronization System (Abstract), <i>Donald Richman</i>	94
BTR29 The CBS Colortron, A Color Picture Tube of Advanced Design (Abstract), <i>N. F. Fyler, W. E. Rowe and C. W. Cain</i>	95
BTR30 Operation of the Chromatron on NTSC Standards (Abstract), <i>J. D. Gow and R. Dorr</i>	96

PGBTR-6, April, 1954

BTR31 Dorman D. Israel.....	1
Factors Affecting the Correlation of TV Picture Quality Between Field and Laboratory Signals, <i>Bernard Amos and Carl Quirk</i>	2
BTR32 High Efficiency, Low-Copper Sweep Yokes with Balanced Transient Response, <i>C. E. Torsch</i>	17

PGBTR-6, April, 1954 (Cont'd.)

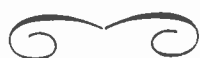
Index Number	Page
BTR33 Extension of the Balanced-Transient Response Principle to Color Television Yokes, <i>C. E. Torsch</i>	33
BTR34 NTSC Ad Hoc Committee on Amateur-Color TV Interference, <i>Earl I. Anderson</i>	36
BTR35 Industry-Amateur Cooperation, <i>George Grammer</i>	39
BTR36 Blights on the Radio Spectrum, <i>George E. Sterling</i>	42
BTR37 Television Receiver Interference Industry Record to Date, <i>L. M. Clement</i>	50
BTR38 Measurement and Standards for Control of Spurious Radiation, <i>Stuart L. Bailey</i>	59
BTR39 The Ignorance of the IF in TV Receivers, <i>D. W. Pugsley</i>	65
BTR40 Spurious Radiation from TV Receivers, <i>K. A. Chittick</i>	71
BTR41 Industry-Wide Cooperation Under the JTAC Spurious Radiation Program, <i>Donald G. Fink</i>	78
BTR42 Study of Noise Reduction by Feedback in Ultra-High Frequency Amplifiers, <i>A. B. Glenn</i>	82

PGBTR-7, July, 1954

Minutes of the Meeting of the PGBTR Administrative Committee.....	1
BTR43 UHF Tuner Design for 6BA4 Amplifier, <i>Ralph S. Brown</i>	4
BTR44 IF Amplifier Design for Color TV Receivers, <i>Jack Avins</i>	14
BTR45 The Measurement of Yoke Astigmatism, <i>R. A. Bloomsberg</i>	26
BTR46 Semiconductor Diodes for TV Receivers, <i>J. P. Roweto</i>	34
BTR47 A New Approach to Series Heater Strings for Television, <i>Frank Roberts</i>	39
BTR48 Interference to Color and Monochrome TV Receivers by Oscillator Radiation and Other CW Signals, <i>W. K. Roberts, L. C. Middlekamp and E. W. Chapin</i>	47

PGBTR-8, October, 1954

BTR49 The Vectroscope and Its Applications in Color TV, FM and Radio Navigation, <i>Kurt Schlesinger</i>	1
BTR50 A Method for Determining Q and Selectivity of Low-loss Parallel Resonant Circuits, <i>Robert C. Skar</i>	14



Professional Group on Circuit Theory

1954

Vol. CT-1, No. 1, March, 1954

<i>Index Number</i>		<i>Page</i>
	Abstracts of the Servomechanism Papers.....	1
CT17	Editorial, Trends in Feedback Systems, <i>Otto J. M. Smith</i>	2
CT18	Nonlinear Control Systems with Random Inputs, <i>Richard C. Booton, Jr.</i>	9
CT19	A Method of Analysis and Synthesis of Closed-Loop Servo Systems Containing Small Discontinuous Nonlinearities, <i>D. T. McRuer and R. G. Halliday</i>	19
CT20	Stability of Feedback Systems Using Dual Nyquist Diagram, <i>Paul Jones</i>	35
CT21	Optimum Lead-Controller Synthesis in Feedback-Control Systems, <i>Louis G. Walters</i>	45
CT22	On the Comparison of Linear and Nonlinear Servomechanism Response, <i>T. M. Stout</i>	49
CT23	Predictor Servomechanisms, <i>Lawrence M. Silva</i>	56
CT24	Conversion of a Brune Cycle with an Ideal Transformer into a Cycle without an Ideal Transformer, <i>F. M. Reza</i>	71
	Correspondence:	
CT25	What is Nature's Error Criterion?, <i>Ernst A. Guillemin</i>	76
CT26	Multi-Loop Nonlinear Systems, <i>K. Klotter</i>	76
CT27	Are Bibliographies Wanted? <i>G. Allan Smith</i>	77
	PGCT News Section.....	78

Vol. CT-1, No. 2, June, 1954

	Abstracts.....	1
CT28	Simultaneous Oscillations in Oscillators, <i>Johannes S. Schaffner</i>	2
CT29	Synthesis of Transfer Functions by Active RC Networks with Feedback Loops, <i>D. B. Armstrong and F. M. Reza</i>	8
CT30	Reciprocity Relations in Active Three-Terminal Elements, <i>Jacob Shekel</i>	17
CT31	Power Gain in Feedback Amplifiers, <i>Sam J. Mason</i>	20
CT32	Spectra of Waves with Periodic Modulation, <i>Leo V. Skinner</i>	26
	Correspondence:	
CT33	What is Nature's Error Criterion? <i>Maurice V. Joyce</i> ...	32
CT34	Network Synthesis and the Moment Problem, <i>Roy C. Spencer</i>	32
CT35	The Champions, <i>H. J. Carlin</i>	33
	PGCT News:	
	Minutes of Administrative Committee Meeting.....	34
	Reports of PGCT Committees.....	35
	Chapter News.....	39
	General Assembly of URSI.....	39

Vol. CT-1, No. 3, September, 1954

Abstracts of Papers in this Issue.....	1
--	---

PGCT-1, September, 1954 (Cont'd.)

<i>Index Number</i>		<i>Page</i>
CT36	Editorial—The International Scientific Radio Union, <i>J. M. Pettit</i>	2
CT37	The Approximation Problem, <i>W. H. Kautz</i>	4
CT38	The Approximation Problem of Network Synthesis, <i>S. Winkler</i>	5
CT39	A General Method for Time Domain Network Synthesis, <i>F. Bailli</i>	21
CT40	Transient Synthesis in the Time Domain, <i>W. H. Kautz</i>	29
CT41	The Introduction of Constraints into Feedback-System Designs, <i>J. H. Westcott</i>	39
CT42	Numerical Analysis for Network Design, <i>J. G. Truxal</i>	49
CT43	A Note on Filter Synthesis, <i>B. J. Bennett</i>	61
CT44	Transient Response of Cascaded-Tuned Circuits, <i>A. I. Talkin</i>	65
	Correspondence:	
CT45	On the Bott-Duffin Synthesis of Driving-Point Impedances, <i>V. Belevitch</i>	68
CT46	A Low-Pass Transformation for Z-Transforms, <i>W. H. Huggins</i>	69
CT47	Error-Criterion vs. Harmonic Content, <i>C. H. Page</i> ...	70
CT48	What Is Nature's Error Criterion? <i>E. A. Guillemin</i> ...	70
	PGCT News Section.....	71

Vol. CT-1, No. 4, December, 1954

CT49	Nonlinear Filters, <i>W. D. White</i>	2
CT50	Erratum: Approximating Band-Pass Attenuation and Phase Functions, <i>V. Grinich</i>	5
CT51	Analysis of Multi-Loop Nonlinear Systems, <i>Y. H. Ku</i>	6
CT52	Steady-State Oscillations in Nonlinear Multi-Loop Circuits, <i>K. Klotter</i>	13
CT53	Communication Theory and Cybernetics, <i>D. Gabor</i> ...	19
CT54	The Measurement and Representation of Nonlinear Systems, <i>R. C. Booton, Jr.</i>	32
CT55	A Theory of Time Series for Waveform Transmission Systems, <i>W. E. Thomson (Reviewed by R. J. Schwarz)</i>	35
CT56	Wave Form Computations by the Time Series Method, <i>N. W. Lewis (Reviewed by R. J. Schwarz)</i>	36
CT57	Synthesis of Ladder Networks to give Butterworth or Chebyshev Response in the Pass Band, <i>E. Green (Reviewed by H. R. Orchard)</i>	37
CT58	Bandpassschaltungen mit Minimaler Spulenzahl (Bandpass Circuits with Minimum Number of Inductors), <i>G. Bosse (Reviewed by V. Belevitch)</i>	37
CT59	Filters en Echelle Elementaires (Elementary Ladder Filters), <i>J. Oswald (Reviewed by V. Belevitch)</i>	38
CT60	The Solution of Passive Electrical Networks by Means of Mathematical Trees, <i>W. S. Percival (Reviewed by R. M. Foster)</i>	39
CT61	Suggestions for an Improved Terminology, <i>W. R. Bennett</i>	42
CT62	On the Definition of "Sensitivity," <i>H. A. Schulke, Jr.</i> ...	42

Professional Group on Communications Systems

1954

Vol. CS-2, No. 1, January, 1954

<i>Index Number</i>		<i>Page</i>
CS12	Discussion of Solar Research Papers, <i>J. L. Callahan</i> ...	1
CS13	Solar Study as a Phase of Radio Systems Engineering, <i>William A. Miller</i> ...	4
CS14	Tidal Forces on the Sun, <i>I. B. Moulton</i>	15
CS15	Radio Weather Forecasting Techniques, <i>John H. Nelson</i>	19
CS16	Solar Spicules and Their Role in Solar Phenomena,	

Transactions Index —8

PGCS-2, January, 1954 (Cont'd.)

<i>Index Number</i>		<i>Page</i>
	<i>J. H. Rush and W. O. Roberts</i>	24
CS17	Development of the LD Radio System, <i>N. F. Schlaack</i> ...	29
CS18	Overseas Radiotelephone Services of A. T. & T. Co., <i>K. P. Stiles</i>	39
CS19	A Codan for A-M Receivers, <i>James B. Rudd</i>	45
CS20	The Tangier Radio Relay System of RCA Communications, Inc., <i>Carl G. Drietsch</i>	65

Continued on page 9

Vol. CS-2, No. 2, July, 1954

(All but the last two papers in this issue were presented at the Microwave Radio Relay Systems Symposium, New York, N. Y., November 5-6, 1953 and appeared also in Transactions of the I.R.E. Professional Group on Microwave Theory and Techniques, Vol. MTT-2, No. 1, April, 1954.)

Index Number	Page
CS21 The Microwave System of the Michigan-Wisconsin Pipeline Company, <i>W. P. Maginnis and H. Place</i> ...	1
CS22 Microwave Site Selection in Undeveloped Country, <i>R. D. Pynn</i>	9
CS23 Microwave Repeater Site Planning and Development, <i>R. D. Chapman</i>	16
CS24 Remote Control of Standby Engine Generator Sets over a Microwave System, <i>Robert L. Halvorson</i>	32
CS25 Application of Compandors to FM Radio Systems with Frequency Division Multiplexing, <i>M. C. Harp, M. H. Kebby, and E. J. Rudisuhle</i>	36
CS26 A Double Side-Band Amplitude-Modulated Multiplex System for Use over Microwave Radio, <i>Nelson B. Sharp</i>	41
CS27 A Microwave System for Trunk Service, <i>J. J. Lenehan</i>	50
CS28 A Microwave Radio System for Pipeline Use (Abstract), <i>Ed Dyke</i>	60
CS29 Microwave and VHF Radio Installation for the Union Electric System, <i>George W. Fox</i>	63
CS30 Microwave Radio Relay Link for Military Use, <i>Sidney Metzger</i>	84
CS31 Transco Microwave System, <i>H. A. Rhodes</i>	89
CS32 Microwave Testing with Millimicrosecond Pulses, <i>A. C. Beck</i>	93
CS33 Theoretical Aspects of Microstrip Waveguides (Abstract), <i>G. A. Deschamps</i>	100
CS34 Considerations in Klystron Design for Microwave Relay Systems, <i>R. W. Olthius</i>	103
CS35 A New 5 Kilowatt HF Multi-Channel Transmitter, <i>F. R. Hill</i>	108
CS36 Radio Transmitters in the American Cable and Radio System, <i>Fullerton D. Webster</i>	122

Vol. CS-2, No. 3, November, 1954

(Section I—IRE Symposium on Global Communications, Washington, D. C., June 23-25, 1954)

CS37 Historical Exhibits at Global Communications Symposium.....	1
CS38 Global Communications Systems of the Armed Services, <i>V. D. Corput, Jr.</i>	3
CS39 Global Communications, <i>Haraden Pratt</i>	7
CS40 Development and Research Trends in Global Communications, <i>G. W. Gilman</i>	11
CS41 Administrative Aspects of Telecommunications, <i>W. A. Porter</i>	14
CS42 The ITU and Global Communications, <i>F. C. de Wolf</i>	18
CS43 International Radio Frequency Management, <i>Paul D. Miles</i>	22
CS44 The Organization and Functions of the C.C.I.R., <i>E. W. Allen</i>	26
CS45 Improving Frequency Management to Facilitate Global	

PGCS-2, November, 1954 (Cont'd.)

Index Number	Page
Communications, <i>D. R. MacQuivey</i>	30
CS46 British Global Communications, <i>A. H. Read</i>	35
CS47 Global Marine Communications, <i>E. M. Webster</i>	38
CS48 Department of the Army Command Communications, <i>W. F. Spanke</i>	43
CS49 Organization and Operation of the Naval Communication System, <i>G. M. Neely</i>	51
CS50 USAF Strategic Communications System, <i>G. M. Higginson</i>	56
CS51 Overseas Air Traffic Control and Meteorological Communication Circuits of the Civil Aeronautics Administration, <i>F. J. Cervanka</i>	63
CS52 International Planning of Global Communications for Aviation, <i>H. R. Adam</i>	67
CS53 Global Air/Ground Radiotelephone Communications, <i>W. W. Lynch</i>	72
CS54 Frequency Propagation Forecasting for Civil Airline Operations on World Air Routes, <i>C. A. Petry</i>	77
CS55 Frequency Propagation Forecasting for Military World Air Route Operations, <i>L. S. F. Meaker</i>	82
CS56 Recent Advances in International Radio Communications, <i>I. K. Given</i>	86
CS57 Global Public Telephone Service, <i>D. D. Donald</i>	93
CS58 The Impact of Submerged Repeaters on Global Telephony, <i>C. S. Lawton</i>	101
(Section II—IRE-AIEE Symposium on Military Communications, New York, N. Y., April 28, 1954)	
CS59 A Private Microwave Radio System for Power Company Use, <i>D. F. Hazen, J. W. Danzer, and G. S. Zilis</i>	113
CS60 Simplified Transmission Engineering in Exchange Cable Plant Design, <i>L. B. Bogan and K. D. Young</i>	119
CS61 Considerations for Development of New Military Carrier Telephone Systems, <i>R. S. Boykin, J. H. Johnston, and S. D. Bedrosian</i>	124
CS62 A New Cable Design for Military Carrier Telephone Systems, <i>H. F. X. Kingsley</i>	127
CS63 New Military Carrier Telephone Systems Equipment Features, <i>J. P. Hoffmann</i>	130
CS64 New Military Carrier Telephone Systems, <i>G. H. Huber, W. F. Miller, and C. W. Schramm</i>	136
CS65 A New Ultrahigh-Frequency Multichannel Military Radio Relay System, <i>J. G. Nordahl</i>	147
CS66 Telegraph Terminal AN/FGC-29 Circuit Design Aspects, <i>J. E. Boughtwood</i>	152
CS67 Telegraph Terminal AN/FGC-29 Equipment Features, <i>F. H. Cusack</i>	157
CS68 A New Multichannel Teletype Terminal for Use on Long-Range High-Frequency Radio Systems, <i>I. Mack and R. H. Levine</i>	161
CS69 Equipment and Mechanical Features of the AN/TRC-24 Radio Set, <i>V. I. Crusier</i>	165
CS70 Considerations for a New Military Radio Relay System, <i>M. L. Ribe and S. P. Brown</i>	168
Section III	
CS71 Steerable Directional High Frequency Antenna, <i>Helmut Brueckmann</i>	174

Professional Group on Component Parts

1954

PGCP-1, March, 1954

Index Number	Page
CP1 A New Profession, Component Part Engineering, <i>L. K. Lee and F. A. Paul</i>	1
CP2 Breakdown and Leakage Resistance Investigation of Metallized Paper Capacitors, <i>John Burnham</i>	3
CP3 A Comparison of 6AK5 and 5654 Tubes, <i>Floyd A. Paul</i>	18
CP4 Some Characteristics and Limitations of Capacitor and Resistor Components, <i>Leon Podolsky and J. K. Sprague</i>	33

PGCP-2, September, 1954

Index Number	Page
WESCON Component Parts Sessions Program.....	1
WESCON Component Parts Technical Papers Committee.....	3
CP5 Short Time Ratings for Paper Capacitors (Abstract), <i>W. M. Allison</i>	4
CP6 Rotating Components and Their Application to Advanced Electronic Systems (Abstract), <i>R. N. Brown</i>	4
CP7 Appraisal of Wirewound Potentiometers, <i>J. A. Csepely</i>	5

Continued on page 10

Transactions Index—9

Index Number		Page
CP8	Relay Characteristics and Application, <i>C. F. Cameron</i>	34
CP9	A Precise, Wide Band, Continuously-Variable Delay Line, <i>N. Gaw, M. Kline and D. Silverman</i>	48
CP10	The User Looks at the Component Parts Problem, <i>A. M. Okun</i>	58
CP11	Packaging of Component Parts for High Intensity Vi-	

Index Number		Page
	bration Environments, <i>M. G. Comuntzis</i>	72
CP12	Electrostrictive Relay, <i>J. W. Crownover</i>	77
CP13	A Temperature Stabilized Transistor Amplifier, <i>R. B. Hurley</i>	93
CP14	Reliable Electronics Through Protective Coating Tech- niques, <i>E. R. Gamson and A. Hennesian</i>	104

Professional Group on Electronic Devices

1954

Vol. ED-1, No. 1, February, 1954

PGED-1, February, 1954 (Cont'd.)

Index Number		Page
	<i>Part I—Radio Fall Meeting Papers</i>	
ED24	The CBS Colortron—A Color Picture Tube of Ad- vanced Design (Abstract), <i>N. F. Fyler, W. E. Rowe, and C. W. Crain</i>	6
ED25	A PNP Triode Alloy Junction Transistor for RF Am- plification (Abstract), <i>J. I. Pankove and C. W. Muel- ler</i>	6
ED26	The Transistor as a Network Element (Abstract), <i>J. T. Bangerl</i>	7
ED27	Some Application Aspects of the Tetrode Transistors (Abstract), <i>L. Dwork, C. Huang, and H. W. Palmer</i>	7
ED28	Investigation of UHF Television Amplifier Tubes, <i>Wen Yuan Pan</i>	8
ED29	The Dougle-Base Diode: a Semiconductor Thyatron Analog, <i>R. W. Aldrich and I. A. Lesk</i>	24
ED30	General Problems in the Use of Electron Tubes, <i>Walter R. Jones</i>	28
ED31	Application Considerations for RCA Commercial Transistors, <i>R. M. Cohen</i>	32
ED32	A Hermetically Sealed PNP Fused Junction Trans- istor for Medium Power Applications, <i>Conrad H. Zierdt, Jr.</i>	47
	<i>Part II—Abstracts of National Conference on Tube Techniques</i>	
ED33	The Use of Radioactive Tracer Techniques in the Study of Cathode Problems, <i>W. G. Shepherd</i>	56
ED34	Cathode Structure of Indirectly Heated, Narrow, Elongated Oxide Cathodes, <i>W. H. Kohl</i>	56
ED35	An Experimental Tube for Measuring the Sublimation Properties of Nickel Alloy Cathodes, <i>Sherman Cryl- zer, Edmund Mockus, and T. B. Gillis</i>	56
ED36	Study of Cathodes of Rare-Earth Oxides, <i>E. N. Wyler and F. C. Todd</i>	57
ED37	Studies on the Mechanism of Operation of the L- Cathode; Part I—Nature of the Emitting Surface, <i>E. S. Rittner and R. H. Ahlert</i>	57
ED38	Studies on the Mechanism of the L-Cathode; Part II— Production and Transport of Barium, <i>W. C. Rutledge and E. S. Rittner</i>	57
EC39	General Survey of the Philips Dispenser Cathodes, <i>O. S. Duffendack</i>	58
ED40	Interface Measurements and Methods Employed, <i>W. R. Jones</i>	58
ED41	The Improvement of Base Adherence on Electron Tubes, <i>A. M. Seybold</i>	58
ED42	Interesting Techniques Employed in Foreign Tubes, <i>F. R. Michael and T. H. Briggs</i>	59
ED43	Three Dimensional Data Presentation, <i>F. R. Michael</i>	59
ED44	Sealing Techniques for Miniature Tubes, <i>A. P. Hasse and C. E. Horton</i>	60
ED45	Vacuum Tube Spacer Materials, <i>J. C. Hickel and A. P. Hasse</i>	60
ED46	Design and Development of a High-Reliability Twin Triode, <i>C. D. McCool</i>	60
ED47	Tube Envelope Temperatures with Heat-Sensitive Paints, <i>Paul Meissner</i>	60
ED48	The Correlation Between True Contact Potential and Vacuum Tube Processing and Characteristics,	

Index Number		Page
	<i>Eugene Maloney</i>	61
ED49	Thermal Effects in Vacuum Tubes, <i>I. E. Levy</i>	61
ED50	Microphonic Reduction in Filamentary Tubes, <i>Vladimír Kenn</i>	61
ED51	Advanced Processing of Receiving-Type Electron Tubes Subjected to High Temperatures, <i>W. Knecht</i>	62
ED52	Novel Techniques Used in the Assembly of Phototubes and Similar Vacuum Devices, <i>C. Day</i>	62
ED53	Tube Analysis Program, <i>R. D. Wilson</i>	62
ED54	A New Frame Grid that Improves Electron Tube Uni- formity and is Adaptable to Automatic Production, <i>Richard C. White and R. E. Booth</i>	63
ED55	The Nature and Prevention of Keep-Alive Malfunction in TR Tubes, <i>John D. Hall and Paul E. Gates</i>	63
ED56	Nuclear Radiation Counter Tubes, <i>E. H. Wakefield</i>	63
ED57	Gaseous Impurities and the Performance of VR tubes, <i>J. E. Drennan and F. C. Todd</i>	64
ED58	Dependence of Microwave Cavity Characteristics on Properties of Enclosed Regions of a DC Discharge, <i>B. J. Udelson and J. E. Creedon</i>	64
ED59	Magnetron Heater Design to Avoid Undesirable Mag- netic Effects, <i>R. L. Vance and C. Maggs</i>	64
ED60	Corrosion Proofing Electronic Apparatus Parts Exposed to Ozone, <i>F. J. Biondi</i>	65
ED61	Synthetic Mica for Vacuum Tube Use, <i>T. E. Hanley</i>	65
ED62	Miniature Magnetron Assembly Techniques, <i>B. D. Kumpfer</i>	65
ED63	Resonant Window Fabrication Techniques, <i>I. Rein- gold and E. V. Edwards</i>	65
ED64	Power Capabilities of Mica Output Seals at 10,000 Megacycles, <i>W. J. McBride, Jr.</i>	65
ED65	Utilization of Stainless in Large Tube Fabrication, <i>E. C. Okress</i>	66
ED66	Fabricating Copper Magnetron Parts with High Preci- sion and Uniformity by a "Coining"-Type Tech- nique, <i>L. J. Caprarola</i>	66
ED67	A Sandwich-Type Metal-to-Ceramic Vacuum Tight Seal, <i>N. E. Pryslak</i>	66
ED68	A Combination High Temperature Hydrogen or Vacu- um Furnace, <i>W. H. Kohl</i>	67
ED69	Manufacturing Techniques Used in the Production of of Magnetrons, <i>K. R. Stephanz</i>	67
ED70	A Ceramic-Insulated, Flush-Mounted Terminal for All-Metal High-Vacuum Tubes, <i>W. H. Kohl</i>	67
ED71	A Demountable Vacuum System for Tube Develop- ment, <i>T. M. Shrader</i>	68
ED72	The Effect of Electron Bombardment on the Secondary Emission of MgO Thin Films, <i>P. Wargo</i>	68
ED73	Flicker Noise Resistance in Vacuum Tubes, <i>A. van der Ziel</i>	69
ED74	Vacuum Tubes Utilizing Self-Sustained Emission from MgO Films, <i>D. Dobischek, J. Freely, and H. Jacobs</i>	69
ED75	Thermionic Emission of Positive Ions from Alumina- Coated Tungsten, <i>F. G. Ullman</i>	69
ED76	Conductivity of Single-Crystal Al ₂ O ₃ , <i>R. B. Morgan</i>	70
ED77	Preliminary Report on the Regrowth of Silicon Through a Low Melting Zone of Silicon-Gold Eutec- tic, <i>J. N. Carman, P. E. Stello and C. A. Bittman</i>	70
ED78	Design and Properties of Target Structures for Storage Tubes, <i>Max Knoll</i>	70

Continued on page 11

PGED-1, February, 1954 (Cont'd.)

Index Number		Page
ED79	Some Problems in the Design of a Direct View Storage Tube, <i>R. P. Stone</i>	71
ED80	A Demountable Gun Tester Which Can Be Baked Out, <i>H. D. Arnett and R. H. Keyser</i>	71
ED81	Accuracy of Filament Centering, <i>Kenneth Robinson</i>	71
ED82	The Application of a Demountable Diode to the Study and Control of Thoria Cathodes, <i>J. Apelbaum</i>	71
ED83	A Vertical Inconel Furnace for Experimental Laboratories, <i>R. P. Allaire</i>	72
ED84	The Measurement of Surface Flatness of Cathodes for Close-Spaced Electron Tubes, <i>C. T. Goddard</i>	72

Vol. ED-1, No. 2, April, 1954

ED85	Microwave Oscillator Stability, <i>George Hetland, Jr. and Robert R. Buss</i>	1
ED86	Air-Coolers for High Power Vacuum Tubes, <i>A. L. London</i>	9
ED87	Recovery Time Measurements on Point-Contact Germanium Diodes, <i>T. E. Firlie, M. E. McMahon and J. F. Roach</i>	27
ED88	Rapid Determination of Some Electrical Properties of Semiconductors, <i>Luther Davis, Jr., Lawrence G. Rubin and W. D. Straub</i>	34
ED89	A Point-Contact Transistor VHF FM Transmitter, <i>D. E. Thomas</i>	43
ED90	Pulsed Operation of a Cold Cathode Thyatron (395A), <i>M. Silver</i>	53
ED91	Ionization Phenomena in Thyratrons, <i>M. Silver</i>	57
ED92	Germanium Photovoltaic Cells, <i>Bernard J. Rothlein and Alan B. Fowler</i>	67
ED93	An Equivalent Circuit for the Noise in VHF Triodes, <i>A. van der Ziel</i>	72

Vol. ED-1, No. 3, August, 1954

ED94	Traveling Wave Tube Characteristics for Finite Values of C, <i>Charles K. Birdsall and George R. Brewer</i>	1
ED95	Transistor Metrology, <i>D. A. Alsberg</i>	12
ED96	A Developmental Voltage-Tunable Microwave Magnetron, <i>Jules S. Needle</i>	18
ED97	Measurement of Klystron Amplifier Parameters, <i>C. M. Veronda</i>	29
ED98	Calculations of Wave Propagation on a Helix in the Attenuation Region, <i>S. E. Webber</i>	35
ED99	Transistors and Their Applications (A Bibliography, 1948-1953), <i>Alan R. Krull</i>	40

Vol. ED-1, No. 4, December, 1954

Welcoming Address, *G. A. Esperson*..... 1

Session I

ED100	Source Noise and Its Effect Upon Electronic Systems, <i>J. B. Wiesner</i>	3
ED101	Microwave Oscillator Requirements for CW Radar, <i>W. W. McLeod, Jr.</i>	11
ED102	The Influence of Noisy Components on the Sensitivity of Microwave Receivers, <i>W. L. Pritchard and K. I. Larkin</i>	22

PGED-1, December, 1954 (Cont'd.)

Index Number		Page
ED103	Stabilization of Microwave Oscillators, <i>E. J. Shelton</i>	30
	<i>Session II</i>	
ED104	Noise Measurements of Microwave Local Oscillators, <i>R. Mueller</i>	42
ED105	Microwave Oscillator Noise Spectrum Measurements, <i>C. G. Dalman and A. S. Rhoads, Jr.</i>	51
ED106	Theory of Phenomenological Models and Direct Measurements of the Fluctuating Output of CW Magnetrons, <i>D. Middleton</i>	56
ED107	Direct Detection Measurements of Noise in CW Magnetrons, <i>W. M. Gottschalk</i>	91
ED108	Development of a Low-Noise X-Band CW Klystron Power Oscillator, <i>R. A. LaPlante</i>	99
ED109	Frequency and Phase Stability Considerations, <i>B. R. Mayo, H. H. Grimm, J. K. Records</i>	107
ED110	A Technique for Measuring FM Noise in Microwave Oscillators (Abstract), <i>E. F. McClain and W. R. Ferris</i>	121
ED111	National Bureau of Standards Noise Comparator (Abstract), <i>C. R. Greenhow</i>	122
ED112	A Magnetron Test Set for MTI Purposes, <i>H. F. Ayres and R. E. Woods</i>	123
ED113	Carcinotron Noise Measurements, <i>R. L. Krulic</i>	131
	<i>Session III A</i>	
ED114	General Sources of Noise in Vacuum Tubes, <i>J. R. Pierce</i>	135
ED115	Microwave Noise Measurements on Electron Beams, <i>L. D. Smullin and C. Fried</i>	168
ED116	The Signal to Noise Ratio in the M-Carcinotron, <i>O. Doehler and G. Convent</i>	184
ED117	The Small Signal Performance of the 416B Planar Triode Between 60 and 4000 mc, <i>L. H. VanOhlsen</i>	189
ED118	Measurements of Traveling-Wave Tube Noise Figure (Abstract), <i>G. E. St. John</i>	200
ED119	Noise Characteristics of a Voltage-Tunable Magnetron, <i>J. A. Boyd</i>	201
ED120	Measurement and Analysis of Triode Noise, <i>W. A. Harris</i>	206
	<i>Session III B</i>	
ED121	Space Charge Waves on an Accelerating Stream of Uniformly Charged Square Laminae (Abstract), <i>W. E. Danielson</i>	215
ED122	Observations on Ion Oscillations in a Cylindrical-Beam Tetrode Under Hard Vacuum Conditions, <i>W. E. Waters, Jr.</i>	216
ED123	Noise Phenomena in the Region of the Potential Minimum, <i>J. R. Whinnery</i>	221
ED124	Limitations on the Noise Figure of Microwave Amplifiers of the Beam Type, <i>H. A. Haus</i>	238
ED125	Theory of Noisy Four-Poles, <i>H. Rothe</i>	258
ED126	Physical Mechanism of Noise Generation in Magnetrons, <i>C. G. Lehr and A. L. Collins</i>	260
	<i>Session IV</i>	
ED127	Summary of Important Points of Paper, <i>W. H. Huggens</i>	270
ED128	Open Discussion Notes.....	274

Professional Group on Electronic Computers

1954

Vol. EC-3, No. 1, March, 1954

Index Number		Page
EC25	System Organization of the DYSEAC, <i>A. L. Leiner and S. N. Alexander</i>	1
EC26	A Time-Sharing Analog Multiplier, <i>H. Freeman and E. Parsons</i>	11
EC27	An Operational-Digital Feedback Divider, <i>M. A. Meyer, B. M. Gordon, and R. N. Nicola</i>	17
EC28	Review Section, <i>H. D. Huskey, Ed.</i>	23

Vol. EC-3, No. 2, June, 1954

Index Number		Page
	PGEC News, <i>Stanley B. Disson, Editor</i>	1
EC29	Logic, Discovery, and the Foundations of Computing Machinery, <i>M. E. Maron</i>	2
EC30	System Design of the SEAC and DYSEAC, <i>A. L. Leiber, W. A. Notz, J. L. Smith, and A. Weinberger</i>	8
EC31	Digital Techniques in Analog Systems, <i>M. A. Meyer</i>	23
EC32	A High Speed Correlator, <i>Harold Bell, Jr., and Vincent C. Rideout</i>	30

Continued on page 12

Transactions Index—11

PGEC-2, June, 1954 (Cont'd.)

<i>Index Number</i>		<i>Page</i>
EC33	A Wide-Band Square-Law Computing Amplifier, <i>Aaron S. Soltes</i>	37
EC34	An Analog Multiplier Using Thyrite, <i>L. D. Kovach and W. Comley</i>	42
EC35	A Sub-Audio Time Delay Circuit, <i>C. D. Morrill</i>	45
	Contributors.....	50
EC36	Review Section, <i>H. D. Huskey, Editor</i>	53

Vol. EC-3, No. 3, September, 1954

	Editorial, <i>W. Buchholz</i>	1
EC37	A Permanent High Speed Store for Use with Digital Computers, <i>R. D. Ryan</i>	2
EC38	Application of Boolean Algebra to Switching Circuit Design and to Error Detection, <i>D. E. Muller</i>	6
EC39	An Algebraic Theory for Use in Digital Computer Design, <i>E. C. Nelson</i>	12
EC40	An Improved Reading System for Magnetically Recorded Digital Data, <i>Samuel Lubkin</i>	22
EC41	A Digital Voltage Encoder, <i>J. R. Zuehlzig</i>	25
EC42	A New Method of Generating Functions, <i>L. G. Polimerou</i>	29
EC43	A Function Generator for the Solution of Engineering Design Problems, <i>C. J. Savant and R. C. Howard</i> ...	34

PGEC-3, September, 1954 (Cont'd.)

<i>Index Number</i>		<i>Page</i>
	News, <i>S. B. Disson, Editor</i>	39
	Contributors.....	39
EC44	Review Section.....	41

Vol. EC-3, No. 4, December, 1954

EC45	The Use of a Reflected Code in Digital Control Systems, <i>F. A. Foss</i>	1
EC46	A Transistorized Pulse Code Modulator, <i>Gordon R. Partridge</i>	7
EC47	A Radio-Frequency Nondestructive Readout for Magnetic-Core Memories, <i>Bernard Widrow</i>	12
EC48	Time-Delay Networks for an Analog Computer, <i>W. J. Cunningham</i>	16
EC49	A Stabilized Driftless Analog Integrator, <i>Howard Hamer</i>	19
EC50	A Desk-Model Electronic Analog Computer, <i>H. A. Rosen and M. W. Fossier</i>	20
EC51	1954 IRE National Convention Computer Session III. Contributors.....	24
	News.....	31
EC52	Review of Current Literature.....	33
	Annual Index.....	45

Professional Group on Engineering Management

1954

PGEM-1, February, 1954

<i>Index Number</i>		<i>Page</i>
	Message to Members of the Professional Group on Engineering Management, <i>T. C. Rives</i>	1
	Message from Publications Committee, <i>C. G. Cambridge</i>	2
EM1	Management and the Engineer, <i>T. C. Rives</i>	3
EM2	An Engineering Incentive Problem, <i>Harold Goldberg</i> ...	7
EM3	Staff Engineer's Part in Control of Design and Development Costs, <i>Harold G. Purinton</i>	10
EM4	How Design Quality Control Can Help Engineering, <i>Charles E. Ellis</i>	17
EM5	An Informational Approach to Organization and System Engineering Design, <i>Jerome Rothstein</i>	25
EM6	The Role of the Military Laboratory in Electronics Research and Development, <i>Herbert Sherman</i>	30
	Constitution of the Professional Group on Engineering Management.....	44

PGEM-1, February, 1954 (Cont'd.)

<i>Index Number</i>		<i>Page</i>
	Bylaws of the Professional Group on Engineering Management.....	48
	Contributors.....	52

PGEM-2, November, 1954

EM7	Control of Cost of Research and Development Projects, <i>H. J. Finison</i>	1
EM8	Human Relations in Engineering Management, <i>Melville Hopkins</i>	16
EM9	Engineers Can Be Managers, <i>Douglas Watson</i>	28
EM10	Management Can Be Taught, <i>Lowell H. Hattery</i>	41
EM11	The Responsibility of Engineering Management, <i>A. F. Coleman</i>	48
EM12	Some Views on Executive Management, <i>T. M. Linville</i>	52
EM13	The Selection and Development of Laboratory Executives, <i>Oliver G. Tallman</i>	59
	Biographical Notes on the Authors.....	66

Professional Group on Information Theory

1954

PGIT-3, March, 1954

<i>Index Number</i>		<i>Page</i>
	Symposium on Statistical Methods in Communication Engineering.....	1
IT38	Passage of Stationary Processes through Linear and Non-Linear Devices, <i>A. J. F. Siegert</i>	4
IT39	Statistical Theory of Signal Detection, <i>David Middleton</i>	26
IT40	Detectability of Random Signals in the Presence of Noise, <i>R. C. Davis</i>	52
IT41	The Response of Linear Systems to Non-Gaussian Noise, <i>B. Gold and G. O. Young</i>	63
IT42	Estimation of Signal Parameters in the Presence of	

PGIT-3, March, 1954 (Cont'd.)

<i>Index Number</i>		<i>Page</i>
	Noise, <i>David Slepian</i>	68
IT43	The Use of the Method of Maximum Likelihood in Estimating Continuous-Modulated Intelligence which has been Corrupted by Noise, <i>Dante C. Youla</i>	90
IT44	Detection of Modulated Noise-like Signals, <i>Ralph Deutsch</i>	106
IT45	Statistically Almost Optimum Nonlinear Network Design (Abstract), <i>Otto J. M. Smith</i>	123
IT46	Simple Games of Strategy Occurring in Communication through Natural Languages, <i>Benoit Mandelbrot</i>	124

Continued on page 13

PGIT-3, March, 1954 (Cont'd.)

<i>Index Number</i>		<i>Page</i>
IT47	Application of Linear Graphs to Communication Problems (Abstract), <i>Arthur E. Laemmel</i>	138
IT48	Minimum-Cost Encoding of Information, <i>Nelson M. Blachman</i>	139
IT49	Generalized Servomechanism Evaluation (Abstract), <i>William P. Caywood, Jr., Richard C. Lyman and William M. Kaufman</i>	150
IT50	Optimum Pulse-Time Determination, <i>A. J. Mullinckrodt and T. E. Sollenberger</i>	151

PGIT-4, September, 1954

	Preface, <i>W. G. Tuller</i>	1
IT51	A New Basic Theorem of Information Theory, <i>A. Feinstein</i>	2
IT52	Binary Coding, <i>M. J. E. Golay</i>	23
IT53	Error-Free Coding, <i>P. Elias</i>	29
IT54	A Class of Multiple-Error-Correcting Codes and the Decoding Scheme, <i>I. S. Reed</i>	38
IT55	Coding for Constant-Data-Rate Systems, <i>R. A. Silverman and M. Balser</i>	50
IT56	Information, Organization and Systems, <i>J. Rothstein</i>	64

PGIT-4, September, 1954 (Cont'd.)

<i>Index Number</i>		<i>Page</i>
IT57	An Information-Theoretical Model of Organizations, <i>M. Kochen</i>	67
IT58	Simulation of Self-Organizing Systems by Digital Computer, <i>B. G. Farley and W. A. Clark</i>	76
IT59	A Study of Ergodicity and Redundancy Based on Inter-symbol Correlation of Finite Range, <i>S. Watanabe</i>	85
IT60	Multivariate Information Transmission, <i>W. J. McGill</i>	93
IT61	Choice and Coding in Information Retrieval Systems, <i>C. N. Mooers</i>	112
IT62	Modern Statistical Approaches to Reception in Communication Theory, <i>D. Van Meter and D. Middleton</i>	119
IT63	A Non-Linear Prediction Theory, <i>R. F. Drenick</i>	146
IT64	The Detection of Signals Perturbed by Scatter and Noise, <i>R. Price</i>	163
IT65	The Theory of Signal Detectability, <i>W. W. Peterson, T. G. Birdsall, and W. C. Fox</i>	171
IT66	The Human Use of Information. I: Signal Detection for the Case of the Signal Known Exactly, <i>Wilson P. Tanner, Jr., and John A. Swets</i>	213
IT67	The Human Use of Information. II: Signal Detection for the Case of an Unknown Signal Parameter, <i>W. P. Tanner, Jr., and R. Z. Norman</i>	222

Professional Group on Instrumentation

1954

PGI-3, April, 1954

<i>Index Number</i>		<i>Page</i>
	Foreword, <i>Ivan Easton</i>	1
I-19	The Application of Counter Techniques to Precision Frequency Measurements, <i>A. F. Boff</i>	2
I-20	Timing Circuits, II, <i>B. Brooks</i>	11
I-21	Impedance Meter—50-1000 MC/S, <i>H. A. Finke</i>	15
I-22	The Measurement of Time Jitter in Trains of Video	

PGI-3, April, 1954 (Cont'd.)

<i>Index Number</i>		<i>Page</i>
	Pulses, <i>John L. Fitch and Robert E. Buss</i>	23
I-23	A Ratiometer, <i>Nicholas L. Pappas</i>	28
I-24	Swept Wide-Range SWR Indicators for 100 to 1350 Megacycles, <i>W. P. Peyser</i>	35
I-25	An Improved Method of Measuring the Current Amplification Factor of Junction Type Transistors, <i>F. R. Stansel</i>	41
I-26	Strain Gage Oscillator, <i>Francis A. Varallo</i>	50

Professional Group on Microwave Theory and Techniques

1954

Vol. MTT-2, No. 1, April, 1954

<i>Index Number</i>		<i>Page</i>
MTT21	The Microwave System of the Michigan-Wisconsin Pipeline Company, <i>W. P. Maginnis and H. Place</i>	1
MTT22	Microwave Site Selection in Undeveloped Country, <i>R. D. Pynn</i>	9
MTT23	Microwave Repeater Site Planning and Development, <i>R. D. Chapman</i>	16
MTT24	Remote Control of Standby Engine Generator Sets over a Microwave System, <i>Robert L. Halvorson</i>	32
MTT25	Application of Companders to FM Radio Systems with Frequency Division Multiplexing, <i>M. C. Harp, M. H. Kebby, and E. J. Rudisuhle</i>	36
MTT26	A Double Side-Band Amplitude-Modulated Multiplex System for Use over Microwave Radio, <i>Nelson B. Tharp</i>	41
MTT27	A Microwave System for Trunk Service, <i>J. J. Lenehan</i>	50
MTT28	A Microwave Radio System for Pipeline Use (Abstract), <i>Ed Dyke</i>	60
MTT29	Microwave and VHF Radio Installation for the Union Electric System, <i>George W. Fox</i>	63
MTT30	Microwave Radio Relay Link for Military Use, <i>Sidney Metzger</i>	84
MTT31	Transco Microwave System, <i>H. A. Rhodes</i>	89
MTT32	Microwave Testing with Millimicrosecond Pulses, <i>A. C. Beck</i>	93

PGMTT-2, April, 1954 (Cont'd.)

<i>Index Number</i>		<i>Page</i>
MTT33	Theoretical Aspects of Microstrip Waveguides (Abstract), <i>G. A. Deschamps</i>	100
MTT34	Considerations in Klystron Design for Microwave Relay Systems, <i>R. W. Olthuis</i>	103
Vol. MTT-2, No. 2, July, 1954		
	Symposium on Modern Advances in Microwave Techniques.....	1
	Frequency Curves.....	1
MTT35	A Traveling-Wave Electron Deflection System, <i>R. C. Honey</i>	2
MTT36	Crystal Checker for Balanced Mixers, <i>P. D. Strum</i>	10
MTT37	A Technique for Stabilizing Microwave Oscillators, <i>Jerome L. Allman</i>	16
MTT38	Use of Crystals in Balanced Mixers, <i>Jesse Taub and Paul J. Giordano</i>	26
MTT39	A Coaxial Line Filled with Two Non-concentric Dielectrics, <i>D. J. Angelakos</i>	39
MTT40	H ₀₁ Mode Circular Waveguide Components, <i>D. A. Lanciani</i>	45
MTT41	Characteristic Impedance of the Shielded-Strip Transmission Line, <i>Seymour B. Cohn</i>	52
MTT42	Bibliography on Directional Couplers, <i>Richard F. Schwarzl</i>	58
MTT43	Addenda.....	64

Continued on page 14

<i>Index Number</i>		<i>Page</i>
MTT44	Optical Methods for the Measurement of Complex Dielectric and Magnetic Constants at Centimeter and Millimeter Wavelengths, <i>T. E. Talpey</i>	1
MTT45	Millimeter Wave Spectroscopic Components, <i>W. C. King</i>	13
MTT46	Harmonics at Millimeter Wavelengths, <i>A. H. Nethercot, Jr.</i>	17
MTT47	Millimeter Wave Generation by Cerenkov Radiation, <i>M. Danos and H. Iashinsky</i>	21
MTT48	Stabilization of Reflex Klystrons by High-Q External Cavities, <i>S. J. Rabinowitz</i>	23
MTT49	Superheterodyne Receiver for the 100 to 150-kmc	

<i>Index Number</i>		<i>Page</i>
	Region, <i>C. M. Johnson</i>	27
MTT50	Magnetron Research at Columbia Radiation Laboratory, <i>M. J. Bernstein and N. M. Kroll</i>	33
MTT51	Low Loss Dielectric Waveguides, <i>M. T. Weiss and E. M. Gyorgy</i>	38
MTT52	An Experimental Broad-Band Helix Traveling-Wave Amplifier for Millimeter Wavelengths, <i>S. D. Robertson</i>	48
MTT53	A Calorimeter for Power Measurements at Millimeter Wavelengths, <i>W. M. Sharpless</i>	45
MTT54	The Use of Flat Waveguide in the Millimeter Range (Abstract), <i>S. Hopfer</i>	54

Professional Group on Nuclear Science

1954

Vol. NS-1, No. 1, September, 1954

<i>Index Number</i>		<i>Page</i>
	Editorial.....	1
NS1	Instrumentation and Control of the Brookhaven Nuclear Reactor, <i>J. E. Binns</i>	2
NS2	Constant Reactor Outlet Temperature Control System, <i>G. S. Stubbs</i>	8
NS3	Electronic Equipment for an Electron Analogue Accelerator, <i>J. G. Cottingham, M. Plotkin and E. C. Raka</i> ..	12

PGNS-1, September, 1954 (Cont'd.)

<i>Index Number</i>		<i>Page</i>
	Program of the Annual National Meeting, Chicago, October 6-7, 1954.....	18
	Abstracts of Contributed Papers for Annual Meeting..	21
	Annual Report of the Secretary.....	25
	Revised Constitution of the PGNS.....	26
	By-Laws.....	29
	Publication Policy.....	31
NS4	Selections from "Nuclear Science Abstracts".....	33

Professional Group on Quality Control

1954

PGQC-3, February, 1954

<i>Index Number</i>		<i>Page</i>
QC12	The Program on Reliability of Electronic Equipment, <i>Lewis M. Clement</i>	1
QC13	Report on U. K. Project to Improve Valves for Military Applications, <i>F. H. Powell</i>	10
QC14	Inspection Procedures for MIL-E-18 Reliable Electron Tubes, <i>R. J. E. Whittier</i>	15
QC15	The Panel on Electron Tubes Program for Coordinating Tube Reliability Activities, <i>L. S. Schwartz</i>	28
QC16	Those Unreliable Thermionic Tubes, <i>Walter R. Jones</i> ..	32

PGQC-4, December, 1954

<i>Index Number</i>		<i>Page</i>
QC17	Developments in Trustworthy-Valve Techniques, <i>E. G. Rowe and P. Welch</i>	1
QC18	Analysis of a Cumulative Results Sampling Plan for Use with Sampling Tables Using Zero Acceptance Numbers, <i>Arnold Cronshagen</i>	14
QC19	Reliability of Quantity Produced Transistors in Lower Power Audio Applications, <i>F. M. Dukal</i>	32
QC20	Reliable Electronics Through Protective Coating Techniques, <i>E. R. Gamson and A. Henesian</i>	40
QC21	Quality in Production, <i>R. Weller</i>	51

Professional Group on Radio Telemetry and Remote Control

1954

PGRTRC-1, August, 1954

<i>Index Number</i>		<i>Page</i>
	Editorial.....	1
TRC1	The Integrated Air Force Missile Test Center Data Processing Facility, <i>Charles F. West</i>	1
TRC2	Basic Design of Commutating Devices, <i>Martin V. Kiebert, Jr.</i>	7
TRC3	High Speed Digital Computers, <i>Marc Shiovitz</i>	12
	Correction.....	16

PGRTRC-1, November, 1954 (Cont'd.)

<i>Index Number</i>		<i>Page</i>
TRC4	Modern Concepts for Digital Computer Input-Output Philosophy, <i>Charles F. West</i>	2
TRC5	Delay Line Controlled Subcarrier Discriminator, <i>Kenneth A. Morgan and Richard F. Blake</i>	7
TRC6	A Temperature-Stable Transistor VCO, <i>Fred M. Riddle</i>	11
TRC7	Telemetering and Information Theory, <i>Frank W. Lehan</i>	15
TRC8	A Slope Modulator for FM Recording of Analog Data on Magnetic Tape, <i>Louis W. Erath and Frank C. Smith, Jr.</i>	20

PGRTRC-2, November, 1954

	Editorial, <i>M. V. Kiebert, Jr.</i>	1
--	--	---

Continued on page 15

Professional Group on Ultrasonics Engineering

1954

PGUE-1, June, 1954

<i>Index Number</i>		<i>Page</i>
	History, Plans and Policies of the PGUE, <i>A. L. Lane</i> .	1
	Our Membership—Who We Are and Where, <i>M. D. Fagen</i>	3
UE1	Ultrasound and Medicine, <i>Julia F. Herrick and Frank H. Krusen</i>	4
UE2	A Noncontact Micro-Displacement Meter, <i>Harold F. Sharaf</i>	14
UE3	Characteristics of Ultrasonic Delay Lines Using Quartz and Barium Titanate Ceramic Transducers, <i>John E. May</i>	26
UE4	Metal Cleaning and Its Improvement by the Use of Ultrasonics, <i>T. J. Kearney</i>	43
UE5	A Temperature Controlled Ultrasonic Solid Acoustic Delay Line (Abstract), <i>E. S. Pennell</i>	48

PGUE-1, June, 1954 (Cont'd.)

<i>Index Number</i>		<i>Page</i>
UE6	An Acoustic Flowmeter Using Electronic Switching, <i>H. P. Kalmus, A. L. Hedrich, and D. R. Pardue</i>	49

PGUE-2, November, 1954

	Chairman's Report 1953-1954, <i>A. L. Lane</i>	1
UE7	Bibliography on Ultrasonic Delay Lines, <i>M. D. Fagen</i>	3
	Introduction.....	9
UE8	An Ultrasonic Machine Tool, <i>Neil Clark, Jr.</i>	10
UE9	Ultrasonic Machining of Tungsten Carbide, <i>Dieter Goetze</i>	19
UE10	Mechanical Impedance Transformers in Relation to Ultrasonics Machining, <i>Lewis Balamuth</i>	23
	Bibliographical Notes on the Authors.....	34
	PGUE Constitution and Bylaws.....	35

Professional Group on Vehicular Communications

1954

PGVC-4, June, 1954

<i>Index Number</i>		<i>Page</i>
VC33	Integrated Mobile-Microwave System, <i>J. R. Neubauer</i>	1
VC34	A Commissioner's Reflections on the Mobile Radio Service, <i>E. M. Webster</i>	13
VC35	Duplex and Multi-Channel Mobile Equipment, <i>W. Ornstein</i>	22
VC36	Mobile Radio System Performance in the United States Forest Service, <i>W. S. Claypool</i>	34

PGVC-4, June, 1954 (Cont'd.)

<i>Index Number</i>		<i>Page</i>
VC37	The Knee of the Nose, <i>R. P. Gifford</i>	40
VC38	Portable Equipment in the Communications System, <i>W. J. Weisz</i>	52
VC39	Mobile and Fixed Radio Relay Operation in the Power Radio Service, <i>G. E. Dodrill and J. F. Atkinson</i>	62
VC40	Problems in Maintenance and Operation of Long-Haul and Distribution Radio Networks, <i>D. E. York</i>	86
VC41	Planning and Operation of the Erie Railroad Main Line Radio Communications System, <i>W. J. Young</i>	91



INDEX TO SUBJECTS

Professional Group on Aeronautical and Navigational Electronics

Air-Ground Communication: AE26, AE29
Channel Allocations: AE29
Information Content of Messages: AE26
Altimeters: AE30, AE31, AE32, AE33, AE34
AN/APN-22: AE33
Flare-Out Unit: AE32
History: AE30
Low-Altitude: AE34
Principles: AE31
Antennas, Aircraft: AE28, AE42
High-Voltage Problems: AE42
Scan Considerations: AE28
Direction Finders: AE43
Ring Goniometer: AE43
Distance Measuring Equipment: AE41
Interrogator: AE41
Feedback in Weapons Systems: AE39
Flare-Out Unit: AE32
Instrument Landing System: AE32
Flare-Out Unit: AE32
Loran Receiver, Airborne: AE27, AE35
Maintenance of Airborne Radio Equipment: AE36
Radar, Airborne: AE37, AE38, AE40
Applications: AE38
Navigation Aid: AE40
Thunderstorm Avoidance: AE37
Ring Goniometers for Direction Finders: AE43
Simulators for Flight-Path Study: AE44
Thunderstorm Avoidance Radar: AE37

Professional Group on Antennas and Propagation

Adcock Antenna Error Evaluation: AP134
Aircraft Antennas, Omnidirectional: AP113
Antenna Feed with Equal *E* and *H* Plane Patterns: AP128
Backscatter Observations, Sweep-Frequency Method: AP120
Corrugated Plane Surface: AP122
Dielectric Sheet Radiators: AP133
Dipole Radiation over Stratified Ground: AP131
Fading of 100 Mc FM Signals: AP116
Field Intensities in Diffraction Zone, VHF: AP117
Folded Antenna Impedance Chart: AP135
Helical Transmission Line Analysis: AP130
Hyperboloid Lens: AP129
Lens, Luneberg: AP124
Meteoric Radio Echoes, Survey: AP123
Paraboloid Reflector: AP129
Recombination Coefficient in Lower Ionosphere: AP125
Refractive Index Measurements, U. S.: AP114
Rhombic Antennas, Mutual Impedance: AP118
Single-Wire Transmission Line Experiments: AP119
Slots, Closely-Spaced: AP127
Stochastic Problems in Wave Propagation: AP115, AP121
Transmission Lines of Composite Section: AP126
VHF Field Intensities in Diffraction Zone: AP117
Waveguide Array for Solar Noise Studies: AP132

Professional Group on Audio

Acoustic Lens as Directional Microphone: A57
Acoustic-Mechanical Circuit Analysis: A75, A82
Amplifiers: A63, A65, A66, A69, A83
Cascade: A66
Dynamic, for Phonograph: A69
Miniature High-Gain: A83
Noise in: A63
Power: A65
Cascade as an Audio Amplifier: A66
Cathode Bias Resistor for Class A, Triode: A79
Circuit Analysis of Mechano-Acoustic Structures: A75, A82
Correction to A75: A82
Electron-Beam Reproducing Head: A62
Film, Magnetic Recording on: A70
Flutter Standards: A72
Frequency Compression-Expansion of Speech: A58
Frequency Modulation Phonograph Pickups: A76
Hearing Aid, Transistor: A67
High Fidelity in Musical Tone Production: A73
Loudspeaker Accessory for Reverberance: A71
Magnetic Recording: A58, A59, A62, A70, A77, A78, A80, A81
Adverse Environments: A77
Electron-Beam Reproducing Head: A62
Navy Standard Recorder-Reproducer: A80
On 35 MM Film: A70
State of the Art: A78
Television Signals: A81
Time Expansion Device: A59
Time or Frequency Compression-Expansion: A58
Magnetic Tape:
Mechanical Components for Handling: A60
Navy 1" Standard: A80
Mechano-Acoustic Circuit Analysis: A75, A82
Microphones: A57, A64, A84
Directional: A57
"Lipstik" Condenser: A84
"Vagabond" Wireless: A64
Noise in Audio Amplifiers: A63
Phonograph Pickups, FM: A76
Resistor, Cathode Bias, for Class A, Triode: A79
Sound Reproduction, Natural: A74
Speech, Time or Frequency Compression-Expansion: A58
Speech, Visible: A68
Time Compression-Expansion of Speech: A58
Time Expansion Device: A59
Transistor Hearing Aid: A67

Professional Group on Broad- cast and Television Receivers

Amplifiers: BTR19, BTR44
IF, for Color Receivers: BTR44
UHF Disc-Seal Triode: BTR19
Colorimeter, Barnes: BTR22, BTR23
Color Television: BTR21, BTR22, BTR23, BTR24, BTR25, BTR26, BTR27, BTR28, BTR29, BTR30, BTR33, BTR34, BTR35, BTR44, BTR48
Amateur Interference: BTR34, BTR35
Barnes Colorimeter: BTR22, BTR23

For Quality Control: BTR23
For Screen Color Determination: BTR22
CBS Colortron: BTR29
Chromatron Operation on NTSC Standards: BTR30
Color Decoder: BTR25
DC Quadricorrelator: BTR28
IF Amplifier Design: BTR44
Interference from Oscillator Radiation: BTR48
NTSC Field Tests: BTR27
Pulsed Envelope Detection: BTR24
Sound-Color Beat: BTR26
Transient Response of Receiver: BTR21
Yokes: BTR33
Electron Tubes: BTR18, BTR19
UHF Disc-Seal Triode: BTR19
UHF Television Amplifiers: BTR18
Generator, Test, for Horizontal Scan: BTR20
Heaters, Series Strings for TV: BTR47
Horizontal Scan Test Generator: BTR20
Interference: BTR34, BTR35, BTR36, BTR37, BTR38, BTR39, BTR40, BTR41, BTR48
Amateur-TV: BTR34, BTR35
JTAC Program: BTR41
Oscillator Radiation: BTR48
Over Radio Spectrum: BTR36
Spurious Radiation from TV Receivers: BTR40
Spurious Radiation Standards: BTR38
Television IF: BTR39
Television Receiver Industry: BTR37
Noise Reduction in UHF Amplifiers: BTR42
Q Determination of Resonant Circuits: BTR50
Quality Control, Barnes Colorimeter: BTR23
Receivers: BTR17
Transistor Broadcast: BTR17
Selectivity Determination of Resonant Circuits: BTR50
Semiconductor Diodes for TV Receivers: BTR46
Television Picture Quality: BTR31
Test Generator for Horizontal Scan: BTR20
Transient Response of TV Receivers: BTR21
Transistor Broadcast Receiver: BTR17
Tuner Design, UHF: BTR43
Vectorscope Applications: BTR49
Yoke Astigmatism: BTR45
Yokes: BTR32, BTR33

Professional Group on Circuit Theory

Amplifiers, Feedback, Power Gain in: CT31
Approximation Problem: CT37, CT38
Band-Pass Attenuation: CT50
Band-Pass Filter Classification: CT58
Bibliographies, Suggestion on: CT27
Bott-Duffin Synthesis of Driving-Point Impedances: CT45
Brune Cycle Without Ideal Transformer: CT24
Cascaded-Tuned Circuit Transient Response: CT44
Champions, The: CT35
Communication Theory and Cybernetics: CT53
Control Systems: CT18, CT21
Optimum Lead-Controller Synthesis: CT21
Random Input, Nonlinear: CT18
Cybernetics: CT53
Error Criterion, Nature's: CT25, CT32, CT47, CT48

Continued on page 17

Feedback Systems: CT17, CT20, CT21, CT29, CT31, CT41
 Constraints in Design: CT41
 Optimum Lead-Controller Synthesis: CT21
 Power Gain in Amplifiers: CT31
 Stability, Using Dual Nyquist Diagrams: CT20
 Transfer Function Synthesis: CT29
 Trends in: CT17
 Filter, Nonlinear: CT49
 Filter Synthesis: CT43
 Ideal Transformer in Brune Cycle: CT24
 International Scientific Radio Union: CT36
 Ladder Filters: CT59
 Ladder Network Synthesis: CT57
 Moment Problem in Network Synthesis: CT34
 Multi-Loop Nonlinear Systems: CT26, CT51, CT52
 Nature's Error Criterion: CT25, CT32, CT47, CT48
 Network Synthesis and the Moment Problem: CT34
 Nonlinear Filters: CT49
 Nonlinear System Response Measurements: CT54
 Numerical Analysis for Network Design: CT42
 Nyquist Diagrams, Dual: CT20
 Oscillations in Multi-Loop Circuits: CT52
 Oscillations, Simultaneous, in Oscillators: CT28
 Passive Network Solution by Mathematical Trees: CT60
 Reciprocity, Three-Terminal Elements: CT30
 Servomechanism: CT17, CT18, CT19, CT20, CT21, CT22, CT23
 Closed-Loop System Analysis: CT19
 Control Systems with Random Inputs: CT18
 Feedback System Trends: CT17
 Optimum Lead-Controller Synthesis: CT21
 Predictor: CT23
 Response: CT22
 Stability, Using Dual Nyquist Diagram: CT20
 Spectra of Waves with Periodic Modulation: CT32
 Time Domain Network Synthesis: CT39
 Time Domain Transient Synthesis: CT40
 Transient Response of Cascaded-Tuned Circuits: CT44
 Transient Synthesis in the Time Domain: CT40
 URSI: CT36
 Waveform Computations: CT56
 Waveform Transmission Systems: CT55
 Z-Transforms: CT46

Professional Group on Communications Systems

Administrative Aspects: CS41
 Air Force: CS50
 Air-Ground Radiotelephone: CS53
 Airline Propagation Forecasting: CS54, CS55
 Antenna: CS71
 Steerable Directional: CS71
 AN/TRC-24 Radio Set: CS69
 Army, Dept. of: CS48
 Aviation Global Communications: CS52
 British Global Communications: CS46
 CCIR: CS44
 Civil Aeronautics Administration: CS51
 Codan for AM Receivers: CS19
 Companders for FM System: CS25
 Double Sideband AM Multiplex System: CS26
 Exchange Cable Plant Design: CS60

Frequency Management, International: CS43, CS45
 Frequency Propagation Forecasting: CS54, CS55
 Generator, Standby, Remote Control of: CS24
 Global Communications: CS37, CS38, CS39, CS40, CS42, CS43, CS45, CS46, CS47, CS52, CS57, CS58
 Air-Ground: CS53
 Armed Services: CS38
 Aviation: CS52
 British: CS46
 Development and Research Trends: CS40
 Frequency Management: CS43, CS45
 Historical Exhibits: CS37
 History: CS39
 ITU: CS42
 Marine: CS47
 Public Telephone: CS57
 Telegraphy: CS58
 History of Global Communications: CS37, CS39
 International Radio Communications: CS56
 ITU: CS42
 Klystron Design Considerations: CS34
 LD Radio System: CS17
 Marine Communications, Global: CS47
 Microstrip Waveguides: CS33
 Military Carrier Telephone Systems: CS61
 Navy: CS49
 Pipeline Microwave System: CS21, CS28
 Private Microwave Radio for Power Co.: CS59
 Radio Relay: CS20, CS30, CS65, CS70
 Military: CS30, CS65, CS70
 Considerations for: CS70
 Microwave: CS30
 UHF: CS65
 Tangier System of RCA: CS20
 Radiotelephone: CS18, CS53
 Air-Ground: CS53
 AT&T Overseas: CS18
 Receivers: CS19
 Codan Anti-Noise Device: CS19
 Repeater Site: CS23
 Repeaters, Submerged: CS58
 Site Selection: CS22, CS23
 Solar Research: CS12, CS13, CS14, CS15, CS16
 Telegraph: CS66, CS67
 Terminal Circuits: CS66
 Terminal Equipment: CS67
 Telegraphy: CS59
 Submerged Repeaters: CS59
 Telephone: CS57, CS61, CS62, CS63
 Cable Design: CS62
 Equipment Features, Military: CS63
 Public, Global: CS57
 System, Military: CS61
 Teletype: CS68
 Multichannel Terminal: CS68
 Testing with Millimicrosecond Pulses: CS32
 Transco Microwave System: CS31
 Transmitter: CS35, CS36
 American Cable and Radio: CS36
 HF Multi-Channel: CS35
 Trunk Service, Microwave System for: CS27
 Union Electric System Installation: CS29
 Waveguide: CS33
 Microstrip: CS33

Professional Group on Component Parts

Amplifier, Transistor: CP13
 Capacitors: CP2, CP4, CP5
 Limitations: CP4
 Metallized Paper: CP2
 Breakdown and Leakage Resistance: CP2
 Short Time Ratings: CP5
 Coating, Protective: CP14
 Component Part Engineering Profession: CP1

Component Parts, User's Problems: CP10
 Delay Line, Continuously Variable: CP9
 Electron Tubes: CP3
 Comparison of 6AK5 and 5654: CP3
 Electrorestrictive Relay: CP12
 Packaging for Vibration: CP11
 Potentiometers: CP7
 Wirewound: CP7
 Protective Coating: CP14
 Relays: CP8, CP12
 Characteristics and Applications: CP8
 Electrostrictive: CP12
 Resistors: CP4
 Limitations: CP4
 Rotating Components, Applications: CP6
 Transistor Amplifier: CP13
 User's Problems: CP10
 Vibration, Packaging for: CP11

Professional Group on Electron Devices

Air Coolers for High Power Tubes: ED86
 Amplifiers: ED28, ED97, ED124
 Klystron, Parameter Measurements: ED97
 Microwave, Noise Figure: ED121
 UHF Tubes: ED28
 Assembly Techniques: ED52
 Automatic Production, Frame Grid for: ED54
 Base Adherence on Tubes: ED41
 Bibliography of Transistors and Applications: ED99
 Cathode: ED33, ED34, ED35, ED36, ED37, ED38, ED39, ED82, ED84, ED128
 L-: ED37, ED38
 Emitting Surface: E37
 Production of Barium: ED38
 Narrow, Elongated: ED34
 Nickel Alloy, Sublimation Measurement: ED35
 Philips Dispenser: ED39
 Philips Impregnated, Noise Figure: ED128
 Radioactive-Tracer Study Method: ED33
 Rare-Earth Oxides: ED36
 Surface Flatness: ED84
 Thoria, Study of: ED82
 Carcinotron: ED113, ED116
 Noise Measurements: ED113
 Signal-to-Noise Ratio: ED116
 Cavity, Microwave: ED58
 CBS Colortron Picture Tube: ED24
 Conductivity of Single Crystal AlO_3 : ED76
 Contact Potential, Tube Processing Effects on: ED48
 Copper Magnetron Parts Fabrication: ED65
 Corrosion Proofing: ED60
 Counter Tube, Radiation: ED56
 Definitions of Noise: ED128
 Diode: ED29, ED82, ED87
 Demountable, for Cathode Study: ED82
 Double-Base: ED29
 Germanium, Recovery Time: ED87
 Electron Beam: ED115
 Noise Measurements: ED115
 Electron Gun: ED80
 Tester: ED80
 Electron Tubes, Problems in Use of: ED30
 Emission: ED72, ED74, ED75
 From Aluminum-Coated Tungsten: ED75
 Secondary: ED72
 MgO Thin Films: ED72
 Self-Sustained from MgO Films: ED74
 Envelope, Tube, Temperatures: ED47
 Filament Centering Accuracy: ED81
 Flicker Noise Resistance: ED73
 Fluctuation Phenomena in Microwave Sources: ED100-ED128
 Foreign Tube Techniques: ED42
 Frequency and Phase Stability, MTI Radar: ED109

Continued on page 18

Furnace: ED68, ED83
 Combination Hydrogen or Vacuum: ED68
 Vertical Iconel: ED83
 Germanium Photovoltaic Cells: ED92
 Grid: ED54
 Frame, for Automatic Production: ED54
 Helix: ED98
 Propagation Calculation: ED98
 Interface Measurements: ED40
 IRE Standards, Tube Noise Measurements: ED128
 Klystron: ED97, ED108
 Amplifier Parameter Measurements: ED97
 Oscillator, Low-Noise: ED108
 Magnesium Oxide Films: ED72, ED74
 Electron Bombardment: ED72
 Self-Sustained Emission: ED74
 Magnetron: ED59, ED62, ED66, ED69, ED96, ED106, ED107, ED112, ED119, ED126
 Fabricating Copper Parts: ED66
 Fluctuation Output, Measurement: ED106
 Heater Design, Magnetic Effects of: ED59
 Manufacturing Techniques: ED69
 Miniature Assembly Techniques: ED62
 Noise Generation: ED126
 Noise Measurements: ED107, ED119
 Test Set: ED112
 Voltage-Tunable: ED96
 Metal-to-Ceramic Seal: ED67
 Mica Seals, Power Capabilities: ED64
 Mica, Synthetic: ED61
 Microphonic Noise Reduction: ED50
 Noise: ED50, ED73, ED93, ED100, ED102, ED104, ED105, ED107, ED108, ED110, ED111, ED113, ED114, ED115, ED116, ED118, ED119, ED120, ED122, ED123, ED124, ED125, ED126, ED127, ED128
 Carcinotron, Measurements: ED113
 Carcinotron, Signal-to-Noise Ratio: ED116
 Cathode: ED128
 Comparator: ED111
 Definitions: ED128
 Discussion of: ED128
 Electron Beam, Measurements: ED115
 Figure, Microwave Amplifiers: ED124
 Figure, Traveling-Wave Tube: ED118
 Flicker, Resistance: ED73
 Four-Pole: ED125
 General Sources in Tubes: ED114
 In Region of Potential Minimum: ED123
 IRE Technical Committee Standardization: ED128
 Low-, Klystron Oscillator: ED108
 Magnetron, Generation of: ED126
 Magnetron, Measurement: ED107, ED119
 Microphonic, in Microwave Components: ED128
 Microphonic, Reduction of: ED50
 Microwave Oscillator, Measurement: ED104, ED105, ED110
 Microwave Receiver Sensitivity: ED102, ED128
 Monitoring: ED128
 Source, Effect on Electronic Systems: ED100
 Specification: ED128
 Summary of Fluctuation Symposium Papers: ED127
 Tetrode Ion Oscillations: ED122
 Triode, Equivalent Circuit for: ED93
 Triode, Measurement and Analysis: ED120
 Oscillator: ED85, ED101, ED103, ED104, ED105, ED108, ED110
 Klystron, Low-Noise: ED108
 Microwave, Noise Measurements: ED104, ED105, ED110
 Microwave, Stability: ED85, ED103
 Radar, Noise Requirements: ED101
 Phototube: ED52

Assembly: ED52
 Photovoltaic Cell, Germanium: ED92
 Processing Receiving Tubes: ED51
 Radioactive-Tracer Study of Cathodes: ED33
 Seals: ED44, ED64, ED67
 Metal-to-Ceramic: ED67
 Mica, at 10,000 Mc: ED64
 Miniature Tube: ED44
 Secondary Emission of MgO Thin Films: ED72
 Semiconductors, Rapid Determination of Properties: ED88
 Silicon: ED77
 Regrowth by Zone Melting Method: ED77
 Space-Charge Waves on an Accelerating Stream: ED121
 Spacer Materials: ED45
 Stainless Steel in Large Tube Fabrication: ED65
 Standards, IRE, Tube Noise Measurements: ED128
 Storage Tubes: ED78, ED79
 Direct View: ED79
 Target Structures: ED78
 Television: ED24, ED28
 CBS Colortron Picture Tube: ED24
 UHF Amplifier Tubes: ED28
 Terminal for All-Metal Tube: ED70
 Tetrode: ED122
 Ion Oscillations: ED122
 Thermal Effects in Tubes: ED49
 Three-Dimensional Data Presentation: ED43
 Thyatron: ED90, ED91
 Ionization in: ED91
 Pulsed Operation: ED90
 Transistor: ED25, ED26, ED27, ED31, ED32, ED89, ED95, ED99
 Bibliography: ED99
 Hermetically Sealed: ED32
 Metrology: ED95
 Network Element: ED26
 PCA Commercial, Applications: ED31
 RF Triode Junction: ED25
 Tetrode Applications: ED27
 VHF FM Transmitter: ED89
 Transmitter: ED89
 Transistor, VHF, FM: ED89
 Traveling-Wave Tube: ED94, ED118
 Characteristics for Finite C: ED94
 Noise Figure Measurements: ED118
 Triode: ED46, ED93, ED117, ED120
 Noise, Equivalent Circuit: ED93
 Noise Measurement and Analysis: ED120
 Planar, Small Signal Performance: ED117
 Twin, High Reliability: ED46
 TR Tube: ED55, ED63
 Failure of Keep-Alive Electrodes: ED55
 Resonant Window Fabrication: ED63
 Tube Analysis Program: ED53
 Twin Triode, High Reliability: ED46
 Vacuum System, Demountable: ED71
 Voltage-Regulator Tube: ED57
 Gaseous Impurities and Performance: ED57
 Window Fabrication for TR Tubes: ED63

Professional Group on Electronic Computers

Amplifier, Square-Law: EC33
 Analog Computer, Desk Model, EC50
 Analog Computing with Digital Techniques: EC31
 Analog Integrator: EC49
 Analog Multiplier: EC26, EC34
 Time-Sharing: EC26
 Using Thyrite: EC34
 Autonomous Computers: EC51
 Boolean Algebra: EC38, EC39
 Computer Design: EC39
 Switching Circuits and Error Detection: EC38
 Code, Reflected Binary: EC45

Correlator, High-Speed: EC32
 Desk Model Analog Computer: EC50
 Digital Computer Design by Algebra: EC39
 Digital Techniques in Analog Systems: EC31
 Discovery by Computers: EC29
 Divider, Digital Feedback: EC27
 DYSEAC: EC25, EC30
 System Design: EC30
 System Organization: EC25
 Encoder, Digital Voltage: EC41
 Feedback Divider, Digital: EC27
 Function Generating Methods: EC42
 Function Generator: EC43
 Generating Functions: EC42, EC43
 Integrator, Analog: EC49
 Logic: EC29
 Modulator, Transistor Pulse-Code: EC46
 Multiplier: EC26, EC34
 Time-Sharing Analog: EC26
 Using Thyrite: EC34
 Reading Magnetically Recorded Data: EC40
 Readout, RF Nondestructive: EC47
 Repairing Computers Automatically: EC51
 Review of Literature: EC28, EC36, EC44, EC52
 SEAC System Design: EC30
 Self-Repairing Computers: EC51
 Storage, Permanent High-Speed: EC37
 Sub-Audio Time Delay Circuit: EC35
 Thyrite for Analog Multiplier: EC34
 Time-Delay Circuit, Sub-Audio: EC35
 Time-Delay Networks for Analog Computers: EC48
 Transistor Pulse-Code Modulator: EC46

Professional Group on Engineering Management

Cost Control in Research and Development: EM7
 Design Quality Control: EM4
 Engineers Can Be Managers: EM9
 Executive Management: EM12
 Human Relations: EM8
 Incentive: EM2
 Informational Approach to Organization: EM5
 Laboratory Executives, Selection: EM13
 Management and the Engineer: EM1
 Management Can Be Taught: EM10
 Military Role in Research and Development: EM6
 Responsibility of Management: EM11
 Selection of Laboratory Executives: EM13
 Staff Engineer's Part in Controlling Cost: EM3
 System Engineering Design: EM5

Professional Group on Information Theory

Binary Coding: IT52
 Coding: IT52, IT53, IT54, IT55, IT61
 Binary: IT52
 Constant-Data-Rate Systems: IT55
 Error-Free: IT53
 Information Retrieval Systems: IT61
 Multiple-Error-Correcting: IT54
 Continuous-Modulated Intelligence Corrupted by Noise: IT43
 Detection: IT39, IT40, IT44, IT64, IT65, IT66, IT67
 Modulated Noise-Like Signals: IT44
 Random Signals in Noise: IT40
 Sensory: IT67
 Signals Perturbed by Scatter and Noise: IT64
 Statistical Theory: IT39
 Theory: IT65
 Visual: IT66
 Encoding, Minimum-Cost: IT48
 Ergodicity, Study of: IT59
 Error-Free Coding: IT53
 Graphs for Communication Studies: IT47

Continued on page 19

Information Retrieval, Coding: IT61
 Language, Statistical Study of: IT46
 Library Indexing: IT61
 Network, Nonlinear, Almost-Optimum: IT45
 Non-Gaussian Noise in Linear Systems: IT41
 Organization and Systems: IT56
 Organizations, Information-Theoretical Model: IT57
 Prediction Theory, Nonlinear: IT63
 Pulse-Time Determination; Optimum: IT50
 Reception, Statistical Approaches to: IT62
 Redundancy, Study of: IT59
 Response of Linear Systems to Non-Gaussian Noise: IT41
 Self-Organizing Systems: IT58
 Servomechanism, Generalized Evaluation: IT49
 Signal Parameters in Noise, Estimating: IT42
 Stationary Processes, Passage of: IT38
 Systems and Organization: IT56
 Theorem for Noisy Channels: IT51
 Transmission, Multivariate Information: IT60

Professional Group on Instrumentation

Counter Techniques for Frequency Measurement: I-19
 Impedance Meter: I-21
 Ratiometer: I-23
 Strain Gage Oscillator: I-26
 SWR Indicators, Swept Wide-Range: I-24
 Time Jitter in Video Pulse Trains: I-22
 Transistor Current Amplification Measurements: I-25
 Tuning Circuits: I-20

Professional Group on Microwave Theory and Techniques

Bibliography on Directional Couplers: MTT42
 Calorimeter for Millimeter Power Measurement: MTT53
 Cerenkov Radiation, Millimeter Wave Generation: MTT47
 Coaxial Line with Two Nonconcentric Dielectrics: MTT39
 Compandors for FM Systems: MTT25
 Crystal Checker for Balanced Mixers: MTT36
 Crystals in Balanced Mixers: MTT38
 Dielectric Constant, Optical Measuring Method: MTT44
 Direction Couplers, Bibliography: MTT42
 Double Sideband AM Multiplex System: MTT26
 Generation of Millimeter Waves: MTT47
 Generator, Standby, Remote Control of: MTT24
 Harmonics at Millimeter Wavelengths: MTT46
 Klystron Design Considerations: MTT34

Klystron Stabilization: MTT48
 Magnetic Constant, Optical Measuring Method: MTT44
 Magnetron Research at Columbia: MTT50
 Microstrip Waveguides: MTT33
 Optical Measurement of Dielectric Constant: MTT44
 Oscillator Stabilizing: MTT37
 Pipeline Microwave System: MTT21, MTT28
 Polarization Chart: MTT43
 Radio Relay for Military Use: MTT30
 Receiver, Superheterodyne, 150 KMC: MTT49
 Repeater Site: MTT23
 Shielded-Strip Transmission Line Impedance: MTT41
 Site Selection: MTT22, MTT23
 Spectroscopic Components: MTT45
 Stabilizing Microwave Oscillators: MTT37
 Superheterodyne Receiver for 150 KMC: MTT49
 Testing with Millimicrosecond Pulses: MTT32
 Transco Microwave System: MTT31
 Transmission Line: MTT41, MTT43
 Chart: MTT43
 Shielded-Strip, Impedance: MTT41
 Traveling-Wave Amplifier, Millimeter Wavelength: MTT52
 Traveling-Wave Electron Deflection System: MTT35
 Trunk Service, Microwave System for: MTT27
 Union Electric System Installation: MTT29
 Waveguide: MTT33, MTT40, MTT51, MTT54
 Circular, Components for: MTT40
 Flat, for Millimeter Range: MTT54
 Low-Loss Dielectric: MTT51
 Microstrip: MTT33

Professional Group on Nuclear Science

Abstracts: NS4
 Accelerator, Electron Analogue: NS3
 Brookhaven Reactor Instrumentation: NS1
 Temperature Control, Reactor Outlet: NS2

Professional Group on Reliability and Quality Control

(formerly Quality Control)

Cumulative Result Sampling Plan: QC18
 Equipment Reliability Program: QC12
 Panel on Electron Tubes Program: QC15
 Protective Coating Techniques: QC20
 Quality in Production: QC21
 Transistor Reliability, Low-Power Audio: QC19
 Tube Reliability: QC13, QC14, QC15, QC16, QC17
 Inspection Procedures: QC14
 Panel on Electron Tubes Coordination Program: QC15

Techniques: QC17
 U.K. Project: QC13
 Unreliability Study: QC16

Professional Group on Telemetry and Remote Control

(formerly Radio Telemetry and Remote Control)

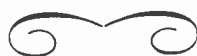
Commutating Device Design: TRC2
 Computer: TRC3, TRC4
 High-Speed Digital: TRC3
 Input-Output Philosophy: TRC4
 Data Processing Facility, Air Force: TRC1
 Discriminator, Subcarrier: TRC5
 Information Theory and Telemetering: TRC7
 Modulator, Slope, for Magnetic Recording: TRC8
 Oscillator, Voltage-Controlled, Transistor: TRC6
 Telemetering and Information Theory: TRC7
 Transistor Oscillator: TRC6

Professional Group on Ultrasonics Engineering

Bibliography on Ultrasonic Delay Lines: UE7
 Delay Lines: UE3, UE5, UE7
 Acoustic, Solid: UE5
 Bibliography: UE7
 Characteristics: UE3
 Flowmeter, Acoustic: UE6
 Machine Tool, Ultrasonic: UE8
 Machining, Mechanical Impedance Transformers for: UE10
 Machining of Tungsten Carbide: UE9
 Medicine and Ultrasound: UE1
 Metal Cleaning with Ultrasonics: UE4
 Micro-Displacement Meter: UE2

Professional Group on Vehicular Communications

Duplex and Multi-Channel Equipment: VC35
 Erie Railroad Communication System: VC41
 Interference: VC37
 Mobile-Microwave System: VC33
 Mobile Radio Service, Frequency Use and Licensing: VC34
 Network Maintenance and Operation Problems: VC40
 Portable Equipment: VC38
 Radio Relay Operation in Power Radio Service: VC39
 U. S. Forest Service Mobile Radio Performance: VC36



NONTECHNICAL INDEX

Professional Group on Aeronautical and Navigational Electronics

Chairman's Report: March, p. 3
East Coast Conference: September, p. 1
National Conference on Airborne Electronics: March, p. 1; June, p. 1

Professional Group on Antennas and Propagation

Editorials

"Why Transactions," by Ernst Weber: March, p. 41

Group News

Administrative Committee: July, p. 91; October, pp. 129, 130
Chapters:
Albuquerque-Los Alamos: January, p. 3; March, p. 44; July, p. 92
Chicago: March, p. 44; October, p. 131
Los Angeles: January, p. 4; July, p. 92
San Diego: January, p. 3
Washington: January, p. 3; July, p. 92
Subdivision: January, pp. 1-2; July, p. 92
Transactions: January, p. 2, 3; March, p. 45; July, pp. 91-92; October, p. 129

Meetings

IRE National Convention, 1954: January, pp. 3, 4
URSI Fall Meeting, 1953: January, p. 4
URSI Spring Meeting, 1954: March, p. 44
Western Electronics Convention, 1954: March, p. 44; July, p. 92

Miscellaneous

Antennas and Waveguide Committee: January, p. 3; July, p. 92
Combining Professional Groups: October, p. 129
Data on Professional Groups: October, p. 130
National Bureau of Standards Relocated: October, p. 130
New Unit for Logarithmic Ratios: pp. 42-44

Personals

Cohn, S.: October, p. 131
Jordan, E. C.: July, p. 93; October, p. 131
Loria, R. M.: October, p. 131
Newman, M. M.: October, p. 131
Rumsey, V. H.: July, p. 93

Professional Group on Audio

Chapters

Cincinnati: July-August, p. 104
Cleveland: May-June, p. 75
Houston: March-April, p. 42; September-October, p. 131

Philadelphia: January-February, p. 1; July-August, p. 104
San Diego: January-February, p. 1

Committees

PGA Committees: November-December, p. 156
Recording and Reproducing: July-August, p. 106
Sound Recording and Reproducing: May-June, p. 76
Tapescripts: January-February, p. 2; September-October, p. 131

Meetings

Department of Defense Magnetic Recording Symposium: January-February, p. 3
IRE National Convention: May-June, p. 73
WESCON: November-December, p. 155

Miscellaneous

How Much Distortion Can You Hear?: March-April, p. 42
RCA Records Color Television on Magnetic Tape: January-February, p. 4

Reports

Chairman, 1953-1954: May-June, p. 71
Editorial Committee, 1953-1954: May-June, p. 72
Treasurer, 1953-1954: May-June, p. 72

Professional Group on Broadcast and Tele- vision Receivers

Administrative Committee Meeting: January, p. 1; July, p. 1
Israel, Dorman D.: April, p. 1

Professional Group on Circuit Theory

Chapters

Albuquerque-Los Alamos: March, p. 79; June, p. 39
Chicago: June, p. 39; September, p. 72; December, p. 41
Formation: March, p. 78
Los Angeles: June, p. 39; September, p. 72; December, p. 41
Philadelphia: March, p. 79; June, p. 39; September, p. 72; December, p. 41
Seattle: June, p. 39; September, p. 72

Committees

Administrative: June, p. 34
Nominations: June, p. 38
Papers and Transactions: June, p. 36
Reviews and Abstracts: June, p. 37
Section-Chapters: June, p. 36
Symposium: June, p. 35

Meetings

URSI Eleventh General Assembly: June, p. 39

Miscellaneous

Future Transactions Topics: September, p. 71
Society for Industrial and Applied Mathematics: March, p. 79

Reports

Chairman: June, p. 35
Papers and Transactions Committee: June, p. 36
Secretary-Treasurer: June, p. 35
Section-Chapters Committee: June, p. 36
Symposium Committee: June, p. 35

Professional Group on Electronic Computers

Administrative Committee: June, p. 1; September, p. 39

Chapters

Akron: September, p. 39
Albuquerque: June, p. 1
Chicago: June, p. 1
Dallas-Fort Worth: September, p. 39
Los Angeles: June, p. 1
New York: June, p. 1
Philadelphia: June, p. 1
Washington: June, p. 1
Eastern Joint Computer Conference: June, p. 1; September, p. 39; December, p. 32
Editorial:
"The Editor" by W. Buchholz: September, p. 1
Membership: June, p. 1; September, p. 39
Report of Chairman: December, p. 31
Standing Committees: September, p. 39
Western Computer Conference, 1955: December, p. 31

Professional Group on Engineering Management

Chairman's Message: February, p. 1
Constitution and By-Laws: February, p. 44
Publications Committee: February, p. 2

Professional Group on Nuclear Science

Annual Meeting Abstracts: September, p. 21
Annual Meeting Program: September, p. 18
Constitution and By-Laws: September, p. 26
Editorial: September, p. 1
Publication Policy: September, p. 31
Secretary's Report: September, p. 25

Professional Group on Ultrasonics Engineering

Chairman's Message: June, p. 1
Chairman's Report: November, p. 1
Constitution and Bylaws: November, p. 35
Membership: June, p. 3

Gulton abstracts

4

MATERIALS RESEARCH • ELECTRONIC COMPONENTS • PRECISION INSTRUMENTS • SYSTEMS ENGINEERING

Variable reluctance pressure gauge has high accuracy over a wide frequency range



Glennite PVR-200-1 Pressure Gauge
Specifications: Frequency response, 0-500 c.p.s.; Natural frequency, 5 KC or higher; Linearity, better than $\pm 1.5\%$ full scale; Pressure ranges, ± 100 mm Hg, ± 5 psi, ± 10 psi; Temperature range, -67° F to $+120^\circ$ F; Size, $11/16"$ D., $5/16"$ high; Weight, 12.5 grams.

The new Glennite PVR-200 series pressure gauges fulfill the requirements of industry for subminiature differential pressure transducers of high accuracy and performance over a wide frequency range.

These rugged gauges are designed to transform differential air or gas pressure into a measurable electrical signal which can be utilized by standard instruments.

New, portable ultrasonic soldering iron eliminates use of flux

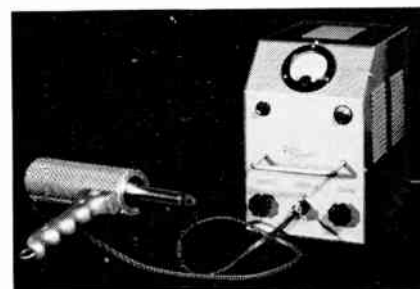
The Glennite Ultrasonic Soldering Iron, Model U-611, is an electrically heated, ultrasonically driven tool for soldering such materials as aluminum and magnesium and their alloys without the use of corrosive flux. It is invaluable for soldering metals or alloys that form refractory oxides and eliminates special surface pre-treatment.

Lightweight, small size and portable, the unit can operate in areas not accessible to bench work. Uses include assembly and installation of wave guides, surface

Exceedingly versatile, they can measure static or slowly varying pressures, as well as those fluctuating at frequencies well into the audio range. These flexible, dynamic test instruments can be easily and accurately calibrated.

Because of their outstanding characteristics, the Glennite PVR-200 gauges permit research where pressure measurement was hitherto difficult or impractical, as in flight testing. In stationary installations, such as wind tunnels, they can be used in confined spaces to measure rapidly varying phenomena which are beyond the capabilities of manometers or the conventional bulky pressure pickup. The transducers are engineered to work with commercially available carrier amplifier systems.

Gulton Mfg. Corp.



Model U-611 Soldering Iron

tinning and filling of voids in aluminum or magnesium castings.

Only 35 watts are needed to drive the soldering tip to sufficiently agitate the molten solder and remove oxide films.

The 9" x 9" x 6" generator, supplied with the soldering iron, has an input of 117 volts at 1 ampere, 60 cycles. It has gain control, off-on power switch and fine frequency tuning control. A gas heated ultrasonic soldering iron, Model U-610, is also available for soldering large components.

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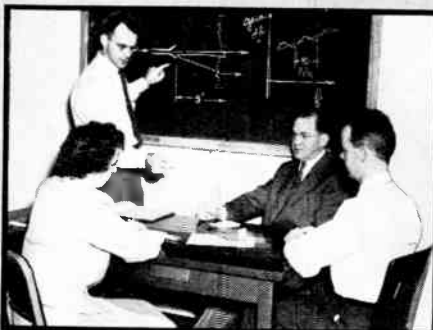
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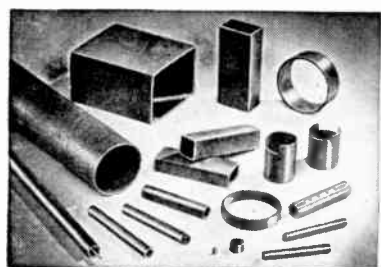
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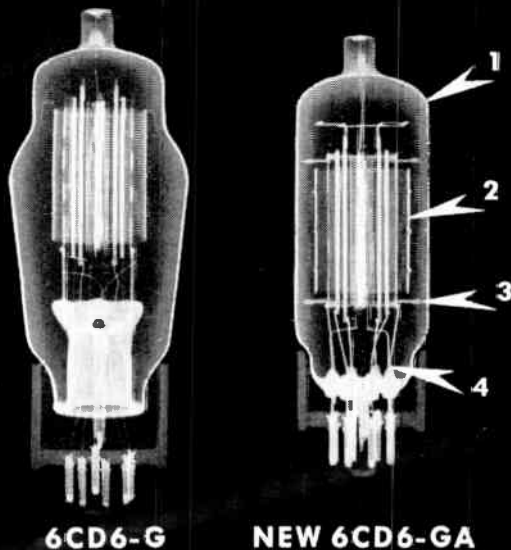
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(Continued on page 84A)

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1. New bulb is straight-side, smaller and sturdier.

2. Redesigned, more shock-resistant tube structure. Redesigned plate, with larger area.

3. Bottom mica, as well as top, now contacts the glass, for greater rigidity. Both micas are completely redesigned to minimize arc-overs.

4. Button-stem base gives shorter and better-separated leads; improves heat conduction.



No price increase! Now one economical tube will serve in both monochrome and color TV sets!

NEW high-rating tube performance, arc-overs cut 'way down . . . yet price stays the same as the prototype 6CD6-G! Plate positive-pulse voltage now is 7,000 volts, against 6,600 volts. Plate dissipation has been increased one-third—from 15 watts to 20 watts.

Every 6CD6-GA gets an arc-over test at absolute max ratings. This built-in, tested-in freedom from tube arcing, with high-capacity performance as shown by the new ratings, makes G.E.'s new sweep tube equally suitable for color TV along with black-and-white.

Consequently, you need specify and stock only

one tube for monochrome and color. You save on inventory . . . and save substantially on tube cost, at the 6CD6-GA's low price. Also, TV quality benefits. Fewer arc-overs mean less horizontal picture streaking.

To high-rating tube performance, add important structural improvements. These make the new 6CD6-GA more shock-resistant—far longer-lived. The tube also takes up less chassis space than before. Side-by-side X-ray pictures above show details of Type 6CD6-GA's new design.

Ask for complete information! *Tube Department, General Electric Co., Schenectady 5, New York.*

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(Continued on page 86A)

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BASIC TOROIDAL MEASUREMENTS

Initial Permeability μ_0 (1Mc)	125
Figure of Merit Q (1Mc)	400
Loss Factor $\frac{1}{\mu_0 Q}$ (1Mc)	.000020
$\mu_0 Q$ (5Mc)	.000050
(10Mc)	.000130
(20Mc)	.000500
μ_0 vs Frequency Characteristics	Good to over 30 Mc
Q vs Frequency Characteristics	Good to over 30 Mc
Curie Temperature ($^{\circ}\text{C}$)	350
Temp. Coeff. of μ_0 (1Mc) $\% / ^{\circ}\text{C}$ (25°C to 70°C)	± 0.10 max.
Temp. Coeff. of Q (Same units as above)	— 0.75
Saturation Flux Density	
Bs (gauss) at Hdc = 25 oersteds	3300
Max. Permeability μ max	400
Coercive Force H (oersteds)	2.10
Residual Magnetism Br	1800

	F-125 DIA. ROD .250" \pm .015"	F-214 DIA. ROD .330" \pm .020"	F-429 WIDTH .725" \pm .025" THICKNESS .125" \pm .030" — .000"
LENGTH	PART NO.	PART NO.	PART NO.
7.520 \pm 7/32	1	6	11
6.250 \pm 3/16	2	7	12
5.300 \pm 5/32	3	8	13
4.625 \pm 1/8	4	9	14
4.100 \pm 1/8	5	10	15

*Camber .011 per inch

TYPICAL ANTENNA ROD MEASUREMENTS

FREQUENCY	Q	C = mmf.
0.6	310	360
0.8	331	200
1.0	325	126
1.2	325	85
1.4	310	63

TEMPERATURE COEFFICIENTS

Antenna Rod No. F-214 (.330 x 7.520"). Standard Test Coil—Space wound solenoid 85 turns #26 AWG. Formex copper, occupying approx. 90% of length of rod and centered on rod. (Resonates at 1 Mc. with 126 mmf.)

$$TC = \frac{\% \Delta \mu_0}{\mu_0} (25^{\circ} \text{ to } 75^{\circ} \text{C})$$

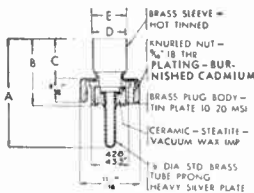
Temp. Coeff. of Rod. ± 1.0 to ± 2.0
Temp. Coeff. of Coil only ≈ 0



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101 Series furnished with 1/4", .290", 5/16", 3/8", or 1/2" ferrule for cable entrance. Knurled nut securely fastens unit together. Plugs have ceramic insulation; sockets bakelite. Assembly meets Navy specifications.

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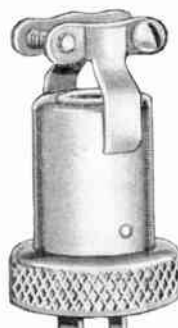
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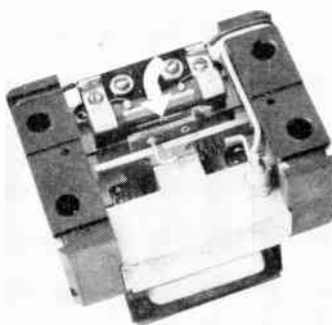
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Institute, University of Michigan, Ann Arbor, Mich.
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(Continued on page 88A)

STABILITY...

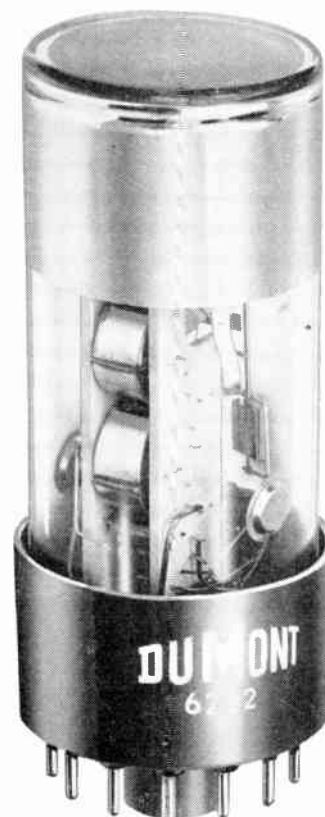
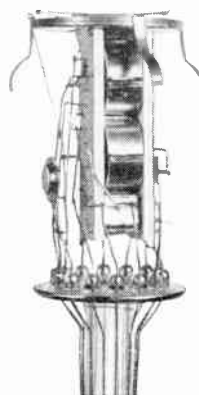
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between cathode and
all other electrodes
Anode Luminous Sensitivity
105 v/stage; 0 cps
145 v/stage; 0 cps
Current Amplification at:
105 v/stage
145 v/stage
Average Anode Current
Peak Anode Current
Tube Diameter
Seated Height to Center
of Window

S11

60 μ A/lumen

13 A/lumen
120 A/lumen

215,000
2,000,000
5 ma
25 ma
2 \pm 1/16 in.
4-7/8 \pm 3/16 in.

The performance features of the Type 6292 are representative of those of the entire line of Du Mont Multiplier phototubes, covering the entire range of sizes from 3/4-inch to 16 inches. All are built to Du Mont's rigid specifications for quality, and are backed by the well known Du Mont guarantee. For full technical details on the Type 6292, or other Du Mont multiplier phototubes, write the *Technical Sales Department, Allen B. Du Mont Laboratories, Inc., 2 Main Avenue, Passaic, N. J.*

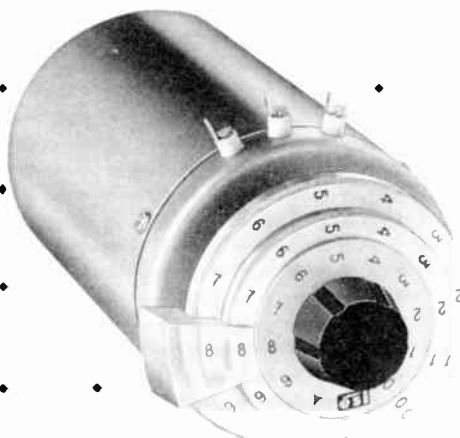
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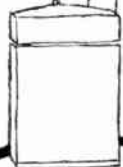
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 Nelson, W. C., 610 Garrett, Pasadena, Calif.
 Nidus, L. S., 1324 Rosedale Ave., New York 72, N. Y.
 O'Brien, E. D., 1060 S. Broadway, Rm. 807, Los Angeles, Calif.
 Oefinger, H. C., Safety Center, 653 Main St., Stamford, Conn.
 Ohlmann, G. A., 1295 Morningside Dr., Sunnyvale, Calif.
 Oliver, E. F., 14 Diana La., N., Fairborn, Ohio
 Phillips, W. E., Jr., 1221 Virginia Ave., Lakewood, Ohio
 Podell, R. L., 1417 Chandler Dr., Fair Lawn, N. J.
 Porat, D. I., Weizmann Institute of Science, Rehovoth, Israel
 Pounds, W. H., 7432 Olive Tree La., Highland, Calif.
 Quirk, J. B., 2133 Sunset Dr., Owensboro, Ky.
 Rastovich, M., 7506 S.E. Reedway, Portland 6, Ore.
 Reading, A., 33 Pine St., Hamilton, Ont., Canada
 Regniere, J. P., 1090 Pere Marquette, Quebec, Que., Canada
 Reilly, R. A., 48 Grant Ave., Somerville, N. J.
 Reimherr, R. N., 4450 Gondar Ave., Long Beach, Calif.
 Rhodes, W. H., 6902 Marlborough Rd., Baltimore 12, Md.
 Risley, M. I., 118 S. 19 St., Omaha 2, Nebr.
 Robbins, L. W., 17 Selfridge Rd., Reading, Mass.
 Robinson, D. E., 4030 Maypole, Chicago 24, Ill.
 Rossi, S., 500 Madison St., Hoboken, N. J.
 Rubin, S., 237 Harvard, Malden 48, Mass.
 Sanders, H. E., Controls, Bldg. 30, General Electric Co., Cincinnati, Ohio
 Sang, W. W., 1202 E. Pontiac St., Fort Wayne, Ind.
 Sarrafian, G. P., 4580 Bordeaux, Dallas, Tex.
 Schmidt, H. L., 409 Dupont St., Toronto 4, Ont., Canada
 Schock, H. E., Jr., Rittenhouse Claridge, Philadelphia 3, Pa.
 Schroer, C. F., 7814 Maplewood Industrial Ct., St. Louis 17, Mo.
 Schwartz, H., Electronic Fabricators, Inc., 682 Broadway, New York 12, N. Y.
 Scott, W. A., 157 Saratoga Ave., Yonkers, N. Y.
 Shaw, J. R., 10343 S. Morgan St., Chicago 43, Ill.
 Shipe, J. J., 7322 Reeds Rd., Overland Park, Kan.
 Shoemaker, H. M., 1362 S. Flower St., Los Angeles 15, Calif.
 Olsen, R. T., 941 Jefferson Ave., Brooklyn, N. Y.
 Osick, W. R., 4141 1/2 Monroe St., Los Angeles, Calif.
 Park, K. R., 64 Avenue De Salaberry, Quebec, Que., Canada
 Paul, R. A., 2445 Laurier Blvd., Sillery, Que., Canada
 Pena, H., Casilla de Correo #4, Lomas de Zamora, Buenos Aires, Argentina
 Peters, J. D., Jr., 11 Hillside Ave., Chelsea 50, Mass.
 Peterson, T. W., 233 Gray Pl., Apt. 402, Scott AFB, Ill.
 Pierson, G. R., 17256 Horace St., Granada Hills, Calif.
 Pinkstaff, J. R., 408 Shannon St., Schenectady, N. Y.
 Plank, P. E., Newton Rd., Baltimore 19, Md.
 Seth, H. P., 3846 W. Grand Ave., Chicago 51, Ill.
 Simon, H., 3033 Brighton 14 St., Brooklyn 35, N. Y.
 Smikle, K. R., 150 32 -115 Rd., Jamaica 36, L. I., N. Y.
 Smith, L. C., 1057 Prospect Ave., Westbury, L. I., N. Y.
 Smith, L. L., Jr., 543 W. Butterfield Rd., Elmhurst, Ill.
 Sokoloff, B. A., 32 Rue D'Alleray, Paris, Seine, France

(Continued on page 90A)

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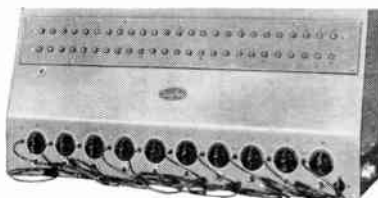
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NETWORK SYNTHESIZER



MODEL
 NS-1



New network synthesizer* and laboratory filter for experimental use. Accelerates network design by eliminating time-consuming design calculations.

Fifty-section delay line permits rapid synthesis of filter characteristics. Ten cathode followers, each with attenuator and polarity selector switch, permit any 10 voltages to be selected and com-

bined, to obtain 10 terms of a Fourier series or any 10-step approximation to a transient response function. Voltages can be added in accordance with harmonic analysis schedule of any selectivity curve.

The NS-1 is simple to operate, and has excellent stability. Controls can be reset, to repeat a desired network. The unit is completely self-powered.

*Described in RCA Review, June 1954.

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For Microphone Heads**

"Vibration Isolation" has helped solve the increasing problem of mechanical vibrations in high fidelity reproduction of sound.

For years, sound engineers have been plagued by mechanical vibrations caused by movement of grips, dollies and other studio equipment. And the progressive development of the high fidelity microphone has increased the importance of eliminating the adverse effects of these disturbances.

Faced with this problem, design engineers of a leading manufacturer of microphones and related electronic equipment consulted with LORD engineers. LORD's 30 years of experience and knowledge in vibration control resulted in a bonded tube-form joint of live rubber which effectively isolated mechanical vibrations from the microphone head.

"Vibration Isolation" is the answer to only one of the many problems presented to and solved by LORD engineers. If you are interested in producing tape recorders, microphones and other types of reproduction equipment, LORD engineers are ready to consult with you. Let them help you produce equipment of the most exacting professional standards with LORD rubber bonded products.

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Sylvania 3-2442

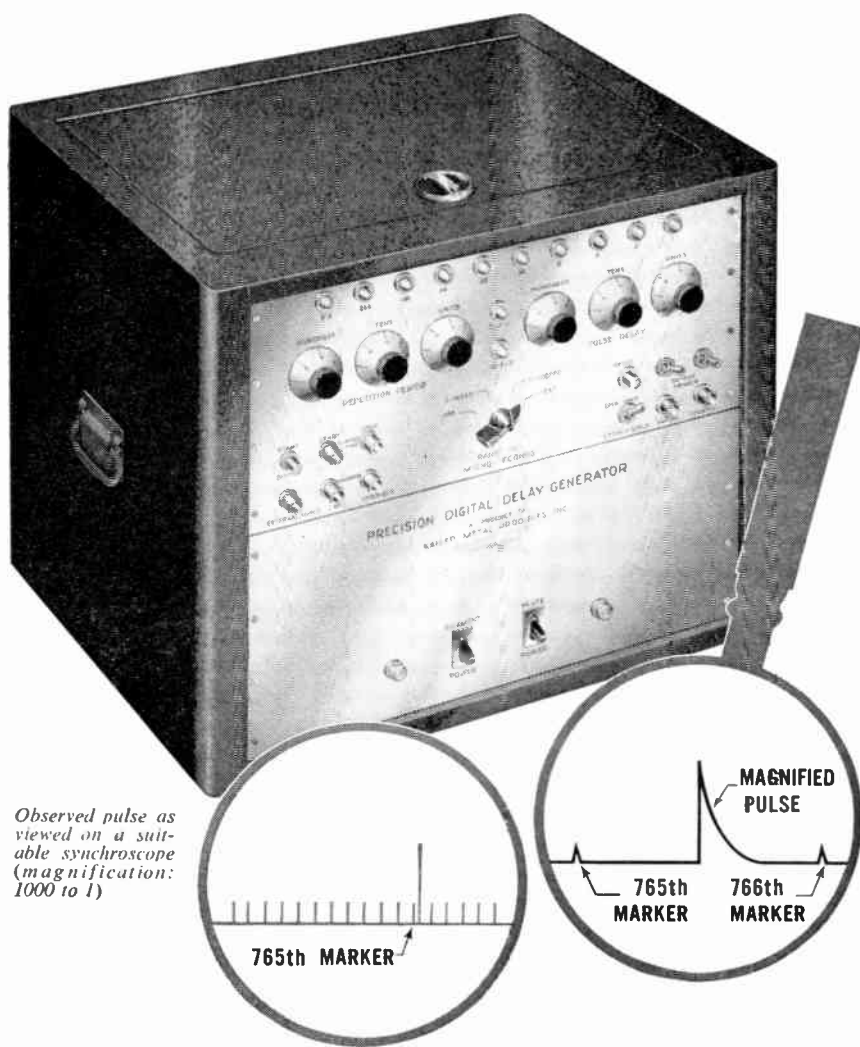
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 Stevens, A. M., Jr., Vauxhall St., Ext., R.F.D. 2, New London, Conn.
 Stevenson, D. D., 133 A Hornet, China Lake, Calif.
 Stimpson, L. D., Jr., 3388 Rosewood Ave., Los Angeles 66, Calif.
 Stripeika, A. J., 6204 Majestic Ave., Oakland 5, Calif.
 Summers, C. R., 28110-B S. Abingdon, Arlington 6, Va.
 Swing, R. E., 21 Notre Dame Rd., Bedford, Mass.
 Sze, T. W., Electrical Engineering Department, University of Pittsburgh, Pittsburgh, Pa.
 Tate, J. P., Jr., 2121 N. Hollister St., Arlington 5, Va.
 Tetters, D. R., 3 B 212, Bell Telephone Laboratories, Whippany, N. J.
 Thalmann, V., Gundeldingerstr. 325, Basel, Switzerland
 Thomas, E. K., R.F.D. 2, 1 Peakham Cir., Sudbury, Mass.
 Thomas, G. W., Box 133, R.F.D. 7, Oklahoma City, Okla.
 Tibbits, A., c/o Young, Trott & Co., Ltd., Hamilton, Bermuda
 Titherington, R. H., Jr., 15 Gibson St., Cambridge 38, Mass.
 Todd, J., 924 Broadway, Boulder, Colo.
 Toler, E. L., Jr., 522 Eighth St., Virginia Beach, Va.
 Tringale, S. R., 141 Hillsdale Rd., W. Somerville, Mass.
 Triplett, J. E., 6410 Knollbrook Dr., Hyattsville, Md.
 Tyksinski, S. P., 3644 N. Leclair Ave., Chicago 41, Ill.
 Tyllinski, F. V., Grayhill, 561 Hillgrove Ave., La Grange, Ill.
 Verstrate, J. W., 22 A Vosmaerstraat, Rotterdam W., Holland
 Vierling, H. D., 7915 Elmhurst Ave., Parkville 14, Md.
 Visocnik, H. R., 1748 S. Fifth St., Columbus 7, Ohio
 Vogel, R. P., 21 Young Ave., Yonkers, N. Y.
 Wainscott, D. M., 110 W. Kirby Ave., Champaign, Ill.
 Walker, D. H., 320 Willow Ave., Frederick, Md.
 Walls, D. M., 3747 Eaton, Kansas City, Kan.
 Warhurst, J. S., 194 Hubbard St., Glastonbury Conn.
 Wessel-Berg, T., Stanford University, Microwave Laboratory, Stanford, Calif.
 Whitley, W. M., 10949—109 St., Edmonton, Alta., Canada
 Wiek, M. H., Jr., 99 Chapel Rd., Havre de Grace, Md.
 Williams, H. C., Det. 4, 9470-TU, Ft. Hauchuca, Ariz.
 Williamson, R. T., 216 Beach 126 St., Belle Harbor, N. Y.
 Willner, W. H., Hazel Trail, Herald Harbor, Crownsville, Md.
 Witlin, J. J., 1332 W. Sixth St., Brooklyn 4, N. Y.
 Woods, E. L., 14035-C Dicky St., Whittier, Calif.
 Worthington, R. L., 255 Prospect St., E. Orange, N. J.
 Wott, H. W., 545 N. Locust St., Oak Harbor, Ohio
 Young, E. A., c/o Hawk Eye Works, 20 Ave. E., Rochester 4, N. Y.
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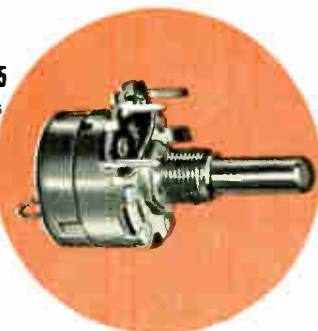


Type UPM-45

For TV preset control applications. Control mounts directly on printed circuit panel with no shaft extension through panel. Recessed screwdriver slot in front of control and 3/8" knurled shaft extension out back of control for finger adjustment. Terminals extend perpendicularly 7/32" from control's mounting surface.

Type GC-U45

Threaded bushing mounting. Terminals extend perpendicularly 7/32" from control's mounting surface. Available with or without associated switches.



Type U70 (Miniaturized)

Threaded bushing mounting. Terminals extend perpendicularly 5/32" from control's mounting surface.



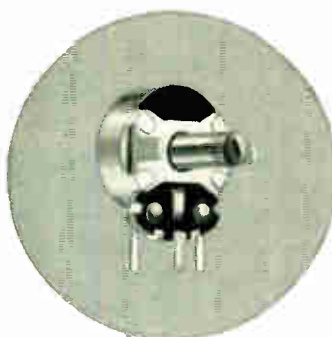
Type YGC-B45

Self-supporting snap-in bracket mounted control. Shaft center spaced 29/32" above printed circuit panel. Terminals extend 1-1/32" from control center.



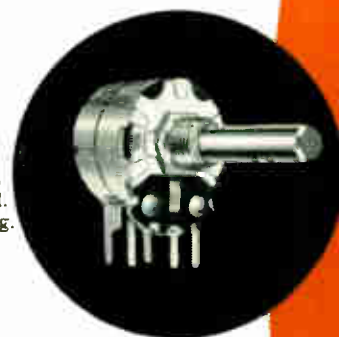
Type XP-45

For TV preset control applications. Control mounts on chassis or supporting bracket by twisting two ears. Available in numerous shaft lengths and types.

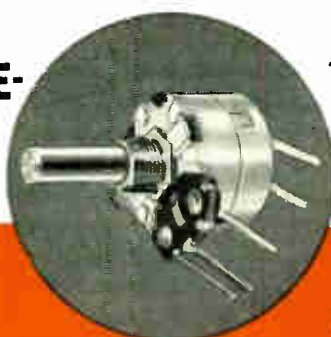


Type XGC-45

For applications using a mounting chassis to support printed circuit panel. Threaded bushing mounting.



VARIABLE RESISTORS FOR SOLDERLESS "WIRE- WRAP" CONNECTIONS



Type WGC-45

Designed for solderless wire-wrapped connections with the use of present wire-wrapping tools. Available with or without switch and in single or dual construction.

The controls illustrated are typical constructions. CTS' years of engineering and technical experience makes available many other types for your automation needs.



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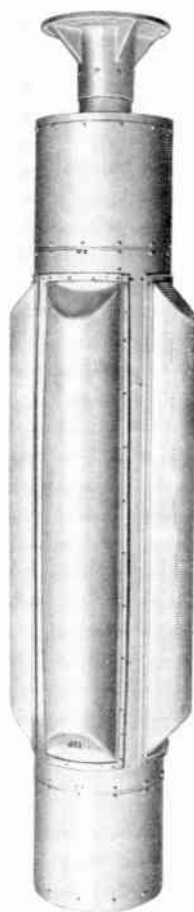
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has these important advantages

- ▶ no moving parts
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- ▶ used successfully both in the U.S.A. and in other countries
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- ▶ shipped completely assembled, and tuned when frequency is specified
- ▶ can be readily retuned to any frequency within its range
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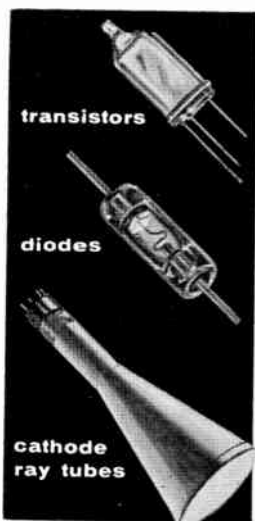
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ALBUQUERQUE-LOS ALAMOS

"Time Conversion Pulse Height Analyzers," by R. J. Watts, Los Alamos Labs., Univ. of California; February 18, 1955.

"Mathematics of Boards, Panels and Committees," by Dr. J. W. McRae, Sandia Corp., and "Problems Encountered in Phase Measurements," by Claxton Foster, Technology Instruments Corp.; January 20, 1955.

ATLANTA

Tapescript: "Method for Time or Frequency Compression-Expansion of Speech," by Messrs. Everitt, Fairbanks and Jaeger, University of Illinois; January 21, 1955.

BALTIMORE

"Transmission Lines for Millimeter Waves," by Dr. D. D. King, Johns Hopkins Radiation Lab.; February 9, 1955.

BEAUMONT-PORT ARTHUR

"Recent Advances in the Reproducing Art," by A. M. Wiggins and H. T. Souther, Electro-Voice, Inc.; January 17, 1955.

CEDAR RAPIDS

Installation of officers; January 29, 1955.

CLEVELAND

"Teaching Old Dogs New Tricks in Electronic Instrumentation," by W. C. Moore, Boonton Radio Corp.; January 27, 1955.

CONNECTICUT VALLEY

"A Resistor Network for Simulating Geological Conditions," by J. H. Baker, Schlumberger Well Surveying Corp., and "The Solution of Simultaneous Equations on a Differential Analyzer," by G. H. Martin; January 20, 1955.

DAVTON

"Air Defense," by Dr. A. G. Hill, MIT Lincoln Laboratory; February 3, 1955.

DENVER

"Transcontinental Microwave Radio Relay Systems," by E. L. Broders and F. D. Borstadt, both of American Tel. and Tel. Company; December 10, 1954.

"Synthesis of Aperture Antennas," by Dr. C. T. Johnk, University of Colorado; January 14, 1955.

DES MOINES-AMES

"The Effect of Utilization of Engineering Manpower," by F. D. Agathe, Allis Chalmers, February 15, 1955.

"Manpower Development," by George Downing, General Electric Company; January 18, 1955.

DETROIT

"Principles of Color Television," by C. N. Hoyer, R.C.A.; January 26, 1955.

"Controls that Think and Act Automatically," by C. R. Molenaar, General Electric Company; February 18, 1955.

EL PASO

"Field Measurements of Guided Missiles," by Lt. Col. W. J. Bronley, Flight Determination Lab., WSPG, N. Mex.; January 27, 1955.

EMPORIUM

"The 600 Mill Line of T.V. Tubes," by A. W. Peterson, Sylvania Electric Products, Inc.; January 25, 1955.

EVANSVILLE-OWENSBORO

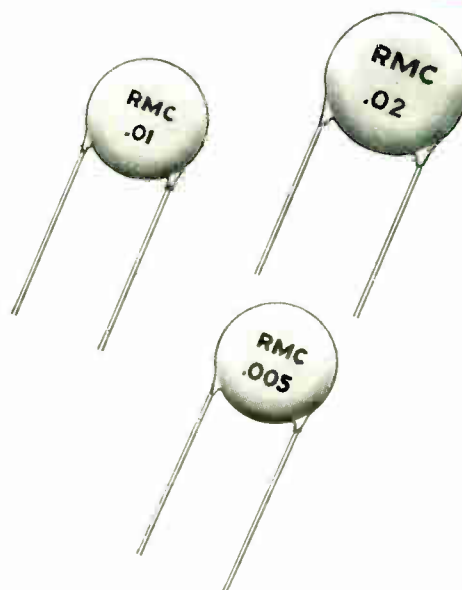
"Application of IBM Equipment to Engineering and Statistical Problems," by Dr. P. Sterbenz, I.B.M.; February 9, 1955.

(Continued on page 94A)

RMC BY-PASS DISCAPS

RMC Type B "Heavy Duty" DISCAPS are designed for all by-pass or filtering applications and meet or exceed RTMA REC-107-A specifications for type Z5Z capacitors

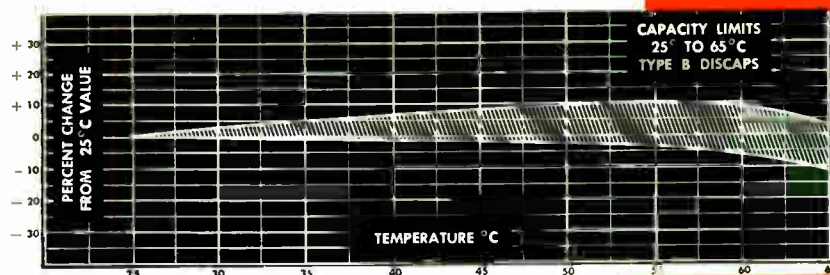
- Rated at 1000 working volts
- Available in any capacity between .00015 MFD and .04 MFD
- Minimum capacity change between $+10^{\circ}\text{C}$ and $+65^{\circ}\text{C}$ (See Curve)
- Heavy duty construction means greater dependability at no extra cost



**PLUG-IN TYPES
NOW
AVAILABLE**



RMC is now producing plug-in DISCAPS designed for printed circuit applications. Available in by-pass, temperature compensating, and stable capacity types, plug-in DISCAPS have the same high specifications featured in standard RMC capacitors. Leads are No. 20 tinned copper (.032 diameter) and are available up to $1\frac{1}{2}$ " in length. Popular range of sizes for all applications.



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CERAMIC
CAPACITORS

RMC

RADIO MATERIALS CORPORATION

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Two RMC Plants Devoted Exclusively to Ceramic Capacitors

**ALLIED
CONTROL**
starts with
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**GARFIELD
WIRE**

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Wire*

When you specify Allied Control, you're asking for the best. For example, take the relay shown here. It's the finest there is — in split-second response, unfailing accuracy, rugged dependability.

Design — construction — materials — all have to be the best for Allied... and the heart of this relay is wound with Garfield Enameled Magnet Wire.

You'll start with the finest in wire, too, when you specify Garfield. Our modern drawing and enameling equipment, our rigid production control and our stringent inspection system are geared to produce only top-quality wire with tolerances closer than NEMA specifications.

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Section Meetings

(Continued from page 92A)

FORT WAYNE

"Telemetry Challenge, 1955," by W. J. Mayo-Wells, Johns Hopkins University; February 3, 1955.

"Illustrating the Technical Report," by Adrian TerLouw, Eastman Kodak Company; February 10, 1955.

HAMILTON

"Microwave Techniques," by Arthur Dinnin, Bell Telephone Co.; January 10, 1955.

HAWAII

"Inspection Tour and Demonstration of CAA Airport Surveillance Radar, ASR-2," by Frank Kadi, Civil Aeronautics Administration; February 10, 1955.

HOUSTON

"Electronics, Engineers and Education," by Dr. J. D. Ryder, President, IRE; February 8, 1955.

HUNTSVILLE

"The Southern Research Institute, Its Objectives and Functions, and Summary of Its Projects," by Sabert Oglesby, Southern Research Institute; December 16, 1954.

"Analysis of Data Recording Systems," by T. L. Greenwood, Redstone Arsenal; January 26, 1955.

INVOKERN

"Low Noise Travelling Wave Tubes," by Dr. D. A. Watkins, Stanford University; December 31, 1954.

ITHACA

"Electronics and Medicine," by Dr. E. B. Wright, University of Rochester; February 1, 1955.

LITTLE ROCK

"Recent Amendments of F.C.C. Rules and Standards Affecting Color Television and AM Broadcasting," by J. G. Roundtree, consulting radio engineer; February 8, 1955.

"Electronics in the Computer Systems," by R. R. Pierce, I.B.M.; January 11, 1955.

LONDON

"The Principles of Color Television," by C. N. Hoyler, RCA Victor Ltd.; January 27, 1955.

"Primary Standard and the Radio Engineer," by Dr. J. T. Henderson, Director, IRE Region 8; February 10, 1955.

LONG ISLAND

"Measurements at DC and Power Frequencies," by John H. Miller, Weston Electrical Instrument Corp.; January 27, 1955.

"Basic Audio and Radio-Frequency Measurements," by W. R. Thurston, General Radio Company; February 3, 1955.

"Computers," by J. Johnson, I.B.M.; February 8, 1955.

"Oscillography," by W. G. Fockler, A. B. DuMont and A. A. Emmerling and H. H. Chamberlain, General Electric Company; February 10, 1955.

LOS ANGELES

"Electronics and Mathematical Analysis in Business Operations," by Dr. Simon Ramo, Ramo Wooldridge Corp.; January 11, 1955.

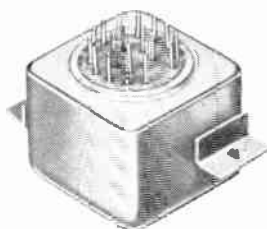
"Instrumentation in Smog Research," by Dr. F. Littman, Pasadena Lab. of S.R.I., "Surface Wave Antennas," by Dr. R. S. Elliott, Hughes Aircraft, and "New Developments in Traveling Wave Tubes and Backward-Wave Tubes," by Dr. D. A. Watkins, Stanford University; February 1, 1955.

(Continued on page 96A)



Airborne Components...

a C. A. C. Specialty



Depicted—6KC 100 Watt Unit
Less than 1.65 cubic inches

POWER TRANSFORMERS

Range—400-6000 cps
Efficiency—up to 95%
Wattage—6mw-200 watts
Temperature—-55 to +155° C.

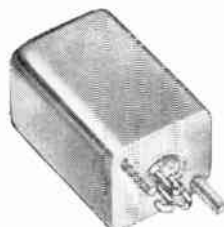


W—3/4"
L—3/4"
H—5/16"

PULSE TRANSFORMERS

Pulse Width—.2-50 microseconds
Rise Time—from .03 microseconds

- Blocking oscillator
- Pulse coupling
- Toroidal construction



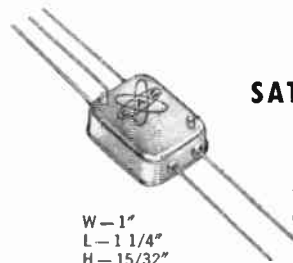
W—23/32"
L—23/32"
H—11/16"

Illustrated
4KC
Band Pass

SUB-MINIATURE FILTERS

For Chassis Mount
Frequency—2.3-35Kc
Impedance in—600-10K Ohms
Impedance out—Grid

- Hermetic Sealed
- Temperature Compensated
- Internal D.C. Isolation
- Balanced or Unbalanced
- Military Specifications



W—1"
L—1 1/4"
H—15/32"

SATURABLE REACTORS

Applications

- Servo Systems
- Data Telemetering
- Remote Frequency Control

Illustrated—High Frequency Reactor Tuned by Varying D. C. Current



MAGNETIC AMPLIFIERS

Wattage (output) .5-200 watts
Response—1 cycle up

W—1 1/4"
L—1 3/4"
H—2 5/32"

Illustrated—Auto Pilot Application for Printed Circuit Mounting



SUB-MINIATURE TUNED CIRCUITS

For Printed Circuit Applications

- Multiple Tuned Transformers
- Delay Lines
- Tuned Circuits

W—1"
L—4 1/4"
H—7/16"

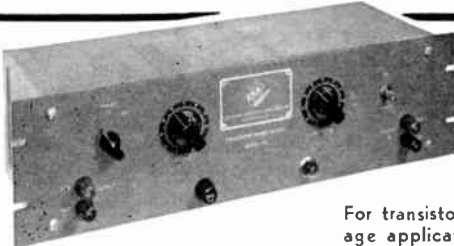
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- Model 110D Price \$179.50
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(metered) Price \$224.50

Specifications

INPUT 95-125v, 60 cps. AC
DUAL VOLTAGE OUTPUT ..Continuously Variable
Either Output 0-1/10/100v*
DC CURRENT (max.) Either Output 100 MA
RIPPLE Less than 0.01%
INTERNAL IMPEDANCE ..Less than 15/20/100 ohms
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Section Meetings

(Continued from page 94A)

MIAMI

Taped recording: "Principles, Design and Operation of Color Television," by John Wentworth, RCA, and question and answer forum conducted by C. X. Castle, WGBS-TV; January 28, 1955.

NEW ORLEANS

Colorcast and demonstration of the color camera and monitoring equipment of WDSU; January 30, 1955.

"Electrons, Engineers and Education," by John D. Ryder, President, IRE; February 7, 1955.

NEW YORK

"The Past, Present and Future of Magnetic Recording," by J. S. Boyers, The National Company; January 5, 1955.

"Limitations on the Production and Measurement of Very Low Pressure," by D. Alpert, Westinghouse Research Labs; February 2, 1955.

NORTH CAROLINA-VIRGINIA

"Recent Developments in Raydist Systems," by A. L. Comstock, Hastings Instrument Company; February 18, 1955.

NORTHERN NEW JERSEY

"Highlights of Antenna Lore," by E. A. Laport, RCA International; February 9, 1955.

OKLAHOMA CITY

"AC Network Calculators," by Miles Maxwell, Westinghouse Electric Corp.; February 16, 1955.

OTTAWA

"Recent Advances in Microwave Tubes," by Dr. J. R. Pierce, Bell Telephone Labs.; January 27, 1955.

PHILADELPHIA

"Electronic Instrumentation for the Brookhaven Nuclear Reactor," by J. Binns, Brookhaven National Laboratory; January 13, 1955.

"Operation Dew Line," (Distant Early Warning) by V. B. Bannall, Western Electric Company; February 2, 1955.

PHOENIX

"Smog" (included movie "The City That Disappears") by Dr. Beardsley Graham, Stanford Research Institute; January 21, 1955.

"The Mission and Technical Philosophy of the Army Electronic Proving Ground," by Dr. Robert Burns, Fort Huachuca; February 18, 1955.

PORTLAND

"Bonneville Power Administration Microwave Communication and Load Dispatching Systems," by E. Warchol and L. W. Danilson; January 20, 1955.

"High Fidelity and Speaker Enclosures," by J. C. Riley, Iron Fireman-Electronics Div.; February 16, 1955.

ROME-UTICA

"Information Theory," by Dr. Stanford Goldman, Syracuse University; February 3, 1955.

SACRAMENTO

"Locked Oscillators," by L. S. Cutler, Gertsch Products Corp.; February 11, 1955.

ST. LOUIS

"Global Communications in the Air Force," by R. P. Mueller, Scott Air Force Base; December 16, 1954.

"The Iatron Variable Persistence Cathode Ray Tube," by Harold Jacobsmeier, Emerson Electric Corp.; January 27, 1955.

(Continued on page 99A)

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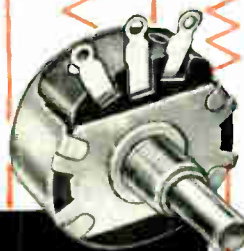
CHICAGO TELEPHONE SUPPLY
Corporation



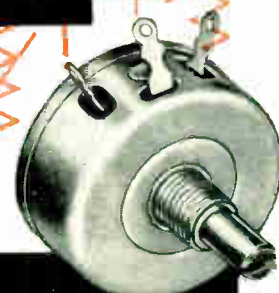
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resistor. The total resistance
can be varied from 3 ohms to
15,000 ohms, depending upon the
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CTS 25 Series
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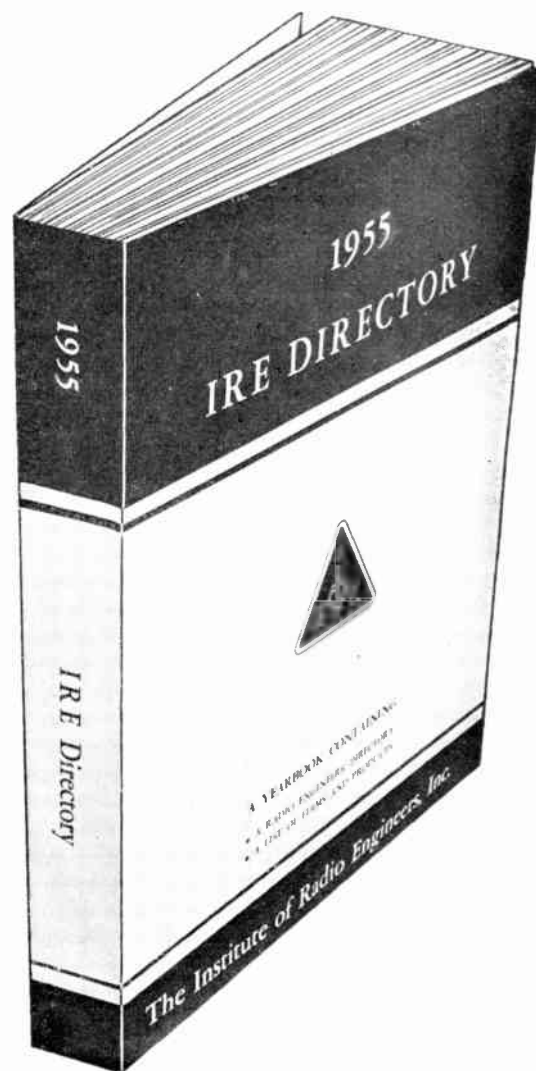
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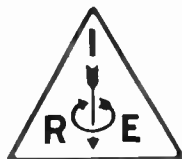
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Section Meetings

(Continued from page 96A)

SAN ANTONIO

"Digital Computers," by W. M. Amoth and W. Robert Hydeman, Remington Rand, December 16, 1954.

"Recent Improvements in the Reproducing Art," by A. M. Wiggins and H. T. Souther, Electro-Voice, Inc.; January 20, 1955.

"High High Frequencies (Generation of Millimeter Waves)," by Dr. J. D. Ryder, President, IRE; February 9, 1955.

SCHENECTADY

"Electrons, Engineers and Education," by Dr. J. Ryder, President, IRE; January 21, 1955.

SYRACUSE

"The Psychological Matrix Rotation Computer," by G. T. Jacobi, General Electric Company; February 7, 1955.

TORONTO

"The Automatic Switching of Long Distance Calls," by F. H. Western; January 17, 1955.

"Problems of Television Receiver Manufacturing in Canada," by R. Munitz, Canadian Westinghouse Company; January 31, 1955.

Students' Night: "Electrical Characteristics of Human Nervous Systems," by L. D. Pengelly, "Magnetic Recording," by J. F. Hanson and "Binary Numbers and Boolean Algebra," by W. F. Elliott, all students; February 10, 1955.

TULSA

"Characteristics of Mechanical Filters," by Bob Lomig, Collins Radio Corp.; January 20, 1955.

TWIN CITIES

"A Case History of a Booster Station for Improved UHF TV Reception," by W. C. Morrison, RCA Labs.; January 20, 1955.

VANCOUVER

"Automatic Computers as Applied to Training Devices," by Dr. E. V. Bohn, University of British Columbia; January 17, 1955.

WASHINGTON, D. C.

"Information Storage in the Protein Molecule," by Dr. George Gamow, George Washington University; February 14, 1955.

WILLIAMSPORT

"A General Discussion of Various Color TV Picture Tubes," by Dr. H. B. Law, RCA; January 19, 1955.

SUBSECTIONS

AMARILLO-LUBBOCK

General meeting; January 13, 1955.

BERKSHIRE COUNTY

General meeting; February 2, 1955.

BUENAVENTURA

"Raydac Computer at Point Mugu," by Dr. L. Fein, Computer Control Company; January 13, 1955.

ERIE

Demonstration lecture on "The Principles of Color Television," by C. N. Hoyler, RCA; January 24, 1955.

LANCASTER

"Control of Costs of Research and Development Projects," by H. J. Finison, National Pneumatic Company, Inc.; January 12, 1955.

MID-HUDSON

"High-Voltage Equipment," by Dr. Victor Wouk, Beta Electric Corp.; February 1, 1955.

(Continued on page 191A)

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For operating tape recorders, dictating machines, amplifiers and other 110-volt radio-audio devices from DC or storage batteries. Used by broadcast studios, program producers, executives, salesmen and other "field workers".

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The preferred power supply for 2-way mobile radio installations. Operates from either 6 or 12-volt batteries. Carter Genemotors are standard equipment in leading makes of auto, aircraft, railroad, utility and marine communications.



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C11	6.3	173	.36'
C2	6.3	171	.44'
C22	5.5	184	.44'
C3	5.4	197	.64'
C33	4.8	220	.64'
C4	4.6	229	1.03'
C44	4.1	252	1.03'

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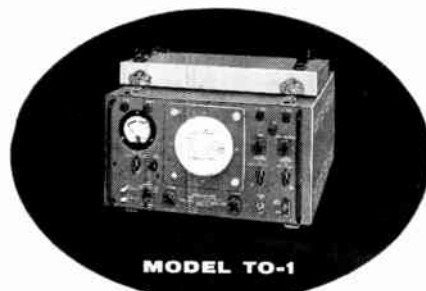
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TV STUDIO MONITOR MODEL M-105



The Polarad Model M-105 is portable — comes in sturdy aluminum case, can be rack mounted as well! And it is one of the finest instruments available to check the picture quality of video signals. Equipped with 12½" aluminized kinescope, capable of presenting highest definition transmitted pictures with exceptionally good "sync" stability over a wide range of operating conditions.

PORTABLE TV WAVE FORM MONITOR



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LOOK AT THESE FEATURES:

1. Can be rack mounted.
2. Can be used for both color and black and white TV.
3. Vertical Amplifier Bandwidth Switch for 2MC, 4MC, 6MC.
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6. Compact and Rugged.

Polarad manufactures a complete line of color TV equipment including a Flying Spot Scanner, Sync Generator, Bar Generator and Color Monitors.

See other Polarad equipment advertised on pages 24A, 39A & 56A.



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UTILIZES STANDARD PLUG-IN DECADES
BASIC UNIT READS OUT TO 10 KC (4 decades)
AIR COOLED (Fan)



(Continued from page 99A)

MONMOUTH

"Color Television," talk and demonstration by I. E. Lempert, Westinghouse Electric Corp.; February 2, 1955.

ORANGE BELT

"General Considerations and Mathematical Development of Reliability in Electronic Systems," by J. H. Parsons and A. Yeiser, Hughes Aircraft Company, and "Your IRE—Let's Discuss It," by John Byrne, Motorola Research Lab.; February 9, 1955.

TUCSON

"Etched Circuitry Processes," by G. McLaughlin and Dr. L. Ott, both of Hughes Aircraft Company; December 16, 1954.



AERONAUTICAL AND NAVIGATIONAL ELECTRONICS

The Dayton Chapter of the Professional Group on Aeronautical and Navigational Electronics met December 2 at the Engineers Club, Chairman Paul Wiegert presiding, Kenneth C. Jordan, Monsanto Chemical Company, presented a paper on "Nuclear-Powered Batteries," in which he outlined the history of atomic batteries from 1878, the time of Faraday's first experiments with radium as a source of electrical power, and described the technical composition and application of the battery. The paper provoked much discussion from the floor.

AUDIO

The Houston Chapter of the Professional Group on Audio met at the Humble Research Center on January 18. Chairman Walter J. Greer presided. A paper on "High-Fidelity Components" was delivered by A. M. Wiggins and Howard T. Souther, Vice President of Engineering and Sales Manager of Electro-Voice, Incorporated.

The meeting of the Cleveland Chapter, at which Chairman Herbert H. Heller presided, was held at WDOK-Cleveland Recording Company Studios on January 20. E. M. Jones, an engineer with the Baldwin Company, presented a paper on "How Much Distortion Can You Hear?"

Hoyt Westcott presided at the meeting of the Albuquerque-Los Alamos Chapter, held on January 17 at the Radiation Therapy Building, Lovelace Clinic, Albuquerque. Ben Sanders, of Sanders Associates, gave a paper on "Ampex Stereophonic System and Ampex 600 Design," and demonstrated the system.

(Continued on page 102A)

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- EXCEED REQUIREMENTS FOR JAN-C-81



CUSTOM TRIMMER ASSEMBLIES

ERIE Style 557 Trimmer is manufactured for Military use and is widely used in Test Equipment and other Industrial Applications. It can be Compactly Mounted in Multiple Groups on practically any desired Phenolic Base Design. Shown here are typical examples of Single and Multiple Space Saving Assemblies.



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MODEL 82

SPECIFICATIONS:

FREQUENCY RANGE: 20 cycles to 200 Kc. in four ranges. 80 Kc. to 50 Mc. in seven ranges.

OUTPUT VOLTAGE: 0 to 50 volts across 7500 ohms from 20 cycles to 200 Kc. 0.1 microvolt to 1 volt across 50 ohms over most of range from 80 Kc. to 50 Mc.

MODULATION: Continuously variable 0 to 50% from 20 cycles to 20 Kc.

POWER SUPPLY: 117 volts, 50/60 cycles, 75 watts.

DIMENSIONS: 15" x 19" x 12". Weight, 50 lbs.

Standard Signal Generator

20 cycles-50 mc.

FEATURES:

- Continuous frequency coverage from 20 cycles to 50 mc.
- Direct-reading individually calibrated dials.
- Low harmonic content.
- Accurate, metered output.
- Mutual inductance type attenuator for high frequency oscillator.
- Stray field and leakage negligible.
- Completely self-contained.

Laboratory Standards



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Professional Group Meetings

(Continued from page 104A)

AUTOMATIC CONTROL

On January 13, The Dallas-Fort Worth Chapter of the Professional Group on Automatic Control met at the Engineering Building, Southern Methodist University. John A. Green, Section Chairman, spoke briefly on IRE general policies and plans. At this meeting, the following officers were elected: Chairman—F. W. Tatum, Head of Electrical Engineering Department, Southern Methodist University; Vice-Chairman—A. R. Teasdale, Chief of Electronic Design, Temco Aircraft Corporation, Dallas; Secretary—H. W. Prier, Lead Systems Design Engineer, Chance Vought Aircraft, Dallas.

BROADCAST TRANSMISSION SYSTEMS

George M. Ives presided at the January 21 meeting of the Chicago Chapter of the Professional Group on Broadcast Transmission Systems. A paper on "The Chromacoder" was presented by Pierre H. Boucheron, Jr., Project Engineer, General Electric Company, Syracuse, N. Y. The Chromacoder is a device for changing a field-sequential color-television signal into a N.T.S.C. color TV signal. Mr. Boucheron outlined its structure and use.

COMMUNICATIONS SYSTEMS

The Washington, D. C., Chapter of the Professional Group on Communications Systems met on January 26 at the Auditorium of the Potomac Electric Power Company. William C. Boese, Chairman, presided. The speaker was Haraden Pratt, Secretary of the Institute of Radio Engineers. Mr. Pratt spoke on the "Birth and Growth of Telecommunications," covering the highlights of the growth of telecommunications, and describing the vicissitudes of such early pioneers in the field as Morse, Collins, and Field. Mr. Pratt's address reflected his intimate knowledge of the history of the early wire, cable, and radio telecommunications organizations and the personalities who directed it.

COMPONENT PARTS

M. P. Feyerherm presided at the September 28 meeting of the Philadelphia Chapter of the Professional Group on Component Parts, held at the Engineers Club of Philadelphia. Alex Bezat, of the Minneapolis-Honeywell Company, spoke on "Electric Insulation at Elevated Temperatures." Mr. Bezat described the Minneapolis-Honeywell program for classification of insulating materials according to useful temperature range, and discussed specific materials in some detail. He also submitted proposals for new industry-wide temperature classes and new test methods.

The Dayton meeting was held at the Engineers Club of Dayton on December 2, with Floyd E. Wenger presiding. A paper on "Computer Components" was pre-

(Continued on page 104A)

COMMON CHARACTERISTICS OF ALL TYPE 2131 GEARED MOTOR GENERATOR UNITS

O.D. of Case.....1.000 inch
Case Length.....3.301
Weight.....7.5 ounces
Frequency.....400 cycles

No. of Poles (Motor).....6
*No Load Speed (Min.).....6500 rpm
Rotor Inertia.....1.1 gram-cm²

*Motor Speed at input to gear train



NEW integral gear head in small servo motors

OUTSTANDING FEATURES OF TYPE 2131 GEARED MOTOR GENERATOR

- New methods of manufacture result in high efficiency
- High torque to inertia ratio to give fast response
- Available for 115 volt—115 volt two phase or single ended tube operation
- High impedance winding for direct plate to plate operation available
- High generator output voltage with excellent signal to noise ratio
- Zero degree phase shift in generator
- All metal parts corrosion resistant
- Extremely wide operating temperature range

*Other models
of one inch O.D. units*

TYPE NO.	DESCRIPTION
2103	Induction Motor
2101	Geared Induction Motor
2028	Motor Generator

Latest catalog and/or complete specification drawings will be sent upon request.

A new line of units has been added to the Kollsman "Special Purpose Motors" family combining precision machining, advanced electrical design and the latest in new materials. An unusual feature of the new line is the integral gear head unit. Contained within a single case is the gear train and motor; or gear train, motor and generator. Gear ratios as high as 300:1 can be supplied.

This new line consists of Induction Motors and Induction Generators supplied separately or combined in a single case one-inch in diameter. The new motors have been designed to give the maximum torque per watt ratio with the minimum rotor inertia. The generators have been designed to give the maximum output voltage with the minimum residual voltage and phase shift.

One of the principal features of the Kollsman "Special Purpose Motors" is the interchangeability of parts which permits numerous electrically different combinations of motor and generator windings within the same case.

INPUT PER PHASE ONLY 1.8 WATTS ELECTRICAL CHARACTERISTICS OF TYPICAL TYPE 2131 GEARED MOTOR GENERATORS

TYPE NO.	MOTOR					GENERATOR		
	EXCITATION FIXED	CONTROL	INPUT PER PHASE	STALL TORQUE	Theoretical Acceleration At Stall	EXCI- TATION FIXED	INPUT	OUTPUT PER 1000 rpm
2131-0411110	26	26	2.3	0.4	25600	26	1.8	.51
2131D-0412120	26	26	4.0	0.6	38500	26	2.2	.68
2131D-0413120	26	26	1.8	0.3	19200	26	2.2	.68
2131-0460600	115	115	4.0	0.6	38500	115	2.6	1.00
2131-0463600	115	55	4.0	0.6	38500	115	2.6	1.00
2131-0470600	115	P-P	4.0	0.6	38500	115	2.6	1.00
	volts	volts	watts	Oz-in	rad/sec ²	volts	watts	volts



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① Check the outstanding engineering design of this modern *printed circuit* Scope. Designed for color TV work, ideal for critical Laboratory applications. Frequency response essentially flat from 5 cycles to 5 Mc down only 1½ db at 3.58 Mc (TV color burst sync frequency). Down only 5 db at 5 Mc. New sweep generator 20-500,000 cycles, 5 times the range usually offered. Will sync wave form display up to 5 Mc and better. Printed circuit boards stabilize performance specifications and cut assembly time in half. Formerly available only in costly Lab type Scope. Features horizontal trace expansion for observation of pulse detail — retrace blanking amplifier — voltage regulated power supply — 3 step frequency compensated vertical input — low capacity nylon bushings on panel terminals — plus a host of other fine features. Combines peak performance and fine engineering features with low kit cost!

Heathkit TV SWEEP GENERATOR KIT ELECTRONIC SWEEP SYSTEM

② A new Heathkit sweep generator covering all frequencies encountered in TV service work (color or monochrome). FM frequencies too! 4 Mc — 220 Mc on fundamentals, harmonics up to 880 Mc. Smoothly controllable all-electronic sweep system. Nothing mechanical to vibrate or wear out. Crystal controlled 4.5 Mc fixed marker and separate variable marker 19-60 Mc on fundamentals and 57-180 Mc on calibrated harmonics. Plug-in crystal included. Blanking and phasing controls — automatic constant amplitude output circuit — efficient attenuation — maximum RF output well over .1 volt — vastly improved linearity. Easily your best buy in sweep generators.

MODEL 0-10
\$695.00
Shpg. Wt. 27 lbs.

MODEL TS-4
\$495.00
Shpg. Wt. 16 lbs.

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Professional Group Meetings

(Continued from page 102A)

sented by Gilbert Devy of the Sprague Electric Company. Mr. Devy described two assembled components, a flip-flop circuitry unit, and a phase-shift device. He stressed the requirements for utmost reliability and stated the aim of satisfying RETMA and military standards wherever such requirements have been formalized or specified.

ELECTRON DEVICES

The San Francisco Chapter of the Professional Group on Electron Devices met at Stanford University on January 5. Chairman S. F. Kaisel presided. G. Alpert, Manager of the Physics Department Research Laboratory of Westinghouse Electric Corporation, delivered a paper on "Ultra-High Vacuum Techniques."

ENGINEERING MANAGEMENT

On January 6 the Dayton Chapter of the Professional Group on Engineering Management met at the Engineers Club under the Chairmanship of Elbert W. Piety. Raymond W. Crowley spoke on "The Rights of Employers in the Inventions of Employees." He traced the origin of common law on such invention rights, and outlined the present position on patent ownership, "shop rights," employer-employee contracts, and the establishment, by Executive Order No. 10096, of a uniform patent policy for the government.

ELECTRONIC COMPUTERS

The Philadelphia Chapter of the Professional Group on Electronic Computers met on January 18 at the Benjamin Franklin Center for Physics, Astronomy, and Mathematics of the University of Pennsylvania. T. H. Bonn was the presiding officer. A paper on "Automatic Programming for Digital Computers" was presented by Dr. John W. Mauchly, Eckert-Mauchly Division, Remington-Rand, Incorporated. Dr. Mauchly discussed the probable development of programming, especially automatic coding, and gave the presently known types of automatic coding techniques, i.e., compiler techniques, generator techniques, interpretive techniques, and analytic techniques.

The Dallas-Fort Worth Chapter met on November 30 at Magnolia Petroleum Company, Field Research Laboratories. Louis B. Wadel occupied the Chair. Lynn D. Mullins and W. F. Baldwin, of the Magnolia Petroleum Company, demonstrated an analog computer used for finding the best method of recovering petroleum from an oil field of known or assumed characteristics. From a four-year history of the field, including logging information from field engineers, prediction of future production under different rates and pressures may be made for a two-year period with an accuracy of five to ten per cent.

(Continued on page 107A)

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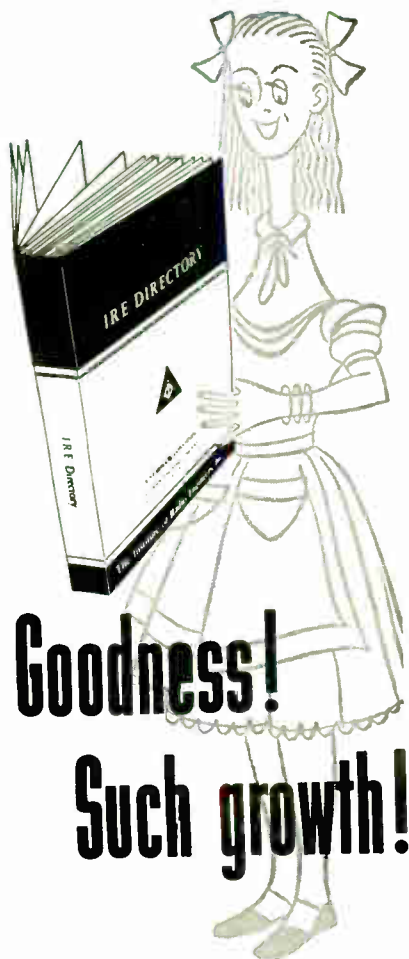
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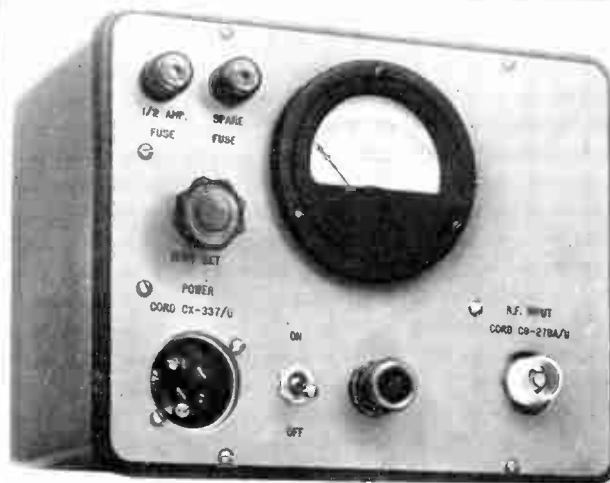
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.2 to 700 MCS
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Type PM-12 shown

These peak reading power meters are designed to accurately measure the peak power of pulsed RF signals in the range of .2 to 700 MC with PRF of 800 to 10,000 pps and pulse duration of .5 microseconds or more, and less than specified maximum average power dissipation.

Model	Power Range	VSWR	Maximum Average Power Dissipation	Connector	Impedance	Supply Voltage	Accuracy	Freq. Response
PM-12	0-50 kw	1.15	60 watts	To Be Specified	51.5	110 volts 60 cps	10%	.2-700MC
PM-18	0-500 kw	1.15	500 watts	To Be Specified	51.5	110 volts 60 cps	10%	.2-700MC

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Professional Group Meetings

(Continued from page 104A)

On January 25, the chapter met again in Fort Worth with D. J. Simmons presiding. On this occasion L. E. Heizer, an aerophysics engineer with Convair, presented a paper on a "Handbook of Analog Computer Circuits and Techniques." Topics discussed in the paper included basic problem set-up procedures, voltage scale factors, time-scale changes, tables of computing circuits, machine applications and limitations, and problem check-out procedures. A sample airplane autopilot problem is set up to illustrate and integrate the various sections of the manual.

At Massa Hall, Hayward, California, the San Francisco Chapter held a meeting on November 16, 1954. Dr. Jerre Noe conducted the meeting. "A Symposium—Computer Maintenance" was presented by Arnold Karpin, Lou Fine, and Preston Hamilton.

The Akron Chapter met on December 20 at Goodyear Hall, with Chairman C. D. Morrill presiding. Professor Robert M. Howe of the University of Michigan presented a paper on "Choosing Computer Components for Large-Scale Simulators." Dr. Howe discussed the type and scope of equations involved in several large-scale simulations, the computer-component requirements for solving these equations, and the relative merits of ac and dc differential analyzers.

INFORMATION THEORY

The Albuquerque-Los Alamos Chapter of the Professional Group on Information Theory met on January 12 at the University of New Mexico. C. H. Bidwell presided. Dr. B. L. Basore spoke to the group on "A Statistical Theory of Target Detection by Pulsed Radar."

With Dr. D. B. Duncan presiding, the Los Angeles Chapter met at the Institute for Numerical Analysis on December 9. There were two speakers. "Passage of Non-Gaussian Noise through Linear Systems" was the name of the paper presented by Dr. Jack Heilfron of Ramo Wooldridge Corporation. Ralph Deutsch, Hughes Aircraft Company, spoke on "A Method of Wiener for Noise through Non-Linear Devices."

MICROWAVE THEORY AND TECHNIQUES

On January 19 the Northern New Jersey Chapter of the Professional Group on Microwave Theory and Techniques held an organizational meeting at which the following officers were elected: Chairman, T. N. Anderson; Vice-Chairman, R. E. White; Secretary, S. Levine; Program Chairman, R. C. McVeety, Jr. A paper, "A Display of X-Band Impedance on an Oscilloscope," was presented by H. L. Bachman of Wheeler Laboratories.

The Buffalo-Niagara Chapter met on October 27 at the University of Buffalo. Dwight Caswell, President of Cascade Research Corporation, spoke to the group on

(Continued on page 108A)



BUT ONLY ONE QUALITY....THE BEST!

AMERICAN BEAUTY makes the finest of Soldering Irons. No second or third grade bears the name which is still the standard of top performance — sixty-one years of it!

DEPENDABLE • DURABLE • EFFICIENT
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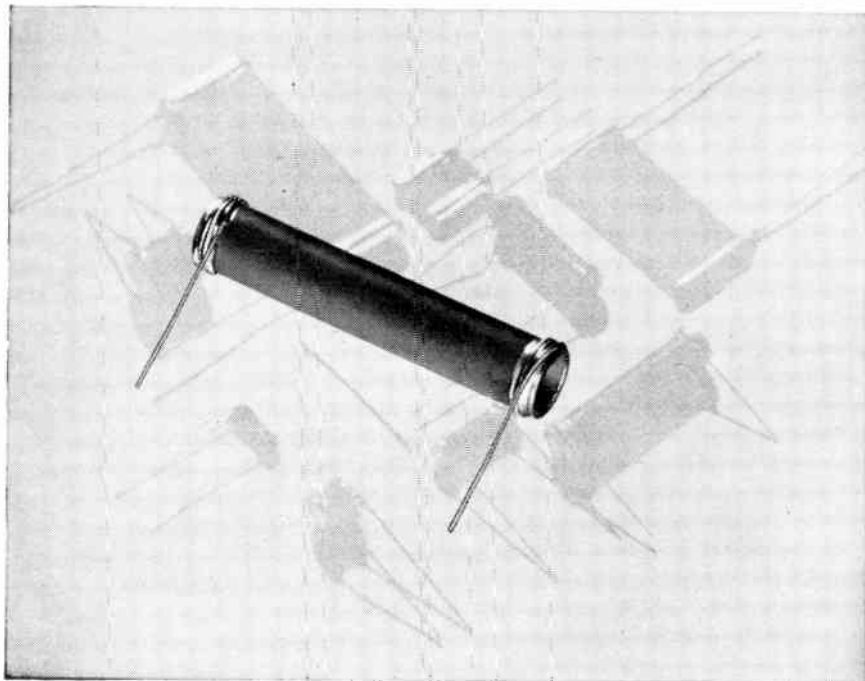
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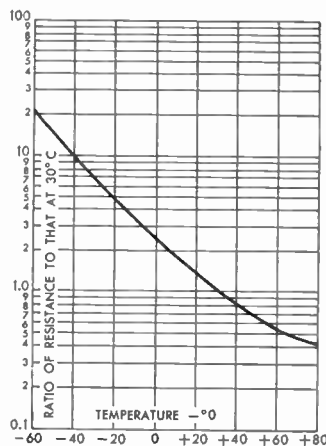
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Resistance: decreases approx. 3% for each degree C temperature rise (see curve)

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Send for Thermistor Inquiry Questionnaire for prompt and accurate estimate.



Above curve shows typical temperature-resistance characteristic of Thermistor. Resistance drops approximately 3% for each degree C temperature rise. As temperature varies up and down, resistance retraces its path precisely, regardless of number of reversals.

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Professional Group Meetings

(Continued from page 107A)

"Ferrites for Microwave Applications." On December 15 the group met jointly with the Buffalo-Niagara Section. "An Airborne Weather Radar for Civil Aircraft," was the subject presented by K. F. Molz of Bendix Radio.

NUCLEAR SCIENCE

The Oak Ridge Chapter of the Professional Group on Nuclear Science met on October 20 with H. E. Walchli presiding. J. E. Broyles, Chief Engineer of Station WTSK, spoke on "Some Operational Aspects of Television Stations." On November 17 the Oak Ridge Chapter met again. W. H. Lee, Chairman of the AIEE Oak Ridge Section, presided. Dr. Robert M. Page, Naval Research Laboratory, presented "Detection of Ice and Hurricanes by Radar."

The Connecticut Valley Chapter met on December 14 at Christopher Columbus Auditorium. Three films were presented to the group: "Operation Greenhouse," "Operation Crossroads," and "A Tale of Two Cities."

TELEMETRY AND REMOTE CONTROL

The Los Angeles Chapter of the Professional Group on Telemetry and Remote Control met on January 18 at the IAS Building. R. E. Rawlins presided and there were two speakers. F. E. Bryan of Douglas Aircraft spoke on "Telemetry of Millivolt Level Signals by PWM-FM," and G. F. Anderson of Radiation, Incorporated, discussed "A High Capacity PTM-AM Telemeter." A summary of telemetering objectives at Convair was presented by E. L. Watkins, Consolidated-Vultee Aircraft Corporation.

VEHICULAR COMMUNICATIONS

The Detroit Chapter met on January 19 at the Engineering Society of Detroit. A. B. Buchanan presided and T. P. Rykala, Michigan Consolidated Gas Company, discussed the company's communication problems as related to its growth.



IRE People

A number of changes in the staff of the Antenna Laboratory at the Ohio State University have been made recently. T. E. Tice (S'46-A'50) has been made Supervisor of the Antenna Laboratory. Dr. Tice received the Ph.D. degree from Ohio State in 1951 and since last spring has been Acting Supervisor of the laboratory; he replaces V. H. Rumsey (SM'50). C. T. Tai (S'44-A'48-SM'51) has joined the Antenna Lab-

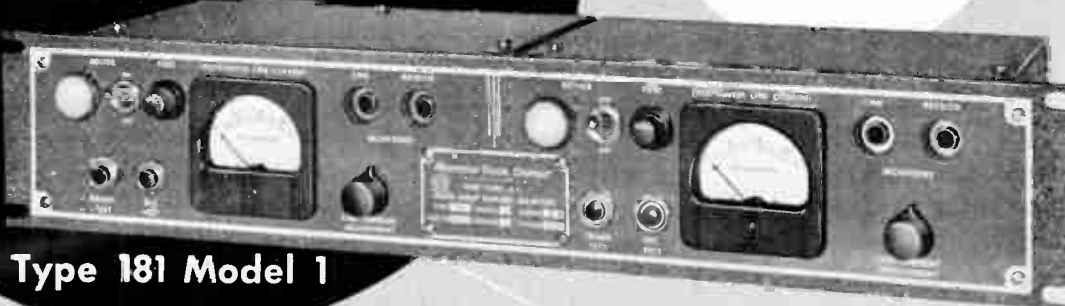
(Continued on page 110A)

NEW!

NORTHERN RADIO DUAL HALF DUPLEX ADAPTER

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**A
Low-Cost
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Communication
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Type 181 Model 1

PURPOSE — The Northern Radio Company Type 181 Dual Half Duplex Adapter couples a 4-wire full duplex tone telegraph system to a half duplex 2-wire D.C. teleprinter loop. This makes possible the half duplex operation of the tone links, and in such a system a teleprinter in any D.C. loop becomes a two-way non-simultaneous system, with any other teleprinter in any other D.C. loop associated with a remote tone station. This provides for an economical two-way communication system between any number of stations which are linked by tone lines or radio links and at each station a maximum of 6 teleprinters which are linked by a standard 2-wire D.C. loop.

EXCLUSIVE BREAK CIRCUIT — In a half duplex circuit, two-way communication cannot be carried on simultaneously and the stations obviously have to take turns in the use of the circuit. Thus in the case of urgent messages, it must be possible for any teleprinter to break into the transmission of any other teleprinter and thereby show its need to take over the circuit. For this reason the Northern Radio Half Duplex Adapter is provided with a Break Circuit which immediately recognizes a break signal and automatically switches the Adapter from its transmit to its receive position. This enhances the half duplex arrangement by permitting a receiving operator to break into the circuit, bringing the system closer to full duplex operation.

EXCLUSIVE MARK RESTORING CIRCUIT — A difficulty generally inherent in Loop operation is that an intentional or accidental space signal may "lock out" the circuit. The Automatic Mark Restoring Circuit built into the Northern Radio Adapter overcomes this trouble by insuring that the tone keyer will be automatically keyed to Mark if Space is sustained for more than 3 seconds.

LONG-TERM UNATTENDED OPERATION — The Adapter is conservatively designed for long-term unattended operation. While both vacuum tubes and sealed relays are used, the relays perform only non-critical switch functions and are not depended upon to repeat telegraph signals. This design takes advantage of the relatively high power efficiency and electrical isolation inherent in a relay without the maintenance requirements usually associated with relay circuits.

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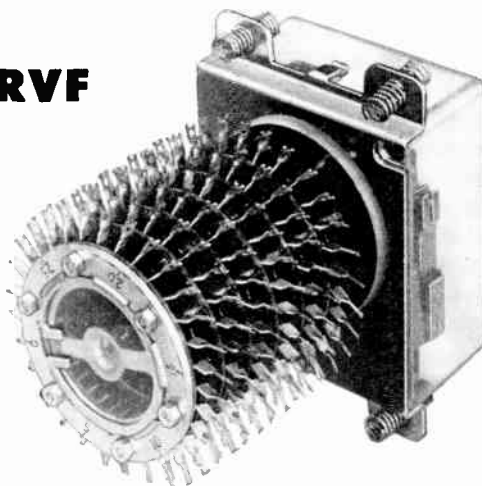
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**30 Points
6 Levels
Single
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or
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12 Levels
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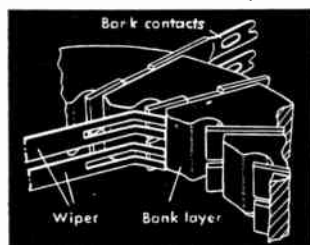
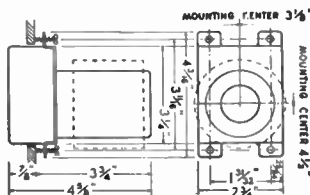
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Shorting type wiper contacts . . . non-bridging . . . connecting two individual adjacent contacts. Interrupting springs of special contact alloy . . . needs no field adjustment. Spring driven switch rotates in one direction . . . eliminates fly-back spring. Switch overtravel is impossible . . . positive stopping at any selected point.

USAGE: Automatic controls . . . Scanning . . . Coding . . . Register Storage . . . Programming . . . Sequence Operation . . . Pulsing . . . Tele-metering . . . Computers.



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(Continued from page 108A)

oratory and been appointed Associate Professor in the Department of Electrical Engineering. Dr. Tai received the Ph.D. degree from Harvard in 1947 and comes to Ohio State from the Stanford Research Institute. Professors **J. D. Kraus** (A'32-M'43-SM'43-F'54), and **G. E. Mueller** (S'39-A'41-SM'46) of the Department of Electrical Engineering are consultants to the Antenna Laboratory. **R. G. Kouyoumjian** (A'53) continues on the Antenna Laboratory staff and also joins the Department of Electrical Engineering as Assistant Professor.



George Haydu, General Manager of the Haydu Brothers Division, Burroughs Corporation, has announced the appointment of **Victor Le Gendre** (M'53) as Chief Engineer of the Plainfield, New Jersey plant.



VICTOR LE GENDRE

Mr. Le Gendre came to Haydu from Chatham Electronics Corporation where he was Design and Development Engineer. In the interim between the war years and his experience with Chatham Electronics Corporation, Mr. Le Gendre was with Tung-Sol Electric, Inc., for five years and National Union Electric for a year and a half. Prior to United States' entry in World War II, Mr. Le Gendre volunteered for the Canadian Army and saw three and a half years of service with the Anti-Aircraft Artillery; he attained the rank of Captain.

Although born in the United States, he was educated in Canada, where he received the B.A. degree from Laval University in Quebec City and the B.S. from University of Ottawa where he taught physics and chemistry for three years. Mr. Le Gendre holds Patent No. 2,654,401 on fine pitch grid winding and has another patent pending on grid winding structures.



The appointment of **G. L. Haller** (A'28-II'36-SM'43-F'50) as Manager of the Laboratories Department of General Electric Company's Electronics Division has been announced. Dr. Haller was Dean of the College of Chemistry and Physics at Pennsylvania State University prior to his appointment and, for the past two years, was also a consultant to the Laboratories Department.

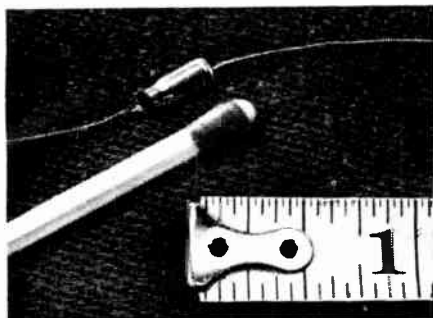


G. L. HALLER

Dr. Haller was born in Pittsburgh, Pa.

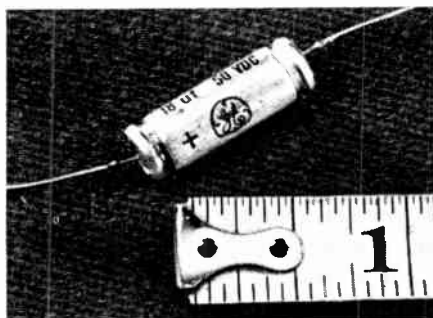
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CAPACITORS by General Electric



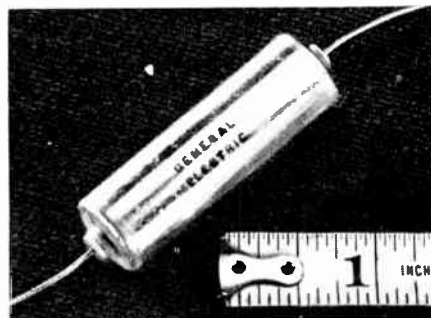
MICRO-MINIATURE

For low voltage d-c miniaturized electronic equipment (hearing aids, walkie-talkies, paging systems). Ideal for transistorized assemblies. **Ratings** 1-8 uf at 4 v. d-c, 1 uf at 8 v. d-c, 0.5 uf at 15 v. d-c. **Tolerance** $\pm 20\%$. **Temp. range** -20 to $+50^{\circ}\text{C}$. BULLETIN GEA-6065.



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For electronic equipment requiring small size, low leakage current, long shelf life, wide temperature range. Plain or etched foil, and polar or non-polar types, suitable for a-c or d-c. **Ratings** 0.25-580 uf, 3.75-150 v. d-c. **Tolerance** $\pm 20\%$ (plain foil), $\pm 15\%$ to $\pm 75\%$ (etched). **Temp. range** -55 to $+85^{\circ}\text{C}$. BULLETIN GEC-808.



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For d-c uses where reliability under severe operating conditions is required (military electronic equipment). **Ratings** 0.001-1 uf at 100, 200, 300, 400 and 600 working v. d-c. (Can be applied to a-c circuits with adequate derating.) **Tolerances** $\pm 5\%$, $\pm 10\%$, or $\pm 20\%$. **Temp. range** -55 to $+125^{\circ}\text{C}$. BULLETIN GEC-987.



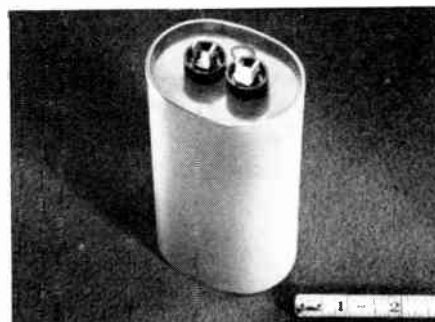
PERMAFIL-IMPREGNATED

Designed to meet requirements of MIL-C-25A, characteristic K specifications, and are suitable for high-temperature operation. **Ratings** 0.05-1 uf at 400 v. d-c. **Tolerance** $\pm 10\%$. **Temp. range** -55 to $+125^{\circ}\text{C}$. BULLETIN GEC-811.



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For motors, filters, communication equipment, luminous-tube transformers, industrial control. **Ratings** dual rated units (a-c or d-c) rated at 0.01-50 uf, at 236-660 v. a-c, 400-1500 v. d-c. Single rated units also available. **Tolerance** $\pm 10\%$. **Temp. range** -55 to $+85^{\circ}\text{C}$. BULLETIN GEC-809.



DRAWN-OVAL

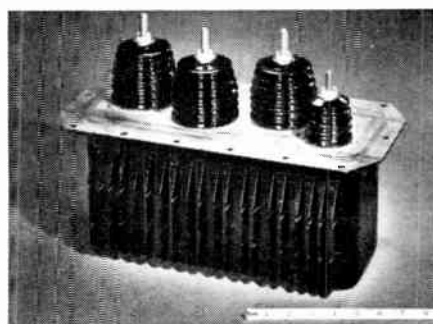
For air conditioning and refrigeration equipment, fluorescent lamp ballasts, business machines, voltage stabilizers. Single, dual or triple-section types. **Ratings** 1-20 uf at 236-660 v. a-c, and 1-15 uf at 600-1500 v. d-c. **Tolerance** $\pm 10\%$. **Temp. range** -30 to $+70^{\circ}\text{C}$. BULLETIN GEA-5777.

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For use in high magnetic fields and high intensity arc discharge. **Ratings**: may be built as high as 2000 joules (watt-seconds). **Tolerance** $\pm 10\%$. BULLETIN GEA-4646.



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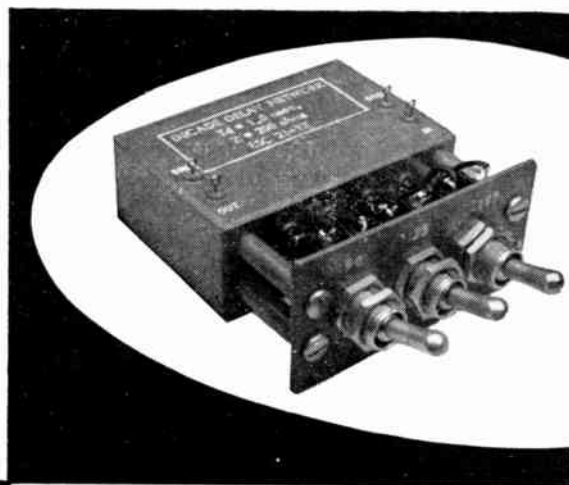
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(Continued from page 110A)

He was graduated from Mercersburg Academy in 1924, and received four degrees from Pennsylvania State University: B.S. in e.e. in 1927, electrical engineer in 1934, M.S. in e.e. in 1935, and Ph.D. in physics in 1942.

He was a radio engineer for Westinghouse Electric and Manufacturing Company, in East Pittsburgh from 1927 to 1929, and audio engineer for E. A. Myers & Sons in Pittsburgh from 1929 to 1933, before returning to Penn State as a graduate assistant. He remained at the university until 1935 when he became a radio engineer at Wright Field for the War Department. From 1942 until 1946, Dr. Haller served in the Signal Corps and later the Air Corps, holding the rank of colonel. For his service, he was awarded the Legion of Merit. He became assistant dean of the College of Chemistry and Physics at Penn State in 1946, and a year later was appointed dean.

Dr. Haller is a member of the Signal Corps Research and Development Advisory Council, the Army Electronic Proving Ground Advisory Council, and the Technical Advisory Panel on Electronics for the Assistant Secretary of Defense. He is a fellow in the American Physical Society, an associate in the Institute of Aeronautical Engineers, and a member of the American Institute of Electrical Engineers, the American Society for Engineering Education, the Franklin Institute, the Newcomen Society of England, Sigma Xi, Tau Beta Pi, Eta Kappa Nu, Sigma Pi Sigma, Phi Lambda Upsilon, Phi Eta Sigma, Pi Mu Epsilon, Alpha Epsilon Delta.



The appointment of R. G. E. Hutter (SM'46-F'54) as Manager of the Physics Laboratory of Sylvania Electric Products Incorporated has been announced. Formerly Manager of the Physical Electronics Branch of the Physics Laboratories, Dr. Hutter has been with the Sylvania Laboratories on Long Island since November, 1944. As a research physicist he has been associated with the field of electron optics, especially the design of cathode ray and traveling wave tubes.



R. G. E. HUTTER

Born in Berlin, Dr. Hutter was a graduate student in physics and mathematics at the University of Berlin from 1930 to 1936. From 1936 to 1938 he was a research physicist in the transmitter laboratories at Telefunken, G.M.B.H. Following that he was Chief Engineer of radio station KZIB, Manila, Philippine Islands. From 1940 to 1941 he was a graduate student in com-

(Continued on page 114A)

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Air-System Sockets

Eimac air-system sockets are custom designed to provide adequate cooling with the most economical blower requirements for several Eimac radial-beam power tetrodes.

4-400A/4000 air-system socket is employed with Eimac tube type 4-400A. Air enters through the bottom of the socket and is guided by a pyrex glass chimney, assuring efficient cooling of the various seals. If desired, this socket may also be used with Eimac 4-125A and 4-250A.

4-1000A/4000 air-system socket is designed for use with Eimac tube type 4-1000A. Air entering the bottom of the socket is guided by a pyrex glass chimney toward the plate seal, assuring correct cooling even during maximum rating operation of the tube.

4X150A/4000 air-system socket provides adequate air cooling and high frequency circuit arrangement for Eimac 4X150A and 4X150D. Air enters the socket through the bottom and is guided by a ceramic chimney.

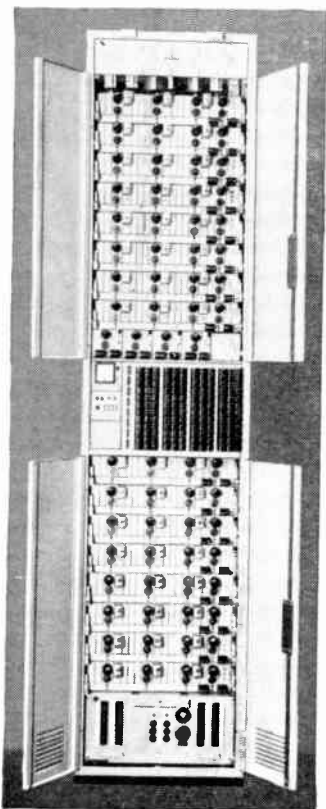
4X150A/4010 socket is identical to the 4X150A/4000 except that this socket is complete with grounded cathode connecting tabs.

Eimac air-system sockets and chimneys are also available as separate units.

**For further information contact our
Technical Services Department.**



EITEL-McCULLOUGH, INC. SAN BRUNO CALIFORNIA
The world's largest manufacturer of transmitting tubes



SIXTY-CHANNEL CARRIER-TELEPHONE SYSTEM OF ADVANCED DESIGN FOR RADIO LINKS

The type F60 carrier-telephone system provides up to 60 channels, in 12-channel groups, on a four-wire basis for transmission over cable pairs or an FM radio system. Transmission is single-sideband suppressed-carrier in the frequency range 12 to 252 kc. Miniaturized plug-in equipment units are used, which also form part of universal carrier-telephone systems of from 3 to 960 channels. Channel band width is 300 to 3400 cycles. Three telephone channels in each group may be replaced by a 10-kc program channel. Built-in ringing and dialing facilities are available. The types FM 60/2000 Radio System, operating in the band 1700 to 2300 mc, FM60/300 Radio System, in the band 235 to 328 mc, and FM24/50 Radio System, in the band 41 to 68 mc, are designed for use with the F60 carrier-telephone system.

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Chicago 30, Ill.



(Continued from page 112A)

munication engineering and physics at Stanford University; in 1941 he became a research associate in the Division of Electron Optics; and in 1944 he received the Ph.D. degree there.

Dr. Hutter is a member of the American Physical Society and of Sigma Xi. He is an Adjunct Professor at the Polytechnic Institute of Brooklyn.



P. C. Sandretto (A'30-M'40-SM'43-F'54), Brigadier General U. S. Air Force Reserve, has been named an Assistant Vice-President of Federal Telecommunication Laboratories, Nutley, New Jersey, a division of International Telephone and Telegraph Corporation. A technical director of FTL, he will act as general coordinator for military research and development projects.



P. C. SANDRETTO

General Sandretto joined the IT&T System in 1946 and held a number of positions in the aeronautical radio research and development activities of the corporation. He was made an Assistant Technical Director of the laboratories in 1948 and was promoted to Technical Director in 1953. Prior to his association with IT&T, he was a member of the technical staff of Bell Telephone Laboratories where he helped design some of the first radio equipment for commercial aircraft in the United States. During World War II, he won recognition for his services with the U. S. Air Force in connection with the planning and establishing of military electronics in the Pacific area.

A writer on aeronautical radio engineering, General Sandretto has served on committees of the Radio Technical Commission for Aeronautics since its inception in 1935. He is a member of the Institute of Navigation, and an associate member of the Institution of Electrical Engineers.



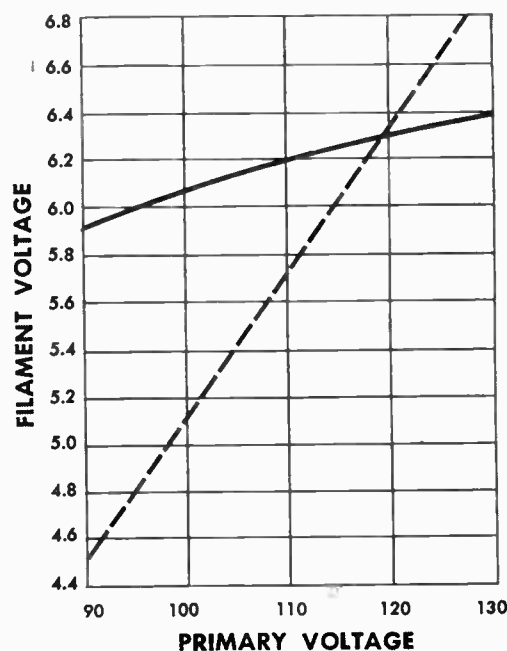
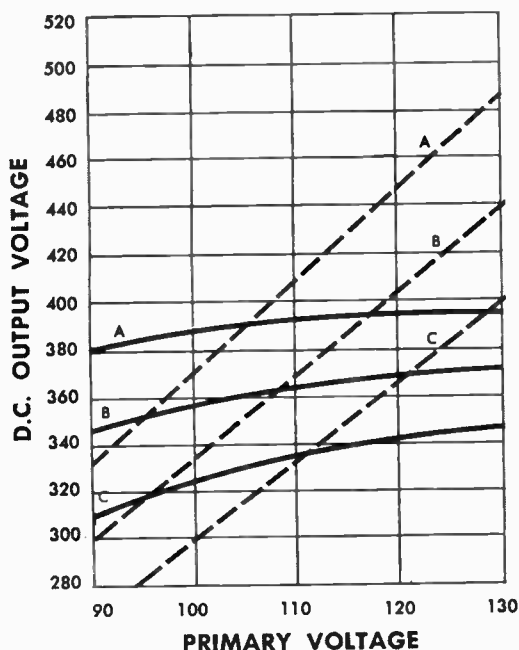
A. E. Abel (A'43-SM'45) has been named Director of Engineering and Research for the Bendix Radio Communications Division of the Bendix Aviation Corporation. In his new post he will direct the activities of the division's 1900 engineering employees engaged in design, test and inspection, research and development, and field engineering work. Mr. Abel formerly served as Assistant Director for more than two years.

A native of Chicago, he was graduated from the Oak Cliff High School in Dallas, Texas, and attended the University of Illinois from 1926-34. He was awarded a Bachelor of Science and a Master of Sci-

(Continued on page 117A)

A: 50% Load
B: 100% Load
C: 150% Load

— SOLA CONSTANT VOLTAGE TRANSFORMER
- - - CONVENTIONAL POWER TRANSFORMER



These curves contrast the plate and filament supply voltages obtained from a Sola and a conventional power transformer when line voltage is varied from 100v to 130v.

Improve Performance of electronic products with built-in regulating power transformer

You can make sure your product will always receive correct plate and filament voltages by building in a Sola Constant Voltage Power Transformer (Type CVE) in place of a conventional, non-regulating power transformer.

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Three stock units (all with high voltage ct, 5.0v and 6.3v regulated windings) are stocked by your electronic distributor. You can order production quantities of special units manufactured to your specification. We invite your inquiry.

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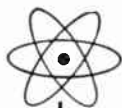
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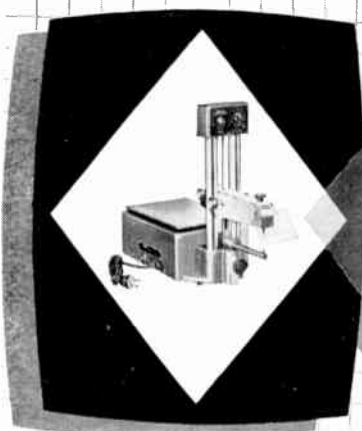
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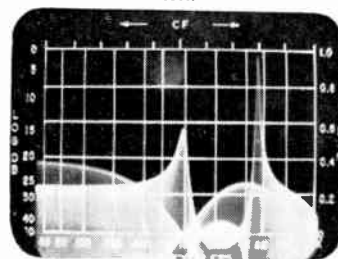
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Assures Convenient
Accurate Analysis of
Frequency Responses
between 40-200,000
cps because of these
Unique Advanced
Engineering Features

DIRECT READING SCREENS: Frequency and amplitude calibrated for slave scopes.
FREQUENCY RANGE: Log 40-20,000 cps or 400-200,000 cps, true decade, selectable. Linear same as above. Two selectable linear sweep-widths calibrated and continuously variable from 20 kc to 100 cps, and 200 kc to 1 kc. Sweepwidth remains constant as calibrated center frequency control is varied anywhere between 0 to 200 kc.
AMPLITUDE SCALES: Linear or 2 decade log.
OUTPUT VOLTAGE: 2 volts into matched 600 ohm load, flat to within 5%. 75 db attenuation in 5 db steps.
RETURN SIGNAL AMPLIFIER AND ATTENUATOR: Covering range of 40 millivolts to 200 volts.
DISTORTION: At least 40 db below maximum output.
SCAN RATES: 1 cps internal or 0.04 to 60 cps with Model TW-1 Triangular Wave Generator.
INTERNAL MARKERS: Passive fundamental frequency null type. Fixed markers at 40, 1000 and 20,000 cps. Variable markers 20 to 200,000 cps in four decade steps.
Separate output for variable marker for use with Signal Alternator Model SW-1 for alternate presentation of marker and response characteristic under test.



Response of an audio band stop filter with null marker showing frequency of one peak.

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makers of
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Analyzer

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Mount Vernon 4-3970

(Continued from page 111.1)

ence degree from that university, both in electrical engineering. During his last two years there he was a special test assistant.

From 1935-1936 Mr. Abel was a Project Engineer for the RCA Manufacturing Company, in Camden, New Jersey, and worked primarily with aircraft transmitters. In 1937 he joined the Bendix Radio Corporation, predecessor of the present division and has been with the organization ever since.

At the conclusion of World War II, Mr. Abel received a Certificate of Commendation from U. S. Navy, Bureau of Ships, and a Certificate of Appreciation from the War Department for outstanding service during the war in radar, communication development, and production.



A. B. Bronwell (A'39-SM'43) has been elected to succeed Alvin E. Cormeny as President of Worcester Polytechnic Institute. On the Northwestern University electrical engineering faculty since 1937, Dr. Bronwell during World War II organized and supervised the Army Signal Corps school in radio and ultra high frequencies, located at Northwestern. Executive secretary of the American Society for Engineering Education for the past seven years, in 1951, at Gen. Matthew B. Ridgway's invitation, he visited Japan as a member of a commission on engineering education.



The appointment of N. H. Mageoch (M'53) to the position of Vice-President for Operations at Daystrom Instrument has been announced. In his new position, he will direct certain phases of product engineering and all activities related to industrial engineering, manufacturing engineering, quality control, production control, assembly, fabrication and installation.

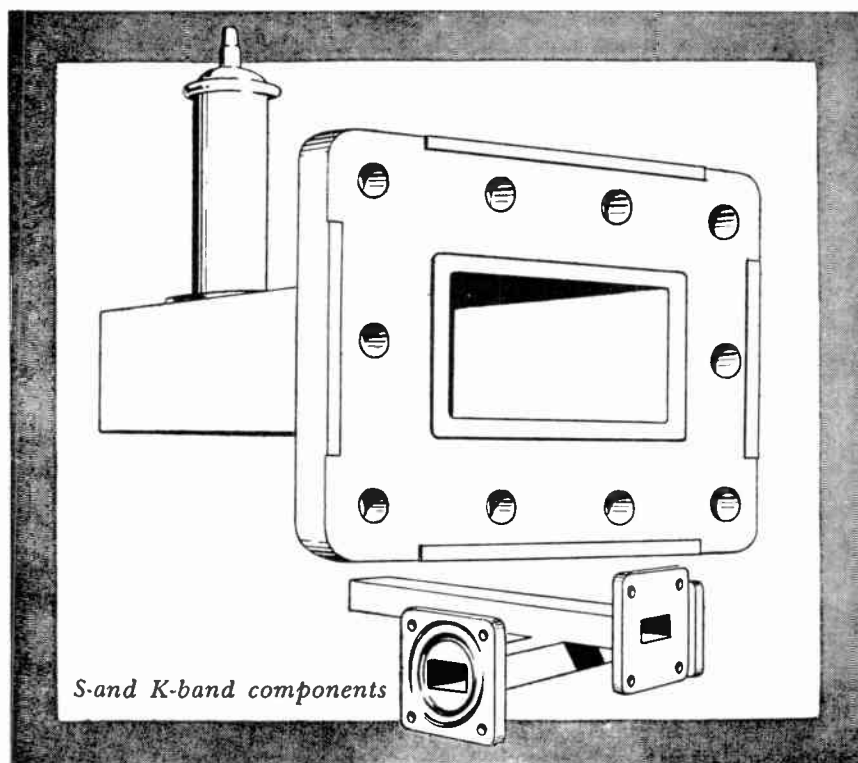


N. H. MAGEOCH

Mr. Mageoch, a graduate of Drexel Institute with post graduate work in applied science at University of Pennsylvania, came to Daystrom Instrument in 1951. He was made Chief Engineer in 1952, Director of Research and Engineering in 1953, and in 1954 Vice-President of Research and Engineering. In this position he directed industrial engineering, equipment installation, inspection and test and the firm's Pilot Plant operation.

Mr. Mageoch is a member of the American Institute of Electrical Engineers, the American Ordnance Association, the Institute of Radio Engineers, and the Association of Computing Machinery.

(Continued on page 118.1)



how small can a wave guide get?

Well, alongside some of the stuff we're working with now, the radar plumbing we used during World War II gets to look like air-conditioning duct. What's more, some of our boys here seem to regard anything below S-band as practically pure D.C. Naturally, we're up to our hips as usual in work on military equipment. However, we do occasionally have some extra creative capacity available, so if you have a problem involving something special in wave guide components (real small ones, too) and like that, maybe we can help. Drop us a line.



L. H. TERPENING COMPANY

DESIGN • RESEARCH • PRODUCTION

Microwave Transmission Lines and Associated Components

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(Continued from page 11A)

A. G. Clavier (M'30-F'39), formerly a Technical Director, has been made an Assistant Vice-President of Federal Telecommunication Laboratories, a division of International Telephone and Telegraph Corporation. Mr. Clavier joined the IT&T System in 1929 as a member of the engineering staff of Laboratoire Central de Telecommunications, an IT&T associate in Paris, and later became assistant director of the company. He was named an Assistant Technical Director of FTL in 1946 and a Technical Director in 1952.



A. G. CLAVIER

The new Assistant Vice-President is recognized in connection with the development of microwave communication. He was associated with the first successful demonstration of microwave transmission across the English Channel in 1931 and directed the project which led to the opening of the world's first microwave radio-telephone link between England and France in 1934. Mr. Clavier has written on high-frequency radio communication and has taught field theory and applications of ultra-high frequencies at the Ecole Supérieure d'Electricité in France.

He was made a Fellow of the American Institute of Electrical Engineers in 1953 for "pioneer work in research, development and engineering in the microwave field." He was chairman of the IRE's Professional Group on Microwave Theory and Techniques in 1953, "Membre Laureat" of the Societe Francaise des Electriciens, and a member of the Institution of Electrical Engineers of Great Britain.

(Continued on page 121A)

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advertising
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June 15, 1955**

From servo-mechanisms to electronic computers, "RADIO" is a way of THINKING!

Far-reaching progress in the radio-electronic field is no "happy accident." Television, electronic computers and the "radiation" power of the atom, which soon will be harnessed to industry were not discovered . . . they were engineered. From "fission" to "computation," these engineering achievements are accomplished through an enormous process of information exchange—the methodical and brilliant teaming together of engineering *thinking* to solve a problem. In radio this work has been done deliberately by a growing engineering society, through its meetings and published proceedings, which unleash the creative minds of men.

In 1954, "Proceedings of the I-R-E" published 1837 text pages, exclusive of product news and departmental features. This is the word-count equivalent of seven 500-page textbooks on radio-electronics for engineers. It exceeds the contents of the next two contemporary publications put together. This "high" in genuine reader service was logically matched by advertising worth over a half-million dollars, by firms investing in the engineers' reading interest and benefiting by it.

simplify custom Installation

The 4200 Sound Effects Filter and 4201 Program Equalizer are now available in component form, as illustrated, for the custom builder.

In addition to the flexibility of installation, all the features and characteristics of the standard models are retained.

The high and low sections of either model may be obtained separately. Complete wiring instructions included.

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Model 4201, Program Equalizer
(Send for Bulletin E)

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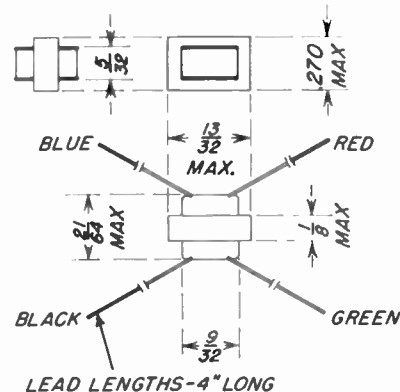
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Field tested—used with transistors by leading manufacturers in large quantities.

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Every
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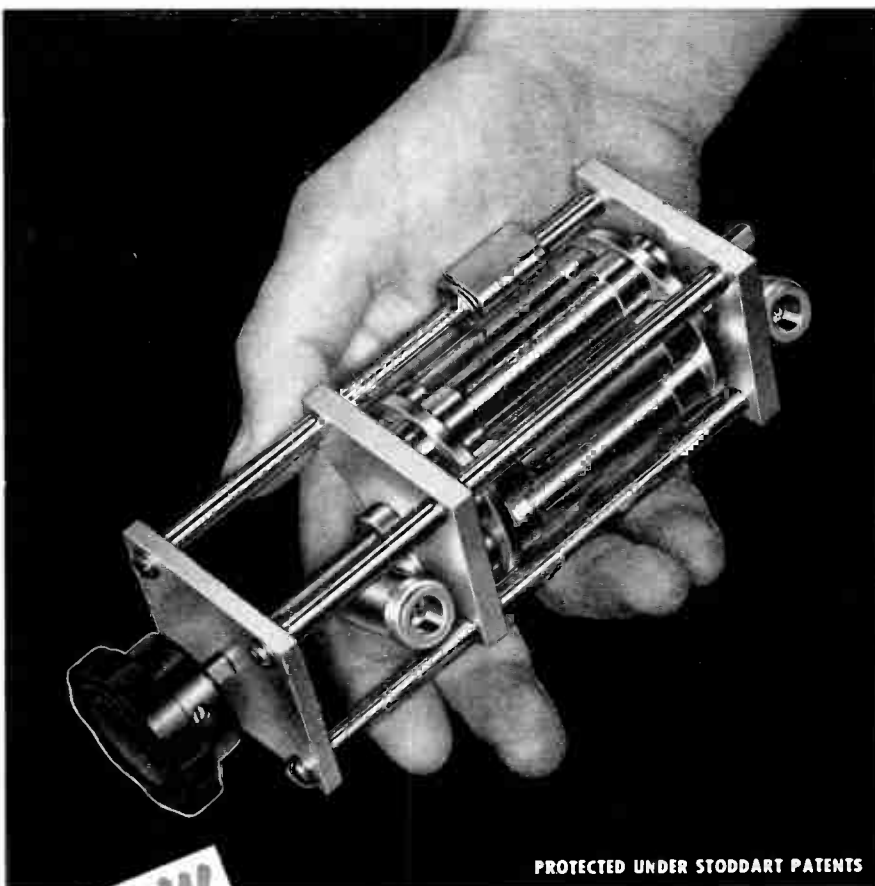
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- Q-Max is widely accepted as the standard for R-F circuit components because it is chemically engineered for this sole purpose.
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- Q-Max is easy to apply, dries quickly and adheres to practically all materials. It is useful over a wide temperature range and serves as a mild flux on tinned surfaces.
- Q-Max is an ideal impregnant for "high" Q coils. Coil "Q" remains nearly constant from wet application to dry finish. In 1, 5 and 55 gallon containers.

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NOW

Precision Attenuation to 3000 mc!

TURRET ATTENUATOR featuring "PULL-TURN-PUSH" action

SINGLE "IN-THE-LINE"
ATTENUATOR PADS
and
50 ohm COAXIAL
TERMINATION



FREQUENCY RANGE:

dc to 3000 mc.

CHARACTERISTIC IMPEDANCE:

50 ohms

CONNECTORS:

Type "N" Coaxial female fittings each end

AVAILABLE ATTENUATION:

Any value from .1 db to 60 db

VSWR:

<1.2, dc to 3000 mc., for all values from 10 to 60 db

<1.5, dc to 3000 mc., for values from .1 to 9 db

ACCURACY:

±0.5 db

POWER RATING:

One watt sine wave power dissipation

Send for free bulletin entitled
"Measurement of RF Attenuation"

Inquiries invited concerning pads or
turrets with different connector styles

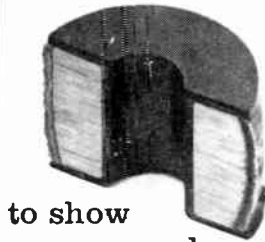
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WE SPLIT RELAY COILS



*Unretouched photographs



to show
you why
P&B RELAYS
are the best

In the coil at left—impregnated by the standard method—the varnish has failed to penetrate beyond the first few strands. The resulting air- and moisture-trapping spaces allowed the strands to pull loose when sawed. *This trapped moisture sets up electrolytic action and causes eventual breakdown.*

Note, though, that the P&B coil at right has no such "empty" spaces. *Centrifugal impregnation—exclusive with P&B in the relay field—solidly imbeds all strands in varnish and protects completely against moisture and electrolysis.*

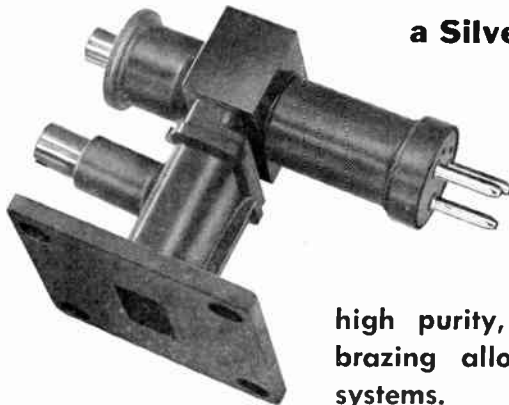
It's just one of many reasons why, when you need a relay . . . of any size, any type, for any application . . . your smartest move is to P&B and Sterling Relays.

Write
Potter & Brumfield Mfg. Co.,
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Another  Product

P&B Relays

no "ghosts" with **WESGO** decarbonized brazing alloy



a Silver-Copper Eutectic
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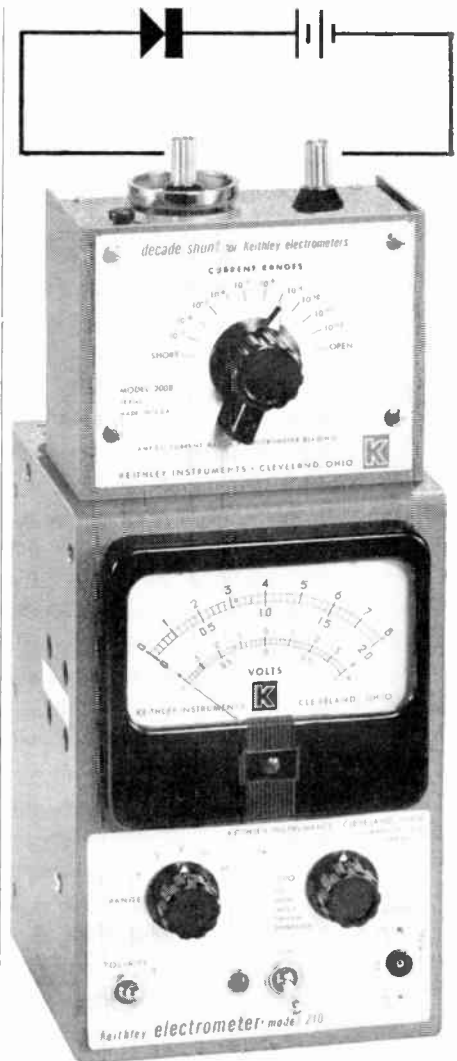
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high purity, low vapor pressure
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systems.

Decarbonization . . . an exclusive process for elimination of carbon and "dirt" ever present in silver brazing alloy wires. No "leakers" due to non-wetting characteristics of carbon. No free carbon particles in electron beam.

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WESTERN GOLD & PLATINUM WORKS
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Quick way to measure low currents

HERE'S a simple way to measure currents down to 10^{-11} ampere. A Keithley Electrometer and Shunt give fast results, accurate with 3%.

Keithley Electrometers are vacuum tube voltmeters with input resistances above 10^9 ohms. They are quickly converted to inexpensive micromicroammeters by clipping an accessory shunt over the input terminals. Current is then read directly by simply combining the current range on the shunt with the voltage reading on the Electrometer.

Features of the equipment include excellent resolution over a range of 10^{-3} to 10^{-11} ampere, polarity sensitivity, and output terminals for recorders. Typical uses include measurement of the inverse current of semiconductors (illustrated), capacitor and insulation leakages, current in photocells, ion chambers, and vacuum tube grids. For a complete catalog, write—

KEITHLEY INSTRUMENTS

3868 Carnegie Avenue
Cleveland 15, Ohio



(Continued from page 118A)

F. J. Gaffney has been appointed recently Vice-President for Engineering of Marion Electrical Instrument Company. In his new position he will direct the development of industrial and aircraft instrumentation.



F. J. GAFFNEY

Mr. Gaffney, well known for his work in electrical measurements, was formerly Director of Engineering for the Guided Missiles Division of Fairchild Engine and Airplane Company. During World War Two he served as head of the Test and Measurements group of the M.I.T. Radiation Laboratory, and, from 1945 until 1953, he was General Manager of the Polytechnic Research and Development Company.

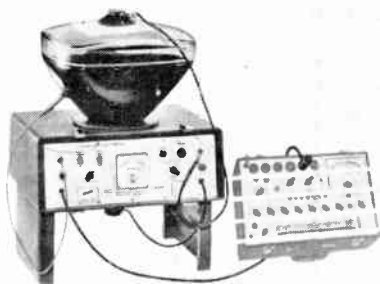
For a number of years he has served as a consultant to the Department of Defense and currently serves as a member of the Steering Committee of the Panel on Electronics in the Office of the Assistant Secretary of Defense for Research and Development. Mr. Gaffney is a member of the AIEE, American Physical Society, American Association for the Advancement of Science, U. S. Committee of the International Scientific Radio Union, Tau Beta Pi, and Sigma Xi.



These manufacturers have invited PROCEEDINGS readers to write for literature and further technical information. Please mention your I.R.E. affiliation.

(Continued from page 36A)

Portable Picture Tube Tester



In order to provide a complete, accurate test of the quality of a TV picture tube at a reasonable price, the Model 590 Picture Tube Tester, has been introduced by Hickok Electrical Instrument Co., 10551 Dupont Ave., Cleveland 8, Ohio. The 590 permits an accurate check of the overall "light" efficiency of a TV picture tube, including brilliance, condition of phosphor, possible ion burns and probable life, and also permits an accurate check for emission, shorts, gas content, leakage and grid control.

(Continued on page 116A)

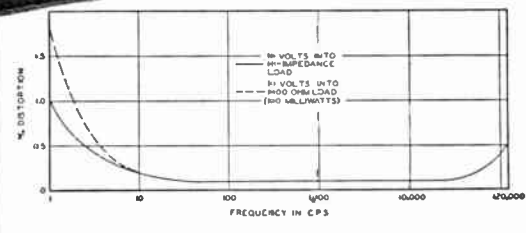
SIE

MODEL M-2
Low Distortion

R-C OSCILLATOR

Unexcelled for:

- ★ Amplifier & filter design
- ★ Galvanometer manufacture
- ★ Vibration Analysis
- ★ Variable frequency standard
- ★ Transformer & servo design



Guaranteed SPECIFICATIONS:

Wide frequency range
1 to 120,000 cycles per second

Accurate dial calibration
within $1\frac{1}{2}\% \pm .1$ cps

Fully regulated power supply
ripple less than .01% of output

Low frequency drift
less than .1% over long term

Excellent amplitude stability
within $\frac{1}{2}$ db. throughout range

High output: **425⁰⁰**
20 volts into 1000 ohms or more
12 volts into 600 ohms
1 volt at 300 ohms constant impedance

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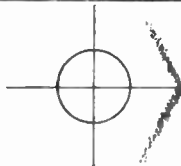
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join the company that
creates history in electronics!



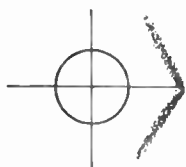
now... you can CREATE with the
WESTINGHOUSE electronics division

It's a long way from KDKA... the world's first radio broadcasting station... to radar, the backbone of America's air defense. It's a long way from the vacuum tube to the transistor.

To the men who have played vital roles in these developments... electronics has been a challenge. To you... it is an opportunity. It is the opportunity to create history in electronics, just as other engineers have done at Westinghouse.

Top-level design and development positions are now open. If you can fill one of these exciting openings... you will find unlimited creative opportunity. You will be well-compensated for your efforts... both in income and benefits, and in the satisfaction of contributing to America's superiority in ground and shipborne radar; fire-control, communications and missile guidance systems.

openings exist for:



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Positions Wanted

By Armed Forces Veterans

In order to give a reasonably equal opportunity to all applicants and to avoid overcrowding of the corresponding column, the following rules have been adopted:

The Institute publishes free of charge notices of positions wanted by I.R.E. members who are now in the Service or have received an honorable discharge. Such notices should not have more than five lines. They may be inserted only after a lapse of one month or more following a previous insertion and the maximum number of insertions is three per year. The Institute necessarily reserves the right to decline any announcement without assignment of reason.

ELECTRONIC—PRODUCT ENGINEER

BEE 1951. Age 33. 13 years in electronics. Design, development, production on fire and missile control radar systems. Video, servo, CRT, data transmission, system integration, human engineering. Liaison between engineering and production. Supervisory and some administrative experience. Self-starter and neck sticker-outer when expediency requires. Seeks position slightly over his head. Will relocate. Present income \$9,000. Box 802 W.

(Continued on page 126A)

ENGINEERS

MICROWAVE

Development of microwave instruments & test equipment

ELECTRONIC

Development of electronic instruments

Precision instrument manufacturer requires men with good academic & practical background. At least 2-3 years design & development experience required. Should exhibit qualities of leadership, with the ability to meet & deal with people. Men who fill the bill will be substantially compensated.

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MISSILE SYSTEMS

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Physicists and engineers at Lockheed Missile Systems Division are engaged in a group effort covering virtually every field of science.



Missile Systems Division scientists and engineers discuss a new missile systems concept in light of tactical requirements. Left to right: Dr. H. H. Hall, nuclear physicist; I. H. Culver, systems development division engineer; Dr. R. J. Havens, research scientist; W. M. Hawkins, chief engineer; Dr. Ernst H. Krause, nuclear physicist and director of research laboratories; S. W. Burriss, experimental operations division engineer; Ralph H. Miner, staff engineering division engineer; and Dr. Eric Durand, nuclear physicist.

Continuing developments are creating new positions for those capable of significant contributions to the technology of guided missiles.

Lockheed

MISSILE SYSTEMS DIVISION
research and engineering staff

LOCKHEED AIRCRAFT CORPORATION • VAN NUYS, CALIFORNIA

ENGINEERS AND

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COMPUTERS
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RADIO SYSTEMS

A whole new program of expansion at RCA—in Research, Systems, Design, Development and Manufacturing—opens a broad variety of permanent positions with all the features that appeal to the alert, creative engineer. These are opportunities with a future...available *today* for the man who wants to move ahead professionally with the world leader in electronics. They include work in fields of phenomenal growth. At the RCA engineering laboratories *listed in the chart on the right*, you'll find the kind of living and working conditions attractive to the professional man and his family.

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Your individual accomplishments and progress are recognized and rewarded through carefully planned advancement programs. Financially as well as professionally, you move ahead at RCA!

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Check the chart below for openings in your field.

FIELDS OF ENGINEERING ACTIVITY	TYPE OF DEGREE AND YEARS OF EXPERIENCE PREFERRED											
	Electrical Engineers			Mechanical Engineers			Physical Science			Chemistry Ceramics Glass Technology Metallurgy		
	1-2	2-3	4+	1-2	2-3	4+	1-2	2-3	4+	1-2	2-3	4+
SYSTEMS (Integration of theory, equipments, and environment to create and optimize major electronic concepts.)												
AIRBORNE FIRE CONTROL			W						W			
DIGITAL DATA HANDLING DEVICES			C			C			C			
MISSILE GUIDANCE			M			M			M			
INERTIAL NAVIGATION			M			M			M			
COMMUNICATIONS			C O F						C O F			
	F							F				
DESIGN • DEVELOPMENT												
COLOR TV TUBES —Electron Optics—Instrumental Analysis—Solid States (Phosphors, High Temperature Phenomena, Photo Sensitive Materials and Glass to Metal Sealing)	L	L	L	L	L	L	L	L		L	L	L
RECEIVING TUBES —Circuitry—Life Test and Rating—Tube Testing—Thermionic Emission	H	H	H		H	H		H			H	H
MICROWAVE TUBES —Tube Development and Manufacture (Traveling Wave—Backward Wave)		H	H	H			H	H			H	H
GAS, POWER AND PHOTO TUBES —Photo Sensitive Devices—Glass to Metal Sealing	L	L	L	L	L		L	L		L	L	
AVIATION ELECTRONICS —Radar—Computers—Servo Mechanisms—Shock and Vibration—Circuitry—Remote Control—Heat Transfer—Sub-Miniaturization—Automatic Flight—Design for Automation—Transistorization		F	M C F			M C F			F C F			
RADAR —Circuitry—Antenna Design—Servo Systems—Gear Trains—Intricate Mechanisms—Fire Control		F	M C F			M C F			F C F			
COMPUTERS —Systems—Advanced Development—Circuitry—Assembly Design—Mechanisms—Programming	C	C F	M C F	C	C F	M C F	C	C F	M C F			
COMMUNICATIONS —Microwave—Aviation—Specialized Military Systems		F	M C F			M C F			F C F			
RADIO SYSTEMS —HF-VHF—Microwave—Propagation Analysis—Telephone, Telegraph Terminal Equipment		O	O F			O F			O F			
MISSILE GUIDANCE —Systems Planning and Design—Radar—Fire Control—Shock Problems—Servo Mechanisms		F	M F			M F			F F			
COMPONENTS —Transformers—Coils—TV Deflection Yokes (Color or Monochrome)—Resistors		C	C			C	C		C	C		
MACHINE DESIGN Mech. and Elec.—Automatic or Semi-Automatic Machines		H	H			H	H		H	H		

Location
Code

C—Camden, N. J.—in Greater Philadelphia near many suburban communities.
F—Florida—on east central coast.
H—Harrison, N. J.—just 18 minutes from downtown New York.

L—Lancaster, Pa.—about an hour's drive west of Philadelphia.
M—Moorestown, N. J.—quiet, attractive community close to Phila.
O—Overseas—domestic and overseas locations.
W—Waltham, Mass.—near the cultural center of Boston.

Please send resume of education and experience,
with location preferred, to:

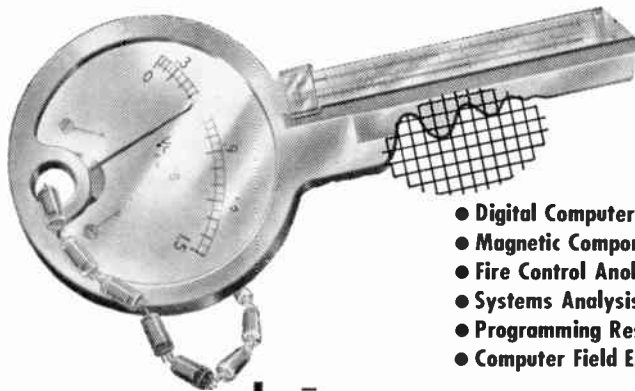
Mr. John R. Weld, Employment Manager
Dept. C-1D, Radio Corporation of America
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New York 20, N.Y.



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Positions Wanted

By Armed Forces Veterans

(Continued from page 122A)

ELECTRONIC ENGINEER

BEE, MS. 5 years electronic engineering experience. Pulse circuits, microwave, radar development. 1 year technical writing. Desires position in electronic research and development. New York City area preferred. Box 803 W.

SALES ENGINEER

BEE, age 27, married. 4 years experience in military electronics, including 2 years application engineer in aviation electronics. Desires a challenging opportunity in a sales capacity. Metropolitan New York area. Box 804 W.

ENGINEER

Twenty-seven years a technical writer and editor on a New York newspaper. Wants public relations, publicity or technical writing-editing job in electronics. Preferably in New York area. Box 805 W.

ADVERTISING & PUBLIC RELATIONS MANAGER

BS Engr./Bus. Admn., MBA Marketing. 10 years progressive experience all phases of industrial marketing. Program planning, budgeting and administration, market survey, agency liaison, media evaluation, copywriting and production. Pamphlets & brochures, catalogues & direct mail, trade shows & technical publicity. Media, industry, community & Government relations. Licensed radio operator with background in radio & electronic equipment promotion. Age 35. Desires career position offering greater responsibilities & advancement. Box 811 W.

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BEE communications option, some graduate courses. Age 31, married, 2 children. 5 years experience in the field of radio-frequency coaxial, and multi-contact audio, power and control connectors and fittings. Desires development and/or production engineering position in this or related fields. Box 812 W.

PHYSICIST

MS Physics, 1952. Age 30, married. Research and development experience in electronics, ion devices, vacuum systems and instrumentation. Prefer location in southwest or Florida. Box 813 W.

SYSTEMS ENGINEER

BSTE January 1949. Age 30, married, 1 child. 6 years diversified experience: radar, automatic data reduction systems, digital computers, telemetering and instrumentation systems. Desires similar project or systems engineering position. Box 814 W.

ELECTRONIC ENGINEER

BEE 1950, MEE expected 1955 from New York University. Age 30, married. 5 years experience as electronic circuit and development engineer on automation. Considerable production and mechanical experience. Desires position in New York City area until June 1955 and then will relocate. Box 817 W.

ELECTRONIC ENGINEER

BEE 1955 (Jan.). Age 34, married. 3 years radio coils, 2 years radar, 2 years VHF communications, 2 years radio and TV service. 2 years dial switching ckts., 3 years TV broadcast lab. (color), 2 years sales. Desires responsible position where broad experience can be utilized. Box 818 W.

(Continued on page 134A)



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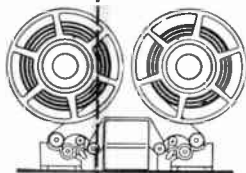
says Wolloce D. Bolton,
Development Engineer at
the Endicott Laboratories

"The way IBM is growing certainly offers a young engineer the opportunity to move ahead—and in work that's interesting," says Wally. "Since I joined IBM in July of '50, right after getting my BS/EE from the University of Pennsylvania, I've been closely associated with a new development in the field of high-speed printing. Now, I'm in charge of the research phase of this program. And in just about every other area around me, I've seen opportunities opening up all the time for other young engineers."



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***Required**—a degree in E.E., M.E., or Physics, or equivalent experience.

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- Microwave Filters
- Flight Simulators
- Servomechanisms
- Electro-Mechanical Design

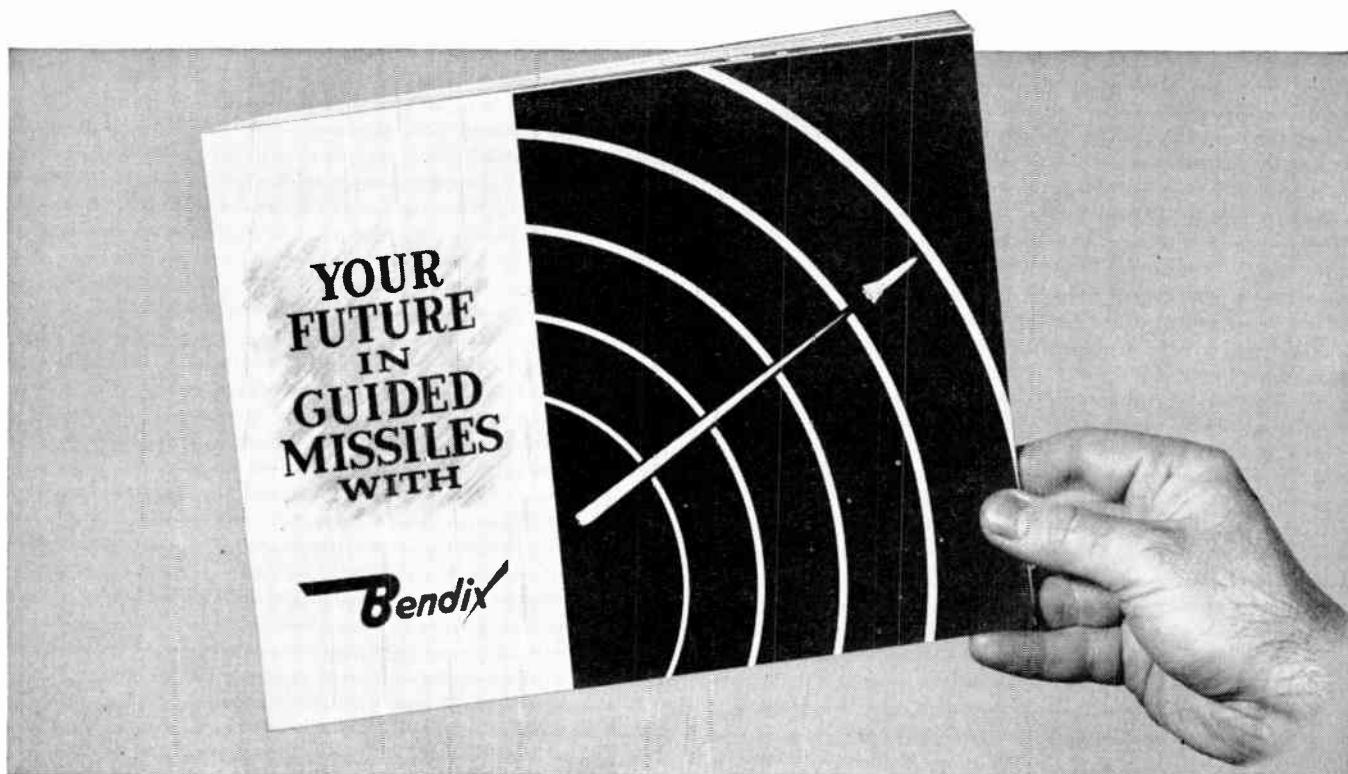
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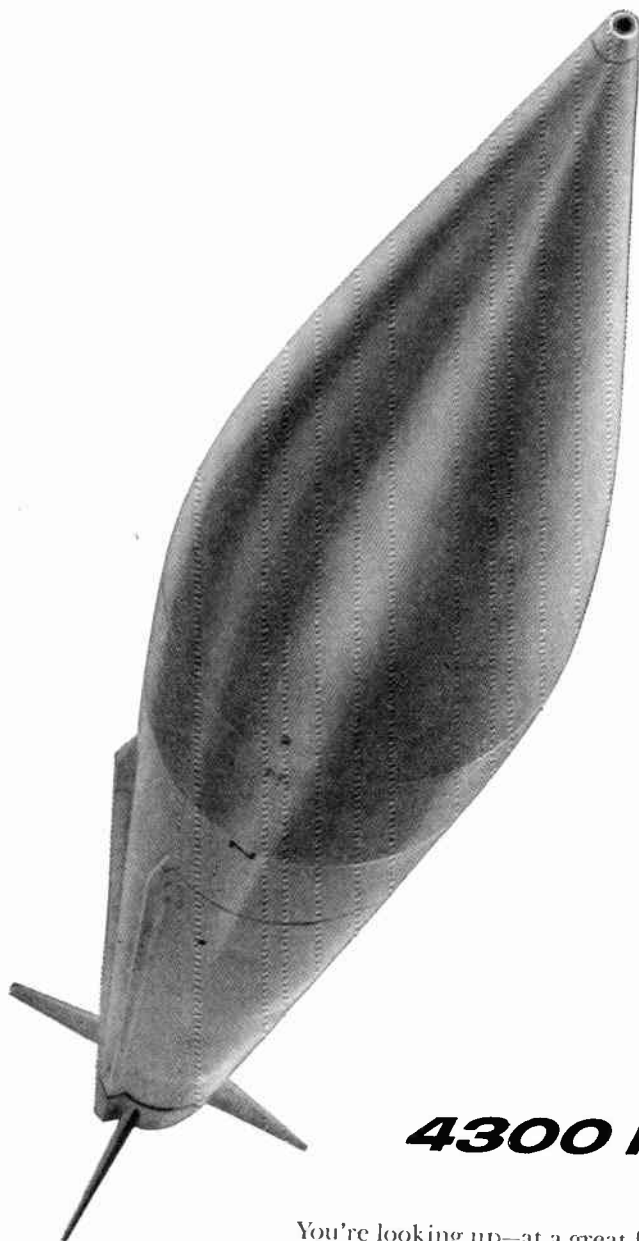
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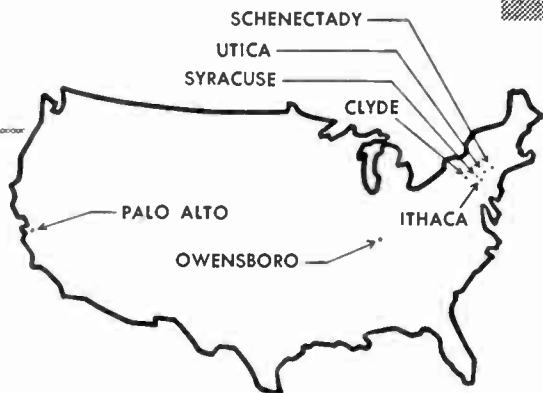
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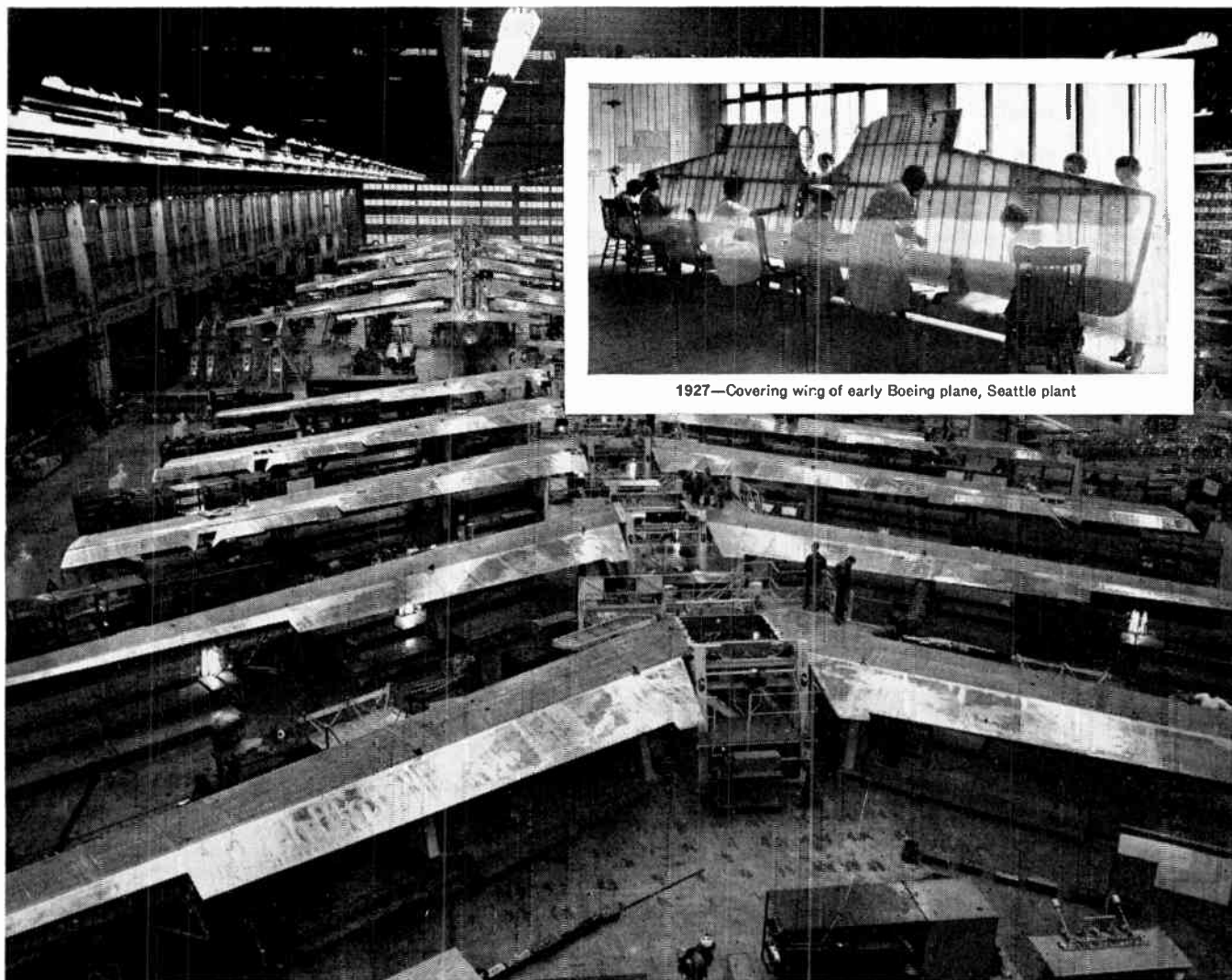
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1927—Covering wing of early Boeing plane, Seattle plant

1955—B-47 Stratojet assembly, Boeing Wichita Division

Boeing offers engineers long-range careers

Back in 1927 engineers designed airplane wings in simple terms of wood and cloth. An airplane wing of today is a complex aerodynamic structure housing a myriad of electrical, mechanical and hydraulic systems.

Yet many Boeing engineers of 1927 are still with the company. They have grown with the science of aviation—capitalized on its potentials, and contributed to its progress.

What engineer in 1927 could foresee the stature of the aviation industry today. They saw only a challenge—an

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Boeing is seeking more engineers of ability—in Research, Design and Production. Today, one out of each seven Boeing employees is an engineer! You'll work on such diverse programs as: The B-52 and B-47 multi-jet bombers. The "707," America's first jet transport. Research in nuclear-powered and supersonic flight. One of the nation's major guided missile

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Boeing has openings for virtually all types of engineers—electrical, civil, mechanical, aeronautical and related fields, as well as for applied physicists and mathematicians with advanced degrees.

For full information on your opportunities at Boeing, send résumé of your education and experience background to:

JOHN C. SANDERS, Staff Engineer—Personnel
Boeing Airplane Company, Dept. G-10, Seattle 14, Wash.

BOEING
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SEMI-CONDUCTOR
APPLICATIONS
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LOGICAL TECHNIQUES
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PRODUCT DEVELOPMENT
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Personnel with ability in the above or allied fields are invited to submit a resume or write for an application. Recent and prospective college graduates with high scholastic standing are also invited to apply. Training will be given at full pay. Liberal company benefits and advancement on merit are offered to those who would like to work and play in sunny Southern California.

National

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ELECTRONICS DIVISION

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Positions Wanted

By Armed Forces Veterans

(Continued from page 126A)

BROADCAST ENGINEER

RCA graduate. 1st Class ticket. Ambitious beginner. Desires position in radio or TV station anywhere, to start at bottom. Anxious to learn all aspects involved in broadcasting. Salary secondary. Box 828 W.

TRANSISTOR ENGINEER

AB., BSEE., MEE., Tau Beta Pi, Eta Kappa Nu, Sigma Xi, pre-doctoral student and EE instructor. 2 years experience design and teaching audio and pulse transistor circuits. Also several years vacuum tube circuit experience. Desires summer job in New York City. Box 829 W.

RADIO ENGINEER

BSEE 1950, Oregon State College, age 28, married. 4½ years experience in airborne electronics and radio aids to navigation, theoretical and practical. Desires position with future with a progressive company in the west, preferably the northwest. Box 830 W.

OPERATIONS RESEARCH

Master's degrees in physics and in electronics. 14 years experience in research engineering and management. analytical mind with keen mathematical ability. Desires position in operations research. Box 831 W.

ENGINEER (no license)

Age 27, single. 3 years Army. 6 years civilian experience in electronics. Speak read and write Spanish fluently. Desires position in Latin America. Box 832 W.

ELECTRONIC ENGINEER

Five years of missile, radar and fire control system study work, and four years of radar and missile component development prior to that. Desires position in southern South America. Box 833 W.

RADIO-TV TECHNICAL DIRECTOR

Six years experience in program production and direction. Education: B.A. in programming and production. Technical background includes control room operations, equipment design, construction, maintenance. 1st phone license. Age 26, married. Completing Army duty as microwave instructor at the Signal School in June 1955. Prefer Chicago or vicinity. Box 834 W.

ELECTRONICS RESEARCH

BEE 1946, MEE. 1950 electronics. Age 29, married, 1 child. 4 years experience hyperbolic radio navigation systems research, creative design, construction, analysis, laboratory and field evaluation. Lieut. Naval Reserve. Member I.R.E., P.G.A.N.E. and I.O.N. Presently Unit head. Desires similar position with advancement opportunity industry or university research program. Box 835 W.

ENGINEER

Instructor of technical electricity and electronics with extensive field and teaching experience. Desires (within 25 miles of Poughkeepsie, N.Y.) a teaching position with some H.S. or college or some phase of technician-engineering work with a private firm. Box 836 W.

(Continued on page 137A)

ELECTRONIC ENGINEERS

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WITH A LEADER IN
WESTERN ELECTRONICS

An expanding program of:

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Engineering Employment Manager

Hoffman

LABORATORIES, INC.

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Professorial appointment at Brown University open for Electrical Engineer with established ability in teaching and advanced research. Will have responsibility for organizing and directing theoretical and experimental research program in his fields of interest.

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D. C. Drucker, Chairman
Division of Engineering
BROWN UNIVERSITY
PROVIDENCE 12, R.I.

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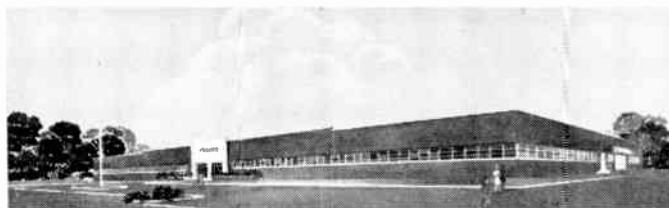
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CONVAIR offers you an imaginative, explorative, energetic engineering department... truly the "engineer's" engineering department to challenge your mind, your skills, your abilities in solving the complex problems of vital, new, long-range programs. You will find salaries, facilities, engineering policies, educational opportunities and personal advantages excellent.



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DEVELOPMENT

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ELECTRONIC EQUIPMENT PACKAGING

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MISSILE SYSTEMS DEVELOPMENT

FLIGHT TESTING

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Professional Staff Appointments

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(Electrical and Electronic)

to design electrical and electronic installations in aircraft and to supervise a small section of electrical and mechanical draftsmen. Applicants should have a degree in Electrical Engineering or Mechanical Engineering with electrical subjects as a strong secondary and preferably 5 years experience in design of electrical and electronic installations in aircraft. Experience in supervision of draftsmen desirable. Salary commensurate with experience and qualifications. Reply stating age, education, experience, personal particulars and salary expected to Box No. 807, Institute of Radio Engineers, 1 East 79th Street, New York 21, N.Y.



Positions Wanted

By Armed Forces Veterans

(Continued from page 134A)

ENGINEER

BSEE 1952 Communications option. 1½ year graduate work in nuclear engineering and EE. 2 years as electronics technician, Navy. 2 years industrial experience in electronic design. Desires job placement in electronics work, possibly numerical controlled machine tools. Box 837 W.



Positions Open

The following positions of interest to I.R.E. members have been reported as open. Apply in writing, addressing reply to company mentioned or to Box No.

The Institute reserves the right to refuse any announcement without giving a reason for the refusal.

PROCEEDINGS of the I.R.E.

1 East 79th St., New York 21, N.Y.

ELECTRONIC ENGINEER

Laboratories located in a small midwestern town have an opening for a man experienced in the field of magnetic recording. Must be capable of building and operating a testing laboratory for the development and quality control of magnetic tape. Box 798.

ELECTRONIC ENGINEER

Edgerton, Germeshausen & Grier, Inc., 160 Brookline Ave., Boston, Mass. has a position open for an electronic engineer experienced in the development of instrumentation including oscillographic pulse techniques and allied circuitry. Send resume or call Personnel Dept. at COpley 7-3520.

ELECTRONICS ENGINEER

Electronics manufacturer on San Francisco peninsula has openings for engineers qualified to handle design and development of specialized electronic circuits involving amplifiers, telemetering and related systems. Must have BSEE and at least 2 years experience in related fields. Salary commensurate with experience. Please send resume to Dalmo Victor Co., 1414 El Camino Real, San Carlos, Calif.

ELECTRONICS ENGINEER

The Civil Aeronautics Administration urgently needs electronics engineers who have had or who wish experience and training in electronics research and development. Salaries range from \$3410 to \$5940 per annum. Write or send application or Standard Form 57, which can be obtained at Post Office, to Personnel Officer, Civil Aeronautics Admn., Development of Evaluation Center, P.O. Box 5767, Indianapolis, Ind.

ENGINEERS IN ELECTRO-ACOUSTICS

ELECTRO-VOICE, INC. has positions open for engineers with degrees in electrical engineering or physics. Positions are open for experienced men in speaker, microphone or phonograph pickup design and development for the recent graduates. Excellent future for the exceptional man. Write to Vice-President, Engineering, Electro-Voice, Inc., Buchanan, Michigan.

(Continued on page 138A)

CAREER OPPORTUNITIES FOR . . .

- mechanical engineers
- electronics engineers
- electrical engineers
- physicists
- mathematicians
- aerodynamicists
- systems engineers

SANDIA CORPORATION, a subsidiary of the Western Electric Company, operates Sandia Laboratory under contract with the Atomic Energy Commission. Sandia engineers and scientists work at the challenging task of designing and developing atomic weapons. Graduate engineers and scientists, with or without applicable experience, will find excellent opportunities in the fields of component development, systems engineering, applied research, testing, and production.

COMPENSATION is competitive with that offered in other industry. Ingenuity and initiative are valued highly, and opportunities for professional advancement are outstanding. Working conditions are excellent, and employee benefits are most liberal.

SANDIA LABORATORY is located in Albuquerque — a modern, cosmopolitan city of 150,000, rich in cultural and recreational attractions and famous for its excellent year-around climate. Adequate housing is easily obtained. For descriptive literature giving more detailed information on Sandia Laboratory and its activities — or to make application for employment — please write:

professional employment
division 1A

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Corporation

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RADIATION Inc. Melbourne, Fla.
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Electronics • Avionics • Instrumentation

SUPERVISOR **Guidance Design**

Attractive supervisory assignment now available for an experienced electronics engineer to direct the analysis, design and flight test of systems for the guidance of pilotless aircraft. Applicants must possess supervisory experience and specialized knowledge of U.H.F. and microwave frequencies.

Requirements include Bachelor degree with advanced study preferable. Minimum of eight years of experience necessary.

Generous travel allowance; insurance, hospitalization and retirement programs.

For detailed information and to arrange for a personal interview, submit resume of education and experience with salary desired to

G. H. Orgelman, Supervisor — Engineering Personnel

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Home Office: Chicago
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with experience in the design, development and production of delay lines, pulse transformers and allied electronic components.

Salaries commensurate with ability and experience.



(Continued from page 137A)

ENGINEER, CHEMIST OR PHYSICIST

Opening for research and product development in the field of semi-conductors and semi-conductor devices. Prefer direct experience with growth of silicon crystals and use for transistors. Also opening for research and product development of thermistors. Direct experience required. With a midwest manufacturer of electrical components. Capable of taking complete responsibility. Excellent working conditions in new laboratory. Reply in detail giving experience, training and salary desired. Box 801.

RADIO ENGINEERS

Consulting engineering firm requires senior and junior engineers for permanent staff positions. Experience in design and development of military radio and radar systems required. Work is of unusual interest and offers sound opportunities for advancement. Location midwest. Salary open. Send summary of qualifications and background to Box 802.

RADIO ENGINEER

As technical consultant to large national trade association in two-way mobile and microwave radio fields. Also act as contact man with FCC and other associations on technical matters and frequency allocations. Location: Washington, D.C. Send complete details in first application. Salary \$600 per month. Reply Box 803.

(Continued on page 141A)

Electrical Engineers and Physicists

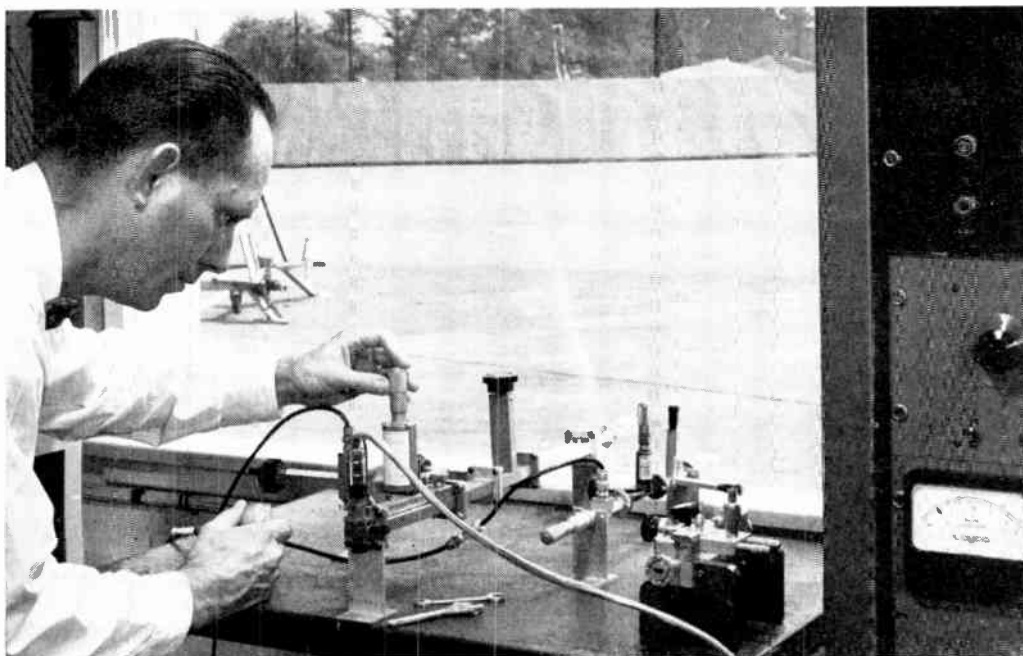
- Radar Simulation
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Industrial Research Laboratories
Division of Aeronca Manufacturing Corp.
Dept. B-4, Hilltop & Frederick Rds.
Baltimore 28, Maryland



Research Specialist Edward Lovick measures reflection coefficient of dielectric materials in the K-band region. Lockheed is expanding K-band studies to meet future radar requirements.

Lockheed expands airborne antenna program

Lockheed's diversified expansion program is causing a major increase in airborne antenna research and development. Antenna design is one of the fastest growing areas at Lockheed, with research and development being applied to: extremely high-speed fighters, advanced jet trainers and jet transports; advanced versions of vertical-rising aircraft, turbo-prop transports, radar search planes (developed and produced exclusively by Lockheed) and a number of significant classified projects.

New positions at Lockheed

The program presents Physicists and Electronic Engineers qualified for airborne antenna design with a wide range of assignments in communication, navigation and microwaves.

In addition to the compensation of challenging work, Lockheed offers you increased salary rates now in effect; generous travel and moving allowances; an opportunity to enjoy Southern California life; and an extremely wide range of extra employee benefits which add approximately 14% to your salary in the form of insurance, retirement pension, etc.



Electronics Research Engineer Irving Alne records radiation antenna patterns. Twenty-two foot plastic tower in background eliminates ground reflections, approximates free space. Tower is of Lockheed design, as are pattern integrator, high gain amplifier, square root amplifier, logarithmic amplifier.



E. O. Richter, Electronics Research department manager (seated), **W. R. Martin**, antenna laboratory group engineer (standing), and **J. L. Rodgers**, electronics research engineer, discuss design of corrugated surface antenna.

Lockheed
AIRCRAFT CORPORATION
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qualified in this area...

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Scientific and Engineering Staff

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ERA DIVISION

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Positions Open

(Continued from page 138A)

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Television engineers for civilian production of both new color and conventional black and white sets. Development and manufacturing engineers are needed. Employer pays our fee and interviewing and relocation expenses. Write Guilford Personnel, 308 American Bldg., Baltimore, Md.

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Wanted for an intriguing job. *Essentials:* knowledge of military electronics; an interest in a skill at writing technical information; freedom to travel; ability to communicate with other engineers; organizing power and punch. *Opportunity:* this job will make your reputation in an important phase of electronics. *Location:* New York City. Box 804.

(Continued on page 143A)



needs

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Engineering Employment Manager

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- Microwave Applications
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Boston, Mass.

Randall A. Kenyon
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SYLVANIA 
ELECTRIC PRODUCTS INC.

Your inquiries will be answered within two weeks.

Creative Engineering Opportunities with Republic

Research

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Familiar with airborne electronic equipment (communications, navigation I.F.F., Radar and Autopilots), preferably with 2 to 4 years aircraft experience. Should be a college graduate. Duties will include system investigations, establishing test procedures and conducting environmental tests on airborne electronic equipment and components.

Computer Engineer

To supervise maintenance and to design special circuitry for computers. Experience with either analogue or digital computers required. College graduate preferred.

Senior Power Plant Engineer

Three to eight years aircraft power plant experience. Capable conducting power plant testing in conjunction with jet engine and induction system analysis. B.S. in M.E. or A.E.

Antenna Engineer

To conduct pattern studies, design prototype antennas and supervise flight tests of new antenna installations. College graduate in Physics, Math or E.E.

Electronic Instrumentation Engineer

Three to five years aircraft instrumentation experience required. Knowledge of transducers, amplifiers and recording equipment used in experimental research testing of hi-speed jet aircraft is essential. Knowledge of servo loop theory as applied to aircraft systems coupled with ability to properly instrument, record and analyze is desirable. Graduate with E.E. degree preferred.



Please address complete resume, outlining details of your technical background, to:
Mr. R. L. Bortner
Administrative Engineer

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(Continued from page 141A)

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Experience in semiconductors preferred; Ph.D. level or equivalent. Salary will reflect applicant's ability, education and experience, and will satisfy the professional man. All replies confidential. Reply: P.O. Box 407, Waltham 54, Mass.

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Experience in germanium essential, silicon experience in addition preferred. Good salary for competent man. All replies confidential. Reply: P.O. Box 407, Waltham 54, Mass.

ENGINEERS

Engineers, interesting, diversified assignments on missile systems development, airborne digital computers, missiles stability and control, radar guidance. Sperry Gyroscope Co. (Division of the Sperry Corp.). Great Neck, Long Island, N.Y.

(Continued on page 145A)



ELECTRONIC ENGINEERS or PHYSICISTS

Experience in communications, navigation, and/or radar systems desirable. A minimum of three years' experience is required.

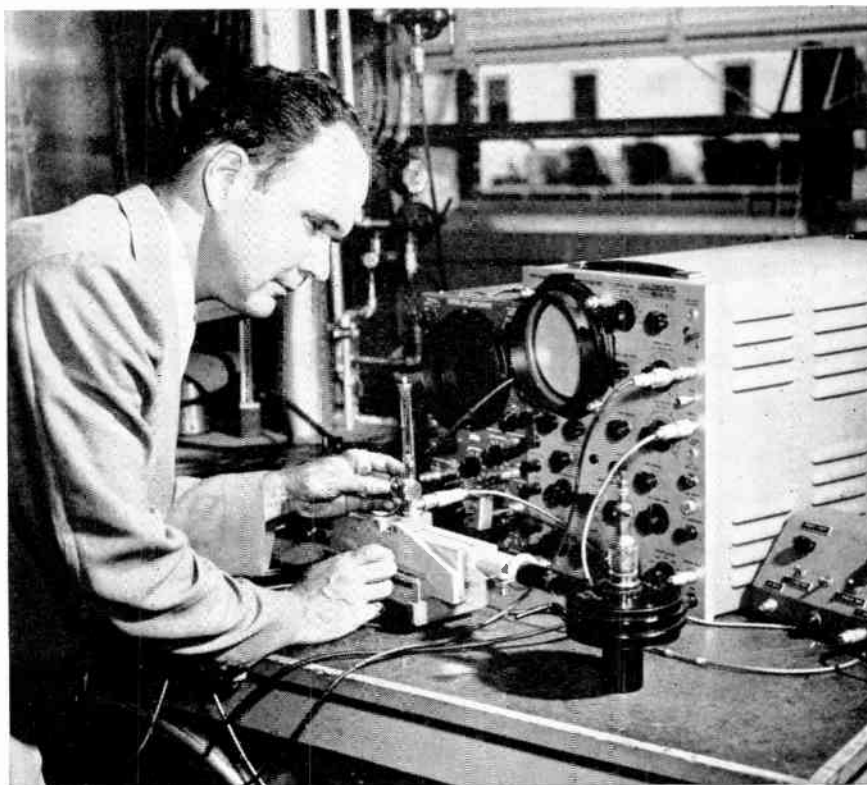
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Men with interests in the above and related fields should submit resumes to Mr. H. F. Ware. Expenses will be paid for qualified applicants who come for interviews. We regret we can consider only U. S. citizens.

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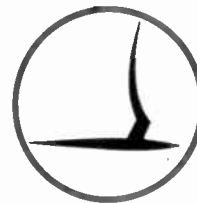
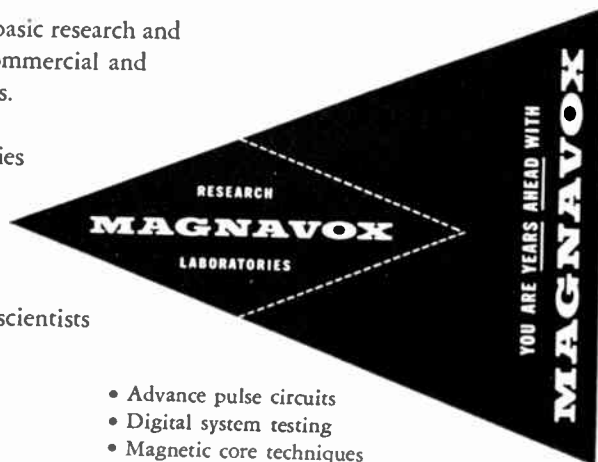
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Aircraft Instrumentation

Radar

Computers

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Must be a competent Electronics Engineer with an avocation for technical writing. Writing will include specifications, test procedures, operating procedures and reports.

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Positions Open

(Continued from page 143A)

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The U. S. Naval Postgraduate School has need for an electronics engineer in the computer laboratory for work with analog and digital computers. Opportunity to learn programming and coding and continue graduate studies. Electronics experience and mathematical interest necessary. Annual salary \$5,060 to \$6,940. Reply: Dept. of Mathematics and Mechanics, USNPS, Monterey, Calif.

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Graduate engineers with experience on tubeless regulated power supplies and magnetic amplifiers are needed. Write company complete resume, or phone Philip Diamond, President, Perkin Engineering Corp., 345 Kansas Street, El Segundo, Calif., OREGON 8-7215.

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their feet
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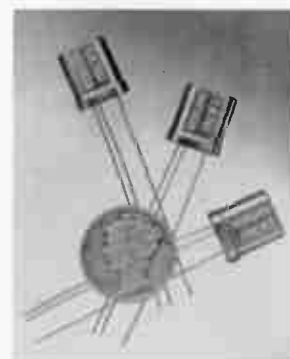


News-New Products

These manufacturers have invited PROCEEDINGS readers to write for literature and further technical information. Please mention your I.R.E. affiliation.

(Continued from page 121A)

Transistors



Raytheon Manufacturing Co., Receiving and Cathode-Ray Tube Operations, 55 Chapel St., Newton 58, Mass., announces three rf fusion-alloy germanium transistors, types CK760, CK761 and CK762 with alpha cutoff frequencies of 5, 10 and 20 mc, respectively. All are hermetically sealed and use a polarized lead arrangement for ease in socketing. Collector capacity for each type averages 14 μf , and extrinsic base resistance for each is about 75 ohms. Further information may be obtained from Technical Information Service, Raytheon.

(Continued on page 149A)

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----- or -----

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National invites engineers who are "Tuned to Tomorrow" to apply now for the following positions:

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MECHANICAL ENGINEERING
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You will participate in the research, development and design of:

COMMUNICATIONS SYSTEMS—Microwave components, transmitters, radar & terminal equipment

COMPUTING DEVICES—Digital & analog techniques

AUDIO DEVICES—Tape recording, high fidelity amplifiers, tuners

INSTRUMENTATION—Sensing elements, high vacuum techniques, specialized circuits & frequency multipliers

RECEIVERS—Specialized diversity, low noise & microwave

COMPONENTS—Hardware, capacitors, inductors

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RESULT: *Important, permanent career opportunities for electronics engineers and radar technical representatives.*

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Less well known is the fact that Gilfillan holds prime long-range research and development contracts in many electronics fields. This is due to the Gilfillan ability to approach a whole problem and arrive at a whole solution in the field of counter measures and guided missiles.

One that can now be announced is Gilfillan's prime contract for the complete guidance system of the Army's "Corporal." Others concern more advanced systems for the missiles of tomorrow. Still other classified projects deal with advanced and unsolved techniques for all branches of the military services.

This adds up to unusually advantageous career opportunities for electronics engineers in Southern California and good openings for radar technical representatives overseas. More details about this pioneer and progressive firm can convince you that your place in the sun is with Gilfillan.

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Personal interview will
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The automation of industrial processes, the elimination of tedious paper work, the safeguarding of human lives and creative energy through split-second sensing, thinking and deciding machines that act with intelligence and discretion are part of the second industrial revolution that is changing the life and work patterns of us all.

ECA's engineers are creating the automatic industrial controls, the electronic business machines, the digital and analog computers that are bringing this revolution into focus day by day. Until they can design a machine that can do it better, these engineers are encouraged to bend their best thoughts to this work in an atmosphere that allows for professional freedom, where there are open channels for the propagation of new ideas, where work executed with imagination is remembered, where there is opportunity to grow in the profession.

As one of the leaders in this change, ECA is daily stretching out into new fields, and enlarging its interest in old ones. Nevertheless, the corporation rests on a sound base of well-established commercial products, which provide the ECA engineer with stability, and assure him of compensation on a high industrial pay scale.

There are now a few positions open for electronic engineers with a good theoretical background and a few years' experience. Address all inquiries to: Mr. W. F. Davis, Dept. 706.



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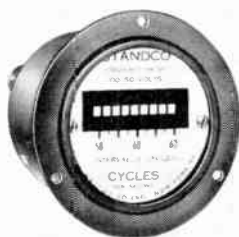


These manufacturers have invited PROCEEDINGS readers to write for literature and further technical information. Please mention your I.R.E. affiliation.

(Continued from page 146A)

New Meters

The Herman H. Sticht Co., Inc., 27 Park Place, New York 7, N. Y., announces their new line of $3\frac{1}{2}$ inch flush panel mounting vibrating reed frequency meters, trade-name, "STANDCO." These instruments are direct-reading frequency meters which are based on the principle of resonance. They consist of a number of steel reeds which are tuned to specific frequencies.



"Standco" panel frequency meters are made in 3 styles of cases, molded bakelite case, metal case and hermetically sealed case. They come with 5, 7, 9, 11, 21, 36 or 41 reeds for normal frequencies of 25, 50, 60 and 400 cps. Other ranges from 15-1500 cps can be supplied. Accuracy of calibration ± 0.5 per cent. Request bulleting 805 from the manufacturer.

Gams Appointed by N. J. Electronics

N. J. Electronics Corp., 345 Carnegie Ave., Kenilworth, N. J., announces the appointment of Theodore C. Gams as Director of Research. He will direct the company's new development program in the field of electronic instruments.



Mr. Gams has been a consultant in industrial electronics, instrumentation, and radar equipment design for the past eight years. Previously he was a lecturer in applied electronics at the Polytechnic Institute in Brooklyn from 1945 to 1952.

(Continued on page 150A)

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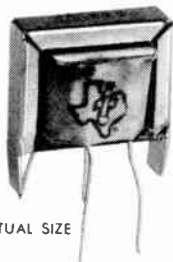
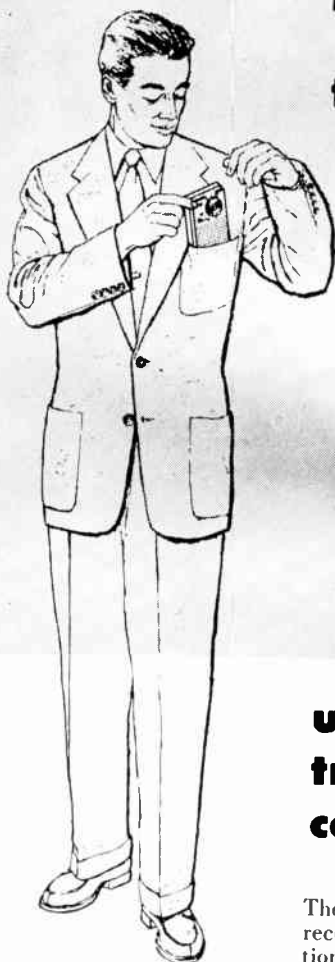
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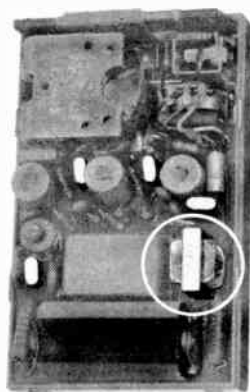
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used in the first transistorized consumer product!

The world's smallest commercial radio receiver makes the most of miniaturization possibilities with a Texas Instruments subminiature transformer and four TI transistors. TI subminiature transformers, such as the one used in the Regency pocket radio, are adaptable to mass production dip-soldering assembly techniques.

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Don't delay your own product miniaturization program. Write today for Bulletin DL-C 424, describing TI subminiature transformers in detail.



Rear view of pocket radio with back removed, showing TI transformer and transistors in relation to other circuit components.



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News-New Products

These manufacturers have invited PROCEEDINGS readers to write for literature and further technical information. Please mention your I.R.E. affiliation.

(Continued from page 149-1)

Hermetic-Seal Bushing

A new rivet-type, hermetic-seal bushing which meets MIL-T-27 specifications and conforms to the MIL-T-27 Twist Test, has been announced by the Helder Bushing & Terminal Co., Inc., through its sales agent Helder Manufacturing Corp., 238 Lewis St., Patterson, N. J.

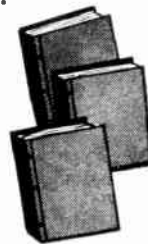


Its insulation resistance, at 45 per cent relative humidity at seal level, is over 500,000 megohms. The manufacturers claim that these terminals can be supplied and installed at a lower price than solder seal terminals of equivalent rating. They further claim they will out-perform any terminal made today. These bushings are available in 5 standard styles, or can be modified to meet the customer's requirements.

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"E" PLANE Bends, 90 deg. flex flanges. **\$7.50**

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MITRED ELBOW, Cast aluminum, 1 1/4" x 3/4" W.G. W.E. Flanges, "E" Plane. **\$3.50**

FLEX. WAVEGUIDE SECTION, 1 ft. long, With UG 40 UG 39 flanges, Attenuation is less than 0.1 db. at 9375 mc, and VSWR is less than 1.02. **\$2.50**

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7-5E3-1-200-67P, 7.5 KV "E" Circuit, 1 microsec. 200 PPS, 67 ohms impedance 3 sections **\$7.50**

7-5E4-16.60, 67P, 7.5 KV "E" Circuit, 4 sections 1 microsec, 60 PPS, 67 ohms impedance. **\$15.00**

7-5E3-3-200-67P, 7.5 KV, "E" Circuit, 3 microsec. 200 PPS, ohms imp 3 sections **\$12.50**

H-616 10KV, 2.2 usec., 375 PPS, 50 ohms imp., **\$27.50**

H-615 10KV, 0.8 usec., 750 PPS, 50 ohms imp., **\$27.50**

KS8865 CHARGING CHOKES: 115-150 H @ .02A, 32 -401H @ .08A, 21KV Test. **\$37.50**

G.E. 25E5-1-350-50 P2T, "E" SKT, 1 Microsec. Pulse @ 350 PPS 50 OHMS Impedance. **\$69.50**

KS9623 CHARGING CHOKES: 161H @ 75 MA, 380 Ohms DCR, 9000 Vac test. **\$14.95**

G.E. 653, 5-2000 50 P2T: 6 KV, "E" Circuit 0.5 usec /2000 PPS/50 ohms/2 sections **\$7.50**

PULSE TRANSFORMERS

GE 2K2748-A, 0.5 usec @ 2000 Pps, Pk. Pwr. out is 32 KV, Impedance 40:100 ohm, 171 volts 2.3 KV Pk. Sec. volts 11.5 KV Pk. Bifilar rated at 1.3 Amp. Fitted with magnetron well. **\$24.50**

K-2745 Primary: 3.1/2.8 KV 50 ohms Z. Secondary: 14/12.6 KV 1025 ohms Z. Pulse Length: 0.25/1.0 usec @ 600/600 PPS, Pk. Power 200/150 KV. Bifilar: 1.3 Amp. Has "built-in" magnetron well. **\$32.50**

K-2461-A, Primary: 3.1/2.6 KV—50 ohms (line), Secondary 14/11.5—1000 ohms Z. Pulse Length: 1 usec @ 600 PPS, Pk. Power Out: 200/130 KV. Bifilar: 1.3 Amp. Fitted with magnetron well **\$29.50**

I. F. AMPLIFIER STRIPS

Model 15: 30 Mc center frequency, Bandwidth 2.5 Mc, gain figure: 65 db, Uses 5 stages of 6AC7's. Has D. C. Restorer and Video Detector, A. F. C. Strip included. Input impedance: 50 ohms, Less tubes **\$17.50**

60 MC, Miniature IF strip, using 6AK5's 60 Mc center Freq. Gain: 95db at Bandwidth of 2.7 Mc, New, Complete with tubes. **\$17.50**

DELAY NETWORKS

D-168184: 0.5 usec. up to 2000 PPS, 1800 ohms **\$4.00**

D-170499: 0.25/5/15 usec, 8 KV., 50 ohms **\$8.50**

D-165997: Delay 1.25 usec **\$2.50**

RCA 2255686-502: 1.7 usec, 1100 ohm impedance **\$2.00**

D-162311: Delay of 0.5 usec, 72 ohms with 4 MC, Bandwidth **\$4.75**

D-168435: Delay 0.5 usec, 555 ohms, 5mc, BW **\$1.50**

D-172578: 416 ohms imp., 0.22 usec, Delay **\$4.75**

D-150979: Oscillating network, Oscillates at 81,955 kc. While "email current" 100ma is interrupted, Has built-in temperature control for stability. Assembled in slide ad can 3" x 3" x 3" Diam. **\$4.35**

MAGNETRONS

Type	Peak Range (MC)	Peak Power Out (KW)	Duty Ratio	Price
2121A	3345-9405	50		\$ 8.75
2122	3267-3333	265		7.50
2126	2992-3019	275	.002	7.40
2127	2965-2992	275	.002	13.50
2129	2914-2939	275	.002	44.95
2131	2820-2860	285	.002	21.50
2132	2780-2820	285	.002	21.50
2138*	3249-3263	5		8.50
2139*	3267-3333	50	.001	8.50
2148	9100-9320	50	.001	21.50
2149	9000-9160	50	.001	54.50
2156*	9215-9275	50	.001	132.50
2162*	2914-3010	35	.002	32.50
3131	24-27 KMC	50	.001	85.95
4134	2740-2780	900	.001	87.50
4138	3550-3600	750	.001	125.00
4142*	670-730	30	.001	16.50
5123	1044-1056	475	.001	49.50
700B	690-700	40	.002	22.50
700D	710-720	40	.002	39.75
706EY	3038-3069	200	.001	32.50
706CY	2976-3007	200	.001	32.50
QK259*	2700-2900	100	CW	65.00
QK60*	2840-3005	800	CW	249.50
QK61*	2975-3170	100	CW	65.00
QK62*	3135-3350	100	CW	65.00

*Packaged with magnet.
†Tunable over indicated range.

TEST EQUIPMENT

- L&N #1553 RATIO BOXES \$275
- ESTERLINE-ANGUS RECORDING MILLIAMMETERS, 0-1 MA \$155

400 CYCLE TRANSFORMERS

(All Primaries 115V, 400 Cycles)		
KS13101	6.3V/15A, 6.3V/0.9A, 6.3V/0.4A, 6.3V/0.2A	\$ 3.85
KS13104	1450VCT/0.283A, 1050VCT/0.217A	7.50
KS9615	6.3V/4A, 3V/1A	1.57
KS9318	6.3V/4A, P/O R-55/ARQ-9	1.35
KS9608	123V/35MA, 1140VCT/0.7A	5.79
352-7102	6.3V/2.5A	1.45
M-7472426	1450V/1.0MA, 2.5V/75A, 6.4V/3.9A, 5V/2A, 6.5V/3A, P/O 1D-39/ APG-13	4.95
352-7039	640 VCT @ 380MA, 6.3V/9A, 6.3V/6A 5V/6A	5.49
702724	9800/8600 @ 32MA	8.95
KS5681	5000V/290MA, 5V/10A	22.50
KS9607	730VCT/1.77A, 1700VCT/1.77A	6.79
352-7273	700VCT/350A, 6.3V/0.9A, 6.3V/2.5A, 6.3V/0.8A, 6.3V/0.4A	6.95
352-7070	2x2.5V/2.5A (2KV TEST) 6.3V/2.25A, 1200/100/750V, @ .005A 1140/1.75MA, 2.5V/1.75A, 2.5V/1.25A—5KV Test	7.45
352-7196	320VCT/50MA, 4.5V/3A, 6.3V/CT 20A, 6.3V/3VCT/6A	3.95
352-7176	2.5/1.75A, 6.3V/2A—5KV Test	4.75
RA6100-1	13V/9A	2.39
901699-501	2.77V @ 4.25A—10KV Test	2.49
901698-501	900V/75MA, 100V/0.1A	3.45
UX8855C	900VCT/0.67A, 5V/3A	3.79
RA6405-1	800VCT/85MA, 5VCT/3A	3.69
T-4862	900VCT/805MA, 5V/3A, 6V/1.75A	4.25
352-7098	2500V 6MA, 300VCT/135MA	5.95
KS9336	110V/50MA TAPPED 625V 2.5V/5A	3.95
M-7474319	6.3V/2.7A, 6.3V/66A, 6.3VCT/21A	4.25
KS8984	27V/4.3A, 6.3V/2.9A, 1.25V/0.2A	2.95
52C080	650VCT/50MA, 6.3VCT/2A, 5VCT/2A	3.75
32332	400VCT/35MA, 6.4V/2.5A, 6.4V/15A	3.85
68G631	150-0-1150V 2MA	2.75
80G198	6VCT/0.0006 KVA	1.75
302433A	6.3V/9.1A, 6.3VCT/6.5A, 2.5V/3.5A, 2.5/3.5A	4.85
KS 9445	592VCT/118MA, 6.3V/8.1A, 5V/2A	5.39
KS9885	6.4/7.5A, 6.4V/3.8A, 6.4/2.5A	4.79
70G3061	300VCT/36MA	2.65

DYNAMOTORS

TYPE	VOLTS	INPUT AMPS	OUTPUT VOLTS	AMPS	Price
35X-059	19	3.8	405	.095	\$4.35
PO5X-15	14	2.8	220	.08	8.95
DA-7A	28	27	1100	.400	15.00
DM 33A	28	7	540	.250	3.95
23350	27	1.75	285	.075	3.95
B-19	12	9.4	275	.110	6.95
			500	.050	
			300	.260	
DA-3A*	28	10	150	.010	6.95
			14.5	5.	
PE 73 CM	28	19	1000	.350	17.50
BD 691	14	2.8	220	.08	8.95
DAG-33A	18	3.2	450	.06	2.50
DM 351	12	2.3	250	.05	6.95
BDAR 93	28	3.25	375	.150	6.95

* Less Filter.
† Used, Excellent.
PE 94-, Brand New **\$5.95**

INVERTERS

800-1B Input 24 vdc, 62 A. Output: 115 V, 800 cy, 7A, 1 phase. Used, excellent **\$18.75**

PE-218H: Input: 25/28 vdc, 92 amp. Output: 115V 350/500 cy 1500 Volt-ampere. **NEW \$32.50**

PE206: Input: 28 vdc, 36 amps. Output: 80 V 800 cy, 500 volt-amp. Dim. 13 x 5 1/2 x 10 1/2. New **\$22.50**

EICOR—ML 3011-5: Input: 13.75V; 18.4A. Output: 115 V/400—, 30, 0.95 PF. New **\$35**

PU 7/AP: Input: 28 vdc/160A. Output: 115 VAC, 400—, 100, 2500 VA., 21.6 Amp. Volt. and Freq. Reg. Used, Exc. **\$75**

MICROWAVE ANTENNAS

3 cm. Horn, 1" x 1 1/2", with twist and 180 deg. bend. With dielectric window. As shown **\$22.50**

AT49/APR—Broadband Conical, 3300 MC, Type N Feed **\$8.95**

Discone Antenna, AS 125 A/H, 1000-3200 mc. Stub supported with type 729 Connector. **\$14.50**

AS114/AP, 10 CM pick up dipole assy, complete w/length of coax and "N" connectors. **\$1.51**

AS46A/APG-4 Yagi Antenna, 5 element array **\$22.50**

30" Parabolic Reflector Spun Aluminum dish **\$4.85**

AN/APA-12—Sector Scan adaptor for APS-2 radar—Complete Kit **\$37.50**

TFS-3, 10 P, Dish, "Chicken Wire" Parabol. Extremely lightweight, portable **\$125.00**

AN-154 3 vertical dipoles working against a rectangular mesh approx. 3'x4', Freq. 140-200 mc. with lobing switch (115V, 60 cy) and portable slatted crate. Extremely rugged **\$27.95**

LP-24 Alford loop, for use with slide-path transmitters (MRN-1, etc.) 100-108 mc. **\$32.50**

POWER TRANSFORMERS

COMBINATION—115V/60 ~ INPUT

CT-133	150-0-150V/65MA, 6.3V/2.5A, 6.3V/0.6A	\$1.79
CT-127	900V/25MA PK. 5V/2A, 2V/7.5A	2.79
CT-008	350-0-350V/120MA, 5VCT/3A, 2.5VCT/12.5A, 2.5VCT/3.5A	4.39
CT-965	78V/0.6A, 6.3V/2A	1.95
CT-004	350-0-350V/90MA, 5VCT/3A, 2.5VCT/12.5A	4.60
CT-002	350-0-350V/50MA, 5VCT/2A, 2.5VCT/7.5A	3.65
CT-479	700V/0.18A (2 X Ind. V. Test) 2.5V 5A/17.80V V. Test	29.50
CT-013	450-0-450V @ 200MA, 10V/1.5A, 2.5V 3.5A 5V/3A	6.95
CT-403	350VCT .026A 5V/3A	2.75
CT-931	585VCT .086A 5V/3A, 6.3V/6A	4.25
CT-929	4200V/.001A, 2.5V/2A, 6.3VCT/6A	5.35

PLATE—115V/60 ~ INPUT

PT 07	400VCT/4.0 AMPS For RA43	\$17.50
PT 034	125V 15MA	1.15
PT 157	660-0-660VAC (500VDC) or 550-0-550VAC (400VDC) at 250 MADC	8.70
PT 167	1400-0-1400 VAC (300MADC) or 1175-0-1175 VAC (1000VDC) at 300 MADC	25.50
PT 168	2100-0-2100 VAC (1750VDC) or 1800-0-1800 VAC (1500VDC) at 300 MADC	33.00
PT 371	210-0-210V at 2.12 Amp.	9.45
PT 133	3140/1570V, 2.36KVA	105.00
PT 801	22,000V/234 MA, 5.35 KVA	135.00
PT 521	7500V .06A, Half-Wave	85.00
PT 913	2500V/12 MA H'SLD	4.95
PT 12A	280VCT/1.2A	3.95
PT-38-2	37.5/40V AT 750 MA	2.15

FILAMENT—115V/60 ~ INPUT

FT-140	50CT @ 10A 25KV Test	17.50
FT-157	4V/16A, 2.5V/2.75A	2.95
FT-101	6V/25A	.79
FT-924	5.25A/21A, 2x7.75V/6.5A	14.95

THERMISTORS

D-164699 Bead Type DCR 1525-2550 Ohms @ 75 Deg. F. Coefficient: 2% Per. Deg. Fahr. Max. Current 25 MA AC/DC **\$1.00**

D-167332 Bead Type, DCR is 1525-2550 Ohms. Rated 25 MA at 825-1.175 VDC **\$1.00**

D-167613 Disk Type DCR: 355 Ohms @ 75 Deg. F. P.M. 2.5%, 1 Watt **\$1.00**

D-166228 Disk Type 7120 Ohms @ 60°F, 4220 Ohms @ 80°F, 2590 Ohms @ 100°F, 1640 Ohms @ 120°F. **\$1.00**

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Industry Research Division Information Service

1. Communications

Audio—Broadcast Microwave—Radar

01 Aircraft & Airport Equip.

- 05 ☐ Airborne receivers
- 10 ☐ Airborne transmitters
- 15 ☐ Airport receivers
- 20 ☐ Airport transmitters
- 25 ☐ GCA equipment, etc.
- 30 ☐ VOR equip., complete

02 Amateur Equipment.

- 05 ☐ Receiving equipment
- 10 ☐ Transmitters

03 Antennas & Accessories.

- 05 ☐ AM broadcast transmitting
- 10 ☐ All-wave receiving
- 15 ☐ Amateur antennas, accessories & rotators
- 20 ☐ Dummy
- 25 ☐ FM broadcast transmitting
- 30 ☐ Microwave
- 35 ☐ Mobile
- 40 ☐ Radar reflectors
- 45 ☐ Rotators
- 50 ☐ Supporting towers
- 55 ☐ TV broadcast transmitting
- 60 ☐ TV multiple systems
- 65 ☐ Tower lighting equipment
- 70 ☐ Tower erection & maintenance service

04 Broadcast Receivers.

- 05 ☐ AM broadcast
- 10 ☐ AM shortwave
- 15 ☐ FM broadcast
- 20 ☐ TV broadcast — B & W
- 25 ☐ TV broadcast — Color
- 30 ☐ TV closed circuit

05 Broadcast Transmitters.

- 05 ☐ AM broadcast
- 10 ☐ FM broadcast
- 15 ☐ Relay broadcast & STL equip.
- 20 ☐ TV broadcast
- 25 ☐ TV closed circuit

06 Carrier Current Equipment

- 05 ☐ Complete terminal equip.
- 10 ☐ Installation & maintenance
- 15 ☐ Repeat & relay equip.

07 Emergency Communications Equipment.

- 05 ☐ Despatching equipment
- 10 ☐ Walkie-talkies

08 Facsimile Equipment.

- 05 ☐ Receiving equipment
- 10 ☐ Sensitized paper
- 15 ☐ Transmitting equipment

09 Industrial Sound Systems.

- 05 ☐ Complete systems
- 10 ☐ Installation & rental contractors

10 Marine Equipment.

- 05 ☐ Contract installers
- 10 ☐ Pleasure craft radiotelephones
- 15 ☐ Shipborne communications equip.

11 Mobile Equipment.

- 05 ☐ Central station receivers
- 10 ☐ Central station transmitters
- 15 ☐ Citizen radio
- 20 ☐ Mobile receivers
- 25 ☐ Mobile transmitters
- 30 ☐ Selective calling equipment

12 Monitor Equipment.

- 05 ☐ Frequency
- 10 ☐ Modulation
- 15 ☐ Remote control monitors
- 20 ☐ Video monitors

13 Radar-Microwave Receivers.

- 05 ☐ Microwave relay
- 10 ☐ Position indicators
- 15 ☐ Radar

14 Radar-Microwave Transmitters.

- 05 ☐ Microwave
- 10 ☐ Microwave relay
- 15 ☐ Radar

15 Recording Equipment.

- 05 ☐ Disk recorders
- 10 ☐ Embossing types of recorders
- 15 ☐ Magnetic tape recorders
- 20 ☐ Magnetic wire recorders

16 Studio Equipment.

- 05 ☐ Consoles & speech input equip.
- 10 ☐ TV cameras
- 15 ☐ TV camera control equipment
- 20 ☐ TV film projectors
- 25 ☐ TV studio lighting equipment
- 30 ☐ Turntables

17 Telegraph & Teleprinter Equipment.

- 05 ☐ Frequency shift converters
- 10 ☐ Frequency shift keyers
- 15 ☐ Receivers
- 20 ☐ Transmitters

2. Components

18 Amplifiers.

- 05 ☐ DC amplifiers
- 10 ☐ Decade
- 15 ☐ Dynamic noise suppressors
- 20 ☐ Geophysical
- 25 ☐ Hearing Aid
- 30 ☐ High fidelity
- 35 ☐ IF amplifiers
- 40 ☐ Inter-coms
- 45 ☐ Magnetic
- 50 ☐ Microwave
- 55 ☐ Phono Pre-Amplifiers
- 60 ☐ Power
- 65 ☐ Public Address
- 70 ☐ RF amplifiers
- 75 ☐ Recording
- 80 ☐ Servo
- 85 ☐ Strain gage
- 88 ☐ TV amplifiers
- 90 ☐ TV boosters
- 92 ☐ Transistor
- 94 ☐ Ultrasonic
- 96 ☐ Wideband

19 Antenna Accessories.

- 05 ☐ Feeder line
- 10 ☐ Insulators
- 15 ☐ Lightning arrestors
- 20 ☐ Mounting hardware
- 25 ☐ Small towers

20 Attenuators.

- 05 ☐ Decade
- 10 ☐ Impedance matching
- 15 ☐ Microwave
- 20 ☐ Volume controlling

21 Batteries.

- 05 ☐ Dry batteries
- 10 ☐ Mercury batteries
- 15 ☐ Storage batteries — lead-acid
- 20 ☐ Storage batteries — nickel-alkali
- 25 ☐ Storage batteries — silver-zinc

22 Blowers & Cooling Fans.

- 05 ☐ Air wheels & housings
- 10 ☐ Complete blower assemblies
- 15 ☐ Fan blades

23 Cabinets, Consoles, & Enclosures.

- 05 ☐ Assembly enclosures
- 10 ☐ Metal cabinets
- 15 ☐ Metal relay rack panels
- 20 ☐ Moulded plastic cabinets
- 25 ☐ Wood cabinets

24 Capacitors: Fixed.

- 05 ☐ Ceramic
- 10 ☐ Decade
- 15 ☐ Electrolytic
- 20 ☐ Feed through
- 25 ☐ Fixed composition
- 30 ☐ Gas filled
- 35 ☐ Glass dielectric
- 40 ☐ High voltage
- 45 ☐ Metallized paper
- 50 ☐ Mica
- 55 ☐ Oil filled
- 60 ☐ Paper
- 65 ☐ Plastic insulated
- 70 ☐ Power factor
- 75 ☐ Precision air
- 80 ☐ Printed circuit
- 85 ☐ Pulse forming networks
- 90 ☐ Silvered mica
- 95 ☐ Tantalum
- 98 ☐ Vacuum

25 Capacitors: Variable.

- 05 ☐ Neutralizing
- 10 ☐ Precision variable
- 15 ☐ Temperature compensating
- 20 ☐ Trimmers, air
- 25 ☐ Trimmers, ceramic
- 30 ☐ Trimmers, glass
- 35 ☐ Trimmers, vacuum
- 40 ☐ Tuning

26 Chassis & Racks.

- 05 ☐ Aluminum chassis
- 10 ☐ Chassis slides
- 15 ☐ Racks, aging
- 20 ☐ Relay racks & cabinets
- 25 ☐ Steel chassis

27 Coils.

- 05 ☐ Audio frequency chokes
- 10 ☐ Deflection yokes
- 15 ☐ Filter
- 20 ☐ Flyback
- 25 ☐ Focusing coils
- 30 ☐ Power chokes
- 35 ☐ RF & IF chokes
- 40 ☐ Toroids
- 45 ☐ Tuning

28 Connectors.

- 05 ☐ AN & MIL standard types
- 10 ☐ Co-axial
- 15 ☐ Hermetically sealed
- 20 ☐ Microphone
- 25 ☐ Multiple circuit connectors
- 30 ☐ Power
- 35 ☐ Slip ring and commutating connectors
- 40 ☐ Sub-miniature
- 45 ☐ Waterproof

29 Converters.

- 05 ☐ AC to DC
- 10 ☐ DC to AC

- 15 ☐ Inverters
- 20 ☐ Rotary
- 25 ☐ Vibrators
- 30 ☐ Voltage

30 Cores.

- 05 ☐ Ceramic
- 10 ☐ Ferrites
- 15 ☐ Powdered metal
- 20 ☐ Tape wound
- 25 ☐ Toroids

31 Crystals & Accessories.

- 05 ☐ Crystal holders
- 10 ☐ Crystal ovens
- 15 ☐ Germanium
- 20 ☐ Oscillating quartz crystals
- 25 ☐ Piezo electric

32 Delay Lines.

- 05 ☐ Decade
- 10 ☐ Distributed constant
- 15 ☐ Distributed parameter
- 20 ☐ Lumped constant
- 25 ☐ Ultrasonic

33 Equalizers.

- 05 ☐ Dialogue equalizers
- 10 ☐ Line equalizers
- 15 ☐ Magnetic recording
- 20 ☐ Recording diameter equalizers
- 25 ☐ Sound effect equalizers

34 Filters.

- 05 ☐ Antenna
- 10 ☐ Band pass & band rejection
- 15 ☐ High pass
- 20 ☐ Loudspeaker dividing networks
- 25 ☐ Mechanical
- 30 ☐ Microwave
- 35 ☐ RF noise reduction
- 40 ☐ TV noise reduction
- 45 ☐ UHF & VHF
- 50 ☐ Variable band pass

35 Fuses & Fuse Holders.

- 05 ☐ Fuses
- 10 ☐ Fuse holders

36 Jacks, Jack Fields & Plugs.

- 05 ☐ Jacks
- 10 ☐ Jack fields
- 15 ☐ Patch cords
- 20 ☐ Plugs

37 Loudspeakers & Headphones.

- 05 ☐ Commercial grade loudspeakers
- 10 ☐ Headphones
- 15 ☐ High-frequency loudspeakers
- 20 ☐ Low-frequency loudspeakers
- 25 ☐ Wide-range, 2 & 3 way systems

38 Magnets.

- 05 ☐ Electro
- 10 ☐ Permanent

39 Meters (Indicating Instruments).

- 05 ☐ Ammeters
- 10 ☐ Elapsed time
- 15 ☐ Frequency indicating
- 20 ☐ Galvanometers
- 25 ☐ Micro-ammeters
- 30 ☐ Milli-ammeters
- 35 ☐ Voltmeters
- 40 ☐ Volume level, (DB & VU)
- 45 ☐ Wattmeters & watt hour meters

40 Microphones & Stands.

- 05 ☐ Carbon
- 10 ☐ Ceramic
- 15 ☐ Condenser
- 20 ☐ Crystal

- 25 ☐ Dynamic
- 30 ☐ Velocity
- 35 ☐ Microphone stands

41 Motors & Motor-Generators.

- 05 ☐ Blower & fan motors
- 10 ☐ Dynamotors
- 15 ☐ Frequency changers
- 20 ☐ Motor-generators
- 25 ☐ Phonograph
- 30 ☐ Servo
- 35 ☐ Synchronous motors
- 40 ☐ Timing motors

42 Phonographs, Pick-ups, Record Changers, etc.

- 05 ☐ Pick-ups—crystal
- 10 ☐ Pick-ups—dynamic
- 15 ☐ Pick-ups—magnetic
- 20 ☐ Pick-ups—reluctance
- 25 ☐ Playback arms
- 30 ☐ Record changers
- 35 ☐ Record players
- 40 ☐ Styli-diamond
- 45 ☐ Styli-metallic
- 50 ☐ Styli-sapphire
- 55 ☐ Tape phonographs & playback equipment

43 Pilot Lights & Assemblies.

- 05 ☐ Brackets, jewels, mounts, sockets
- 10 ☐ Incandescent lights
- 15 ☐ Neon lights

44 Power Supplies.

- 05 ☐ High voltage, (kilovolt)
- 10 ☐ Klystron
- 15 ☐ Low voltage
- 20 ☐ Medium voltage
- 25 ☐ Microwave
- 30 ☐ Primary power sources, fuel-engine driven
- 35 ☐ Transistor
- 40 ☐ Voltage regulated power sources

45 Printed Circuits.

- 05 ☐ Etched or screened prewired chassis
- 10 ☐ Printed electronic assemblies
- 15 ☐ Solder dipped prewired chassis

46 Recording Accessories.

- 05 ☐ Blank disks
- 10 ☐ Cutting heads
- 15 ☐ Cutting styli
- 20 ☐ Magnetic recording, playback & erase heads
- 25 ☐ Magnetic recording tape-acetate base
- 30 ☐ Magnetic recording tape-paper base
- 35 ☐ Magnetic recording wire
- 40 ☐ Stylus resharpening
- 45 ☐ Suction pumps & equipment
- 50 ☐ Tape splicing equipment

47 Rectifiers.

- 05 ☐ Copper oxide
- 10 ☐ Germanium
- 15 ☐ Selenium diodes
- 20 ☐ Selenium stacks
- 25 ☐ Silicon

48 Rectifiers, Vacuum Tube.

- 05 ☐ Grid controlled
- 10 ☐ Mercury vapor
- 15 ☐ Receiving
- 20 ☐ Transmitting

49 Relays.

- 05 ☐ Co-axial
- 10 ☐ D'Arsonval
- 15 ☐ Differential
- 20 ☐ Frequency selective

- 25 ☐ Hermetically sealed
- 30 ☐ Impulse
- 35 ☐ Keying
- 40 ☐ Latching
- 45 ☐ Mercury contact
- 50 ☐ Polarized
- 55 ☐ Power control & overload
- 60 ☐ Sensitive
- 65 ☐ Stepping
- 70 ☐ Sub-miniature
- 75 ☐ Telephone
- 80 ☐ Time delay

50 Resistors.

- 05 ☐ Carbon fixed, moulded
- 10 ☐ Carbon, variable
- 15 ☐ Deposited carbon
- 20 ☐ Potentiometers
- 25 ☐ Precision film
- 30 ☐ Printed circuit
- 35 ☐ Rheostats
- 40 ☐ Thermistors
- 45 ☐ Vacuum sealed
- 50 ☐ Very high megohm
- 55 ☐ Wirewound, fixed
- 60 ☐ Wirewound, variable
- 65 ☐ Wirewound, precision

51 Semi-Conductors.

- 05 ☐ Germanium diodes
- 10 ☐ Transistors, junction
- 15 ☐ Transistors, point contact

52 Sockets.

- 05 ☐ Capacitor, plug-in sockets
- 10 ☐ Crystal
- 15 ☐ Diode
- 20 ☐ Receiving tube
- 25 ☐ Transistor
- 30 ☐ Transmitting tube
- 35 ☐ UL approved industrial types

53 Switches & Contacts.

- 05 ☐ Band
- 10 ☐ Coaxial
- 15 ☐ Contacts
- 20 ☐ Foot
- 25 ☐ Key
- 30 ☐ Mercury
- 35 ☐ Momentary contact
- 40 ☐ Power
- 45 ☐ Precision snap-acting
- 50 ☐ Rotary
- 55 ☐ Slide
- 60 ☐ Solenoids
- 65 ☐ Spring-return
- 70 ☐ Tap
- 75 ☐ Thermally operated
- 80 ☐ Toggle & push button

54 Thermostats.

- 05 ☐ Bellows
- 10 ☐ Bimetal
- 15 ☐ Hermetically sealed

55 Transformers.

- 05 ☐ Audio
- 10 ☐ Current limiting
- 15 ☐ 400 cps
- 20 ☐ Geophysical
- 25 ☐ Hermetically sealed
- 30 ☐ Horizontal output & flyback
- 35 ☐ Intermediate frequency
- 40 ☐ Isolation
- 45 ☐ Power components
- 50 ☐ Precision matched
- 55 ☐ Pulse
- 60 ☐ Radio Frequency
- 65 ☐ Saturable
- 70 ☐ TV deflection
- 75 ☐ TV impedance matching
- 80 ☐ Toroidal
- 85 ☐ Ultrasonic
- 90 ☐ Variable
- 95 ☐ Voltage regulating

Transistors. See: 51

56 Tuners.

- 05 ☐ AM tuners
- 10 ☐ FM tuners
- 15 ☐ Klystron
- 20 ☐ Microwave
- 25 ☐ TV tuners
- 30 ☐ UHF & VHF

57 Vacuum Tubes, Receiving.

- 05 ☐ Cathode-ray & TV picture
- 10 ☐ Miniature
- 15 ☐ Standard types, glass & metal
- 20 ☐ Sub-miniature

58 Vacuum Tubes, Special Purpose.

- 05 ☐ Electronmeter tubes
- 10 ☐ Geiger-Mueller
- 15 ☐ Phototubes
- 20 ☐ Pirani tubes
- 25 ☐ Thyratrons
- 30 ☐ Voltage regulators & ballasts
- 35 ☐ X-Ray

59 Vacuum Tubes, Transmitting.

- 05 ☐ ATR & TR types
- 10 ☐ Air-cooled power tubes
- 15 ☐ Iconoscopes
- 20 ☐ Image orthicons
- 25 ☐ Klystrons
- 30 ☐ Magnetrons
- 35 ☐ Monoscope tubes
- 40 ☐ Traveling wave
- 45 ☐ Water-cooled power tubes

60 Vibration Controls.

- 05 ☐ Vibration & shock mounts
- 10 ☐ Isolation pads

61 Voltage Regulators.

- 05 ☐ Automatic
- 10 ☐ Manually operated
- 15 ☐ Varistors

62 Waveguides & Accessories.

- 05 ☐ Couplings
- 10 ☐ Flexible waveguides
- 15 ☐ Hybrid junctions
- 20 ☐ Rigid waveguides
- 25 ☐ Sampling lines, reflectors, etc.
- 30 ☐ Switches
- 35 ☐ Transmission lines

3. Controls—Instruments—Test Equipment

63 Audio Frequency Test Equipment.

- 05 ☐ Beat frequency oscillators
- 10 ☐ Distortion & noise analyzers
- 15 ☐ Intermodulation meters
- 20 ☐ Output power meters
- 25 ☐ RC oscillators
- 30 ☐ Spectrum analyzers
- 35 ☐ Square wave generators
- 40 ☐ Telegraph test
- 45 ☐ Voltmeters-audio frequency

64 Automatic Control Equip.

- 05 ☐ Burglar-alarm & protective systems
- 10 ☐ Combustion & smoke- nuisance controls
- 15 ☐ Electronic remote controls
- 20 ☐ Photo electric
- 25 ☐ Positioning equipment
- 30 ☐ Production control, counting & sorting equipment
- 35 ☐ Qualitative controls
- 40 ☐ Quantitative controls
- 45 ☐ Vacuum controls
- 50 ☐ Variable speed controls

65 Computers & Counters.

- 05 ☐ Analog to digital converters
- 10 ☐ Computers, analog
- 15 ☐ Computers, digital
- 20 ☐ Counters, continuous
- 25 ☐ Counters, pre-set
- 30 ☐ Data reduction equipment
- 35 ☐ Drums, magnetic
- 40 ☐ Electronic calculators

66 General Laboratory Equipment & Supplies.

- 05 ☐ Atmospheric test chambers
- 10 ☐ Chemicals
- 15 ☐ General laboratory equipment
- 20 ☐ RF shielding materials
- 25 ☐ Shielded enclosures

67 General Test Equipment.

- 05 ☐ Accelerometers
- 10 ☐ Bridges, capacitance
- 15 ☐ Bridges, impedance
- 20 ☐ Bridges, inductance
- 25 ☐ Bridges, resistance
- 30 ☐ Crystal testers
- 35 ☐ Decades, capacitance
- 40 ☐ Decades, inductance
- 45 ☐ Decades, resistance
- 50 ☐ Diode testers
- 55 ☐ Insulation testers
- 60 ☐ Multimeters
- 65 ☐ Null indicators
- 70 ☐ Ohm & megohm meters
- 72 ☐ Pyrometers
- 74 ☐ Spectrometers
- 76 ☐ Strain gauges
- 78 ☐ Stroboscopes
- 80 ☐ Transistor testers
- 82 ☐ Ultrasonic generators
- 84 ☐ Ultrasonic testers
- 86 ☐ Vacuum tube testers
- 88 ☐ Vacuum tube voltmeters
- 90 ☐ Vibration analyzers

68 Geophysical Apparatus.

- 05 ☐ Complete exploration equipment
- 10 ☐ Control equipment
- 15 ☐ Microphones
- 20 ☐ Recording galvanometers

69 Graphic Recorders.

- 05 ☐ Charts & papers
- 10 ☐ Ink writing
- 15 ☐ Inkless types
- 20 ☐ Fixed speed types
- 25 ☐ Variable speed types

70 Induction Heating Equip.

- 05 ☐ Brazing, soldering & welding
- 10 ☐ Di-electric heating
- 15 ☐ Induction heating

71 Laboratory Standards of Frequency & Time.

- 05 ☐ Elapsed time measuring equip.
- 10 ☐ Primary standards of frequency
- 15 ☐ Primary standards of time
- 20 ☐ Secondary standards of frequency
- 25 ☐ Secondary standards of time
- 30 ☐ Tuning forks

72 Medical Equipment.

- 05 ☐ Cardiographs
- 10 ☐ Diathermy
- 15 ☐ Electronic cauterizing
- 20 ☐ Electro-encephalographs
- 25 ☐ Stimulators

73 Microwave & Radar Test Equipment.

- 05 ☐ Bolometers
- 10 ☐ Frequency meters
- 15 ☐ Marker generators
- 20 ☐ Oscillators
- 25 ☐ Pulse generators
- 30 ☐ Slotted lines & measuring accessories

- 35 ☐ SWR meters
- 40 ☐ Time delay generators
- 45 ☐ Wavemeters

74 Nuclear Equipment.

- 05 ☐ Dosimeters
- 10 ☐ Ionization chambers
- 15 ☐ Pulse count integrators
- 20 ☐ Radiation detectors
- 25 ☐ Scalers
- 30 ☐ Survey meters

75 Oscilloscopes—Cathode-Ray.

- 05 ☐ General purpose
- 10 ☐ Oscilloscope cameras
- 15 ☐ Oscillosynchrosopes
- 20 ☐ Recording oscillographs
- 25 ☐ Special UHF & pulse analysis equipment

76 Radio Frequency Test Equipment.

- 05 ☐ Field strength meters
- 10 ☐ Impedance & admittance meters
- 15 ☐ Marker generators
- 20 ☐ Multi-vibrators
- 25 ☐ Oscillators
- 30 ☐ Phase meters
- 35 ☐ "Q" meters
- 40 ☐ RF power output meters
- 45 ☐ Signal generators, AM
- 50 ☐ Signal generators, FM
- 55 ☐ Sweep generators

77 Servo-Mechanisms.

- 05 ☐ Controls
- 10 ☐ Gears, gear-trains, etc.
- 15 ☐ Servo amplifiers
- 20 ☐ Servo motors & generators

78 Telemetry Equipment.

- 05 ☐ Data recorders
- 10 ☐ Data storage mechanisms
- 15 ☐ FM/FM modulators

79 Television Test Equipment.

- 05 ☐ Synchronizing generators
- 10 ☐ TV calibrators
- 15 ☐ TV marker generators
- 20 ☐ TV signal generators
- 25 ☐ TV sweep generators

4. Materials & Services

80 Cable & Wire.

- 05 ☐ Aluminum
- 10 ☐ Co-axial cables
- 15 ☐ Copper, bare
- 20 ☐ Enamel coated
- 25 ☐ Low-loss, high-frequency types
- 30 ☐ Magnet wire
- 35 ☐ Precious & rare metals wire
- 40 ☐ Pre-formed wire harnesses
- 45 ☐ Rubber insulated
- 50 ☐ Shielded
- 55 ☐ Synthetic insulated
- 60 ☐ Wire mesh
- 65 ☐ Woven wire braid

81 Ceramics.

- 05 ☐ Coil forms
- 10 ☐ Custom fabrication
- 15 ☐ Insulators
- 20 ☐ Proprietary forms

82 Consulting Engineers.

- 05 ☐ Acoustical
- 10 ☐ Electrical
- 15 ☐ Electronic
- 20 ☐ Mechanical
- 25 ☐ Radio
- 30 ☐ Television

83 Core Materials.

- 05 ☐ Laminations
- 10 ☐ Metallic powders

84 Fabricators & Services.

- 05 ☐ Contract fabrication & assembly
- 10 ☐ Electro-plating
- 15 ☐ Embedment
- 20 ☐ Fungus & moisture proofing
- 25 ☐ Hardening & heat-treating
- 30 ☐ Hermetic sealing service
- 35 ☐ Metal stamping & spinning
- 40 ☐ Packaging
- 45 ☐ Plastic fabricators
- 50 ☐ Plastic molders

85 Gasses & Vapors.

- 05 ☐ Acetylene
- 10 ☐ Argon
- 15 ☐ Hydrogen
- 20 ☐ Krypton
- 25 ☐ Neon
- 30 ☐ Oxygen
- 35 ☐ Xenon

86 Hardware & Findings.

- 03 ☐ Adhesive labels & tapes
- 06 ☐ Brushes, motor
- 09 ☐ Bushings & bearings
- 12 ☐ Cable clips
- 15 ☐ Cans
- 18 ☐ Dials & dial assemblies
- 21 ☐ Fasteners
- 24 ☐ Flexible shafts
- 27 ☐ Gaskets
- 30 ☐ Grilles
- 33 ☐ Grommets
- 36 ☐ Lacing cord & tape
- 39 ☐ Machine screw products
- 42 ☐ Markers & tags
- 45 ☐ Nameplates
- 48 ☐ Paper preforms, for coil foundations
- 51 ☐ Retaining rings
- 54 ☐ Rivets
- 57 ☐ Screws, nuts & washers
- 60 ☐ Springs
- 63 ☐ Strain reliefs
- 66 ☐ Terminal boards & straps
- 69 ☐ Terminals, rivet or stud
- 72 ☐ Terminals, wire end
- 75 ☐ Test clips

87 Hermetic Seals.

- 05 ☐ Ceramic to metal seals
- 10 ☐ Complete headers
- 15 ☐ Glass to metal seals

88 Insulating Materials.

- 05 ☐ Cambric sheeting
- 10 ☐ Cambric tubing
- 15 ☐ Ceramic
- 20 ☐ Cotton tapes
- 25 ☐ Fiberglas tapes
- 30 ☐ Fiberglas tubing
- 35 ☐ Glass bonded mica
- 38 ☐ Paper
- 40 ☐ Plastic
- 45 ☐ RF shielding materials
- 50 ☐ Rubber
- 55 ☐ Synthetics

89 Laboratories & Custom Builders.

- 05 ☐ Individual item builders
- 10 ☐ Pilot run plants
- 15 ☐ Research & development labs.
- 20 ☐ Testing & certifying organizations

90 Lacquers, Paints, Compounds & Waxes.

- 05 ☐ Coil dope
- 10 ☐ Conducting paint
- 15 ☐ Finishing & protecting lacquer
- 20 ☐ Fungus preventatives

- 25 ☐ Moisture proofing
- 28 ☐ Phosphors
- 30 ☐ Sealing & potting compounds
- 35 ☐ Silicones

91 Machinery & Tools.

- 05 ☐ Aluminizers
- 10 ☐ Coil winding machines
- 15 ☐ Dehydrators
- 20 ☐ General production machinery & tools
- 25 ☐ Marking & engraving machines
- 30 ☐ Soldering irons & guns
- 35 ☐ Vacuum tube machinery
- 40 ☐ Vacuum pumps
- 45 ☐ Vibration exciters
- 50 ☐ Welders & brazers

92 Metals.

- 05 ☐ Alloys
- 10 ☐ Copper
- 15 ☐ Ferrous
- 20 ☐ Non-ferrous, excluding copper
- 25 ☐ Powdered
- 30 ☐ Precious & rare

93 Moulded Products.

- 05 ☐ Insulators
- 10 ☐ Knobs & parts
- 15 ☐ Proprietary mouldings

94 Plastics.

- 05 ☐ Extrusions
- 10 ☐ Raw powders
- 15 ☐ Rods
- 20 ☐ Sheets
- 25 ☐ Tubes

95 Solder.

- 05 ☐ Acid core
- 10 ☐ Fluxes
- 15 ☐ Plain
- 20 ☐ Precious metal
- 25 ☐ Pre-forms
- 30 ☐ Rosin core

96 Vacuum Tube Parts.

- 05 ☐ Anodes
- 10 ☐ Envelopes, glass
- 15 ☐ Envelopes, metal
- 20 ☐ Getters
- 25 ☐ Grids & grid wire
- 30 ☐ Guns & gun parts
- 32 ☐ Lead wires
- 35 ☐ Phosphors
- 40 ☐ Pins & Prongs
- 42 ☐ Shaft locks
- 45 ☐ Spacers & insulators
- 50 ☐ Stampings
- 55 ☐ Tube caps
- 60 ☐ Tube clamps
- 65 ☐ Tube shields
- 70 ☐ Tubing-cathode

Education & Publishing

97 Education.

- 05 ☐ Technical instruction — home study
- 10 ☐ Technical instruction — resident

98 Publishing.

- 05 ☐ Book publishers
- 10 ☐ Data sheet & manual preparation
- 15 ☐ Magazines

Distributional Functions

99 Distribution.

- 05 ☐ Export & import
- 10 ☐ Jobbers & wholesalers
- 15 ☐ Manufacturers representatives
- 20 ☐ Sales & service
- 25 ☐ Surplus dealers

INDEX AND DISPLAY ADVERTISERS

Meetings with Exhibits	6A
Industrial Engineering Notes	8A
News-New Products	34A
Membership	41A
Section Meetings	92A
Professional Group Meetings	101A
IRE People	108A
Positions Wanted by Armed Forces Veterans	122A
Positions Open	137A

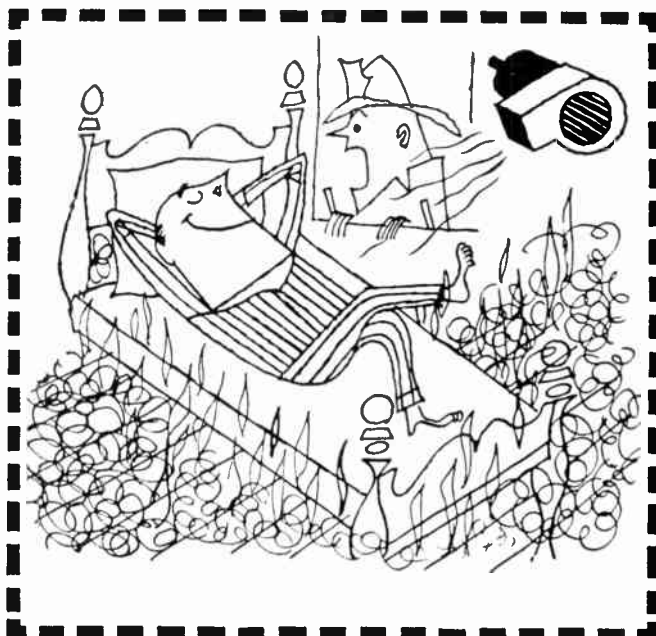
DISPLAY ADVERTISERS

Acme Electric Corporation	102A
Admiral Corporation	42A
Advance Electronics Company	157A
Aerovox Corporation	45A
Air Marine Motors, Inc.	157A
Airborne Instruments Laboratory, Inc.	62A
Airtron, Inc.	56A
Alford Manufacturing Co., Inc.	92A
Allen-Bradley Company	32A
Allied Radio Corp.	116A
American Electrical Heater Co.	107A
American Lava Corporation	61A
American Phenolic Corporation	8A

American Television & Radio Co.	46A
Amperite Company	64A
Arnold Engineering Company	44A

Ballantine Laboratories, Inc.	68A
Barber Laboratories, Alfred W.	151A
Bell Aircraft Corporation (Empl.)	145A
Bell Telephone Laboratories	4A
Bendix Aviation Corp., Guided Missile Section (Empl.)	129A
Bendix Aviation Corp., Pacific Division	73A
Bendix Aviation Corp., Pacific Div. (Empl.) ...	141A
Bendix Aviation Corp., Radio Communication Div. (Empl.)	126A
Bendix Aviation Corp., Red Bank Division	48A
Berkeley Division of Beckman Instruments, Inc.	159A
Bodnar Industries, Inc.	94A
Boeing Airplane Company (Empl.)	133A
Bogue Electric Manufacturing Co.	27A
Brown University (Empl.)	135A
Bussmann Manufacturing Co.	26A

California Institute of Technology, Jet Propulsion Lab. (Empl.)	142A
Cambridge Thermionic Corp.	28A
Capitol Radio Engineering Institute	114A
Carter Motor Company	99A
Chance-Vought Aircraft (Empl.)	138A
Chicago Telephone Supply Corp.	91A
Clarostat Mfg. Co., Inc.	63A
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Cohn Corporation, Sigmund	96A
Communication Accessories Co.	95A
Communication Measurements Lab., Inc.	106A
Communication Products Co.	119A



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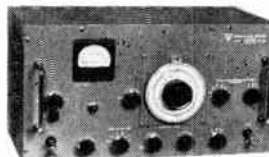
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TYPE
205A

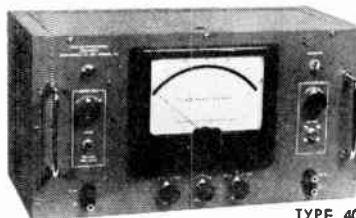
- MEASURES TIME DELAY WITH 1% ACCURACY
- MEASURES PHASE DELAY WITH 1° ACCURACY
- 10 KC TO 15 MEGACYCLES, 0.01 VOLT SENSITIVITY

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PRICE \$585.00

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- 0.3 VOLT SENSITIVITY
- NO AMBIGUITY AT ZERO DEGREE
- 0.25° RELATIVE ACCURACY
- PHASE READING INDEPENDENT OF SIGNAL AMPLITUDES

Type 405 Phase Meter has a frequency range of 8 cps to 100 kc, phase range 0-36, 0-90, and 0-180 degrees; a switch is provided for 180-216, 180-270, and 180-360 degrees. The accuracy is $\pm \frac{1}{4}$ degree relative and 1 degree absolute in any range. The input impedance is 2.7 megohms shunted with 20 uuf on both channels. The meter scale is $6\frac{1}{2}$ " long, thus a fraction of $\frac{1}{4}$ degree can be read easily.
Price \$485. F.O.B., Passaic, N.J. For 0.0001 cps to 1000 cps, use our U-L Phase Counter.

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Communications Equipment Co.152A
Content, Edward J.151A
Convair Div., General Dynamics Corp.
(Empl.)136A, 146A
Cornell Aeronautical Lab., Inc. (Empl.)144A
Cornell-Dubilier Electric Co.Cover 3
Crosby Laboratories, Inc.151A

Daven Company35A
Daystrom Instrument Div., Daystrom, Inc.69A
Detectron Corporation100A
Donner Scientific Co.14A
Driver-Harris Company97A
DuMont Labs., Inc., Allen B. (Empl.)145A
DuMont Labs., Inc., Allen B.87A

ESC Corporation112A
Eastern Industries, Inc.158A
Eitel-McCullough, Inc.40A, 113A
Electrical Industries25A
Electrical & Physical Instrument Corp.60A
Electro-Impulse Laboratory103A
Electro-Measurements, Inc.88A
Electronic Research Associates, Inc.96A, 151A
Electronics Corp. of America (Empl.)148A
Elk Electronic Laboratories, Inc.151A
Engineering Associates84A
Engineering Research Associates, Div. of Remington Rand, Inc. (Empl.)126A, 130A, 140A
Erie Resistor Corporation101A

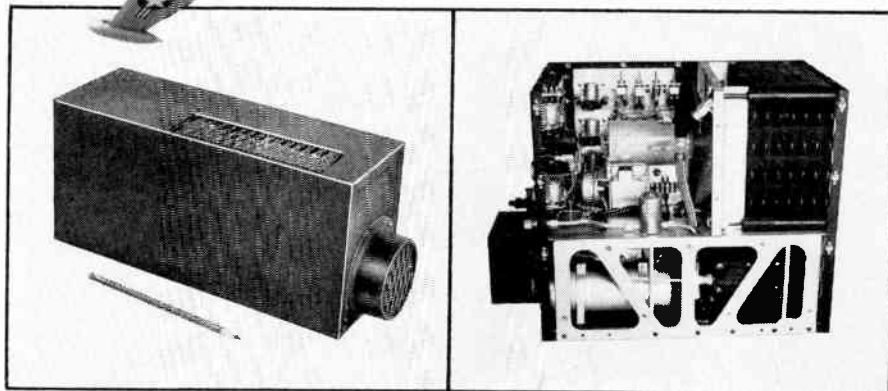
Fairchild Camera & Instrument Corp.66A
Fansteel Metallurgical Corp.43A
Farnsworth Electronics Corp. (Empl.)146A
Federal Telecommunication Labs. (Empl.)131A
Ford Instrument Company58A
Franklin Institute (Empl.)143A
Frederick & Associates, Carl L.150A
Frederick Research Corp.151A
Frequency Standards16A

Garde Manufacturing Co.46A
Garfield Wire Division, The Overlakes Corp.94A
General Ceramics Corp.85A
General Electric Co., Apparatus Dept.10A & 11A
General Electric Co., Capacitor Sales Div.111A
General Electric Co. (Empl.)131A
General Electric Co., Tube Div.54A & 55A, 83A
General Precision Lab., Inc. (Empl.)144A
General Radio CompanyCover 4
Gilfillan Brothers, Inc. (Empl.)147A
Godley Company, Paul151A
Gordon Associates151A
Gudeman Co. of Calif., Inc. (Empl.)138A
Gulton Industries, Inc.81A

Heath Company104A
Heppner Manufacturing Co.72A
Hermetic Seal Products Co.65A
Hewlett-Packard Co.22A & 23A
Hickok Electrical Instrument Co.50A
Highland Engineering Co.151A
Hitemp Wires, Inc.105A
Hoffman Laboratories, Inc. (Empl.)135A
Hoffman Laboratories, Inc.49A
Hogan Laboratories, Inc.151A
Hopkins Engineering Co.84A



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Complete Refrigeration Cooling Systems using various gases and liquids as cooling media in closed-cycle operation, are Eastern specialties. Within the conditions shown at the right, these compact airborne units can be supplied complete with one heat exchanger, or with several exchangers in different locations as a centralized compressor unit.

We welcome inquiries regarding custom made or adapted units which may solve your specific cooling problems, meeting appropriate government specifications.

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Operating range from below 0°F. to 100°F.
Pressurized evaporators available with units.
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Normal aircraft power sources can be used.

Write for data on Eastern's Cooling Unit line, included in Eastern Aviation Catalog No. 330. Related Pressurization Equipment and Hydraulic Products are also described in this catalog.



Eastern

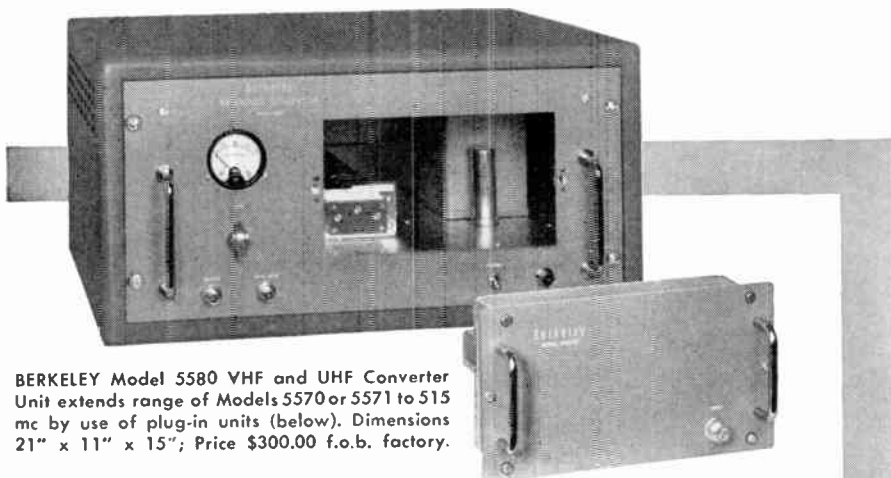


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Measure Frequency to 515 mc

READ IT DIGITALLY, PRINT IT AUTOMATICALLY! Add a Model 5580 VHF-UHF Converter and 1452 Printer (below) to a BERKELEY Frequency Meter*—get the most convenient, inexpensive means yet devised for frequency measurement to 515 mc. Exclusive BERKELEY Modular design uses low cost fixed-band plug-in units in place of costly wide-band amplifiers. Accuracy of measurement is ± 1 cycle, \pm crystal stability (1 part in 10^7).

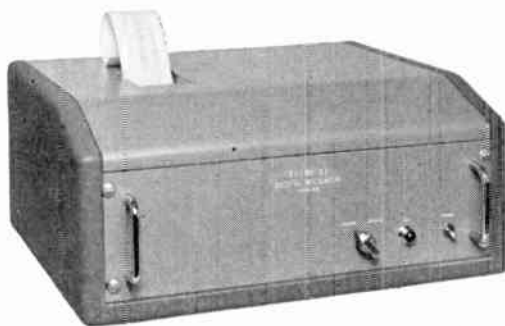
*Model 5580 connects directly to BERKELEY Model 5570 or 5571.



BERKELEY Model 5580 VHF and UHF Converter Unit extends range of Models 5570 or 5571 to 515 mc by use of plug-in units (below). Dimensions 21" x 11" x 15"; Price \$300.00 f.o.b. factory.

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Model 1452 prints 6 digits (8 or 10 on special order) on standard adding machine tape. Only 19" wide x 10½" high x 14" deep, weighs 60 lbs. Price, \$750.00 f.o.b. factory.

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Write for complete specifications and data; please address Dept. N-4

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M-52

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Hughes Research & Development Labs.
.....132A, 140A, 148A
Hughey & Phillips24A
Hycon Manufacturing Co.29A
Hycor Company, Inc.118A

Industrial Research Labs. (Empl.)138A
Institute of Radio Engineers
.....1A, 20A, 98A, 99A, 153A-156A
Interference Testing & Research Lab., Inc.151A
International Business Machines Corp. (Empl.) 127A

Johns Hopkins University, Applied Physics Lab.
(Empl.)136A
Jones Div., Howard B., Cinch Mfg. Corp.86A

Kahle Engineering Company92A
Kahn, Leonard R.151A
Kaiser Metal Products, Inc.90A
Kay Electric Company7A
Kearfott Company, Inc.36A
Keithley Instruments120A
Kessler Company, Frank118A
Kip Electronics Corp.151A
Klipsch & Associates88A
Kollsman Instrument Corp.103A
Kollsman Instrument Corp. (Empl.)132A

Lapp Insulator Company, Inc.76A
Librascope, Inc.149A
Lockheed Aircraft Corp. (Empl.)123A, 139A
Lord Manufacturing Co.89A

Magnavox Research Laboratories (Empl.)144A
Maigne Company, O. J.94A
Mallory & Co., Inc., P. R.70A
Marion Electrical Instrument Co.24A
Martin Company, Glenn L. (Empl.)130A
Measurements Corporation102A, 151A
Melpar, Inc. (Empl.)128A
Microwave Consultants151A
Midland Manufacturing Co., Inc.53A
Millen Manufacturing Co., Inc., James106A
Model Engineering & Mfg. Co., Inc.77A

National Cash Register Co. (Empl.)128A
National Cash Register Co., Electronics Div.
(Empl.)134A
National Company, Inc. (Empl.)147A
Ney Company, J. M.86A
North Electric Mfg. Co.110A
Northern Radio Company, Inc.109A

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Oster Manufacturing Co., John2A
Ostlund, Evert M.151A

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Panoramic Radio Products, Inc.116A
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Pickles, Sidney151A
Plastic Mold & Engineering Co.104A

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Pyramid Electric Company79A

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Radio Materials Corp.93A
Radio Sonic Corporation151A
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Raytheon Manufacturing Co., Receiving Tube Div.
.....18A & 19A, 21A
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Resistance Products Co.78A
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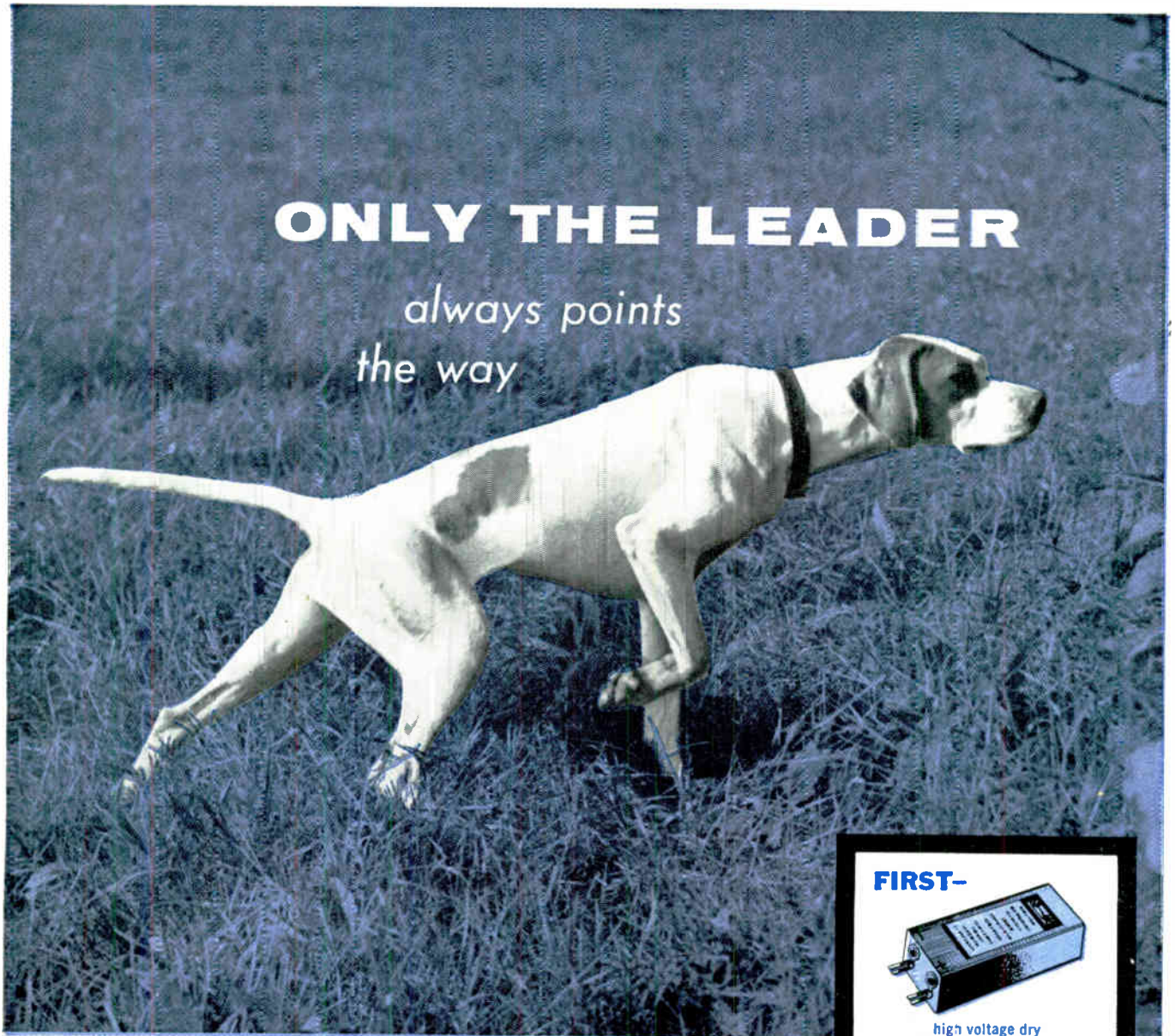
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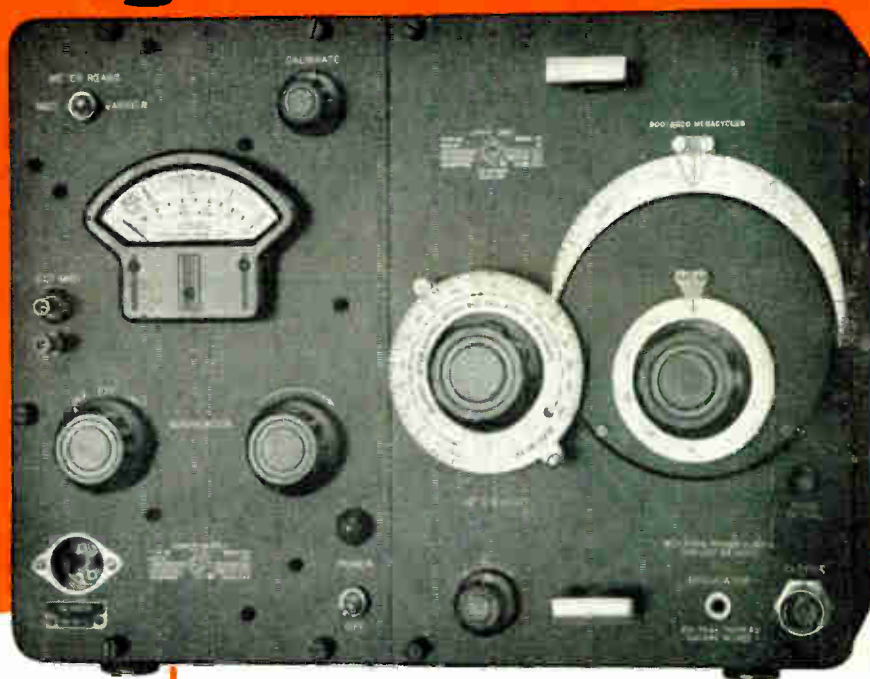
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